Introduction
The proliferation of wireless transceivers in portable applications has led to increased attention to an electronic circuits’ ability to operate in the vicinity of high frequency radio transmitters. For example, in a Bluetooth headset using op amp for the voice band audio, special shielding in conjunction with good layout practice and RF suppression are needed to ensure interference-free operation of the headset. In this example, and in other gigahertz radio systems, the close proximity of the radio antenna to low frequency amplifier sub assemblies, can result in the demodulation of the radio signal causing a disruptive interference in the receiving circuit.

Test Description
Radio transceivers transmit voice and data signals by modulating a high frequency RF carrier signal with a much lower frequency signal. Sensitive circuits located near the antenna (Figure 1) must be designed to reject interference caused by RF signals bleeding into the audio circuits. Layout and proximity to the transmitter can result in several different receiving sites on the circuit board that can cause interference at different frequencies.

This application note describes a test platform using standard equipment found in most high frequency analog labs, that can effectively characterize the interference sensitivity of an amplifier sub circuit.

The test platform described was built to troubleshoot a Bluetooth headset with an excessive noise problem at the headset amplifier output caused by the RF transmitter. Although the noise caused by the Bluetooth transmitter was readily observed at the amplifier output, the frequency hopping RF combined with the complex encoded modulation of the voice signal produced an interference signal that was too complex to analyze.

The test platform re-creates the interference environment by replacing the complex Bluetooth signal with a swept RF frequency source modulated by a 1kHz modulation signal. The 1kHz modulation signal is used to track the source of RF entry and the signal path to the audio amplifier’s output. This enables the user to observe the interference under closely controlled conditions of the RF carrier field strength, carrier frequency and modulation frequency.

Test Results
Using this test platform and procedures, comparisons were made between two different op amps in the same audio headset amplifier circuit. One amplifier had a bipolar input stage and the other a MOS input stage. The results showed different parts of the circuit in an amplifier assembly can act as receivers at different RF frequencies, thus producing multiple interference sources acting at the same time. Susceptibility differences between the two types of amplifiers are reported as well as the results of circuit suppression techniques.

Test Platform
Modulated Sweep Test Hardware
The RF modulated test platform (Figure 2) was built using a HP8753D network analyzer (0.1MHz to 6GHz) as a variable RF source, a sine wave modulating signal source, and a HP3310A function generator to synchronize the sweep generator with the oscilloscope. A 20 second sweep time was used for both the sweep generator and the oscilloscope horizontal time base. The function generator was set to generate a repetitive 0.05Hz trigger for the scope and the network analyzer, so the scope horizontal display could be calibrated to the RF frequency sweep (Figure 2A).
Audio Amplifier Test Boards

The audio amplifier test board implements a single-ended input to differential output headphone amplifier using a dual op amp operating from a single +5V supply (Figure 3). The first stage is connected unity gain (AV = 1) and the second stage creates the inverse output (AV = -1) for an output differential gain of 2. Figure 4 shows the two identical evaluation boards that were used to implement the audio amplifier schematic shown in Figure 3. One board has the bipolar input op amp and the other the MOSFET input op amp. Figure 5 shows the distance between the op amp and the antenna during the RFI tests (~1/4 in). Figure 1 shows the distance between the op amp and the antenna in an actual Bluetooth headset (~1 in).
FIGURE 3. SCHEMATIC OF AUDIO TEST CIRCUIT

RFI SUPPRESSION CAPS CAN REDUCE LOW FREQUENCY CONDUCTIVE INTERFERENCE BUT MAY CAUSE AMPLIFIER INSTABILITY

FIGURE 4. MODIFIED EVALUATION BOARDS TO SIMULATE CUSTOMERS CIRCUIT

FIGURE 5. DISTANCE BETWEEN DIE AND ANTENNA DURING OUR LAB TESTS
Interference Model

Figure 6 illustrates the concept model of the demodulation experiments. This model illustrates how the Bluetooth carrier is stripped off leaving behind the low frequency signal. Because the frequency of the Bluetooth RF carrier is 20 times higher than the bandwidth of the op amp under test (120MHz), the amplifier acts as the demodulator and filter resulting in a carrier-free low frequency replica of the modulating signal appearing as an interference signal at the output of the amplifier as shown in Figure 10.

