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# H8/38524 Group

Hardware Manual

Renesas 16-Bit Single-Chip Microcomputer H8 Family / H8/300H Super Low Power Series

> H8/38524 H8/38523 H8/38522 H8/38521 H8/38520

Renesas Electronics

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## General Precautions in the Handling of MPU/MCU Products

The following usage notes are applicable to all MPU/MCU products from Renesas. For detailed usage notes on the products covered by this manual, refer to the relevant sections of the manual. If the descriptions under General Precautions in the Handling of MPU/MCU Products and in the body of the manual differ from each other, the description in the body of the manual takes precedence.

1. Handling of Unused Pins

Handle unused pins in accord with the directions given under Handling of Unused Pins in the manual.

- The input pins of CMOS products are generally in the high-impedance state. In operation with an unused pin in the open-circuit state, extra electromagnetic noise is induced in the vicinity of LSI, an associated shoot-through current flows internally, and malfunctions occur due to the false recognition of the pin state as an input signal become possible. Unused pins should be handled as described under Handling of Unused Pins in the manual.
- 2. Processing at Power-on

The state of the product is undefined at the moment when power is supplied.

 The states of internal circuits in the LSI are indeterminate and the states of register settings and pins are undefined at the moment when power is supplied.

In a finished product where the reset signal is applied to the external reset pin, the states of pins are not guaranteed from the moment when power is supplied until the reset process is completed.

In a similar way, the states of pins in a product that is reset by an on-chip power-on reset function are not guaranteed from the moment when power is supplied until the power reaches the level at which resetting has been specified.

3. Prohibition of Access to Reserved Addresses

Access to reserved addresses is prohibited.

- The reserved addresses are provided for the possible future expansion of functions. Do
  not access these addresses; the correct operation of LSI is not guaranteed if they are
  accessed.
- 4. Clock Signals

After applying a reset, only release the reset line after the operating clock signal has become stable. When switching the clock signal during program execution, wait until the target clock signal has stabilized.

- When the clock signal is generated with an external resonator (or from an external oscillator) during a reset, ensure that the reset line is only released after full stabilization of the clock signal. Moreover, when switching to a clock signal produced with an external resonator (or by an external oscillator) while program execution is in progress, wait until the target clock signal is stable.
- 5. Differences between Products

Before changing from one product to another, i.e. to one with a different type number, confirm that the change will not lead to problems.

— The characteristics of MPU/MCU in the same group but having different type numbers may differ because of the differences in internal memory capacity and layout pattern. When changing to products of different type numbers, implement a system-evaluation test for each of the products.

## How to Use This Manual

#### 1. Objective and Target Users

This manual was written to explain the hardware functions and electrical characteristics of this LSI to the target users, i.e. those who will be using this LSI in the design of application systems. Target users are expected to understand the fundamentals of electrical circuits, logic circuits, and microcomputers.

This manual is organized in the following items: an overview of the product, descriptions of the CPU, system control functions, and peripheral functions, electrical characteristics of the device, and usage notes.

When designing an application system that includes this LSI, take all points to note into account. Points to note are given in their contexts and at the final part of each section, and in the section giving usage notes.

The list of revisions is a summary of major points of revision or addition for earlier versions. It does not cover all revised items. For details on the revised points, see the actual locations in the manual.

The following documents have been prepared for the H8/38524 Group. Before using any of the documents, please visit our web site to verify that you have the most up-to-date available version of the document.

Document Type	Contents	Document Title	Document No.
Data Sheet	Overview of hardware and electrical characteristics	_	_
Hardware Manual	Hardware specifications (pin assignments, memory maps, peripheral specifications, electrical characteristics, and timing charts) and descriptions of operation	H8/38524 Group Hardware Manual	This manual
Software Manual	Detailed descriptions of the CPU and instruction set	H8/300H Series Software Manual	REJ09B0213
Application Note	Examples of applications and sample programs	The latest versions are available web site.	ailable from our
Renesas Technical Update	Preliminary report on the specifications of a product, document, etc.	-	



#### 2. Description of Numbers and Symbols

Aspects of the notations for register names, bit names, numbers, and symbolic names in this manual are explained below.

È A	Decimal: 1234 Notation for active-low An overbar on the name indicates that a signal or pin is active-low. [Example] WDTOVF (4) (2) (4) (2)
È A	An overbar on the name indicates that a signal or pin is active-low. [Example] $\overline{\text{WDTOVF}}$
	14.2.2 Compare Match Control/Status Register_0, _1(CMCSR_0, CMCSR_1)
	CMCSR indicates compare-m <del>atch gens</del> ration, enables or disables interrupts, and selects the counter out clock. Generation of a WDTOVF signal or interrunt initializes the TCNT value to 0.
	14.3 Opt-
	14.3.1 Interval Count Operation
	When an internal clock is selected with the CKS1 and CKS0 bits in CMCSR and the STR bit in CMSTR is set to 1, CMCNT starts incrementing using the selected clock. When the values in CMCNT and the compare match constant register (CMCOR) match, CMCNT if cleared to H'0000) and the CMF flag in CMCSR is set to 1. When the CKS1 and CKS0 bits are set to B01 at this time.
	a f/4 clock is selected.
	a f/4 clock is selected. Rev. 0,50, 10/04, page 416 of 914
	CMSTR is set to 1, CMCNT starts incrementing using the selected clock. When the values in CMCNT and the compare match constant register (CMCOR) match, CMCNT is cleared to H'0000 and the CMF flag in CMCSR is set to 1. When the CKS1 and CKS0 bits are set to <u>B'01 at this time</u> ,

#### 3. Description of Registers

Each register description includes a bit chart, illustrating the arrangement of bits, and a table of bits, describing the meanings of the bit settings. The standard format and notation for bit charts and tables are described below.





#### 4. Description of Abbreviations

The abbreviations used in this manual are listed below.

• Abbreviations used in this manual

Abbreviation	Description
ACIA	Asynchronous communication interface adapter
bps	Bits per second
CRC	Cyclic redundancy check
DMA	Direct memory access
DMAC	Direct memory access controller
GSM	Global System for Mobile Communications
Hi-Z	High impedance
IEBus	Inter Equipment Bus (IEBus is a trademark of NEC Electronics Corporation.)
I/O	Input/output
IrDA	Infrared Data Association
LSB	Least significant bit
MSB	Most significant bit
NC	No connection
PLL	Phase-locked loop
PWM	Pulse width modulation
SFR	Special function register
SIM	Subscriber Identity Module
UART	Universal asynchronous receiver/transmitter
VCO	Voltage-controlled oscillator



## 5. List of Product Specifications

Below is a table listing the product specifications for each group.

		-	H8/38524 Group
ltem		Flash Memory	Mask ROM
Memory	ROM	16 K, 32 Kbytes	8 K, 12 K, 16 K, 24 K, 32 Kbytes
	RAM	1 Kbyte	512 bytes, 1 Kbyte
Operating	4.5 to 5.5 V	20 MHz	20 MHz
voltage and	2.7 to 5.5 V	20 MHz	20 MHz
operating	1.8 to 5.5 V	_	
frequency	2.7 to 3.6 V	_	
	1.8 to 3.6 V	_	
I/O ports	Input	9	9
	Output	6	6
	I/O	50	50
Timers	Clock (timer A)	1	1
	Reload (timer C)	1	1
	Compare (timer F)	1	1
	Capture (timer G)	1	1
	AEC	1	1
	WDT	_	
	WDT (discrete)	1	1
SCI	UART/Synchronous	1 ch	1 ch
A-D (resolution	× input channels)	10 bit $\times$ 8 ch	10 bit × 8 ch
LCD	seg	32	32
	com	4	4
External interrupt (internal wakeup)		13(8)	13(8)
POR (power-on reset)		1	1
LVD (low-voltag	ge detection circuit)	1	1
Package		FP-80A	FP-80A
		TFP-80C	TFP-80C
Operating	temperature	Standard specificat	ions: –20 to 75°C, WTR: –40 to 85°C

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## Section 1 Overview

## 1.1 Features

Microcontrollers of the H8/38524 Group are CISC (complex instruction set computer) microcontrollers whose core is an H8/300H CPU, which has an internal 32-bit architecture. The H8/300H CPU provides upward compatibility with the H8/300 CPUs of other Renesas Technology-original microcontrollers.

As peripheral functions, each LSI of this Group includes various timer functions that realize lowcost configurations for end systems. The power consumption of these modules can be kept down dynamically by power-down mode.

#### 1.1.1 Application

Examples of the applications of this LSI include motor control, power meter, and health equipment.



## 1.1.2 Overview of Specifications

Table 1.1 lists the functions of H8/38524 Group products in outline.

#### Table 1.1 Overview of Functions

Classification	Module/ Function	Description			
Memory	ROM	ROM lineup: Flash memory version and mask Rom version			
		• ROM capacity: 8 K, 12 K, 16 K, 24 K, and 32 Kbytes			
	RAM	RAM capacity: 512 and 1024 bytes			
CPU	CPU	H8/300H CPU (CISC type)			
		Upward compatibility for H8/300 CPU at object level			
		Sixteen 16-bit general registers			
		Eight addressing modes			
		64-Kbyte address space			
		Program: 64 Kbytes available			
		Data: 64 Kbytes available			
		<ul> <li>62 basic instructions, classifiable as bit arithmetic and logic instructions, multiply and divide instructions, bit manipulation instructions, and others</li> </ul>			
		• Minimum instruction execution time: 400 ns (for an ADD instruction while system clock $\phi$ = 5 MHz and $V_{cc}$ = 2.7 to 3.6 V)			
		• On-chip multiplier ( $16 \times 16 \rightarrow 32$ bits)			
	Operating mode	Normal mode			
	MCU	Mode: Single-chip mode			
	operating mode	Low power consumption state (transition driven by the SLEEP			
	mode	instruction)			
Interrupt (source)	Interrupt controller	<ul> <li>Thirteen external interrupt pins (IRQAEC, IRQ4, IRQ3, IRQ1, IRQ0, WKP7 to WKP0)</li> </ul>			
	(INTC)	Nine internal interrupt sources			
		Independent vector addresses			

	Module/	2/		
Classification	Function	Description		
Clock	Clock pulse	Two clock generation circuits available		
	generator (CPG)	<ul> <li>Separate clock signals are provided for each of functional modules</li> </ul>		
		• Includes frequency division circuit, so the operating frequency is selectable		
		• Seven low-power-consumption modes: Active (medium speed) mode, sleep (high speed or medium speed) mode, subactive mode, subsleep mode, standby mode, and watch mode		
		Equipped with an on-chip oscillator		
A/D converter	A/D converter (ADC)	<ul> <li>10-bit resolution × eight input channels</li> <li>Sample and hold function included</li> <li>Conversion time: 12.4 μs per channel (with φ at 5-MHz operation) / 6.2 μs per channel (with φ at 10-MHz operation)</li> </ul>		
		<ul> <li>A/D conversion can be started by external trigger input</li> </ul>		
Timer	10-bit PWM			
		Four conversion periods selectable		
		Pulse division method for less ripple		
	Timer A	8-bit timer		
		Interval timer functionality: Eight internal clock sources are selectable		
		Clock time base functionality: Four overflow periods are selectable		
		Generates an interrupt upon overflow		
	Timer C	8-bit timer		
		Eight clocks are selectable		
		Auto-reload function supported		
		Generates an interrupt upon overflow		
		Up/down-counter switching is possible		
	Timer F	• 16-bit timer (also can be used as two independent 8-bit timers)		
		Five clocks are selectable		
		Output compare function supported		
		Toggle output function supported		
		Two interrupt sources: Compare match and overflow		



	Module/			
Classification	Function	Description		
Timer	Timer G	8-bit timer		
		Four counter input clocks are selectable		
		Input capture functions supported (a built-in noise canceller)		
		Level detection at counter overflow is possible		
		Counter clearing option		
		Two interrupt sources: Input capture and overflow		
	Asynchron-	• 16-bit pulse timer (also can be used as 8 bits × two channels)		
	ous event counter (AEC)	Can count asynchronously-input external events		
Watchdog timer	Watchdog timer (WDT)	8 bits $\times$ one channel (selectable from ten counter input clocks)		
Serial interface	Serial communi-	<ul> <li>For both asynchronous and clock synchronous serial communications</li> </ul>		
	cation interface 3	Full-duplex communications capability		
		Select the desired bit rate		
	(SCI3)	Six interrupt sources		
I/O ports		Nine CMOS input-only pins		
		Six CMOS output-only pins		
		<ul> <li>50 CMOS input/output pins</li> </ul>		
		Six large-current-drive pins (port 9)		
		27 pull-up resistors		
		Seven open drains		
LCD (Liquid	LCD	<ul> <li>A maximum of 32 segment pins and four common pins</li> </ul>		
Crystal Display)	controller/	<ul> <li>Choice of four duty cycles (static, 1/2, 1/3, or 1/4)</li> </ul>		
drive	driver	<ul> <li>LCD RAM capacity: 8 bits × 16 bytes (128 bits)</li> </ul>		
		Word access to LCD RAM		
		<ul> <li>All four segment output pins can be used individually as port pins</li> </ul>		
		• Common output pins not used because of the duty cycle can be used for common double-buffering (parallel connection)		
		Display possible in operating modes other than standby mode		
		Choice of 11 frame frequencies		
		<ul> <li>Built-in power supply split-resistance, supplying LCD drive power</li> </ul>		
		A or B waveform selectable by software		
		Removal of split-resistance can be controlled in software		

Classification	Module/ sification Function Description			
Measures in power supply drops	Power-on reset and low-voltage	<ul> <li>Power-on reset circuit: An internal reset signal can be issued at power-on by connecting an external capacitor</li> </ul>		
	detection circuits	<ul> <li>Low-voltage detection circuit: Monitors the power supply voltage and issues an internal reset signal or interrupt if the voltage goes below or above a specified range</li> </ul>		
Internal power Power supply step- supply down circuit circuit		• The internal power supply can be fixed at a constant level of approximately 3.0 V, independently of the voltage of the power supply connected to the external $V_{cc}$ pin		
		<ul> <li>It is also possible to use the same level of external power supply voltage and internal power supply voltage without using the internal power supply step-down circuit</li> </ul>		
Package		<ul> <li>QFP-80: package code: FP-80A (package dimensions: 14 × 14 mm, pin pitch: 0.65 mm)</li> </ul>		
		TQFP-80: package code: TFP-80C (package dimensions: 12 × 12 mm, pin pitch: 0.50 mm)		
Operating freque	ency/	Operating frequency: 2 to 20 MHz		
Power supply vo	oltage	Power supply voltage:		
		Vcc = 2.7 to 5.5 V, AVcc = 2.7 to 5.5 V		
		Supply current:		
		Flash memory version: 4.0 mA (typ.) (Vcc = 5.0 V, AVcc = 5.0 V, $\phi$ = 10 MHz)		
		Mask ROM version: 3.3 mA (typ.) (Vcc = 5.0 V, AVcc = 5.0 V, $\phi$ = 10 MHz)		
Operating peripl		<ul> <li>–20 to +75°C (regular specifications)</li> </ul>		
temperature (°C	;)	<ul> <li>–40 to +85°C (wide-range specifications)</li> </ul>		



## **1.2** List of Products

Table 1.2 and figure 1.1 show the list of products and the structure of a product number, respectively.

Group	Product Type	ROM Size	RAM Size	Package	Remarks
H8/38524 Group	HD64F38524	32 Kbytes	1 Kbyte	FP-80A, TFP-80C	Flash memory version
	HD64338524	32 Kbytes	1 Kbyte		Mask ROM version
	HD64338523	24 Kbytes	1 Kbyte		Mask ROM version
	HD64F38522	16 Kbytes	1 Kbyte		Flash memory version
	HD64338522	16 Kbytes	1 Kbyte		Mask ROM version
	HD64338521	12 Kbytes	512 bytes		Mask ROM version
	HD64338520	8 Kbytes	512 bytes		Mask ROM version

#### Table 1.2List of Products





Figure 1.1 How to Read the Product Name Code



## 1.3 Block Diagram



Figure 1.2 Block Diagram of H8/38524 Group













#### **Pin Functions** 1.5

## Table 1.3Pin Functions

		Pin No.		
Туре	Symbol	FP-80A, TFP-80C	I/O	Functions
Power source pins	$V_{cc}$	52	Input	<b>Power supply:</b> All $V_{cc}$ pins should be connected to the system power supply.
	V <sub>ss</sub>	8 (= AV <sub>ss</sub> ), 53	Input	<b>Ground:</b> All $V_{ss}$ pins should be connected to the system power supply (0 V).
	AV <sub>cc</sub>	1	Input	Analog power supply: This is the power supply pin for the A/D converter. When the A/D converter is not used, connect this pin to the system power supply.
	AV <sub>ss</sub>	8 (= V <sub>ss</sub> )	Input	<b>Analog ground:</b> This is the A/D converter ground pin. It should be connected to the system power supply (0V).
	V <sub>1</sub>	51	Input	LCD power supply: These are the
	V <sub>2</sub>	50		power supply pins for the LCD controller/driver.
	V <sub>3</sub>	49		contioner/driver.
	CV <sub>cc</sub>	4	Input	<b>Power supply:</b> This is the internal step-down power supply pin. To ensure stability, a capacitor with a rating of about 0.1 $\mu$ F should be connected between this pin and the V <sub>ss</sub> pin.
Clock pins	OSC <sub>1</sub>	10	Input	These pins connect to a crystal or
	OSC <sub>2</sub>	9	Output	<ul> <li>ceramic oscillator, or can be used to input an external clock. See section 4, Clock Pulse Generators, for a typical connection diagram.</li> </ul>
	<b>X</b> <sub>1</sub>	6	Input	These pins connect to a 32.768-kHz
	X <sub>2</sub>	7	Output	crystal oscillator. See section 4, Clock Pulse Generators, for a typical connection diagram.

		Pin No.		
Туре	Symbol	FP-80A, TFP-80C	- I/O	Functions
System control	RES	12	Input	<b>Reset:</b> When this pin is driven low, the chip is reset.
	TEST	11	Input	<b>Test pin:</b> This pin is reserved and cannot be used. It should be connected to $V_{ss}$ .
Interrupt pins	IRQ, IRQ, IRQ, IRQ₄	72 76 5 3	Input	<b>IRQ interrupt request 4, 3, 1, and 0:</b> These are input pins for edge-sensitive external interrupts, with a selection of rising or falling edge.
	IRQAEC	60	Input	Asynchronous event counter event signal: This is an interrupt input pin for enabling asynchronous event input.
				This must be fixed at $V_{cc}$ or GND because the oscillator is selected by the input level during resets. Refer to section 4, Clock Pulse Generators, for information on the selection method.
	WKP <sub>7</sub> to WKP <sub>0</sub>	20 to 13	Input	Wakeup interrupt request 7 to 0: These are input pins for rising or falling-edge-sensitive external interrupts.
Timer	AEVL	68	Input	Asynchronous event counter event
	AEVH	67		input: These are event input pins for input to the asynchronous event counter.
	TMIC	76	Input	<b>Timer C event input:</b> This is an event input pin for input to the timer C counter.
	UD	61	Input	<b>Timer C up/down select:</b> This pin selects up- or down-counting for the timer C counter. The counter operates as a down-counter when this pin is high, and as an up-counter when low.
	TMIF	5	Input	<b>Timer F event input:</b> This is an event input pin for input to the timer F counter.

		Pin No.		
Туре	Symbol	FP-80A, TFP-80C	I/O	Functions
Timer	TMOFL	62	Output	<b>Timer FL output:</b> This is an output pin for waveforms generated by the timer FL output compare function.
	TMOFH	63	Output	<b>Timer FH output:</b> This is an output pin for waveforms generated by the timer FH output compare function.
	TMIG	2	Input	<b>Timer G capture input:</b> This is an input pin for timer G input capture.
10-bit PWM	PWM1	54	Output	10-bit PWM output: These are output
	PWM2	55		pins for waveforms generated by the channel 1 and 2 10-bit PWMs.
I/O ports	P1,	5	I/O	Port 1: This is a 3-bit I/O port. Input or
	P1 <sub>4</sub>	3		output can be designated for each bit by means of port control register 1
	P1 <sub>3</sub>	2		(PCR1).
	P3 <sub>7</sub> to P3 <sub>0</sub>	68 to 61	I/O	<b>Port 3:</b> This is an 8-bit I/O port. Input or output can be designated for each bit by means of port control register 3 (PCR3).
				If the on-chip emulator is used, pins 33, 34, and 35 are reserved for the emulator and not available to the user.
	P4 <sub>3</sub>	72	Input	Port 4 (bit 3): This is a 1-bit input port.
	P4 <sub>2</sub> to P4 <sub>0</sub>	71 to 69	I/O	<b>Port 4 (bits 2 to 0):</b> This is a 3-bit I/O port. Input or output can be designated for each bit by means of port control register 4 (PCR4).
	P5 <sub>7</sub> to P5 <sub>0</sub>	20 to 13	I/O	<b>Port 5:</b> This is an 8-bit I/O port. Input or output can be designated for each bit by means of port control register 5 (PCR5).
	P6 <sub>7</sub> to P6 <sub>0</sub>	28 to 21	I/O	<b>Port 6:</b> This is an 8-bit I/O port. Input or output can be designated for each bit by means of port control register 6 (PCR6).

		Pin No.		
Туре	Symbol	FP-80A, TFP-80C	- I/O	Functions
I/O ports	$P7_7$ to $P7_0$	36 to 29	I/O	<b>Port 7:</b> This is an 8-bit I/O port. Input or output can be designated for each bit by means of port control register 7 (PCR7).
	P8 <sub>7</sub> to P8 <sub>0</sub>	44 to 37	I/O	<b>Port 8:</b> This is an 8-bit I/O port. Input or output can be designated for each bit by means of port control register 8 (PCR8).
	P9 <sub>5</sub> to P9 <sub>0</sub>	59 to 54	Output	<b>Port 9:</b> This is a 6-bit output port. If the on-chip emulator is used, pin 95 is reserved for the emulator and not available to the user. In the case of the flash memory version, pin 95 should not be left open in the user mode, and should instead be pulled up to high level.
	$PA_{3}$ to $PA_{0}$	45 to 48	I/O	<b>Port A:</b> This is a 4-bit I/O port. Input or output can be designated for each bit by means of port control register A (PCRA).
	PB <sub>7</sub> to PB <sub>0</sub>	80 to 73	Input	Port B: This is an 8-bit input port.
Serial com- munication	RXD <sub>32</sub>	70	Input	SCI3 receive data input: This is the SCI3 data input pin.
interface (SCI)	TXD <sub>32</sub>	71	Output	SCI3 transmit data output: This is the SCI3 data output pin.
	SCK <sub>32</sub>	69	I/O	<b>SCI3 clock I/O:</b> This is the SCI3 clock I/O pin.
A/D converter	AN <sub>7</sub> to AN <sub>0</sub>	80 to 73	Input	<b>Analog input channels 7 to 0:</b> These are analog data input channels to the A/D converter.
	ADTRG	3	Input	<b>A/D converter trigger input:</b> This is the external trigger input pin to the A/D converter.
LCD controller/	$COM_4$ to $COM_1$	45 to 48	Output	LCD common output: These are the LCD common output pins.
driver	$SEG_{32}$ to $SEG_1$	44 to 13	Output	LCD segment output: These are the LCD segment output pins.



		Pin No.		
Туре	Symbol	FP-80A, TFP-80C	I/O	Functions
Low-voltage detection circuit (LVD)	$V_{\text{ref}}$	57	Input	LVD reference voltage input: This is the LVD reference voltage input pin.
	extD	73	Input	LVD power supply drop detect voltage input: This is the LVD power supply drop detect voltage input pin.
	extU	74	Input	LVD power supply rise detect voltage input: This is the LVD power supply rise detect voltage input pin.
NC	NC		—	NC pin


# Section 2 CPU

This LSI has an H8/300H CPU with an internal 32-bit architecture that is upward-compatible with the H8/300 CPU, and supports only normal mode, which has a 64-Kbyte address space.

Upward-compatible with H8/300 CPUs Can execute H8/300 CPUs object programs Additional eight 16-bit extended registers 32-bit transfer and arithmetic and logic instructions are added Signed multiply and divide instructions are added.
General-register architecture Sixteen 16-bit general registers also usable as sixteen 8-bit registers and eight 16-bit registers, or eight 32-bit registers
Sixty-two basic instructions 8/16/32-bit data transfer and arithmetic and logic instructions Multiply and divide instructions
Fight addressing modes

RENESAS

- Register direct [Rn] Register indirect [@ERn] Register indirect with displacement [@(d:16,ERn) or @(d:24,ERn)] Register indirect with post-increment or pre-decrement [@ERn+ or @-ERn] Absolute address [@aa:8, @aa:16, @aa:24] Immediate [#xx:8, #xx:16, or #xx:32]
- Program-counter relative [@(d:8,PC) or @(d:16,PC)]
- Memory indirect [@@aa:8]
- 64-Kbyte address space
- High-speed operation

All frequently-used instructions execute in one or two states

8/16/32-bit register-register add/subtract : 2 state

- $8 \times 8$ -bit register-register multiply : 14 states
- 16 ÷ 8-bit register-register divide : 14 states
- $16 \times 16$ -bit register-register multiply : 22 states
- $32 \div 16$ -bit register-register divide : 22 states

• Power-down state

Transition to power-down state by SLEEP instruction



### 2.1 Address Space and Memory Map

The memory map of the H8/38524 is shown in figure 2.1(1), that of the H8/38523 in figure 2.16(2), that of the H8/38522 in figure 2.1(3), that of the H8/38521 in figure 2.1(4), and that of the H8/38520 in figure 2.1(5).



Notes: 1. Not available to the user if the on-chip emulator is used.

2. Used by the programming control program when programming flash memory. Also, not available to the user if the on-chip emulator is used.

### Figure 2.1(1) H8/38524 Memory Map

RENESAS



Figure 2.1(2) H8/38523 Memory Map



the on-chip emulator is used.

Figure 2.1(3) H8/38522 Memory Map





Figure 2.1(4) H8/38521 Memory Map



Figure 2.1(5) H8/38520 Memory Map



### 2.2 Register Configuration

The H8/300H CPU has the internal registers shown in figure 2.2. There are two types of registers; general registers and control registers. The control registers are a 24-bit program counter (PC), and an 8-bit condition-code register (CCR).



Figure 2.2 CPU Registers



#### 2.2.1 General Registers

The H8/300H CPU has eight 32-bit general registers. These general registers are all functionally identical and can be used as both address registers and data registers. When a general register is used as a data register, it can be accessed as a 32-bit, 16-bit, or 8-bit register. Figure 2.3 illustrates the usage of the general registers. When the general registers are used as 32-bit registers or address registers, they are designated by the letters ER (ER0 to ER7).

The ER registers divide into 16-bit general registers designated by the letters E (E0 to E7) and R (R0 to R7). These registers are functionally equivalent, providing a maximum of sixteen 16-bit registers. The E registers (E0 to E7) are also referred to as extended registers.

The R registers divide into 8-bit registers designated by the letters RH (R0H to R7H) and RL (R0L to R7L). These registers are functionally equivalent, providing a maximum of sixteen 8-bit registers.

The usage of each register can be selected independently.



Figure 2.3 Usage of General Registers

General register ER7 has the function of the stack pointer (SP) in addition to its general-register function, and is used implicitly in exception handling and subroutine calls. Figure 2.4 shows the relationship between the stack pointer and the stack area.





Figure 2.4 Relationship between Stack Pointer and Stack Area

#### 2.2.2 Program Counter (PC)

This 24-bit counter indicates the address of the next instruction the CPU will execute. The length of all CPU instructions is 2 bytes (one word), so the least significant PC bit is ignored. (When an instruction is fetched, the least significant PC bit is regarded as 0). The PC is initialized when the start address is loaded by the vector address generated during reset exception-handling sequence.

### 2.2.3 Condition-Code Register (CCR)

This 8-bit register contains internal CPU status information, including an interrupt mask bit (I) and half-carry (H), negative (N), zero (Z), overflow (V), and carry (C) flags. The I bit is initialized to 1 by reset exception-handling sequence, but other bits are not initialized.

Some instructions leave flag bits unchanged. Operations can be performed on the CCR bits by the LDC, STC, ANDC, ORC, and XORC instructions. The N, Z, V, and C flags are used as branching conditions for conditional branch (Bcc) instructions.

For the action of each instruction on the flag bits, see appendix A.1, Instruction List.



Bit	Bit Name	Initial Value	R/W	Description	
7	I	1	R/W	Interrupt Mask Bit	
				Masks interrupts other than NMI when set to 1. NMI is accepted regardless of the I bit setting. The I bit is set to 1 at the start of an exception-handling sequence.	
6	UI	Undefined	R/W	User Bit	
				Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions.	
5	Н	Undefined	R/W	Half-Carry Flag	
				When the ADD.B, ADDX.B, SUB.B, SUBX.B, CMP.B, or NEG.B instruction is executed, this flag is set to 1 if there is a carry or borrow at bit 3, and cleared to 0 otherwise. When the ADD.W, SUB.W, CMP.W, or NEG.W instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 11, and cleared to 0 otherwise. When the ADD.L, SUB.L, CMP.L, or NEG.L instruction is executed, the H flag is set to 1 if there is a carry or borrow at bit 27, and cleared to 0 otherwise.	
4	U	Undefined	R/W	User Bit	
				Can be written and read by software using the LDC, STC, ANDC, ORC, and XORC instructions.	
3	Ν	Undefined	R/W	Negative Flag	
				Stores the value of the most significant bit of data as a sign bit.	
2	Z	Undefined	R/W	Zero Flag	
				Set to 1 to indicate zero data, and cleared to 0 to indicate non-zero data.	
1	V	Undefined	R/W	Overflow Flag	
				Set to 1 when an arithmetic overflow occurs, and cleared to 0 at other times.	
0	С	Undefined	R/W	Carry Flag	
				Set to 1 when a carry occurs, and cleared to 0 otherwise. Used by:	
				Add instructions, to indicate a carry	
				Subtract instructions, to indicate a borrow	
				Shift and rotate instructions, to indicate a carry	
				The carry flag is also used as a bit accumulator by bit manipulation instructions.	



## 2.3 Data Formats

The H8/300H CPU can process 1-bit, 4-bit (BCD), 8-bit (byte), 16-bit (word), and 32-bit (longword) data. Bit-manipulation instructions operate on 1-bit data by accessing bit n (n = 0, 1, 2, ..., 7) of byte operand data. The DAA and DAS decimal-adjust instructions treat byte data as two digits of 4-bit BCD data.

#### 2.3.1 General Register Data Formats

Data Format Data Type **General Register** 0 7 RnH 6 5 4 3 2 1 0 Don't care 7 1-bit data RnL Don't care 7 6 5 4 3 2 1-bit data 7 4 3 4-bit BCD data **RnH** Upper Lower Don't care 4 3 0 4-bit BCD data RnL Don't care Upper Lower C Byte data RnH Don't care MSB LSB Byte data RnL Don't care MSB LSB

Figure 2.5 shows the data formats in general registers.

Figure 2.5 General Register Data Formats (1)



Figure 2.5 General Register Data Formats (2)



#### 2.3.2 Memory Data Formats

Figure 2.6 shows the data formats in memory. The H8/300H CPU can access word data and longword data in memory, however word or longword data must begin at an even address. If an attempt is made to access word or longword data at an odd address, an address error does not occur, however the least significant bit of the address is regarded as 0, so access begins the preceding address. This also applies to instruction fetches.

When ER7 (SP) is used as an address register to access the stack area, the operand size should be word or longword.

Data Type	Address	Data Format
		7 0
1-bit data	Address L	7 6 5 4 3 2 1 0
Byte data	Address L	MSB LLSB
Word data	Address 2M Address 2M+1	MSB LSB
Longword data	Address 2N Address 2N+1 Address 2N+2	
	Address 2N+3	

Figure 2.6 Memory Data Formats

### 2.4 Instruction Set

### 2.4.1 Table of Instructions Classified by Function

The H8/300H CPU has 62 instructions. Tables 2.2 to 2.9 summarize the instructions in each functional category. The notation used in tables 2.2 to 2.9 is defined in table 2.1.

Symbol	Description
Rd	General register (destination)*
Rs	General register (source)*
Rn	General register*
ERn	General register (32-bit register or address register)
(EAd)	Destination operand
(EAs)	Source operand
CCR	Condition-code register
Ν	N (negative) flag in CCR
Z	Z (zero) flag in CCR
V	V (overflow) flag in CCR
С	C (carry) flag in CCR
PC	Program counter
SP	Stack pointer
#IMM	Immediate data
disp	Displacement
+	Addition
_	Subtraction
×	Multiplication
÷	Division
^	Logical AND
V	Logical OR
$\oplus$	Logical XOR
$\rightarrow$	Move
7	NOT (logical complement)

 Table 2.1
 Operation Notation



Symbol	Description
:3/:8/:16/:2	4 3-, 8-, 16-, or 24-bit length
Note: *	General registers include 8-bit registers (R0H to R7H, R0L to R7L), 16-bit registers (R0

General registers include 8-bit registers (R0H to R7H, R0L to R7L), 16-bit registers (R0 to R7, E0 to E7), and 32-bit registers/address register (ER0 to ER7).

### Table 2.2 Data Transfer Instructions

Instruction	n Size*	Function
MOV	B/W/L	$(EAs) \rightarrow Rd, Rs \rightarrow (EAd)$ Moves data between two general registers or between a general register and memory, or moves immediate data to a general register.
MOVFPE	В	$(EAs) \rightarrow Rd$ Cannot be used in this LSI.
MOVTPE	В	$Rs \rightarrow (EAs)$ Cannot be used in this LSI.
POP	W/L	$@SP+ \rightarrow Rn$ Pops a general register from the stack. POP.W Rn is identical to MOV.W @SP+, Rn. POP.L ERn is identical to MOV.L $@SP+$ , ERn.
PUSH	W/L	$Rn \rightarrow @-SP$ Pushes a general register onto the stack. PUSH.W Rn is identical to MOV.W Rn, @-SP. PUSH.L ERn is identical to MOV.L ERn, @-SP.
Note: *	Refers to the	operand size.
	B: Byte	
	W: Word	
	L: Longword	



Instructio	on Size*	Function
ADD SUB	B/W/L	$Rd \pm Rs \rightarrow Rd$ , $Rd \pm \#IMM \rightarrow Rd$ Performs addition or subtraction on data in two general registers, or on immediate data and data in a general register (immediate byte data cannot be subtracted from byte data in a general register. Use the SUBX or ADD instruction.)
ADDX SUBX	В	$\begin{array}{l} Rd\pmRs\pmC\toRd,  Rd\pm\#IMM\pmC\toRd\\ Performs \text{ addition or subtraction with carry on byte data in two general registers, or on immediate data and data in a general register.} \end{array}$
INC DEC	B/W/L	$Rd \pm 1 \rightarrow Rd$ , $Rd \pm 2 \rightarrow Rd$ Increments or decrements a general register by 1 or 2. (Byte operands can be incremented or decremented by 1 only.)
ADDS SUBS	L	$\begin{array}{ll} Rd\pm 1\toRd, & Rd\pm 2\toRd, & Rd\pm 4\toRd\\ Adds \text{ or subtracts the value 1, 2, or 4 to or from data in a 32-bit register.} \end{array}$
DAA DAS	В	Rd (decimal adjust) $\rightarrow$ Rd Decimal-adjusts an addition or subtraction result in a general register by referring to the CCR to produce 4-bit BCD data.
MULXU	B/W	$Rd \times Rs \rightarrow Rd$ Performs unsigned multiplication on data in two general registers: either 8 bits $\times$ 8 bits $\rightarrow$ 16 bits or 16 bits $\times$ 16 bits $\rightarrow$ 32 bits.
MULXS	B/W	$\begin{array}{l} Rd\timesRs\toRd\\ Performs \text{ signed multiplication on data in two general registers: either 8}\\ bits\times8 \ bits\to16 \ bits \ or \ 16 \ bits\times16 \ bits\to32 \ bits. \end{array}$
DIVXU	B/W	$\begin{array}{l} \text{Rd} \div \text{Rs} \rightarrow \text{Rd} \\ \text{Performs unsigned division on data in two general registers: either 16} \\ \text{bits} \div 8 \text{ bits} \rightarrow 8 \text{-bit quotient and 8-bit remainder or 32 bits} \div 16 \text{ bits} \rightarrow \\ 16 \text{-bit quotient and 16-bit remainder.} \end{array}$
Note: *	Refers to the	operand size.
	B: Byte	
	W: Word	
	L: Longword	

 Table 2.3
 Arithmetic Operations Instructions (1)



Instructio	n Size*	Function
DIVXS	B/W	Rd $\div$ Rs $\rightarrow$ Rd Performs signed division on data in two general registers: either 16 bits $\div$ 8 bits $\rightarrow$ 8-bit quotient and 8-bit remainder or 32 bits $\div$ 16 bits $\rightarrow$ 16-bit quotient and 16-bit remainder.
CMP	B/W/L	Rd – Rs, Rd – #IMM Compares data in a general register with data in another general register or with immediate data, and sets CCR bits according to the result.
NEG	B/W/L	$0 - Rd \rightarrow Rd$ Takes the two's complement (arithmetic complement) of data in a general register.
EXTU	W/L	Rd (zero extension) $\rightarrow$ Rd Extends the lower 8 bits of a 16-bit register to word size, or the lower 16 bits of a 32-bit register to longword size, by padding with zeros on the left.
EXTS	W/L	Rd (sign extension) $\rightarrow$ Rd Extends the lower 8 bits of a 16-bit register to word size, or the lower 16 bits of a 32-bit register to longword size, by extending the sign bit.
Note: *	Refers to the B: Byte W: Word L: Longword	operand size.

#### Table 2.3 Arithmetic Operations Instructions (2)



Instructio	n Size*	Function
AND	B/W/L	$Rd \wedge Rs \rightarrow Rd$ , $Rd \wedge \#IMM \rightarrow Rd$ Performs a logical AND operation on a general register and another general register or immediate data.
OR	B/W/L	$Rd \lor Rs \rightarrow Rd$ , $Rd \lor \#IMM \rightarrow Rd$ Performs a logical OR operation on a general register and another general register or immediate data.
XOR	B/W/L	$Rd \oplus Rs \rightarrow Rd$ , $Rd \oplus #IMM \rightarrow Rd$ Performs a logical exclusive OR operation on a general register and another general register or immediate data.
NOT	B/W/L	¬ (Rd) → (Rd) Takes the one's complement (logical complement) of general register contents.
Note: *	Refers to the	operand size.
	B: Byte	
	W: Word	
	L: Longword	

### Table 2.4 Logic Operations Instructions

#### Table 2.5Shift Instructions

Instructio	n Size*	Function
SHAL SHAR	B/W/L	Rd (shift) $\rightarrow$ Rd Performs an arithmetic shift on general register contents.
SHLL SHLR	B/W/L	Rd (shift) $\rightarrow$ Rd Performs a logical shift on general register contents.
ROTL ROTR	B/W/L	Rd (rotate) $\rightarrow$ Rd Rotates general register contents.
ROTXL ROTXR	B/W/L	Rd (rotate) $\rightarrow$ Rd Rotates general register contents through the carry flag.
Note: *	Refers to the	operand size.