Interference Testing

The oscilloscope was used to measure the differential outputs during the 20 second RF sweep from 0.1MHz to 6GHz (Figure 2A). The RF output power of the HP8753D was adjusted to the standard Bluetooth output power level of 0dBm. The modulation level was set to ~80%, which resulted in some distortion caused by the HP8753D network analyzer. (Figure 2B).

The ability to select a single modulating frequency and to control the RF frequency and level of the carrier signal provides the tool necessary to carry out a variety of controlled experiments needed to partition the interference sources. Close inspection of circuit behavior was possible by sweeping the carrier frequency and measuring the modulating signal at the output of the audio amplifier circuit. (Figures 8, 9, 11, 12).

Isolating Interference Sources

A significant observation from the frequency sweep tests showed two distinct frequency bands of interference; a low frequency band from 0.1MHz to 30MHz and a high frequency band from ~1GHz to 4GHz.
The amplifier circuit design and layout used for this test suggests an interference behavioral model shown in Figure 7. For the purpose of this discussion, conducted interference is defined as interference received by the PCB interconnect and external components (i.e., resistors, caps) and conducted onto the pins of the op amp. Radiated interference is defined as inductive coupling of the radiated RF signal onto the package pins, the bond wires, and the active circuitry of the op amp.

Circuit board layout and components external to the op amp are most likely responsible for the interference in the 0.1MHz to 30MHz. The long PCB traces are effective antennae at the low frequency end of the sweep. The received RF signal in the 0.1MHz to 30MHz range is passed to the op amp on the inverting and non-inverting inputs. Eliminating the low frequency source of interference may not be important if the transceiver frequency is the GHz range (including harmonics).

Sources of Radiated Interference

The interference in the high frequency band was caused by the op amp. The most likely path is the reception of the radiated RF by the op amp package pins, bond wires and active circuitry on the die coupling the modulated 2.4GHz signal as a common-mode input to the differential pair. The mechanism for the demodulation and filtering is the input differential pair.

Sources of Conducted Interference

The interference in the low frequency band was caused by the PCB board layout and external components. Different amplifier designs with the same frequency response and gain can produce different levels of susceptibility to conducted interference depending on their input circuitry. An amplifier with differential (inverting and non-inverting) inputs can have input circuitry arranged so the demodulated signal is presented to the amplifier as a common-mode signal. If the amplitudes and phase of both input signals are the same, then the resulting interference is reduced by the common-mode rejection ratio (CMRR) of the amplifier.

To test this, impedance differences at the op amp input stage due to the circuit schematic were compensated by adjusting the impedance of the input terminals. This was done by adding resistance to the inverting input terminal (Figure 3) which resulted in matching the input impedances which enabled more of the demodulated signal to be cancelled by the input differential pair. This reduced the susceptibility of the op amp to conducted interference from the PC board. Reference section titled “Results of RF Interference Experiments” on page 7 for lab results of this test.

Results

Bipolar

Figure 8 shows the result of a frequency sweep from 100kHz to 6GHz of the bipolar op amp. Across the frequency sweep, the output of Channel “A” (unity gain configuration) shows an interference peak of 25mVP-P at ~2GHz with J1 input grounded (Figure 3). The lower trace shows the output of Channel “B” (inverting gain of one) with a maximum of 30mVp-p in the frequency band of 500kHz to ~30MHz. From 30MHz to 6GHz there is a 5mVP-P interference peak at ~800MHz and a series of peaks of ~25mVP-P over the 1.8GHz to 3GHz range. It is obvious from Figure 8 why the bipolar op amp had a problem with the Bluetooth application.