B: Byte

W: Word

L: Longword



Instruction	Size*	Function	
BSET	В	$1 \rightarrow (\text{-bit-No.> of -EAd>})$ Sets a specified bit in a general register or memory operand to 1. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.	
BCLR	В	$0 \rightarrow$ ( <bit-no.> of <ead>) Clears a specified bit in a general register or memory operand to 0. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.>	
BNOT	В	¬ ( <bit-no.> of <ead>) → (<bit-no.> of <ead>) Inverts a specified bit in a general register or memory operand. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.></ead></bit-no.>	
BTST	В	¬ ( <bit-no.> of <ead>) → Z Tests a specified bit in a general register or memory operand and sets or clears the Z flag accordingly. The bit number is specified by 3-bit immediate data or the lower three bits of a general register.</ead></bit-no.>	
BAND	В	$C \land ( of ) \rightarrow C$ ANDs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.	
BIAND	В	$C \land \neg$ ( <bit-no.> of <ead>) <math>\rightarrow C</math> ANDs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>	
		The bit number is specified by 3-bit immediate data.	
BOR	В	$C \lor ( of ) \rightarrow C$ ORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.	
BIOR	В	$C \lor \neg$ ( <bit-no.> of <ead>) <math>\rightarrow C</math> ORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>	
		The bit number is specified by 3-bit immediate data.	
Note: * F	Refers to the	e operand size.	

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#### Table 2.6 **Bit Manipulation Instructions (1)**

B: Byte

Instruction	Size*	Function	
BXOR	В	$C \oplus (\text{-bit-No.> of } \text{-EAd>}) \rightarrow C$ XORs the carry flag with a specified bit in a general register or memory operand and stores the result in the carry flag.	
BIXOR	В	$C \oplus \neg$ ( <bit-no.> of <ead>) <math>\rightarrow C</math> XORs the carry flag with the inverse of a specified bit in a general register or memory operand and stores the result in the carry flag.</ead></bit-no.>	
		The bit number is specified by 3-bit immediate data.	
BLD	В	( <bit-no.> of <ead>) <math display="inline">\rightarrow</math> C Transfers a specified bit in a general register or memory operand to the carry flag.</ead></bit-no.>	
BILD	В	¬ ( <bit-no.> of <ead>) → C Transfers the inverse of a specified bit in a general register or memory operand to the carry flag.</ead></bit-no.>	
		The bit number is specified by 3-bit immediate data.	
BST	В	$C \rightarrow$ ( <bit-no.> of <ead>) Transfers the carry flag value to a specified bit in a general register or memory operand.</ead></bit-no.>	
BIST	В	$\neg$ C $\rightarrow$ ( <bit-no.> of <ead>) Transfers the inverse of the carry flag value to a specified bit in a general register or memory operand.</ead></bit-no.>	
		The bit number is specified by 3-bit immediate data.	
	Refers to the B: Byte	e operand size.	

 Table 2.6
 Bit Manipulation Instructions (2)



Table 2.7	Branch	Instructions
-----------	--------	--------------

Instruction	Size	Function			
Bcc*	—	Branches to a specified address if a specified condition is true. The branching conditions are listed below.			
		Mnemonic	Description	Condition	
		BRA(BT)	Always (true)	Always	
		BRN(BF)	Never (false)	Never	
		BHI	High	$C \lor Z = 0$	
		BLS	Low or same	C ∨ Z = 1	
		BCC(BHS)	Carry clear (high or same)	C = 0	
		BCS(BLO)	Carry set (low)	C = 1	
		BNE	Not equal	Z = 0	
		BEQ	Equal	Z = 1	
		BVC	Overflow clear	V = 0	
		BVS	Overflow set	V = 1	
		BPL	Plus	N = 0	
		BMI	Minus	N = 1	
		BGE	Greater or equal	$N \oplus V = 0$	
		BLT	Less than	$N \oplus V = 1$	
		BGT	Greater than	$Z \vee (N \oplus V) = 0$	
		BLE	Less or equal	$Z_{\vee}(N \oplus V) = 1$	
JMP	_	Branches unco	nditionally to a specified	address.	
BSR		Branches to a subroutine at a specified address.			
JSR		Branches to a subroutine at a specified address. Returns from a subroutine			
RTS					
Note: * Bo	cc is the g	jeneral name for conditional branch instructions.			

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Instructio	n Size*	Function	
RTE	_	Returns from an exception-handling routine.	
SLEEP	—	Causes a transition to a power-down state.	
LDC	B/W	$(EAs) \rightarrow CCR$ Moves the source operand contents to the CCR. The CCR size is one byte, but in transfer from memory, data is read by word access.	
STC	B/W	$CCR \rightarrow$ (EAd) Transfers the CCR contents to a destination location. The condition code register size is one byte, but in transfer to memory, data is written by word access.	
ANDC	В	CCR $\land$ #IMM $\rightarrow$ CCR Logically ANDs the CCR with immediate data.	
ORC	В	$CCR \lor \#IMM \rightarrow CCR$ Logically ORs the CCR with immediate data.	
XORC	В	$CCR \oplus \#IMM \rightarrow CCR$ Logically XORs the CCR with immediate data.	
NOP	—	$PC + 2 \rightarrow PC$ Only increments the program counter.	
Note: *	Refers to the	e operand size.	
	B: Byte		
	W: Word		

#### Table 2.8 System Control Instructions



Instruction	Size	Function
EEPMOV.B	_	if R4L ≠ 0 then Repeat @ER5+ → @ER6+, R4L-1 → R4L Until R4L = 0 else next;
EEPMOV.W	_	if R4 $\neq$ 0 then Repeat @ER5+ $\rightarrow$ @ER6+, R4-1 $\rightarrow$ R4 Until R4 = 0 else next;
		Transfers a data block. Starting from the address set in ER5, transfers data for the number of bytes set in R4L or R4 to the address location set in ER6.
		Execution of the next instruction begins as soon as the transfer is completed.

### Table 2.9 Block Data Transfer Instructions



#### 2.4.2 Basic Instruction Formats

H8/300H CPU instructions consist of 2-byte (1-word) units. An instruction consists of an operation field (op), a register field (r), an effective address extension (EA), and a condition field (cc).

Figure 2.7 shows examples of instruction formats.

#### (1) Operation Field

Indicates the function of the instruction, the addressing mode, and the operation to be carried out on the operand. The operation field always includes the first four bits of the instruction. Some instructions have two operation fields.

#### (2) Register Field

Specifies a general register. Address registers are specified by 3 bits, and data registers by 3 bits or 4 bits. Some instructions have two register fields. Some have no register field.

#### (3) Effective Address Extension

8, 16, or 32 bits specifying immediate data, an absolute address, or a displacement. A24-bit address or displacement is treated as a 32-bit data in which the first 8 bits are 0 (H'00).

#### (4) Condition Field

Specifies the branching condition of Bcc instructions.

(1) O	(1) Operation field only					
	ор			NOP, RTS, etc.		
(2) O	peration field and register fie	_				
	ор	rn	rm	ADD.B Rn, Rm, etc.		
(3) O	(3) Operation field, register fields, and effective address extension					
	ор	rn	rm	MOV.B @(d:16, Rn), Rm		
	EA(disp)					
(4) Operation field, effective address extension, and condition field						
	ор сс	EA	(disp)	BRA d:8		

#### Figure 2.7 Instruction Formats

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### 2.5 Addressing Modes and Effective Address Calculation

The following describes the H8/300H CPU. In this LSI, the upper eight bits are ignored in the generated 24-bit address, so the effective address is 16 bits.

### 2.5.1 Addressing Modes

The H8/300H CPU supports the eight addressing modes listed in table 2.10. Each instruction uses a subset of these addressing modes. Addressing modes that can be used differ depending on the instruction. For details, refer to appendix A.4, Combinations of Instructions and Addressing Modes.

Arithmetic and logic instructions can use the register direct and immediate modes. Data transfer instructions can use all addressing modes except program-counter relative and memory indirect. Bit-manipulation instructions use register direct, register indirect, or the absolute addressing mode (@aa:8) to specify an operand, and register direct (BSET, BCLR, BNOT, and BTST instructions) or immediate (3-bit) addressing mode to specify a bit number in the operand.

No.	Addressing Mode	Symbol
1	Register direct	Rn
2	Register indirect	@ERn
3	Register indirect with displacement	@(d:16,ERn)/@(d:24,ERn)
4	Register indirect with post-increment Register indirect with pre-decrement	@ERn+ @-ERn
5	Absolute address	@aa:8/@aa:16/@aa:24
6	Immediate	#xx:8/#xx:16/#xx:32
7	Program-counter relative	@(d:8,PC)/@(d:16,PC)
8	Memory indirect	@@aa:8

#### Table 2.10 Addressing Modes

#### (1) Register Direct—Rn

The register field of the instruction specifies an 8-, 16-, or 32-bit general register containing the operand. R0H to R7H and R0L to R7L can be specified as 8-bit registers. R0 to R7 and E0 to E7 can be specified as 16-bit registers. ER0 to ER7 can be specified as 32-bit registers.

#### (2) Register Indirect—@ERn

The register field of the instruction code specifies an address register (ERn), the lower 24 bits of which contain the address of the operand on memory.

#### (3) Register Indirect with Displacement—@(d:16, ERn) or @(d:24, ERn)

A 16-bit or 24-bit displacement contained in the instruction is added to an address register (ERn) specified by the register field of the instruction, and the lower 24 bits of the sum the address of a memory operand. A 16-bit displacement is sign-extended when added.

#### (4) Register Indirect with Post-Increment or Pre-Decrement—@ERn+ or @-ERn

• Register indirect with post-increment—@ERn+

The register field of the instruction code specifies an address register (ERn) the lower 24 bits of which contains the address of a memory operand. After the operand is accessed, 1, 2, or 4 is added to the address register contents (32 bits) and the sum is stored in the address register. The value added is 1 for byte access, 2 for word access, or 4 for longword access. For the word or longword access, the register value should be even.

• Register indirect with pre-decrement—@-ERn

The value 1, 2, or 4 is subtracted from an address register (ERn) specified by the register field in the instruction code, and the lower 24 bits of the result is the address of a memory operand. The result is also stored in the address register. The value subtracted is 1 for byte access, 2 for word access, or 4 for longword access. For the word or longword access, the register value should be even.

#### (5) Absolute Address—@aa:8, @aa:16, @aa:24

The instruction code contains the absolute address of a memory operand. The absolute address may be 8 bits long (@aa:8), 16 bits long (@aa:16), 24 bits long (@aa:24)

For an 8-bit absolute address, the upper 16 bits are all assumed to be 1 (H'FFFF). For a 16-bit absolute address the upper 8 bits are a sign extension. A 24-bit absolute address can access the entire address space.

The access ranges of absolute addresses for this LSI are those shown in table 2.11, because the upper 8 bits are ignored.



Absolute Address	Access Range	
8 bits (@aa:8)	H'FF00 to H'FFFF	
16 bits (@aa:16)	H'0000 to H'FFFF	
24 bits (@aa:24)	H'0000 to H'FFFF	

#### Table 2.11 Absolute Address Access Ranges

#### (6) Immediate—#xx:8, #xx:16, or #xx:32

The instruction contains 8-bit (#xx:8), 16-bit (#xx:16), or 32-bit (#xx:32) immediate data as an operand.

The ADDS, SUBS, INC, and DEC instructions contain immediate data implicitly. Some bit manipulation instructions contain 3-bit immediate data in the instruction code, specifying a bit number.

#### (7) Program-Counter Relative—@(d:8, PC) or @(d:16, PC)

This mode is used in the BSR instruction. An 8-bit or 16-bit displacement contained in the instruction is sign-extended and added to the 24-bit PC contents to generate a branch address. The PC value to which the displacement is added is the address of the first byte of the next instruction, so the possible branching range is -126 to +128 bytes (-63 to +64 words) or -32766 to +32768 bytes (-16383 to +16384 words) from the branch instruction. The resulting value should be an even number.

#### (8) Memory Indirect—@@aa:8

This mode can be used by the JMP and JSR instructions. The instruction code contains an 8-bit absolute address specifying a memory operand. This memory operand contains a branch address. The memory operand is accessed in words, generating a 16-bit branch address. Figure 2.8 shows how to specify branch address for in memory indirect mode. The upper bits of the absolute address are all assumed to be 0, so the address range is 0 to 255 (H'0000 to H'00FF).

Note that the first part of the address range is also the exception vector area.





Figure 2.8 Branch Address Specification in Memory Indirect Mode



#### 2.5.2 Effective Address Calculation

Table 2.12 indicates how effective addresses are calculated in each addressing mode. In this LSI the upper 8 bits of the effective address are ignored in order to generate a 16-bit effective address.

 Table 2.12
 Effective Address Calculation (1)



No	Addressing Mode and Instruction Format	Effective Address Calculation	Effective Address (EA)
5	Absolute address @aa:8 op abs		23 8 7 0 H'FFFF
	@aa:16 op abs		23 16 15 0 Sign extension
	@aa:24 op abs		23 0 []
6	Immediate #xx:8/#xx:16/#xx:32 op IMM		Operand is immediate data.
7	Program-counter relative @(d:8,PC)/@(d:16,PC) op disp	23 0 PC contents 23 0 Sign disp	23 0
8	Memory indirect @@aa:8	23 8 7 0 H'0000 abs 15 0 Memory contents	23 16 15 0 H'00

#### Table 2.12 Effective Address Calculation (2)

[Legend]

r, rm,rn : Register field

op : Operation field

disp : Displacement

IMM : Immediate data

abs : Absolute address



### 2.6 Basic Bus Cycle

CPU operation is synchronized by a system clock ( $\phi$ ) or a subclock ( $\phi_{sub}$ ). The period from a rising edge of  $\phi$  or  $\phi_{sub}$  to the next rising edge is called one state. A bus cycle consists of two states or three states. The cycle differs depending on whether access is to on-chip memory or to on-chip peripheral modules.

#### 2.6.1 Access to On-Chip Memory (RAM, ROM)

Access to on-chip memory takes place in two states. The data bus width is 16 bits, allowing access in byte or word size. Figure 2.9 shows the on-chip memory access cycle.



Figure 2.9 On-Chip Memory Access Cycle

#### 2.6.2 On-Chip Peripheral Modules

On-chip peripheral modules are accessed in two states or three states. The data bus width is 8 bits or 16 bits depending on the register. For description on the data bus width and number of accessing states of each register, refer to section 16.1, Register Addresses (Address Order). Registers with 16-bit data bus width can be accessed by word size only. Registers with 8-bit data bus width can be accessed by word size. When a register with 8-bit data bus width is accessed by word size, a bus cycle occurs twice. In two-state access, the operation timing is the same as that for on-chip memory.

Figure 2.10 shows the operation timing in the case of three-state access to an on-chip peripheral module.



Figure 2.10 On-Chip Peripheral Module Access Cycle (3-State Access)

### 2.7 CPU States

There are four CPU states: the reset state, program execution state, program halt state, and exception-handling state. The program execution state includes active (high-speed or medium-speed) mode and subactive mode. For the program halt state, there are sleep (high-speed or medium-speed) mode, standby mode, watch mode, and subsleep mode. These states are shown in figure 2.11. Figure 2.12 shows the state transitions. For details on program execution state and program halt state, refer to section 5, Power-Down Modes. For details on exception handling, refer to section 3, Exception Handling.



### Figure 2.11 CPU Operating States





Figure 2.12 State Transitions

#### 2.8 Usage Notes

#### 2.8.1 Notes on Data Access to Empty Areas

The address space of this LSI includes empty areas in addition to the ROM, RAM, and on-chip I/O registers areas available to the user. When data is transferred from CPU to empty areas, the transferred data will be lost. This action may also cause the CPU to malfunction. When data is transferred from an empty area to CPU, the contents of the data cannot be guaranteed.

#### 2.8.2 EEPMOV Instruction

EEPMOV is a block-transfer instruction and transfers the byte size of data indicated by R4L, which starts from the address indicated by R5, to the address indicated by R6. Set R4L and R6 so that the end address of the destination address (value of R6 + R4L) does not exceed H'FFFF (the value of R6 must not change from H'FFFF to H'0000 during execution).



#### 2.8.3 Bit-Manipulation Instruction

The BSET, BCLR, BNOT, BST, and BIST instructions read data from the specified address in byte units, manipulate the data of the target bit, and write data to the same address again in byte units. Special care is required when using these instructions in cases where two registers are assigned to the same address, or when a bit is directly manipulated for a port or a register containing a write-only bit, because this may rewrite data of a bit other than the bit to be manipulated.

#### (1) Bit manipulation for two registers assigned to the same address

#### Example 1: Bit manipulation for the timer load register and timer counter

Figure 2.13 shows an example of a timer in which two timer registers are assigned to the same address. When a bit-manipulation instruction accesses the timer load register and timer counter of a reloadable timer, since these two registers share the same address, the following operations takes place.

- 1. Data is read in byte units.
- 2. The CPU sets or resets the bit to be manipulated with the bit-manipulation instruction.
- 3. The written data is written again in byte units to the timer load register.

The timer is counting, so the value read is not necessarily the same as the value in the timer load register. As a result, bits other than the intended bit in the timer counter may be modified and the modified value may be written to the timer load register.



Figure 2.13 Example of Timer Configuration with Two Registers Allocated to Same Address

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# Example 2: When the BSET instruction is executed for port 5

P57 and P56 are input pins, with a low-level signal input at P57 and a high-level signal input at P56. P55 to P50 are output pins and output low-level signals. An example to output a high-level signal at P50 with a BSET instruction is shown below.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	Low level
PCR5	0	0	1	1	1	1	1	1
PDR5	1	0	0	0	0	0	0	0

• Prior to executing BSET instruction

### • BSET instruction executed

BSET #0, @PDR5

The BSET instruction is executed for port 5.

• After executing BSET instruction

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	High level
PCR5	0	0	1	1	1	1	1	1
PDR5	0	1	0	0	0	0	0	1

- Description on operation
- 1. When the BSET instruction is executed, first the CPU reads port 5.

Since P57 and P56 are input pins, the CPU reads the pin states (low-level and high-level input).

P55 to P50 are output pins, so the CPU reads the value in PDR5. In this example PDR5 has a value of H'80, but the value read by the CPU is H'40.

2. Next, the CPU sets bit 0 of the read data to 1, changing the PDR5 data to H'41.



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3. Finally, the CPU writes H'41 to PDR5, completing execution of BSET instruction. As a result of the BSET instruction, bit 0 in PDR5 becomes 1, and P50 outputs a high-level signal. However, bits 7 and 6 of PDR5 end up with different values. To prevent this problem, store a copy of the PDR5 data in a work area in memory. Perform the bit manipulation on the data in the work area, then write this data to PDR5.

υ	
#H'80,	ROL
ROL,	@RAM0
ROL,	@PDR5
	ROL,

• Prior to executing BSET instruction

The PDR5 value (H'80) is written to a work area in memory (RAM0) as well as to PDR5.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	Low level
PCR5	0	0	1	1	1	1	1	1
PDR5	1	0	0	0	0	0	0	0
RAM0	1	0	0	0	0	0	0	0

• BSET instruction executed

BSET #0, @RAMO

The BSET instruction is executed designating the PDR5 work area (RAM0).

# • After executing BSET instruction

MOV.B @RAMO, ROL MOV.B ROL, @PDR5

The work area (RAM0) value is written to PDR5.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	High level
PCR5	0	0	1	1	1	1	1	1
PDR5	1	0	0	0	0	0	0	1
RAM0	1	0	0	0	0	0	0	1



# (2) Bit Manipulation in a Register Containing a Write-Only Bit

# Example 3: BCLR instruction executed designating port 5 control register PCR5

P57 and P56 are input pins, with a low-level signal input at P57 and a high-level signal input at P56. P55 to P50 are output pins that output low-level signals. An example of setting the P50 pin as an input pin by the BCLR instruction is shown below. It is assumed that a high-level signal will be input to this input pin.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	Low level
PCR5	0	0	1	1	1	1	1	1
PDR5	1	0	0	0	0	0	0	0

• Prior to executing BCLR instruction

### BCLR instruction executed

BCLR #0, @PCR5

The BCLR instruction is executed for PCR5.

# • After executing BCLR instruction

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Output	Output	Output	Output	Output	Output	Output	Input
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	High level
PCR5	1	1	1	1	1	1	1	0
PDR5	1	0	0	0	0	0	0	0

- Description on operation
- 1. When the BCLR instruction is executed, first the CPU reads PCR5. Since PCR5 is a write-only register, the CPU reads a value of H'FF, even though the PCR5 value is actually H'3F.
- 2. Next, the CPU clears bit 0 in the read data to 0, changing the data to H'FE.
- 3. Finally, H'FE is written to PCR5 and BCLR instruction execution ends.

As a result of this operation, bit 0 in PCR5 becomes 0, making P50 an input port. However, bits 7 and 6 in PCR5 change to 1, so that P57 and P56 change from input pins to output pins. To prevent this problem, store a copy of the PDR5 data in a work area in memory and manipulate data of the bit in the work area, then write this data to PDR5.

### • Prior to executing BCLR instruction

MOV.B	#H'3F,	ROL
MOV.B	ROL,	@RAM0
MOV.B	ROL,	@PCR5

The PCR5 value (H'3F) is written to a work area in memory (RAM0) as well as to PCR5.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	Low level
PCR5	0	0	1	1	1	1	1	1
PDR5	1	0	0	0	0	0	0	0
RAM0	0	0	1	1	1	1	1	1

#### • BCLR instruction executed

BCLR #0, @RAMO

The BCLR instructions executed for the PCR5 work area (RAM0).

# • After executing BCLR instruction

MOV.B	@RAM0,	ROL
MOV.B	ROL,	@PCR5

The work area (RAM0) value is written to PCR5.

	P57	P56	P55	P54	P53	P52	P51	P50
Input/output	Input	Input	Output	Output	Output	Output	Output	Output
Pin state	Low level	High level	Low level	Low level	Low level	Low level	Low level	High Ievel
PCR5	0	0	1	1	1	1	1	0
PDR5	1	0	0	0	0	0	0	0
RAM0	0	0	1	1	1	1	1	0

# Section 3 Exception Handling

# 3.1 Overview

Exception handling is performed when a reset or interrupt occurs. Table 3.1 shows the priorities of these two types of exception handling.

 Table 3.1
 Exception Handling Types and Priorities

Priority	Exception Source	Time of Start of Exception Handling
High	Reset	Exception handling starts as soon as the reset state is cleared
Low	Interrupt	When an interrupt is requested, exception handling starts after execution of the present instruction or the exception handling in progress is completed

# 3.2 Reset

# 3.2.1 Overview

A reset is the highest-priority exception. The internal state of the CPU and the registers of the onchip peripheral modules are initialized.

# 3.2.2 Reset Sequence

As soon as the  $\overline{\text{RES}}$  pin goes low, all processing is stopped and the chip enters the reset state.

To make sure the chip is reset properly, observe the following precautions.

- At power on: Hold the  $\overline{\text{RES}}$  pin low until the clock pulse generator output stabilizes.
- Resetting during operation: Hold the  $\overline{\text{RES}}$  pin low for at least 10 system clock cycles.

Reset exception handling takes place as follows.

- The CPU internal state and the registers of on-chip peripheral modules are initialized, with the I bit of the condition code register (CCR) set to 1.
- The PC is loaded from the reset exception handling vector address (H'0000 to H'0001), after which the program starts executing from the address indicated in PC.



When system power is turned on or off, the  $\overline{\text{RES}}$  pin should be held low.

Figure 3.1 shows the reset sequence starting from  $\overline{\text{RES}}$  input.

See section 14.3.1, Power-On Reset Circuit.



Figure 3.1 Reset Sequence



# 3.2.3 Interrupt Immediately after Reset

After a reset, if an interrupt were to be accepted before the stack pointer (SP: R7) was initialized, PC and CCR would not be pushed onto the stack correctly, resulting in program runaway. To prevent this, immediately after reset exception handling all interrupts are masked. For this reason, the initial program instruction is always executed immediately after a reset. This instruction should initialize the stack pointer (e.g. MOV.W #xx: 16, SP).

# 3.3 Interrupts

# 3.3.1 Overview

The interrupt sources include 13 external interrupts (WKP<sub>7</sub> to WKP<sub>0</sub>, IRQ<sub>4</sub>, IRQ<sub>3</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub>, IRQAEC) and 9 internal interrupts from on-chip peripheral modules. Table 3.2 shows the interrupt sources, their priorities, and their vector addresses. When more than one interrupt is requested, the interrupt with the highest priority is processed.

The interrupts have the following features:

- Internal and external interrupts can be masked by the I bit in CCR. When the I bit is set to 1, interrupt request flags can be set but the interrupts are not accepted.
- IRQ<sub>4</sub>, IRQ<sub>3</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub>, and WKP<sub>7</sub> to WKP<sub>0</sub> can be set to either rising edge sensing or falling edge sensing, and IRQAEC can be set to either rising edge sensing, falling edge sensing, or both edge sensing.



Interrupt Source	Interrupt	Vector Number	Vector Address	Priority
RES	Reset	0	H'0000 to H'0001	High
Watchdog timer				Ť
IRQ, LVDI	IRQ <sub>0</sub>	4	H'0008 to H'0009	_
	Low-voltage detect interrupt	_		-
	IRQ,	5	H'000A to H'000B	_
	IRQAEC	6	H'000C to H'000D	_
<b>IRQ</b> <sub>3</sub>	IRQ <sub>3</sub>	7	H'000E to H'000F	_
<b>IRQ</b> ₄	IRQ₄	8	H'0010 to H'0011	
WKP WKP WKP WKP WKP WKP WKP	WKP <sup>0</sup> WKP <sup>1</sup> WKP <sup>2</sup> WKP <sup>3</sup> WKP <sup>4</sup> WKP <sup>5</sup> WKP <sup>6</sup> WKP <sup>7</sup>	9	H'0012 to H'0013	_
Timer A	Timer A overflow	11	H'0016 to H'0017	-
Asynchronous event counter	Asynchronous event counter overflow	12	H'0018 to H'0019	-
Timer C	Timer C overflow or underflow	13	H'001A to H'001B	-
Timer FL	Timer FL compare match Timer FL overflow	14	H'001C to H'001D	-
Timer FH	Timer FH compare match Timer FH overflow	15	H'001E to H'001F	-
Timer G	Timer G input capture Timer G overflow	16	H'0020 to H'0021	_
SCI3	SCI3 transmit end SCI3 transmit data empty SCI3 receive data full SCI3 overrun error SCI3 framing error SCI3 parity error	18	H'0024 to H'0025	
A/D	A/D conversion end	19	H'0026 to H'0027	¥
(SLEEP instruction executed)	Direct transfer	20	H'0028 to H'0029	Low

# Table 3.2 Interrupt Sources and Their Priorities

Notes: Vector addresses H'0002 to H'0007, H'0014 to H'0015, and H'0022 to H'0023 are reserved and cannot be used.

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## 3.3.2 Interrupt Control Registers

Table 3.3 lists the registers that control interrupts.

#### Table 3.3 Interrupt Control Registers

Name	Abbreviation	R/W	Initial Value	Address
IRQ edge select register	IEGR	R/W	_	H'FFF2
Interrupt enable register 1	IENR1	R/W	_	H'FFF3
Interrupt enable register 2	IENR2	R/W	—	H'FFF4
Interrupt request register 1	IRR1	R/W*	—	H'FFF6
Interrupt request register 2	IRR2	R/W*	—	H'FFF7
Wakeup interrupt request register	IWPR	R/W*	H'00	H'FFF9
Wakeup edge select register	WEGR	R/W	H'00	H'FF90

Note: \* Write is enabled only for writing of 0 to clear a flag.

### (1) IRQ Edge Select Register (IEGR)

Bit	7	6	5	4	3	2	1	0
				IEG4	IEG3	—	IEG1	IEG0
Initial value	1	1	1	0	0		0	0
Read/Write				R/W	R/W	W	R/W	R/W

IEGR is an 8-bit read/write register used to designate whether pins  $\overline{IRQ}_4$ ,  $\overline{IRQ}_3$ ,  $\overline{IRQ}_1$ , and  $\overline{IRQ}_0$  are set to rising edge sensing or falling edge sensing. For the IRQAEC pin edge sensing specifications, see section 9.7, Asynchronous Event Counter (AEC).

### Bits 7 to 5—Reserved

Bits 7 to 5 are reserved: they are always read as 1 and cannot be modified.



# Bit 4—IRQ<sub>4</sub> Edge Select (IEG4)

Bit 4 selects the input sensing of the  $\overline{IRQ}_4$  pin and  $\overline{ADTRG}$  pin.

Bit 4 IEG4	Description	
0	Falling edge of $\overline{IRQ}_4$ and $\overline{ADTRG}$ pin input is detected	(initial value)
1	Rising edge of $\overline{IRQ}_4$ and $\overline{ADTRG}$ pin input is detected	

### Bit 3—IRQ<sub>3</sub> Edge Select (IEG3)

Bit 3 selects the input sensing of the  $\overline{IRQ}_3$  pin and TMIF pin.

Bit 3 IEG3	Description	
0	Falling edge of $\overline{\text{IRQ}}_{_{\!3}}$ and TMIF pin input is detected	(initial value)
1	Rising edge of $\overline{IRQ}_{_{\!3}}$ and TMIF pin input is detected	

#### Bit 2—Reserved

Bit 2 is reserved: it can only be written with 0.

### Bit 1—IRQ, Edge Select (IEG1)

Bit 1 selects the input sensing of the  $\overline{IRQ}_1$  pin and TMIC pin.

Bit 1 IEG1	Description	
0	Falling edge of $\overline{IRQ}_1$ and TMIC pin input is detected	(initial value)
1	Rising edge of $\overline{IRQ}_1$ and TMIC pin input is detected	

# Bit 0—IRQ<sub>0</sub> Edge Select (IEG0)

Bit 0 selects the input sensing of pin  $\overline{IRQ}_0$ .

Bit 0 IEG0	Description	
0	Falling edge of $\overline{IRQ}_{_0}$ pin input is detected	(initial value)
1	Rising edge of $\overline{IRQ}_0$ pin input is detected	

Bit	7	6	5	4	3	2	1	0
	IENTA		IENWP	IEN4	IEN3	IENEC2	IEN1	IEN0
Initial value	0		0	0	0	0	0	0
Read/Write	R/W	W	R/W	R/W	R/W	R/W	R/W	R/W

#### (2) Interrupt Enable Register 1 (IENR1)

IENR1 is an 8-bit read/write register that enables or disables interrupt requests.

#### Bit 7—Timer A Interrupt Enable (IENTA)

Bit 7 enables or disables timer A overflow interrupt requests.

Bit 7 IENTA	Description	
0	Disables timer A interrupt requests	(initial value)
1	Enables timer A interrupt requests	

#### Bit 6—Reserved

Bit 6 is reserved: it can only be written with 0.

#### Bit 5—Wakeup Interrupt Enable (IENWP)

Bit 5 enables or disables WKP<sub>7</sub> to WKP<sub>0</sub> interrupt requests.

Bit 5 IENWP	Description	
0	Disables $\overline{WKP}_{7}$ to $\overline{WKP}_{0}$ interrupt requests	(initial value)
1	Enables $\overline{WKP}_7$ to $\overline{WKP}_0$ interrupt requests	

### Bits 4 and 3—IRQ<sub>4</sub> and IRQ<sub>3</sub> Interrupt Enable (IEN4 and IEN3)

Bits 4 and 3 enable or disable IRQ<sub>4</sub> and IRQ<sub>3</sub> interrupt requests.

Bit n IENn	Description	
0	Disables interrupt requests from pin IRQn	(initial value)
1	Enables interrupt requests from pin IRQn	

(n = 4 or 3)



# Bit 2—IRQAEC Interrupt Enable (IENEC2)

Bit 2 enables or disables IRQAEC interrupt requests.

Bit 2 IENEC2	Description	
0	Disables IRQAEC interrupt requests	(initial value)
1	Enables IRQAEC interrupt requests	

# Bits 1 and 0—IRQ, and IRQ, Interrupt Enable (IEN1 and IEN0)

Bits 1 and 0 enable or disable  $IRQ_1$  and  $IRQ_0$  interrupt requests.

Bit n IENn	Description	
0	Disables interrupt requests from pin IRQn	(initial value)
1	Enables interrupt requests from pin IRQn	

(n = 1 or 0)



Bit	7	6	5	4	3	2	1	0
	IENDT	IENAD	_	IENTG	IENTFH	IENTFL	IENTC	IENEC
Initial value	0	0	_	0	0	0	0	0
Read/Write	R/W	R/W	W	R/W	R/W	R/W	R/W	R/W

#### (3) Interrupt Enable Register 2 (IENR2)

IENR2 is an 8-bit read/write register that enables or disables interrupt requests.

# Bit 7—Direct Transfer Interrupt Enable (IENDT)

Bit 7 enables or disables direct transfer interrupt requests.

Bit 7 IENDT	Description	
0	Disables direct transfer interrupt requests	(initial value)
1	Enables direct transfer interrupt requests	

#### Bit 6—A/D Converter Interrupt Enable (IENAD)

Bit 6 enables or disables A/D converter interrupt requests.

Bit 6 IENAD	Description	
0	Disables A/D converter interrupt requests	(initial value)
1	Enables A/D converter interrupt requests	

#### Bit 5—Reserved

Bit 5 is reserved bit: it can only be written with 0.



## Bit 4—Timer G Interrupt Enable (IENTG)

Bit 4 enables or disables timer G input capture or overflow interrupt requests.

Bit 4 IENTG	Description	
0	Disables timer G interrupt requests	(initial value)
1	Enables timer G interrupt requests	

#### Bit 3—Timer FH Interrupt Enable (IENTFH)

Bit 3 enables or disables timer FH compare match and overflow interrupt requests.

Bit 3 IENTFH	Description	
0	Disables timer FH interrupt requests	(initial value)
1	Enables timer FH interrupt requests	

### Bit 2—Timer FL Interrupt Enable (IENTFL)

Bit 2 enables or disables timer FL compare match and overflow interrupt requests.

Bit 2 IENTFL	Description	
0	Disables timer FL interrupt requests	(initial value)
1	Enables timer FL interrupt requests	

### Bit 1—Timer C Interrupt Enable (IENTC)

Bit 1 enables or disables timer C overflow and underflow interrupt requests.

Bit 1 IENTC	Description	
0	Disables timer C interrupt requests	(initial value)
1	Enables timer C interrupt requests	

### Bit 0—Asynchronous Event Counter Interrupt Enable (IENEC)

Bit 0 enables or disables asynchronous event counter interrupt requests.

Bit 0 IENEC	Description	
0	Disables asynchronous event counter interrupt requests	(initial value)
1	Enables asynchronous event counter interrupt requests	

For details of SCI3 interrupt control, see section 10.2.6, Serial control register 3 (SCR3).

### (4) Interrupt Request Register 1 (IRR1)

Bit	7	6	5	4	3	2	1	0
	IRRTA			IRRI4	IRRI3	IRREC2	IRRI1	IRRI0
Initial value	0		1	0	0	0	0	0
Read/Write	R/(W)*	W		R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*

Note: \* Only a write of 0 for flag clearing is possible

IRR1 is an 8-bit read/write register, in which a corresponding flag is set to 1 when a timer A, IRQAEC,  $IRQ_4$ ,  $IRQ_3$ ,  $IRQ_1$ , or  $IRQ_0$  interrupt is requested. The flags are not cleared automatically when an interrupt is accepted. It is necessary to write 0 to clear each flag.

#### Bit 7—Timer A Interrupt Request Flag (IRRTA)

Bit 7 IRRTA	Description	
0	Clearing conditions: When IRRTA = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When the timer A counter value overflows	

#### Bit 6—Reserved

Bit 6 is reserved; it can only be written with 0.



# Bit 5—Reserved

Bit 5 is reserved; it is always read as 1 and cannot be modified.

# Bits 4 and 3—IRQ<sub>4</sub> and IRQ<sub>3</sub> Interrupt Request Flags (IRRI4 and IRRI3)

Bit n IRRIn	Description	
0	Clearing conditions: (initial va When IRRIn = 1, it is cleared by writing 0	lue)
1	Setting conditions: When pin IRQn is designated for interrupt input and the designated signal edge input	is

(n = 4 or 3)

### Bit 2—IRQAEC Interrupt Request Flag (IRREC2)

Bit 2 IRREC2	Description	
0	Clearing conditions: When IRREC2 = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When pin IRQAEC is designated for interrupt input and the designated is input	signal edge

# Bits 1 and 0—IRQ<sub>1</sub> and IRQ<sub>0</sub> Interrupt Request Flags (IRRI1 and IRRI0)

Bit n IRRIn	Description	
0	Clearing conditions: When IRRIn = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When pin IRQn is designated for interrupt input and the designated signated signat	gnal edge is

(n = 1 or 0)



Bit	7	6	5	4	3	2	1	0
	IRRDT	IRRAD	_	IRRTG	IRRTFH	IRRTFL	IRRTC	IRREC
Initial value	0	0	_	0	0	0	0	0
Read/Write	R/(W)*	R/(W)*	W	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*

#### (5) Interrupt Request Register 2 (IRR2)

Note: \* Only a write of 0 for flag clearing is possible

IRR2 is an 8-bit read/write register, in which a corresponding flag is set to 1 when a direct transfer, A/D converter, Timer G, Timer FH, Timer FL, Timer C, or asynchronous event counter interrupt is requested. The flags are not cleared automatically when an interrupt is accepted. It is necessary to write 0 to clear each flag.

#### Bit 7—Direct Transfer Interrupt Request Flag (IRRDT)

Bit 7 IRRDT	Description	
0	Clearing conditions: When IRRDT = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When a direct transfer is made by executing a SLEEP instruction whi SYSCR2	le DTON = 1 in

#### Bit 6—A/D Converter Interrupt Request Flag (IRRAD)

Bit 6 IRRAD	Description	
0	Clearing conditions: When IRRAD = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When A/D conversion is completed and ADSF is cleared to 0 in ADSF	3

### Bit 5—Reserved

Bit 5 is reserved: it can only be written with 0.



# Bit 4—Timer G Interrupt Request Flag (IRRTG)

Bit 4 IRRTG	Description	
0	Clearing conditions: When IRRTG = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When the TMIG pin is designated for TMIG input and the designated input, and when TCG overflows while OVIE is set to 1 in TMG	signal edge is

# Bit 3—Timer FH Interrupt Request Flag (IRRTFH)

Bit 3 IRRTFH	Description	
0	Clearing conditions: When IRRTFH = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When TCFH and OCRFH match in 8-bit timer mode, or when TCF (TC and OCRF (OCRFL, OCRFH) match in 16-bit timer mode	CFL, TCFH)

# Bit 2—Timer FL Interrupt Request Flag (IRRTFL)

Bit 2 IRRTFL	Description	
0	Clearing conditions: When IRRTFL = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When TCFL and OCRFL match in 8-bit timer mode	

# Bit 1—Timer C Interrupt Request Flag (IRRTC)

Bit 1 IRRTC	Description	
0	Clearing conditions: When IRRTC = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When the timer C counter value overflows or underflows	

Bit 0 IRREC	Description	
0	Clearing conditions: When IRREC = 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When ECH overflows in 16-bit counter mode, or ECH or ECL overflow counter mode	vs in 8-bit

### Bit 0—Asynchronous Event Counter Interrupt Request Flag (IRREC)

#### (6) Wakeup Interrupt Request Register (IWPR)

Bit	7	6	5	4	3	2	1	0
	IWPF7	IWPF6	IWPF5	IWPF4	IWPF3	IWPF2	IWPF1	IWPF0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/(W)*							

Note: \* Only a write of 0 for flag clearing is possible

IWPR is an 8-bit read/write register containing wakeup interrupt request flags. When one of pins  $\overline{WKP}_7$  to  $\overline{WKP}_0$  is designated for wakeup input and a rising or falling edge is input at that pin, the corresponding flag in IWPR is set to 1. A flag is not cleared automatically when the corresponding interrupt is accepted. Flags must be cleared by writing 0.