Figure 10 shows the interference on Channel A and Channel B outputs with a fixed carrier frequency of 3.9GHz and a 1kHz 80% modulation. 3.9GHz was chosen because the bipolar input amp has an interference peak at this frequency.
Interference below 30MHz is conducted and within the bandwidth of the both the bipolar and MOSFET op amps. This interference results from PCB layout and not the device itself, and therefore not important to this discussion. The interference in the GHz range is the focus of this evaluation.

**MOS**

Figure 9 shows the result of a frequency sweep from 100kHz to 6GHz of the MOSFET Input op amp. Across the frequency sweep, the output of Channel “A” (unity gain configuration) shows a total peak-peak voltage of less than 3mVp-p with J1 input grounded (Figure 3). The lower trace shows the output of Channel “B” (inverting gain of one) with a maximum of 10mVp-p in the frequency band of 500kHz to ~5MHz. From 5MHz to 6GHz the total peak-peak voltage is less than 3mVp-p.

From Figure 9, it is obvious why the MOSFET input op amp did not have a problem with the Bluetooth application.
Results of RF Interference Experiments

1. Matching the input impedances using external series resistor.

Adding a 50k resistor in series with the inverting input of CHA reduced high frequency interference and increased the low frequency interference as shown in Figure 11.

In the GHz region (Figure 11), the reduction in the demodulated interference level varied from ~5mVP-p to 20mVP-p depending on the orientation of the antenna. This orientation sensitivity is thought to cause a mismatch in the signal level coupled on to the inverting and non-inverting bond wires. This difference in signal level presents a differential signal to the input pair which appears at the amplifiers output.

2. Using capacitors on the input pins of the amplifiers to kill the signal before it get inside the part.

10pF capacitors were added to the input pins of Channel “A” to ground. A series 10k resistor was added to the inverting input to kill an oscillation caused by adding the capacitors. It was not possible to add capacitors to the inverting amplifiers input without oscillations. Channel “A” in Figure 12 shows a reduction in the interference across the entire frequency band. The reduction in conducted interference up to 30MHz most likely resulted from the input filters created by the capacitors.
CONCLUSION
The Modulated Sweep test platform is an effective way to generate a controlled RF interference source to measure the RFI rejection ability of a amplifier circuit. It was shown that circuit layout, circuit design and the amplifiers architecture are all contributing factors to RF interference.

The RF sweep feature enables evaluation of low frequency and high frequency carrier interference. The system also has the ability to test a single frequency so that specific solutions can be evaluated.

In the Bluetooth example, the interference was caused by demodulation of the high frequency signal within the input stage of the amplifiers. Bipolar input stages are active at 2.4GHz frequency and demodulate the RF signal to the amplifier output. MOS input stages are not active at 2.4GHz and do not make good demodulators.

If the slower speed and higher voltage noise of the MOS input stage gives acceptable performance, then the preferred input stage for RF tolerant parts would be the MOS input stage. If higher speed and lower voltage noise is required, then care must be taken in the design to minimize input impedance differences and shielding of the op amp to minimize the radiated interference at the output of the amplifier.
Notice

1. Descriptions of circuits, software and other related information in this document are provided only to illustrate the operation of semiconductor products and application examples. You are fully responsible for the incorporation or any other use of the circuits, software, and information in the design of your product or system. Renesas Electronics disclaims any and all liability for any losses and damages incurred by you or third parties arising from the use of these circuits, software, or information.

2. Renesas Electronics hereby expressly disclaims any warranties against and liability for infringement or any other claims involving patents, copyrights, or other intellectual property rights of third parties, by or arising from the use of Renesas Electronics products or technical information described in this document, including but not limited to, the product data, drawings, charts, programs, algorithms, and application examples.

3. No license, express, implied or otherwise, is granted hereby under any patents, copyrights or other intellectual property rights of Renesas Electronics or others.

4. You shall not alter, modify, copy, or reverse engineer any Renesas Electronics product, whether in whole or in part. Renesas Electronics disclaims any and all liability for any losses or damages incurred by you or third parties arising from such alteration, modification, copying or reverse engineering.