#### Bits 7 to 0—Wakeup Interrupt Request Flags (IWPF7 to IWPF0)

Bit n IWPFn	Description	
0	Clearing conditions: When IWPFn= 1, it is cleared by writing 0	(initial value)
1	Setting conditions: When pin $\overline{WKP}_{\!_{n}}$ is designated for wakeup input and a rising or falling at that pin	edge is input

(n = 7 to 0)



## (7) Wakeup Edge Select Register (WEGR)

Bit	7	6	5	4	3	2	1	0
	WKEGS7	WKEGS6	WKEGS5	WKEGS4	WKEGS3	WKEGS2	WKEGS1	WKEGS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

WEGR is an 8-bit read/write register that specifies rising or falling edge sensing for pins  $\overline{WKPn}$ .

WEGR is initialized to H'00 by a reset.

# Bit n—WKPn Edge Select (WKEGSn)

Bit n selects  $\overline{WKP}$ n pin input sensing.

Bit n WKEGSn	Description	
0	WKPn pin falling edge detected	(initial value)
1	WKPn pin rising edge detected	

(n = 7 to 0)



# 3.3.3 External Interrupts

There are 13 external interrupts: WKP7 to WKP0, IRQ4, IRQ3, IRQ1, IRQ0, and IRQAEC.

# (1) Interrupts WKP<sub>7</sub> to WKP<sub>0</sub>

Interrupts WKP7 to WKP0 are requested by either rising or falling edge input to pins  $\overline{WKP}_7$  to  $\overline{WKP}_0$ . When these pins are designated as pins  $\overline{WKP}_7$  to  $\overline{WKP}_0$  in port mode register 5 and a rising or falling edge is input, the corresponding bit in IWPR is set to 1, requesting an interrupt. Recognition of wakeup interrupt requests can be disabled by clearing the IENWP bit to 0 in IENR1. These interrupts can all be masked by setting the I bit to 1 in CCR.

When WKP7 to WKP0 interrupt exception handling is initiated, the I bit is set to 1 in CCR. Vector number 9 is assigned to interrupts WKP7 to WKP0. All eight interrupt sources have the same vector number, so the interrupt-handling routine must discriminate the interrupt source.

# (2) Interrupts IRQ<sub>4</sub>, IRQ<sub>3</sub>, IRQ<sub>1</sub> and IRQ<sub>0</sub>

Interrupts IRQ4, IRQ3, IRQ1, and IRQ0 are requested by input signals to pins  $\overline{IRQ}_4$ ,  $\overline{IRQ}_3$ ,  $\overline{IRQ}_1$ , and  $\overline{IRQ}_0$ . These interrupts are detected by either rising edge sensing or falling edge sensing, depending on the settings of bits IEG4, IEG3, IEG1, and IEG0 in IEGR.

When these pins are designated as pins  $\overline{IRQ}_4$ ,  $\overline{IRQ}_3$ ,  $\overline{IRQ}_1$ , and  $\overline{IRQ}_0$  in port mode register B, 2, and 1 and the designated edge is input, the corresponding bit in IRR1 is set to 1, requesting an interrupt. Recognition of these interrupt requests can be disabled individually by clearing bits IEN4, IEN3, IEN1, and IEN0 to 0 in IENR1. These interrupts can all be masked by setting the I bit to 1 in CCR.

When IRQ4, IRQ3, IRQ1, and IRQ0 interrupt exception handling is initiated, the I bit is set to 1 in CCR. Vector numbers 8, 7, 5, and 4 are assigned to interrupts IRQ4, IRQ3, IRQ1, and IRQ0. The order of priority is from IRQ0 (high) to IRQ4 (low). Table 3.2 gives details.



# (3) IRQAEC Interrupt

The IRQAEC interrupt is requested by an input signal to pin IRQAEC and IECPWM (output of PWM for AEC). When the IRQAEC input pin is to be used as an external interrupt, set ECPWME in AEGSR to 0. This interrupt is detected by rising edge, falling edge, or both edge sensing, depending on the settings of bits AIEGS1 and AIEGS0 in AEGSR.

When bit IENEC2 in IENR1 is 1 and the designated edge is input, the corresponding bit in IRR1 is set to 1, requesting an interrupt.

When IRQAEC interrupt exception handling is initiated, the I bit is set to 1 in CCR. Vector number 6 is assigned to the IRQAEC interrupt exception handling. Table 3.2 gives details.

# 3.3.4 Internal Interrupts

There are 9 internal interrupts that can be requested by the on-chip peripheral modules. When a peripheral module requests an interrupt, the corresponding bit in IRR1 or IRR2 is set to 1. Recognition of individual interrupt requests can be disabled by clearing the corresponding bit in IENR1 or IENR2. All these interrupts can be masked by setting the I bit to 1 in CCR. When internal interrupt handling is initiated, the I bit is set to 1 in CCR. Vector numbers from 20 to 18 and 16 to 11 are assigned to these interrupts. Table 3.2 shows the order of priority of interrupts from on-chip peripheral modules.



# 3.3.5 Interrupt Operations

Interrupts are controlled by an interrupt controller. Figure 3.2 shows a block diagram of the interrupt controller. Figure 3.3 shows the flow up to interrupt acceptance.



Figure 3.2 Block Diagram of Interrupt Controller

Interrupt operation is described as follows.

- When an interrupt condition is met while the interrupt enable register bit is set to 1, an interrupt request signal is sent to the interrupt controller.
- When the interrupt controller receives an interrupt request, it sets the interrupt request flag.
- From among the interrupts with interrupt request flags set to 1, the interrupt controller selects the interrupt request with the highest priority and holds the others pending. (Refer to table 3.2 for a list of interrupt priorities.)
- The interrupt controller checks the I bit of CCR. If the I bit is 0, the selected interrupt request is accepted; if the I bit is 1, the interrupt request is held pending.

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- If the interrupt request is accepted, after processing of the current instruction is completed, both PC and CCR are pushed onto the stack. The state of the stack at this time is shown in figure 3.4. The PC value pushed onto the stack is the address of the first instruction to be executed upon return from interrupt handling.
- The I bit of CCR is set to 1, masking further interrupts.
- The vector address corresponding to the accepted interrupt is generated, and the interrupt handling routine located at the address indicated by the contents of the vector address is executed.
- Notes: 1. When disabling interrupts by clearing bits in an interrupt enable register, or when clearing bits in an interrupt request register, always do so while interrupts are masked (I = 1).
  - 2. If the above clear operations are performed while I = 0, and as a result a conflict arises between the clear instruction and an interrupt request, exception processing for the interrupt will be executed after the clear instruction has been executed.





Figure 3.3 Flow up to Interrupt Acceptance

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Figure 3.4 Stack State after Completion of Interrupt Exception Handling

Figure 3.5 shows a typical interrupt sequence.





Figure 3.5 Interrupt Sequence



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# 3.3.6 Interrupt Response Time

Table 3.4 shows the number of wait states after an interrupt request flag is set until the first instruction of the interrupt handler is executed.

# Table 3.4Interrupt Wait States

Item	States	Total	
Waiting time for completion of executing instruction*	1 to 13	15 to 27	
Saving of PC and CCR to stack	4		
Vector fetch	2		
Instruction fetch	4		
Internal processing	4		

Note: \* Not including EEPMOV instruction.



# 3.4 Application Notes

# 3.4.1 Notes on Stack Area Use

When word data is accessed in the LSI, the least significant bit of the address is regarded as 0. Access to the stack always takes place in word size, so the stack pointer (SP: R7) should never indicate an odd address. Use PUSH Rn (MOV.W Rn, @–SP) or POP Rn (MOV.W @SP+, Rn) to save or restore register values.

Setting an odd address in SP may cause a program to crash. An example is shown in figure 3.6.



Figure 3.6 Operation when Odd Address is Set in SP

When CCR contents are saved to the stack during interrupt exception handling or restored when RTE is executed, this also takes place in word size. Both the upper and lower bytes of word data are saved to the stack; on return, the even address contents are restored to CCR while the odd address contents are ignored.



# 3.4.2 Notes on Rewriting Port Mode Registers

When a port mode register is rewritten to switch the functions of external interrupt pins and when the value of ECPWME in AEGSR is rewritten to switch between selection/non-selection of IRQAEC, the following points should be observed.

When an external interrupt pin function is switched by rewriting the port mode register that controls pins  $\overline{IRQ}_4$ ,  $\overline{IRQ}_3$ ,  $\overline{IRQ}_1$ ,  $\overline{IRQ}_0$ ,  $\overline{WKP}_7$  to  $\overline{WKP}_0$ , the interrupt request flag may be set to 1 at the time the pin function is switched, even if no valid interrupt is input at the pin. Be sure to clear the interrupt request flag to 0 after switching pin functions. When the value of ECPWME in AEGSR that sets selection/non-selection of IRQAEC is rewritten, the interrupt request flag may be set to 1, even if a valid edge has not arrived on the selected IRQAEC or IECPWM (PWM output for AEC). Therefore, be sure to clear the interrupt request flag to 0 after switching the pin function. Table 3.5 shows the conditions under which interrupt request flags are set to 1 in this way.

Interrupt Request Flags Set to 1		Conditions					
IRR1	IRRI4	When PMR1 bit IRQ4 is changed from 0 to 1 while pin $\overline{IRQ}_4$ is low and IEGR bit IEG4 = 0.					
		When PMR1 bit IRQ4 is changed from 1 to 0 while pin $\overline{IRQ}_4$ is low and IEGR bit IEG4 = 1.					
	IRRI3	When PMR1 bit IRQ3 is changed from 0 to 1 while pin $\overline{IRQ}_{_3}$ is low and IEGR bit IEG3 = 0.					
		When PMR1 bit IRQ3 is changed from 1 to 0 while pin $\overline{IRQ}_{_3}$ is low and IEGR bit IEG3 = 1.					
	IRREC2	When an edge as designated by AIEGS1 and AIEGS0 in AEGSR is detected because the values on the IRQAEC pin and of IECPWM at switching are different (e.g., when the rising edge has been selected and ECPWME in AEGSR is changed from 1 to 0 while pin IRQAEC is low and IECPWM = 1).					
	IRRI1	When PMRB bit IRQ1 is changed from 0 to 1 while pin $\overline{IRQ}_1$ is low and IEGR bit IEG1 = 0.					
		When PMRB bit IRQ1 is changed from 1 to 0 while pin $\overline{IRQ}_1$ is low and IEGR bit IEG1 = 1.					
	IRRI0	When PMR2 bit IRQ0 is changed from 0 to 1 while pin $\overline{IRQ}_{_0}$ is low and IEGR bit IEG0 = 0.					
		When PMR2 bit IRQ0 is changed from 1 to 0 while pin $\overline{IRQ}_{_0}$ is low and IEGR bit IEG0 = 1.					

#### Table 3.5 Conditions under which Interrupt Request Flag is Set to 1

Interrupt Request Flags Set to 1		Conditions
IWPR	IWPF7	When PMR5 bit WKP7 is changed from 0 to 1 while pin $\overline{\text{WKP}}_7$ is low and WEGR bit WKEGS7 = 0.
		When PMR5 bit WKP7 is changed from 1 to 0 while pin $\overline{\text{WKP}}_7$ is low and WEGR bit WKEGS7 = 1.
	IWPF6	When PMR5 bit WKP6 is changed from 0 to 1 while pin $\overline{\text{WKP}}_{6}$ is low and WEGR bit WKEGS6 = 0.
		When PMR5 bit WKP6 is changed from 1 to 0 while pin $\overline{\text{WKP}}_6$ is low and WEGR bit WKEGS6 = 1.
	IWPF5	When PMR5 bit WKP5 is changed from 0 to 1 while pin $\overline{\text{WKP}}_{s}$ is low and WEGR bit WKEGS5 = 0.
		When PMR5 bit WKP5 is changed from 1 to 0 while pin $\overline{\text{WKP}}_{_5}$ is low and WEGR bit WKEGS5 = 1.
	IWPF4	When PMR5 bit WKP4 is changed from 0 to 1 while pin $\overline{\text{WKP}}_4$ is low and WEGR bit WKEGS4 = 0.
		When PMR5 bit WKP4 is changed from 1 to 0 while pin $\overline{\text{WKP}}_4$ is low and WEGR bit WKEGS4 = 1.
	IWPF3	When PMR5 bit WKP3 is changed from 0 to 1 while pin $\overline{\text{WKP}}_{3}$ is low and WEGR bit WKEGS3 = 0.
		When PMR5 bit WKP3 is changed from 1 to 0 while pin $\overline{WKP}_{3}$ is low and WEGR bit WKEGS3 = 1.
	IWPF2	When PMR5 bit WKP2 is changed from 0 to 1 while pin $\overline{\text{WKP}}_2$ is low and WEGR bit WKEGS2 = 0.
		When PMR5 bit WKP2 is changed from 1 to 0 while pin $\overline{\text{WKP}}_2$ is low and WEGR bit WKEGS2 = 1.
	IWPF1	When PMR5 bit WKP1 is changed from 0 to 1 while pin $\overline{WKP}_1$ is low and WEGR bit WKEGS1 = 0.
		When PMR5 bit WKP1 is changed from 1 to 0 while pin $\overline{WKP}_1$ is low and WEGR bit WKEGS1 = 1.
	IWPF0	When PMR5 bit WKP0 is changed from 0 to 1 while pin $\overline{\text{WKP}}_0$ is low and WEGR bit WKEGS0 = 0.
		When PMR5 bit WKP0 is changed from 1 to 0 while pin $\overline{WKP}_0$ is low and WEGR bit WKEGS0 = 1.

Figure 3.7 shows the procedure for setting a bit in a port mode register and clearing the interrupt request flag.



When switching a pin function, mask the interrupt before setting the bit in the port mode register (or AEGSR). After accessing the port mode register (or AEGSR), execute at least one instruction (e.g., NOP), then clear the interrupt request flag from 1 to 0. If the instruction to clear the flag is executed immediately after the port mode register (or AEGSR) access without executing an intervening instruction, the flag will not be cleared.

An alternative method is to avoid the setting of interrupt request flags when pin functions are switched by keeping the pins at the high level so that the conditions in table 3.5 do not occur.

However, the procedure in Figure 3.7 is recommended because IECPWM is an internal signal and determining its value is complicated.



Figure 3.7 Port Mode Register (or AEGSR) Setting and Interrupt Request Flag Clearing Procedure



# 3.4.3 Method for Clearing Interrupt Request Flags

Use the recommended method, given below when clearing the flags of interrupt request registers (IRR1, IRR2, IWPR).

• Recommended method

Use a single instruction to clear flags. The bit control instruction and byte-size data transfer instruction can be used. Two examples of program code for clearing IRRI1 (bit 1 of IRR1) are given below.

BCLR #1, @IRR1:8 MOV.B R1L, @IRR1:8 (set the value of R1L to B'11111101)

• Example of a malfunction

When flags are cleared with multiple instructions, other flags might be cleared during execution of the instructions, even though they are currently set, and this will cause a malfunction.

Here is an example in which IRRI0 is cleared and disabled in the process of clearing IRRI1 (bit 1 of IRR1).

MOV.B @IRR1:8,R1L ..... IRRI0 = 0 at this time
AND.B #B'1111101,R1L .... Here, IRRI0 = 1
MOV.B R1L,@IRR1:8 ..... IRRI0 is cleared to 0

In the above example, it is assumed that an IRQ0 interrupt is generated while the AND.B instruction is executing.

The IRQ0 interrupt is disabled because, although the original objective is clearing IRRI1, IRRI0 is also cleared.





# Section 4 Clock Pulse Generators

# 4.1 Overview

Clock oscillator circuitry (CPG: clock pulse generator) is provided on-chip, including both a system clock pulse generator and a subclock pulse generator. The system clock pulse generator consists of a system clock oscillator and system clock dividers. The subclock pulse generator consists of a subclock oscillator circuit and a subclock divider.

In the H8/38524 Group, the system clock pulse generator includes an on-chip oscillator.

#### 4.1.1 Block Diagram

Figure 4.1 shows a block diagram of the clock pulse generators.



Figure 4.1 Block Diagram of Clock Pulse Generators

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# 4.1.2 System Clock and Subclock

The basic clock signals that drive the CPU and on-chip peripheral modules are  $\phi$  and  $\phi_{sub}$ . Four of the clock signals have names:  $\phi$  is the system clock,  $\phi_{sub}$  is the subclock,  $\phi_{osc}$  is the oscillator clock, and  $\phi_w$  is the watch clock.

The clock signals available for use by peripheral modules are  $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ ,  $\phi/16$ ,  $\phi/32$ ,  $\phi/64$ ,  $\phi/128$ ,  $\phi/256$ ,  $\phi/512$ ,  $\phi/1024$ ,  $\phi/2048$ ,  $\phi/4096$ ,  $\phi/8192$ ,  $\phi_w$ ,  $\phi_w/2$ ,  $\phi_w/4$ ,  $\phi_w/8$ ,  $\phi_w/16$ ,  $\phi_w/32$ ,  $\phi_w/64$ , and  $\phi_w/128$ . The clock requirements differ from one module to another.

# 4.1.3 Register Descriptions

Table 4.1 lists the registers that control the clock pulse generators.

### Table 4.1 Clock Pulse Generator Control Registers

Name	Abbreviation	R/W	Initial Value	Address
Clock pulse generator control register	OSCCR	R/W	—	H'FFF5

## (1) Clock Pulse Generator Control Register (OSCCR)

Bit	7	6	5	4	3	2	1	0
	SUBSTP		—	_		IRQAECF	OSCF	_
Initial value	0	0	0	0	0	—	—	0
Read/Write	R/W	R	R/W	R/W	R/W	R	R	R/W

OSCCR is an 8-bit read/write register that contains the flag indicating the selection of system clock oscillator or on-chip oscillator, indicates the input level of the IRQAEC pin during resets, and controls whether the subclock oscillator operates or not.

# Bit 7—Subclock Oscillator Stop Control (SUBSTP)

Bit 7 controls whether the subclock oscillator operates or not. It can be set to 1 only in the active mode (high-speed/medium-speed). Setting bit 7 to 1 in the subactive mode will cause the LSI to stop operating.


Bit 7 SUBSTP	Description	
0	Subclock oscillator operates	(initial value)
1	Subclock oscillator stopped	

#### Bit 6—Reserved

This bit is reserved. It is always read as 0 and cannot be written to.

#### Bits 5 to 3—Reserved

These bits are read/write enabled reserved bits.

#### Bit 2—IRQAEC Flag (IRQAECF)

This bit indicates the IRQAEC pin input level set during resets.

Bit 2 IRQAECF	Description
0	IRQAEC pin set to GND during resets
1	IRQAEC pin set to $V_{cc}$ during resets

#### Bit 1—OSC Flag (OSCF)

This bit indicates the oscillator operating with the system clock pulse generator.