5. Renesas Electronics products are classified according to the following two quality grades: “Standard” and “High Quality.” The intended applications for each Renesas Electronics product depends on the product’s quality grade, as indicated below.

- “Standard”: Computers, office equipment, communications equipment, test and measurement equipment, audio and visual equipment, home electronic appliances, machine tools, personal electronic equipment, industrial robots, etc.
- “High Quality”: Transportation equipment (automobiles, trains, ships, etc.); traffic control (traffic lights); large-scale communication equipment; key financial terminal systems; safety control equipment; etc.

Unless expressly designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not intended or authorized for use in products or systems that may pose a direct threat to human life or bodily injury (artificial life support devices or systems; surgical implantations, etc.), or may cause serious property damage (space system; undersea repeaters; nuclear power control systems; aircraft control systems; key plant systems; military equipment; etc.). Renesas Electronics disclaims any and all liability for any damages or losses incurred by you or any third parties arising from the use of any Renesas Electronics product that is inconsistent with any Renesas Electronics data sheet, user’s manual or other Renesas Electronics document.

6. When using Renesas Electronics products, refer to the latest product information (data sheets, user’s manuals, application notes, “General Notes for Handling and Using Semiconductor Devices” in the reliability handbook, etc.), and ensure that usage conditions are within the ranges specified by Renesas Electronics with respect to maximum ratings, operating power supply voltage range, heat dissipation characteristics, installation, etc. Renesas Electronics disclaims any and all liability for any malfunctions, failure or accident arising out of the use of Renesas Electronics products outside of such specified ranges.

7. Although Renesas Electronics endeavors to improve the quality and reliability of Renesas Electronics products, semiconductor products have specific characteristics, such as the occurrence of failure at a certain rate and malfunctions under certain use conditions. Unless designated as a high reliability product or a product for harsh environments in a Renesas Electronics data sheet or other Renesas Electronics document, Renesas Electronics products are not subject to radiation resistance design. You are responsible for implementing safety measures to guard against the possibility of bodily injury, injury or damage caused by fire, and/or danger to the public in the event of a failure or malfunction of Renesas Electronics products, such as safety design for hardware and software, including but not limited to, redundancy, fire control and malfunction prevention, appropriate treatment for aging degradation or any other appropriate measures. Because the evaluation of microcomputer software alone is very difficult and impractical, you are responsible for evaluating the safety of the final products or systems manufactured by you.

8. Please contact a Renesas Electronics sales office for details as to environmental matters such as the environmental compatibility of each Renesas Electronics product. You are responsible for carefully and sufficiently investigating applicable laws and regulations that regulate the inclusion or use of controlled substances, including without limitation, the EU RoHS Directive, and using Renesas Electronics products in compliance with all these applicable laws and regulations. Renesas Electronics disclaims any and all liability for damages or losses occurring as a result of your noncompliance with applicable laws and regulations.

9. Renesas Electronics products and technologies shall not be used for or incorporated into any products or systems whose manufacture, use, or sale is prohibited under any applicable domestic or foreign laws or regulations. You shall comply with any applicable export control laws and regulations promulgated and administered by the governments of any countries asserting jurisdiction over the parties or transactions.

10. It is the responsibility of the buyer or distributor of Renesas Electronics products, or any other party who distributes, diagnoses of, or otherwise sells or transfers the product to a third party, to notify such third party in advance of the contents and conditions set forth in this document.

11. This document shall not be reprinted, reproduced or duplicated in any form, in whole or in part, without prior written consent of Renesas Electronics.

12. Please contact a Renesas Electronics sales office if you have any questions regarding the information contained in this document or Renesas Electronics products.

(Note 1) “Renesas Electronics” as used in this document means Renesas Electronics Corporation and also includes its directly or indirectly controlled subsidiaries.

(Note 2) “Renesas Electronics products” means any product developed or manufactured by or for Renesas Electronics.