Bit 1 OSCF	Description
0	System clock oscillator operating (on-chip oscillator stopped)
1	On-chip oscillator operating (system clock oscillator stopped)

#### Bit 0—Reserved

This bit is reserved. Never write 1 to this bit, as it can cause the LSI to malfunction.



# 4.2 System Clock Generator

Clock pulses can be supplied to the system clock divider either by connecting a crystal or ceramic oscillator, or by providing external clock input. As shown in figure 4.1, the selection between a system clock oscillator and an on-chip oscillator is supported. See section 4.2 (5), On-Chip Oscillator Selection Method, for information on selecting the on-chip oscillator.

#### (1) Connecting a Crystal Oscillator

Figure 4.2 shows a typical method of connecting a crystal oscillator.





Figure 4.3 shows the equivalent circuit of a crystal oscillator. An oscillator having the characteristics given in table 4.2 should be used.



Figure 4.3 Equivalent Circuit of Crystal Oscillator

Frequency (MHz)	4	4.193	
RS max (Ω)	100	100	
C <sub>o</sub> max (pF)	16	16	

#### Table 4.2 Crystal Oscillator Parameters

#### (2) Connecting a Ceramic Oscillator

Figure 4.4 shows a typical method of connecting a ceramic oscillator.



Figure 4.4 Typical Connection to Ceramic Oscillator



# (3) Notes on Board Design

When generating clock pulses by connecting a crystal or ceramic oscillator, pay careful attention to the following points.

Avoid running signal lines close to the oscillator circuit, since the oscillator may be adversely affected by induction currents. (See figure 4.5.)

The board should be designed so that the oscillator and load capacitors are located as close as possible to pins  $OSC_1$  and  $OSC_2$ .



Figure 4.5 Board Design of Oscillator Circuit

Note: The circuit parameters above are recommended by the crystal or ceramic oscillator manufacturer.

The circuit parameters are affected by the crystal or ceramic oscillator and floating capacitance when designing the board. When using the oscillator, consult with the crystal or ceramic oscillator manufacturer to determine the circuit parameters.



#### (4) External Clock Input Method

Connect an external clock signal to pin  $OSC_1$ , and leave pin  $OSC_2$  open. Figure 4.6 shows a typical connection.



Figure 4.6 External Clock Input (Example)

Frequency	Oscillator Clock ( $\phi_{osc}$ )
Duty cycle	45% to 55%

#### (5) On-Chip Oscillator Selection Method

The on-chip oscillator is selected by setting the IRQAEC pin input level during resets.\* Table 4.3 lists the methods for selecting the system clock oscillator and the on-chip oscillator. The IRQAEC pin input level set during resets must be fixed at  $V_{cc}$  or GND, based on the oscillator to be selected. It is not necessary to connect an oscillator to pins OSC1 and OSC2 if the on-chip oscillator is selected. In this case, pin OSC1 should be fixed at  $V_{cc}$  or GND.

- Note: The system clock oscillator must be selected in order to program or erase flash memory as part of operations such as on-board programming. Also, when using the on-chip emulator, an oscillator should be connected, or an external clock input, even if the on-chip oscillator is selected.
  - \* Other than watchdog timer or low-voltage detect circuit reset.

#### Table 4.3 System Clock Oscillator and On-Chip Oscillator Selection Methods

IRQAEC pin input level (during resets)	0	1
System clock oscillator	Enabled	Disabled
On-chip oscillator	Disabled	Enabled



# 4.3 Subclock Generator

# (1) Connecting a 32.768 kHz Crystal Oscillator

Clock pulses can be supplied to the subclock divider by connecting a 32.768 kHz crystal oscillator, as shown in figure 4.7. Follow the same precautions as noted under 3. notes on board design for the system clock in section 4.2.



Figure 4.7 Typical Connection to 32.768 kHz Crystal Oscillator (Subclock)

Figure 4.8 shows the equivalent circuit of the 32.768 kHz crystal oscillator.



Figure 4.8 Equivalent Circuit of 32.768 kHz Crystal Oscillator

When using a resonator other than the above, ensure optimal conditions by conducting sufficient evaluation of consistency in cooperation with the manufacturer of the resonator. Even if the above resonators or products equivalent to them are implemented, their oscillation characteristics are affected by the board design. Be sure to use the actual board to evaluate consistency as a system.

The consistency as a system has to be verified not only in a reset state (i.e., the  $\overline{\text{RES}}$  pin is driven low) but also in a state where a reset state has been exited (i.e., the low-level  $\overline{\text{RES}}$  signal has been driven high).

### (2) Pin Connection when Not Using Subclock

When the subclock is not used, connect pin  $X_1$  to GND and leave pin  $X_2$  open, as shown in figure 4.9.



Figure 4.9 Pin Connection when not Using Subclock

### (3) Method for Disabling Subclock Oscillator

The subclock oscillator can be disabled by programs by setting the SUBSTP bit in the OSCCR register to 1. The register setting to disable the subclock oscillator should be made in the active mode. When restoring operation of the subclock oscillator after it has been disabled using the OSCCR register, it is necessary to wait for the oscillation stabilization time (typ: 8s) to elapse before using the subclock.



# 4.4 Prescalers

This LSI is equipped with two on-chip prescalers having different input clocks (prescaler S and prescaler W). Prescaler S is a 13-bit counter using the system clock ( $\phi$ ) as its input clock. Its prescaled outputs provide internal clock signals for on-chip peripheral modules. Prescaler W is a 5-bit counter using a 32.768 kHz signal divided by 4 ( $\phi_w$ /4) as its input clock. Its prescaled outputs are used by timer A as a time base for timekeeping.

#### (1) Prescaler S (PSS)

Prescaler S is a 13-bit counter using the system clock ( $\phi$ ) as its input clock. It is incremented once per clock period.

Prescaler S is initialized to H'0000 by a reset, and starts counting on exit from the reset state.

In standby mode, watch mode, subactive mode, and subsleep mode, the system clock pulse generator stops. Prescaler S also stops and is initialized to H'0000.

The CPU cannot read or write prescaler S.

The output from prescaler S is shared by timer A, timer C, timer F, timer G, SCI3, the A/D converter, the LCD controller, watchdog timer, and the 10-bit PWM. The divider ratio can be set separately for each on-chip peripheral function.

In active (medium-speed) mode the clock input to prescaler S is  $\phi$ osc/16,  $\phi$ osc/32,  $\phi$ osc/64, or  $\phi$ osc/128.

#### (2) Prescaler W (PSW)

Prescaler W is a 5-bit counter using a 32.768 kHz signal divided by 4 ( $\phi_w/4$ ) as its input clock.

Prescaler W is initialized to H'00 by a reset, and starts counting on exit from the reset state.

Even in standby mode, watch mode, subactive mode, or subsleep mode, prescaler W continues functioning so long as clock signals are supplied to pins X1 and X2.

Prescaler W can be reset by setting 1s in bits TMA3 and TMA2 of timer mode register A (TMA).

Output from prescaler W can be used to drive timer A, in which case timer A functions as a time base for timekeeping.

# 4.5 Note on Oscillators

Oscillator characteristics are closely related to board design and should be carefully evaluated by the user, referring to the examples shown in this section. Oscillator circuit constants will differ depending on the oscillator element, stray capacitance in its interconnecting circuit, and other factors. Suitable constants should be determined in consultation with the oscillator element manufacturer. Design the circuit so that the oscillator element never receives voltages exceeding its maximum rating.



Figure 4.10 Example of Crystal and Ceramic Oscillator Element Arrangement

Figure 4.11 (1) shows an example measuring circuit with the negative resistance suggested by the resonator manufacturer. Note that if the negative resistance of the circuit is less than that suggested by the resonator manufacturer, it may be difficult to start the main oscillator.

If it is determined that oscillation is not occurring because the negative resistance is lower than the level suggested by the resonator manufacturer, the circuit may be modified as shown in figure 4.11 (2) through (4). Which of the modification suggestions to use and the capacitor capacitance should be decided based upon an evaluation of factors such as the negative resistance and the frequency deviation.







# 4.5.1 Definition of Oscillation Stabilization Wait Time

Figure 4.12 shows the oscillation waveform (OSC2), system clock ( $\phi$ ), and microcomputer operating mode when a transition is made from standby mode, watch mode, or subactive mode, to active (high-speed/medium-speed) mode, with an oscillator element connected to the system clock oscillator.

As shown in figure 4.12, as the system clock oscillator is halted in standby mode, watch mode, and subactive mode, when a transition is made to active (high-speed/medium-speed) mode, the sum of the following two times (oscillation stabilization time and wait time) is required.

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#### 1. Oscillation stabilization time $(t_{rc})$

The time from the point at which the system clock oscillator oscillation waveform starts to change when an interrupt is generated, until the amplitude of the oscillation waveform increases and the oscillation frequency stabilizes.

#### 2. Wait time

The time required for the CPU and peripheral functions to begin operating after the oscillation waveform frequency and system clock have stabilized.

The wait time setting is selected with standby timer select bits 2 to 0 (STS2 to STS0) (bits 6 to 4 in system control register 1 (SYSCR1)).



Figure 4.12 Oscillation Stabilization Wait Time



When standby mode, watch mode, or subactive mode is cleared by an interrupt or reset, and a transition is made to active (high-speed/medium-speed) mode, the oscillation waveform begins to change at the point at which the interrupt is accepted. Therefore, when an oscillator element is connected in standby mode, watch mode, or subactive mode, since the system clock oscillator is halted, the time from the point at which this oscillation waveform starts to change until the amplitude of the oscillation waveform increases and the oscillation frequency stabilizes—that is, the oscillation stabilization time—is required.

The oscillation stabilization time in the case of these state transitions is the same as the oscillation stabilization time at power-on (the time from the point at which the power supply voltage reaches the prescribed level until the oscillation stabilizes), specified by "oscillation stabilization time  $t_{rc}$ " in the AC characteristics.

Meanwhile, once the system clock has halted, a wait time of at least 8 states is necessary in order for the CPU and peripheral functions to operate normally.

Thus, the time required from interrupt generation until operation of the CPU and peripheral functions is the sum of the above described oscillation stabilization time and wait time. This total time is called the oscillation stabilization wait time, and is expressed by equation (1) below.

Oscillation stabilization wait time = oscillation stabilization time + wait time =  $t_{rc}$  + (8 to 131,072 states) .....(1)

Therefore, when a transition is made from standby mode, watch mode, or subactive mode, to active (high-speed/medium-speed) mode, with an oscillator element connected to the system clock oscillator, careful evaluation must be carried out on the installation circuit before deciding on the oscillation stabilization wait time. In particular, since the oscillation stabilization time is affected by installation circuit constants, stray capacitance, and so forth, suitable constants should be determined in consultation with the oscillator element manufacturer.



#### 4.5.2 Notes on Use of Crystal Oscillator Element (Excluding Ceramic Oscillator Element)

When a microcomputer operates, the internal power supply potential fluctuates slightly in synchronization with the system clock. Depending on the individual crystal oscillator element characteristics, the oscillation waveform amplitude may not be sufficiently large immediately after the oscillation stabilization wait time, making the oscillation waveform susceptible to influence by fluctuations in the power supply potential. In this state, the oscillation waveform may be disrupted, leading to an unstable system clock and erroneous operation of the microcomputer.

If erroneous operation occurs, change the setting of standby timer select bits 2 to 0 (STS2 to STS0) (bits 6 to 4 in system control register 1 (SYSCR1)) to give a longer wait time.

For example, if erroneous operation occurs with a wait time setting of 16 states, check the operation with a wait time setting of 8,192 states or more.

If the same kind of erroneous operation occurs after a reset as after a state transition, hold the  $\overline{\text{RES}}$  pin low for a longer period.

# 4.6 Usage Note

When using the on-chip emulator, system clock precision is necessary for programming or erasing the flash memory. However, the on-chip oscillator frequency can vary due to changes in conditions such as voltage or temperature. Consequently, even if the on-chip oscillator is selected when using the on-chip emulator, pins OSC1 and OSC2 should be connected to an oscillator, or an external clock should be supplied. In this case, the LSI uses the on-chip oscillator when user programs are being executed and the system clock oscillator when programming or erasing flash memory. The process is controlled by the on-chip emulator.





# Section 5 Power-Down Modes

# 5.1 Overview

The LSI has nine modes of operation after a reset. These include eight power-down modes, in which power dissipation is significantly reduced. Table 5.1 gives a summary of the nine operating modes.

Table 5.1	<b>Operating Modes</b>
-----------	------------------------

Operating Mode	Description
Active (high-speed) mode	The CPU and all on-chip peripheral functions are operable on the system clock in high-speed operation
Active (medium-speed) mode	The CPU and all on-chip peripheral functions are operable on the system clock in low-speed operation
Subactive mode	The CPU and all on-chip peripheral functions are operable on the subclock in low-speed operation
Sleep (high-speed) mode	The CPU halts. On-chip peripheral functions are operable on the system clock
Sleep (medium-speed) mode	The CPU halts. On-chip peripheral functions operate at a frequency of 1/128, 1/64, 1/32, or 1/16 of the system clock frequency
Subsleep mode	The CPU halts. The time-base function of timer A, timer C, timer F, timer G, SCI3, AEC, and LCD controller/driver are operable on the subclock
Watch mode	The CPU halts. The time-base function of timer A, timer F, timer G, AEC and LCD controller/driver are operable on the subclock
Standby mode	The CPU and all on-chip peripheral functions halt
Module standby mode	Individual on-chip peripheral functions specified by software enter standby mode and halt

Of these nine operating modes, all but the active (high-speed) mode are power-down modes. In this section the two active modes (high-speed and medium speed) will be referred to collectively as active mode.

Figure 5.1 shows the transitions among these operation modes. Table 5.2 indicates the internal states in each mode.

RENESAS





#### Mode Transition Conditions (1)

	LSON	MSON	SSBY	TMA3	DTON
*а	0	0	0	*	0
*b	0	1	0	*	0
*c	1	*	0	1	0
*d	0	*	1	0	0
*e	*	*	1	1	0
*f	0	0	0	*	1
*g	0	1	0	*	1
*h	0	1	1	1	1
*i	1	*	1	1	1
*j	0	0	1	1	1
				*: Do	on't care

#### Mode Transition Conditions (2)

	Interrupt Sources
*1	Timer A, Timer F, Timer G interrupt, IRQ0 interrupt,
	WKP7 to WKP0 interrupts
*2	Timer A, Timer C, Timer F, Timer G, SCl3 interrupt, IRQ4, IRQ3, IRQ1 and IRQ0 interrupts, IRQAEC, WKP7 to WKP0 interrupts, AEC
*3	All interrupts
*4	IRQ1 or IRQ0 interrupt, WKP7 to WKP0 interrupts

- Notes: 1. A transition between different modes cannot be made to occur simply because an interrupt request is generated. Make sure that interrupts are enabled.
  - 2. Details on the mode transition conditions are given in the explanations of each mode, in sections 5.2 to 5.8.

#### Figure 5.1 Mode Transition Diagram

RENESAS

		Active	Mode	Sleep	Mode				
Function		High- Speed	Medium- Speed	High- Speed	Medium- Speed	Watch Mode	Subactive Mode	Subsleep Mode	Standby Mode
System clo	ock oscillator	Functions	Functions	Functions	Functions	Halted	Halted	Halted	Halted
Subclock c	scillator	Functions	Functions	Functions	Functions	Functions	Functions	Functions	Functions
CPU	Instructions	Functions	Functions	Halted	Halted	Halted	Functions	Halted	Halted
operations	RAM	-		Retained	Retained	Retained	-	Retained	Retained
	Registers	-							
	I/O ports	-							Retained <sup>*1</sup>
External	IRQ <sub>0</sub>	Functions	Functions	Functions	Functions	Functions	Functions	Functions	Functions
interrupts	IRQ,	-				Retained <sup>*6</sup>	-		
	IRQAEC	-							Retained*6
	IRQ <sub>3</sub>	-							
	IRQ₄	-							
	WKP <sub>0</sub>	Functions	Functions	Functions	Functions	Functions	Functions	Functions	Functions
	WKP <sub>1</sub>	-							
	WKP <sub>2</sub>	-							
	WKP <sub>3</sub>	_							
	WKP <sub>4</sub>	-							
	WKP₅	-							
	WKP <sub>6</sub>	-							
	WKP <sub>6</sub> WKP <sub>7</sub>	-							
Peripheral	0	Functions	Functions	Functions	Functions	Functions*5	Functions*5	Functions <sup>*5</sup>	Retained
Peripheral functions	WKP <sub>7</sub>	Functions	Functions	Functions	Functions	Functions <sup>*5</sup> Functions <sup>*8</sup>	Functions <sup>*5</sup> Functions	Functions <sup>*5</sup> Functions	Retained Functions*
	WKP, Timer A Asynchronous event counter	Functions	Functions	Functions	Functions	Functions*8	Functions	Functions	Functions*
	WKP <sub>7</sub> Timer A Asynchronous	Functions	Functions	Functions	Functions				
	WKP, Timer A Asynchronous event counter	Functions	Functions	Functions	Functions	Functions <sup>*®</sup> Retained Functions/	Functions/ Retained <sup>*2</sup> Functions/	Functions/ Retained <sup>*2</sup> Functions/	Functions*
	WKP <sub>7</sub> Timer A Asynchronous event counter Timer C	Functions	Functions	Functions	Functions	Functions*® Retained	Functions/ Functions/ Retained <sup>*2</sup>	Functions Functions/ Retained <sup>*2</sup>	Functions*
	WKP, Timer A Asynchronous event counter Timer C WDT	Functions	Functions	Functions	Functions	Functions <sup>*8</sup> Retained Functions/ Retained <sup>*10</sup>	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*7</sup>	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*10</sup>	Functions* Retained Functions/ Retained*11
	WKP, Timer A Asynchronous event counter Timer C WDT Timer F	Functions	Functions	Functions	Functions	Functions <sup>*8</sup> Retained Functions/ Retained <sup>*10</sup> Functions/	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*7</sup> Functions/	Functions Functions/ Retained <sup>*2</sup> Functions/ Functions/	Functions* Retained Functions/ Retained*11
	WKP, Timer A Asynchronous event counter Timer C WDT Timer F Timer G	Functions 	Functions	Functions	Functions	Functions <sup>*</sup> 8 Retained Functions/ Retained <sup>*10</sup> Functions/ Retained <sup>*9</sup>	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*7</sup> Functions/ Retained <sup>*9</sup> Functions/	Functions/ Retained*2 Functions/ Retained*10 Functions/ Retained*9 Functions/	Functions <sup>*</sup> Retained Functions/ Retained <sup>*11</sup> Retained
	WKP, Timer A Asynchronous event counter Timer C WDT Timer F Timer G SCI3	Functions	Functions	Functions	Functions	Functions <sup>*</sup> <sup>6</sup> Retained Functions/ Retained <sup>*10</sup> Functions/ Retained <sup>*9</sup> Reset	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*7</sup> Functions/ Retained <sup>*9</sup> Functions/ Retained <sup>*3</sup>	Functions/ Retained*2 Functions/ Retained*10 Functions/ Retained*9 Functions/ Retained*3	Functions* Retained Functions/ Retained* Retained Reset
	WKP, Timer A Asynchronous event counter Timer C WDT Timer F Timer G SCI3 PWM	Functions 	Functions	Functions	Functions	Functions <sup>*</sup> <sup>6</sup> Retained Functions/ Retained <sup>*10</sup> Functions/ Retained Reset Retained	Functions/ Retained <sup>*2</sup> Functions/ Retained <sup>*7</sup> Functions/ Retained <sup>*9</sup> Functions/ Retained <sup>*3</sup> Retained	Functions/ Retained*2 Functions/ Retained*10 Functions/ Retained*9 Functions/ Retained*3 Retained	Functions <sup>*</sup> Retained Functions/ Retained <sup>*</sup> Retained Reset Retained

#### **Internal State in Each Operating Mode** Table 5.2

- - 3. Functions if  $\phi_w/2$  is selected as the internal clock; otherwise halted and retained.
  - 4. Functions if  $\phi_w$ ,  $\phi_w/2$  or  $\phi_w/4$  is selected as the operating clock; otherwise halted and retained.
  - 5. Functions if the timekeeping time-base function is selected.
  - 6. External interrupt requests are ignored. Interrupt request register contents are not altered.
  - 7. Operates when  $\phi_u/32$  is selected as the internal clock or the on-chip oscillator is selected; otherwise stops and stands by.
  - 8. Incrementing is possible, but interrupt generation is not.
  - 9. Functions if  $\phi_w/4$  is selected as the internal clock; otherwise halted and retained.
  - 10. Operates when  $\phi_u/32$  is selected as the internal clock or the on-chip oscillator is selected; otherwise stops and stands by.
  - 11. Operates only when the on-chip oscillator is selected; otherwise stops and stands by.



#### 5.1.1 System Control Registers

The operation mode is selected using the system control registers described in table 5.3.

### Table 5.3 System Control Registers

Name	Abbreviation	R/W	Initial Value	Address
System control register 1	SYSCR1	R/W	H'07	H'FFF0
System control register 2	SYSCR2	R/W	H'F0	H'FFF1

#### (1) System Control Register 1 (SYSCR1)

Bit	7	6	5	4	3	2	1	0
	SSBY	STS2	STS1	STS0	LSON		MA1	MA0
Initial value	0	0	0	0	0	1	1	1
Read/Write	R/W	R/W	R/W	R/W	R/W	—	R/W	R/W

SYSCR1 is an 8-bit read/write register for control of the power-down modes.

Upon reset, SYSCR1 is initialized to H'07.

#### Bit 7—Software Standby (SSBY)

This bit designates transition to standby mode or watch mode.

Bit 7	
SSBY	Description
0	• When a SLEEP instruction is executed in active mode, (initial value) a transition is made to sleep mode
	• When a SLEEP instruction is executed in subactive mode, a transition is made to subsleep mode
1	When a SLEEP instruction is executed in active mode, a transition is made to standby mode or watch mode
	When a SLEEP instruction is executed in subactive mode, a transition is made to watch mode

### Bits 6 to 4—Standby Timer Select 2 to 0 (STS2 to STS0)

These bits designate the time the CPU and peripheral modules wait for stable clock operation after exiting from standby mode or watch mode to active mode due to an interrupt. The designation should be made according to the operating frequency so that the waiting time is at least equal to the oscillation stabilization time.

Bit 6 STS2	Bit 5 STS1	Bit 4 STS0	Description			
0	0	0	Wait time = 8,192 states	s (initial value)		
0	0	1	Wait time = 16,384 state	es		
0	1	0	Wait time = 32,768 states			
0	1	1	Wait time = 65,536 states			
1	0	0	Wait time = 131,072 states			
1	0	1	Wait time = 2 states	(External clock input mode)		
1	1	0	Wait time = 8 states			
1	1	1	Wait time = 16 states			

Note: If an external clock is being input, set standby timer select to external clock mode before mode transition. Also, do not set standby timer select to external clock mode if no external clock is used. 8,192 states (STS2 = STS1 = STS0 = 0) is recommended if the on-chip oscillator is used.

# Bit 3—Low Speed on Flag (LSON)

This bit chooses the system clock ( $\phi$ ) or subclock ( $\phi_{sub}$ ) as the CPU operating clock when watch mode is cleared. The resulting operation mode depends on the combination of other control bits and interrupt input.

Bit 3 LSON	Description	
0	The CPU operates on the system clock ( $\phi$ )	(initial value)
1	The CPU operates on the subclock ( $\phi_{\mbox{\tiny SUB}}$	

### Bit 2—Reserved

Bit 2 is reserved: it is always read as 1 and cannot be modified.



#### Bits 1 and 0—Active (Medium-Speed) Mode Clock Select (MA1, MA0)

Bits 1 and 0 choose  $\phi_{osc}/128$ ,  $\phi_{osc}/64$ ,  $\phi_{osc}/32$ , or  $\phi_{osc}/16$  as the operating clock in active (medium-speed) mode and sleep (medium-speed) mode. MA1 and MA0 should be written in active (high-speed) mode or subactive mode.

Bit 1 MA1	Bit 0 MA0	Description	
0	0	$\phi_{osc}/16$	
0	1	φ <sub>osc</sub> /32	
1	0	φ <sub>osc</sub> /64	
1	1	φ <sub>osc</sub> /128	(initial value)

#### (2) System Control Register 2 (SYSCR2)

Bit	7	6	5	4	3	2	1	0
				NESEL	DTON	MSON	SA1	SA0
Initial value	1	1	1	1	0	0	0	0
Read/Write	_		_	R/W	R/W	R/W	R/W	R/W

SYSCR2 is an 8-bit read/write register for power-down mode control.

#### Bits 7 to 5—Reserved

These bits are reserved; they are always read as 1, and cannot be modified.

#### Bit 4—Noise Elimination Sampling Frequency Select (NESEL)

This bit selects the frequency at which the watch clock signal  $(\phi_w)$  generated by the subclock pulse generator is sampled, in relation to the oscillator clock  $(\phi_{osc})$  generated by the system clock pulse generator. When  $\phi_{osc} = 2$  to 20 MHz, clear NESEL to 0.

Bit 4 NESEL	Description	
0	Sampling rate is $\phi_{osc}/16$	
1	Sampling rate is $\phi_{osc}/4$	(initial value)

### Bit 3—Direct Transfer on Flag (DTON)

This bit designates whether or not to make direct transitions among active (high-speed), active (medium-speed) and subactive mode when a SLEEP instruction is executed. The mode to which the transition is made after the SLEEP instruction is executed depends on a combination of other control bits.

Bit 3 DTON	Description
0	• When a SLEEP instruction is executed in active mode, (initial value) a transition is made to standby mode, watch mode, or sleep mode
	• When a SLEEP instruction is executed in subactive mode, a transition is made to watch mode or subsleep mode
1	<ul> <li>When a SLEEP instruction is executed in active (high-speed) mode, a direct transition is made to active (medium-speed) mode if SSBY = 0, MSON = 1, and LSON = 0, or to subactive mode if SSBY = 1, TMA3 = 1, and LSON = 1</li> <li>When a SLEEP instruction is executed in active (medium-speed) mode, a direct transition is made to active (high-speed) mode if SSBY = 0, MSON = 0, and LSON = 0, or to subactive mode if SSBY = 1, TMA3 = 1, and LSON = 1</li> <li>When a SLEEP instruction is executed in subactive mode, a direct transition is made to active (high-speed) mode if SSBY = 1, TMA3 = 1, and LSON = 1</li> <li>When a SLEEP instruction is executed in subactive mode, a direct transition is made to active (high-speed) mode if SSBY = 1, TMA3 = 1, LSON = 0, and MSON = 0, or to active (medium-speed) mode if SSBY = 1, TMA3 = 1, LSON = 0, and MSON = 1</li> </ul>

#### Bit 2-Medium Speed on Flag (MSON)

After standby, watch, or sleep mode is cleared, this bit selects active (high-speed) or active (medium-speed) mode.

Bit 2 MSON	Description	
0	Operation in active (high-speed) mode	(initial value)
1	Operation in active (medium-speed) mode	



#### Bits 1 and 0—Subactive Mode Clock Select (SA1, SA0)

These bits select the CPU clock rate  $(\phi_w/2, \phi_w/4, \text{ or } \phi_w/8)$  in subactive mode. SA1 and SA0 cannot be modified in subactive mode.

Bit 1 SA1	Bit 0 SA0	Description	
0	0	φ <sub>w</sub> /8	(initial value)
0	1	φ <sub>w</sub> /4	
1	*	φ <sub>w</sub> /2	

\*: Don't care

# 5.2 Sleep Mode

#### 5.2.1 Transition to Sleep Mode

#### 1. Transition to sleep (high-speed) mode

The system goes from active mode to sleep (high-speed) mode when a SLEEP instruction is executed while the SSBY and LSON bits in SYSCR1 are cleared to 0, the MSON and DTON bits in SYSCR2 are cleared to 0. In sleep mode CPU operation is halted but the on-chip peripheral functions. CPU register contents are retained.

2. Transition to sleep (medium-speed) mode

The system goes from active mode to sleep (medium-speed) mode when a SLEEP instruction is executed while the SSBY and LSON bits in SYSCR1 are cleared to 0, the MSON bit in SYSCR2 is set to 1, and the DTON bit in SYSCR2 is cleared to 0. In sleep (medium-speed) mode, as in sleep (high-speed) mode, CPU operation is halted but the on-chip peripheral functions are operational. The clock frequency in sleep (medium-speed) mode is determined by the MA1 and MA0 bits in SYSCR1. CPU register contents are retained.

Furthermore, it sometimes acts with half state early timing at the time of transition to sleep (medium-speed) mode.



### 5.2.2 Clearing Sleep Mode

Sleep mode is cleared by any interrupt (timer A, timer C, timer F, timer G, asynchronous event counter, IRQAEC, IRQ<sub>4</sub>, IRQ<sub>3</sub>, IRQ<sub>1</sub>, IRQ<sub>0</sub>, WKP<sub>7</sub> to WKP<sub>0</sub>, SCI3, A/D converter), or by input at the  $\overline{\text{RES}}$  pin.

• Clearing by interrupt

When an interrupt is requested, sleep mode is cleared and interrupt exception handling starts. A transition is made from sleep (high-speed) mode to active (high-speed) mode, or from sleep (medium-speed) mode to active (medium-speed) mode. Sleep mode is not cleared if the I bit of the condition code register (CCR) is set to 1 or the particular interrupt is disabled in the interrupt enable register.

To synchronize the interrupt request signal with the system clock, up to  $2/\phi(s)$  delay may occur after the interrupt request signal occurrence, before the interrupt exception handling start.

• Clearing by  $\overline{\text{RES}}$  input

When the  $\overline{\text{RES}}$  pin goes low, the CPU goes into the reset state and sleep mode is cleared.

# 5.2.3 Clock Frequency in Sleep (Medium-Speed) Mode

Operation in sleep (medium-speed) mode is clocked at the frequency designated by the MA1 and MA0 bits in SYSCR1.



# 5.3 Standby Mode

### 5.3.1 Transition to Standby Mode

The system goes from active mode to standby mode when a SLEEP instruction is executed while the SSBY bit in SYSCR1 is set to 1, the LSON bit in SYSCR1 is cleared to 0, and bit TMA3 in TMA is cleared to 0. In standby mode the clock pulse generator stops, so the CPU and on-chip peripheral modules stop functioning, but as long as the rated voltage is supplied, the contents of CPU registers, on-chip RAM, and some on-chip peripheral module registers are retained. On-chip RAM contents will be further retained down to a minimum RAM data retention voltage. The I/O ports go to the high-impedance state.

# 5.3.2 Clearing Standby Mode

Standby mode is cleared by an interrupt (IRQ<sub>1</sub> or IRQ<sub>0</sub>), WKP<sub>7</sub> to WKP<sub>0</sub> or by input at the  $\overline{\text{RES}}$  pin.

• Clearing by interrupt

When an interrupt is requested, the system clock pulse generator starts. After the time set in bits STS2 to STS0 in SYSCR1 has elapsed, a stable system clock signal is supplied to the entire chip, standby mode is cleared, and interrupt exception handling starts. Operation resumes in active (high-speed) mode if MSON = 0 in SYSCR2, or active (medium-speed) mode if MSON = 1. Standby mode is not cleared if the I bit of CCR is set to 1 or the particular interrupt is disabled in the interrupt enable register.

• Clearing by RES input

When the  $\overline{\text{RES}}$  pin goes low, the system clock pulse generator starts. After the pulse generator output has stabilized, if the  $\overline{\text{RES}}$  pin is driven high, the CPU starts reset exception handling. Since system clock signals are supplied to the entire chip as soon as the system clock pulse generator starts functioning, the  $\overline{\text{RES}}$  pin should be kept at the low level until the pulse generator output stabilizes.

(Unit: ms)

#### 5.3.3 Oscillator Stabilization Time after Standby Mode is Cleared

Bits STS2 to STS0 in SYSCR1 should be set as follows.

• When an oscillator is used

The table below gives settings for various operating frequencies. Set bits STS2 to STS0 for a wait time at least as long as the oscillation stabilization time.

					(Chite his)
STS2	STS1	STS0	Wait Time	5 MHz	2 MHz
0	0	0	8,192 states	1.638	4.1
		1	16,384 states	3.277	8.2
	1	0	32,768 states	6.554	16.4
		1	65,536 states	13.108	32.8
1	0	0	131,072 states	26.216	65.5
		1	2 states (Use prohibited with other than external clock)	0.0004	0.001
	1	0	8 states	0.002	0.004
		1	16 states	0.003	0.008

#### Table 5.4 Clock Frequency and Stabilization Time

- When an external clock is used STS2 = 1, STS1 = 0, and STS0 = 1 should be set. Other values possible use, but CPU sometimes will start operation before wait time completion.
- When the on-chip oscillator is used

8,192 states (STS2 = STS1 = STS0 = 0) is recommended if the on-chip oscillator is used.





#### 5.3.4 Standby Mode Transition and Pin States

When a SLEEP instruction is executed in active (high-speed) mode or active (medium-speed) mode while bit SSBY is set to 1 and bit LSON is cleared to 0 in SYSCR1, and bit TMA3 is cleared to 0 in TMA, a transition is made to standby mode. At the same time, pins go to the high-impedance state (except pins for which the pull-up MOS is designated as on). Figure 5.2 shows the timing in this case.



Figure 5.2 Standby Mode Transition and Pin States



#### 5.3.5 Notes on External Input Signal Changes before/after Standby Mode

- When external input signal changes before/after standby mode or watch mode
  When an external input signal such as IRQ, WKP, or IRQAEC is input, both the high- and
  low-level widths of the signal must be at least two cycles of system clock φ or subclock φ<sub>sub</sub>
  (referred to together in this section as the internal clock). As the internal clock stops in standby
  mode and watch mode, the width of external input signals requires careful attention when a
  transition is made via these operating modes. Ensure that external input signals conform to the
  conditions stated in 3, Recommended timing of external input signals, below
- 2. When external input signals cannot be captured because internal clock stops The case of falling edge capture is illustrated in figure 5.3.

As shown in the case marked "Capture not possible," when an external input signal falls immediately after a transition to active (high-speed or medium-speed) mode or subactive mode, after oscillation is started by an interrupt via a different signal, the external input signal cannot be captured if the high-level width at that point is less than 2 t<sub>eve</sub> or 2 t<sub>subeve</sub>.

3. Recommended timing of external input signals

To ensure dependable capture of an external input signal, high- and low-level signal widths of at least 2  $t_{eye}$  or 2  $t_{subcye}$  are necessary before a transition is made to standby mode or watch mode, as shown in "Capture possible: case 1" in figure 5.3.

External input signal capture is also possible with the timing shown in "Capture possible: case 2" and "Capture possible: case 3" in figure 5.3, in which a 2  $t_{cyc}$  or 2  $t_{subcyc}$  level width is secured.





Figure 5.3 External Input Signal Capture when Signal Changes before/after Standby Mode or Watch Mode

4. Input pins to which these notes apply:  $\overline{IRQ}_4$ ,  $\overline{IRQ}_3$ ,  $\overline{IRQ}_1$ ,  $\overline{IRQ}_0$ ,  $\overline{WKP}_7$  to  $\overline{WKP}_0$ , IRQAEC, TMIC, TMIF, TMIG,  $\overline{ADTRG}$ .

# 5.4 Watch Mode

# 5.4.1 Transition to Watch Mode

The system goes from active or subactive mode to watch mode when a SLEEP instruction is executed while the SSBY bit in SYSCR1 is set to 1 and bit TMA3 in TMA is set to 1.

In watch mode, operation of on-chip peripheral modules is halted except for timer A, timer F, timer G, AEC and the LCD controller/driver (for which operation or halting can be set) is halted. As long as a minimum required voltage is applied, the contents of CPU registers, the on-chip RAM and some registers of the on-chip peripheral modules, are retained. I/O ports keep the same states as before the transition.

### 5.4.2 Clearing Watch Mode

Watch mode is cleared by an interrupt (timer A, timer F, timer G, IRQ0, or WKP7 to WKP0) or by input at the RES pin.

• Clearing by interrupt

When watch mode is cleared by interrupt, the mode to which a transition is made depends on the settings of LSON in SYSCR1 and MSON in SYSCR2. If both LSON and MSON are cleared to 0, transition is to active (high-speed) mode; if LSON = 0 and MSON = 1, transition is to active (medium-speed) mode; if LSON = 1, transition is to subactive mode. When the transition is to active mode, after the time set in SYSCR1 bits STS2 to STS0 has elapsed, a stable clock signal is supplied to the entire chip, watch mode is cleared, and interrupt exception handling starts. Watch mode is not cleared if the I bit of CCR is set to 1 or the particular interrupt is disabled in the interrupt enable register.

• Clearing by RES input

Clearing by RES pin is the same as for standby mode; see Clearing by RES input in section 5.3.2, Clearing Standby Mode.

### 5.4.3 Oscillator Stabilization Time after Watch Mode is Cleared

The wait time is the same as for standby mode; see section 5.3.3, Oscillator Stabilization Time after Standby Mode is Cleared.

### 5.4.4 Notes on External Input Signal Changes before/after Watch Mode

See section 5.3.5, Notes on External Input Signal Changes before/after Standby Mode.



# 5.5 Subsleep Mode

### 5.5.1 Transition to Subsleep Mode

The system goes from subactive mode to subsleep mode when a SLEEP instruction is executed while the SSBY bit in SYSCR1 is cleared to 0, LSON bit in SYSCR1 is set to 1, and TMA3 bit in TMA is set to 1. In subsleep mode, operation of on-chip peripheral modules other than the A/D converter and PWM is in active state. As long as a minimum required voltage is applied, the contents of CPU registers, the on-chip RAM and some registers of the on-chip peripheral modules are retained. I/O ports keep the same states as before the transition.

### 5.5.2 Clearing Subsleep Mode

Subsleep mode is cleared by an interrupt (timer A, timer C, timer F, timer G, asynchronous event counter, SCI3, IRQAEC,  $IRQ_4$ ,  $IRQ_3$ ,  $IRQ_1$ ,  $IRQ_0$ ,  $WKP_7$  to  $WKP_0$ ) or by a low input at the RES pin.

• Clearing by interrupt

When an interrupt is requested, subsleep mode is cleared and interrupt exception handling starts. Subsleep mode is not cleared if the I bit of CCR is set to 1 or the particular interrupt is disabled in the interrupt enable register.

To synchronize the interrupt request signal with the system clock, up to  $2/\phi_{sub}(s)$  delay may occur after the interrupt request signal occurrence, before the interrupt exception handling start.

• Clearing by RES input

Clearing by RES pin is the same as for standby mode; see Clearing by RES input in section 5.3.2, Clearing Standby Mode.

# 5.6 Subactive Mode

# 5.6.1 Transition to Subactive Mode

Subactive mode is entered from watch mode if a timer A, timer F, timer G,  $IRQ_0$ , or  $WKP_7$  to  $WKP_0$  interrupt is requested while the LSON bit in SYSCR1 is set to 1. From subsleep mode, subactive mode is entered if a timer A, timer C, timer F, timer G, asynchronous event counter, SCI3, IRQAEC,  $IRQ_4$ ,  $IRQ_3$ ,  $IRQ_1$ ,  $IRQ_0$ , or  $WKP_7$  to  $WKP_0$  interrupt is requested. A transition to subactive mode does not take place if the I bit of CCR is set to 1 or the particular interrupt is disabled in the interrupt enable register.

# 5.6.2 Clearing Subactive Mode

Subactive mode is cleared by a SLEEP instruction or by a low input at the RES pin.

• Clearing by SLEEP instruction

If a SLEEP instruction is executed while the SSBY bit in SYSCR1 is set to 1 and TMA3 bit in TMA is set to 1, subactive mode is cleared and watch mode is entered. If a SLEEP instruction is executed while SSBY = 0 and LSON = 1 in SYSCR1 and TMA3 = 1 in TMA, subsleep mode is entered. Direct transfer to active mode is also possible; see section 5.8, Direct Transfer, below.

• Clearing by RES pin

Clearing by RES pin is the same as for standby mode; see Clearing by RES input in section 5.3.2, Clearing Standby Mode.

# 5.6.3 Operating Frequency in Subactive Mode

The operating frequency in subactive mode is set in bits SA1 and SA0 in SYSCR2. The choices are  $\phi_w/2$ ,  $\phi_w/4$ , and  $\phi_w/8$ .



# 5.7 Active (Medium-Speed) Mode

### 5.7.1 Transition to Active (Medium-Speed) Mode

If the MSON bit in SYSCR2 is set to 1 while the LSON bit in SYSCR1 is cleared to 0, a transition to active (medium-speed) mode results from  $IRQ_0$ ,  $IRQ_1$  or  $WKP_7$  to  $WKP_0$  interrupts in standby mode, timer A, timer F, timer G,  $IRQ_0$ , or  $WKP_7$  to  $WKP_0$  interrupts in watch mode, or any interrupt in sleep mode. A transition to active (medium-speed) mode does not take place if the I bit of CCR is set to 1 or the particular interrupt is disabled in the interrupt enable register. Furthermore, it sometimes acts with half state early timing at the time of transition to active (medium-speed) mode.

### 5.7.2 Clearing Active (Medium-Speed) Mode

Active (medium-speed) mode is cleared by a SLEEP instruction.

• Clearing by SLEEP instruction

A transition to standby mode takes place if the SLEEP instruction is executed while the SSBY bit in SYSCR1 is set to 1, the LSON bit in SYSCR1 is cleared to 0, and the TMA3 bit in TMA is cleared to 0. The system goes to watch mode if the SSBY bit in SYSCR1 is set to 1 and bit TMA3 in TMA is set to 1 when a SLEEP instruction is executed.

When both SSBY and LSON are cleared to 0 in SYSCR1 and a SLEEP instruction is executed, sleep mode is entered. Direct transfer to active (high-speed) mode or to subactive mode is also possible. See section 5.8, Direct Transfer, below for details.

• Clearing by  $\overline{\text{RES}}$  pin

When the  $\overline{\text{RES}}$  pin is driven low, a transition is made to the reset state and active (medium-speed) mode is cleared.

# 5.7.3 Operating Frequency in Active (Medium-Speed) Mode

Operation in active (medium-speed) mode is clocked at the frequency designated by the MA1 and MA0 bits in SYSCR1.

# 5.8 Direct Transfer

# 5.8.1 Overview of Direct Transfer

The CPU can execute programs in three modes: active (high-speed) mode, active (medium-speed) mode, and subactive mode. A direct transfer is a transition among these three modes without the stopping of program execution. A direct transfer can be made by executing a SLEEP instruction while the DTON bit in SYSCR2 is set to 1. After the mode transition, direct transfer interrupt exception handling starts.

If the direct transfer interrupt is disabled in interrupt enable register 2 (IENR2), a transition is made instead to sleep mode or watch mode. Note that if a direct transition is attempted while the I bit in CCR is set to 1, sleep mode or watch mode will be entered, and it will be impossible to clear the resulting mode by means of an interrupt.

- Direct transfer from active (high-speed) mode to active (medium-speed) mode When a SLEEP instruction is executed in active (high-speed) mode while the SSBY and LSON bits in SYSCR1 are cleared to 0, the MSON bit in SYSCR2 is set to 1, and the DTON bit in SYSCR2 is set to 1, a transition is made to active (medium-speed) mode via sleep mode.
- Direct transfer from active (medium-speed) mode to active (high-speed) mode When a SLEEP instruction is executed in active (medium-speed) mode while the SSBY and LSON bits in SYSCR1 are cleared to 0, the MSON bit in SYSCR2 is cleared to 0, and the DTON bit in SYSCR2 is set to 1, a transition is made to active (high-speed) mode via sleep mode.
- Direct transfer from active (high-speed) mode to subactive mode When a SLEEP instruction is executed in active (high-speed) mode while the SSBY and LSON bits in SYSCR1 are set to 1, the DTON bit in SYSCR2 is set to 1, and the TMA3 bit in TMA is set to 1, a transition is made to subactive mode via watch mode.
- Direct transfer from subactive mode to active (high-speed) mode When a SLEEP instruction is executed in subactive mode while the SSBY bit in SYSCR1 is set to 1, the LSON bit in SYSCR1 is cleared to 0, the MSON bit in SYSCR2 is cleared to 0, the DTON bit in SYSCR2 is set to 1, and the TMA3 bit in TMA is set to 1, a transition is made directly to active (high-speed) mode via watch mode after the waiting time set in SYSCR1 bits STS2 to STS0 has elapsed.



- Direct transfer from active (medium-speed) mode to subactive mode When a SLEEP instruction is executed in active (medium-speed) mode while the SSBY and LSON bits in SYSCR1 are set to 1, the DTON bit in SYSCR2 is set to 1, and the TMA3 bit in TMA is set to 1, a transition is made to subactive mode via watch mode.
- Direct transfer from subactive mode to active (medium-speed) mode
   When a SLEEP instruction is executed in subactive mode while the SSBY bit in SYSCR1 is set to 1, the LSON bit in SYSCR1 is cleared to 0, the MSON bit in SYSCR2 is set to 1, the DTON bit in SYSCR2 is set to 1, and the TMA3 bit in TMA is set to 1, a transition is made directly to active (medium-speed) mode via watch mode after the waiting time set in SYSCR1 bits STS2 to STS0 has elapsed.

# 5.8.2 Direct Transition Times

# 1. Time for direct transition from active (high-speed) mode to active (medium-speed) mode

A direct transition from active (high-speed) mode to active (medium-speed) mode is performed by executing a SLEEP instruction in active (high-speed) mode while bits SSBY and LSON are both cleared to 0 in SYSCR1, and bits MSON and DTON are both set to 1 in SYSCR2. The time from execution of the SLEEP instruction to the end of interrupt exception handling (the direct transition time) is given by equation (1) below.

```
Direct transition time = { (Number of SLEEP instruction execution states) + (number of internal processing states) } × (tcyc before transition) + (number of interrupt exception handling execution states) × (tcyc after transition)
```

......(1)

Example: Direct transition time =  $(2 + 1) \times 2 \text{tosc} + 14 \times 16 \text{tosc} = 230 \text{tosc}$  (when  $\phi/8$  is selected as the CPU operating clock)

[Legend] tosc: OSC clock cycle time tcyc: System clock (φ) cycle time

### 2. Time for direct transition from active (medium-speed) mode to active (high-speed) mode

A direct transition from active (medium-speed) mode to active (high-speed) mode is performed by executing a SLEEP instruction in active (medium-speed) mode while bits SSBY and LSON are both cleared to 0 in SYSCR1, and bit MSON is cleared to 0 and bit DTON is set to 1 in SYSCR2. The time from execution of the SLEEP instruction to the end of interrupt exception handling (the direct transition time) is given by equation (2) below.

```
Direct transition time = { (Number of SLEEP instruction execution states) + (number of internal processing states) } × (tcyc before transition) + (number of interrupt exception handling execution states) × (tcyc after transition)
```

Example: Direct transition time =  $(2 + 1) \times 16$ tosc +  $14 \times 2$ tosc = 76tosc (when  $\phi/8$  is selected as the CPU operating clock)

[Legend] tosc: OSC clock cycle time tcyc: System clock (φ) cycle time

#### 3. Time for direct transition from subactive mode to active (high-speed) mode

A direct transition from subactive mode to active (high-speed) mode is performed by executing a SLEEP instruction in subactive mode while bit SSBY is set to 1 and bit LSON is cleared to 0 in SYSCR1, bit MSON is cleared to 0 and bit DTON is set to 1 in SYSCR2, and bit TMA3 is set to 1 in TMA. The time from execution of the SLEEP instruction to the end of interrupt exception handling (the direct transition time) is given by equation (3) below.

Direct transition time = { (Number of SLEEP instruction execution states) + (number of internal
processing states) } × (tsubcyc before transition) + { (wait time set in
STS2 to STS0) + (number of interrupt exception handling execution
states) $\} \times (tcyc after transition)$ (3)

Example: Direct transition time =  $(2 + 1) \times 8tw + (8192 + 14) \times 2tosc = 24tw + 16412tosc$  (when  $\frac{\psi}{8}$  is selected as the CPU operating clock, and wait time = 8192 states)

RENESAS

[Legend]

tosc:	OSC clock cycle time
-------	----------------------

- tw: Watch clock cycle time
- tcyc: System clock ( $\phi$ ) cycle time

tsubcyc: Subclock ( $\phi_{SUB}$ ) cycle time

# 4. Time for direct transition from subactive mode to active (medium-speed) mode

A direct transition from subactive mode to active (medium-speed) mode is performed by executing a SLEEP instruction in subactive mode while bit SSBY is set to 1 and bit LSON is cleared to 0 in SYSCR1, bits MSON and DTON are both set to 1 in SYSCR2, and bit TMA3 is set to 1 in TMA. The time from execution of the SLEEP instruction to the end of interrupt exception handling (the direct transition time) is given by equation (4) below.

Direct transition time = { (Number of SLEEP instruction execution states) + (number	per of internal
processing states) $\} \times (tsubcyc before transition) + \{ (wait the transition) + \} $	time set in
STS2 to STS0) + (number of interrupt exception handling execution	
states) } × (tcyc after transition)	(4)

Example: Direct transition time =  $(2 + 1) \times 8$ tw +  $(8192 + 14) \times 16$ tosc = 24tw + 131296tosc (when  $\phi$ w/8 or  $\phi$ /8 is selected as the CPU operating clock, and wait time = 8192 states)

[Legend]

tosc:	OSC clock cycle time
tw:	Watch clock cycle time

tcyc: System clock ( $\phi$ ) cycle time

tsubcyc: Subclock ( $\phi_{SUB}$ ) cycle time

# 5.8.3 Notes on External Input Signal Changes before/after Direct Transition

- Direct transition from active (high-speed) mode to subactive mode Since the mode transition is performed via watch mode, see section 5.3.5, Notes on External Input Signal Changes before/after Standby Mode.
- Direct transition from active (medium-speed) mode to subactive mode Since the mode transition is performed via watch mode, see section 5.3.5, Notes on External Input Signal Changes before/after Standby Mode.
- Direct transition from subactive mode to active (high-speed) mode Since the mode transition is performed via watch mode, see section 5.3.5, Notes on External Input Signal Changes before/after Standby Mode.
- Direct transition from subactive mode to active (medium-speed) mode Since the mode transition is performed via watch mode, see section 5.3.5, Notes on External Input Signal Changes before/after Standby Mode.
# 5.9 Module Standby Mode

### 5.9.1 Setting Module Standby Mode

Module standby mode is set for individual peripheral functions. All the on-chip peripheral modules can be placed in module standby mode. When a module enters module standby mode, the system clock supply to the module is stopped and operation of the module halts. This state is identical to standby mode.

Module standby mode is set for a particular module by setting the corresponding bit to 0 in clock stop register 1 (CKSTPR1) or clock stop register 2 (CKSTPR2). (See table 5.5.)

### 5.9.2 Clearing Module Standby Mode

Module standby mode is cleared for a particular module by setting the corresponding bit to 1 in clock stop register 1 (CKSTPR1) or clock stop register 2 (CKSTPR2). (See table 5.5.)

Following a reset, clock stop register 1 (CKSTPR1) and clock stop register 2 (CKSTPR2) are both initialized to H'FF.

Register Name	Bit Name		Operation		
CKSTPR1	TACKSTP	1	Timer A module standby mode is cleared		
		0	Timer A is set to module standby mode		
	TCCKSTP	1	Timer C module standby mode is cleared		
		0	Timer C is set to module standby mode		
	TFCKSTP 1		Timer F module standby mode is cleared		
		0	Timer F is set to module standby mode		
	TGCKSTP	1	Timer G module standby mode is cleared		
		0	Timer G is set to module standby mode		
	ADCKSTP	1	A/D converter module standby mode is cleared		
		0	A/D converter is set to module standby mode		
	S32CKSTP	1	SCI3 module standby mode is cleared		
		0	SCI3 is set to module standby mode		

Table 5.5	Setting and Clearin	g Module Standby	Mode by Cl	ock Stop Register
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Register Name	Bit Name	Operation		
CKSTPR2	LDCKSTP	1	LCD module standby mode is cleared	
		0	LCD is set to module standby mode	
	PW1CKSTP	1	PWM1 module standby mode is cleared	
		0	PWM1 is set to module standby mode	
	WDCKSTP	1	Watchdog timer module standby mode is cleared	
		0	Watchdog timer is set to module standby mode	
	AECKSTP	1	Asynchronous event counter module standby mode is cleared	
		0	Asynchronous event counter is set to module standby mode	
	PW2CKSTP	1	PWM2 module standby mode is cleared	
		0	PWM2 is set to module standby mode	
	LVDCKSTP	1	LVD module standby mode is cleared	
		0	LVD is set to module standby mode	

Note: For details of module operation, see the sections on the individual modules.



# Section 6 ROM

# 6.1 Overview

The H8/38524 has 32 Kbytes of on-chip mask ROM, the H8/38523 has 24 Kbytes, the H8/38522 has 16 Kbytes, the H8/38521 has 12 Kbytes, and the H8/38520 has 8 Kbytes. The ROM is connected to the CPU by a 16-bit data bus, allowing high-speed two-state access for both byte data and word data. Flash memory versions of the H8/38524 and H8/38522 are available. The former has 32 Kbytes, and the latter 16 Kbytes, of flash memory.



# 6.2 Flash Memory Overview

### 6.2.1 Features

The features of the 32-Kbyte or 16-Kbyte flash memory built into the flash memory versions are summarized below.

- Programming/erase methods
  - The flash memory is programmed 128 bytes at a time. Erase is performed in single-block units. On the HD64F38524, the flash memory is configured as follows: 1 Kbyte × 4 blocks, 28 Kbytes × 1 block. On the HD64F38522, the flash memory is configured as follows: 1 Kbyte × 4 blocks, 12 Kbytes × 1 block. To erase the entire flash memory, each block must be erased in turn.

Note: The system clock oscillator must be used when programming or erasing the flash memory.

- Reprogramming capability
  - The HD64F38524 and HD64F38522 can be reprogrammed up to 1,000 times.
- On-board programming
  - On-board programming/erasing can be done in boot mode, in which the boot program built into the chip is started to erase or program of the entire flash memory. In normal user program mode, individual blocks can be erased or programmed.
- Programmer mode
  - Flash memory can be programmed/erased in programmer mode using a PROM programmer, as well as in on-board programming mode.
- Automatic bit rate adjustment
  - For data transfer in boot mode, this LSI's bit rate can be automatically adjusted to match the transfer bit rate of the host.
- Programming/erasing protection
  - Sets software protection against flash memory programming/erasing.
- Power-down mode
  - The power supply circuit is partly halted in the subactive mode and can be read in the power-down mode.

### 6.2.2 Block Diagram



Figure 6.1 Block Diagram of Flash Memory



### 6.2.3 Block Configuration

Figure 6.2 shows the block configuration of the flash memory. The thick lines indicate erasing units, the narrow lines indicate programming units, and the values are addresses. In versions with 32 Kbytes of flash memory, the flash memory is divided into 1 Kbyte  $\times$  4 blocks and 28 Kbytes  $\times$  1 block. In versions with 16 Kbytes of flash memory, the flash memory is divided into 1 Kbyte  $\times$  4 blocks and 12 Kbytes  $\times$  1 block. Erasing is performed in these units. Programming is performed in 128-byte units starting from an address with lower eight bits H'00 or H'80.

	H'0000	H'0001	H'0002	← Programming unit: 128 bytes →	H'007F
Erase unit	H'0080	H'0081	H'0082		H'00FF
1 Kbyte		1 1 1			
		1   	1 I		1 1
	H'0380	H'0381	H'0382		H'03FF
	H'0400	H'0401	H'0402	← Programming unit: 128 bytes →	H'047F
Erase unit	H'0480	H'0481	H'0482		H'04FF
1 Kbyte		     			
-	H'0780	H'0781	H'0782		H'07FF
ľ	H'0800	H'0801	H'0802	← Programming unit: 128 bytes →	H'087F
Erase unit	H'0880	H'0881	H'0882		H'08FF
1 Kbyte		1 1 1			1 1 1
-	H'0B80	H'0B81	H'0B82		H'0BFF
F	H'0C00	H'0C01	H'0C02	← Programming unit: 128 bytes →	H'0C7F
Erase unit	H'0C80	H'0C81	H'0C82		H'0CFF
1 Kbyte		-       			
-	H'0F80	H'0F81	H'0F82		H'0FFF
ľ	H'1000	H'1001	H'1002	← Programming unit: 128 bytes →	H'107F
Erase unit	H'1080	H'1081	H'1082		H'10FF
28 Kbytes					
-	H'7F80	H'7F81	H'7F82		H'7FFF

Figure 6.2(1) Block Configuration of 32-Kbyte Flash Memory

]	H'0000	H'0001	H'0002	← Programming unit: 128 bytes →	H'007F
Erase unit	H'0080	H'0081	H'0082	, , , , , , , , , , , , , , , , , , ,	H'00FF
1 Kbyte					
	H'0380	H'0381	H'0382		H'03FF
	H'0400	H'0401	H'0402	← Programming unit: 128 bytes →	H'047F
Erase unit	H'0480	H'0481	H'0482		H'04FF
1 Kbyte					
ŀ	H'0780	H'0781	H'0782		
-				<b>D</b>	H'07FF
	H'0800	H'0801	H'0802	← Programming unit: 128 bytes →	H'087F
Erase unit	H'0880	H'0881	H'0882		H'08FF
1 Kbyte					
-	H'0B80	H'0B81	H'0B82		H'0BFF
	H'0C00	H'0C01	H'0C02	← Programming unit: 128 bytes →	H'0C7F
Erase unit	H'0C80	H'0C81	H'0C82		H'0CFF
1 Kbyte					
	H'0F80	H'0F81	H'0F82		H'0FFF
	H'1000	H'1001	H'1002	← Programming unit: 128 bytes →	H'107F
Erase unit	H'1080	H'1081	H'1082		H'10FF
12 Kbytes					
	H'3F80	H'3F81	H'3F82		H'3FFF

### Figure 6.2(2) Block Configuration of 16-Kbyte Flash Memory



### 6.2.4 Register Configuration

Table 6.1 lists the register configuration to control the flash memory when the built in flash memory is effective.

### Table 6.1 Register Configuration

Register Name	Abbreviation	R/W	Initial Value	Address
Flash memory control register 1	FLMCR1	R/W	H'00	H'F020
Flash memory control register 2	FLMCR2	R	H'00	H'F021
Flash memory power control register	FLPWCR	R/W	H'00	H'F022
Erase block register	EBR	R/W	H'00	H'F023
Flash memory enable register	FENR	R/W	H'00	H'F02B

Note: FLMCR1, FLMCR2, FLPWCR, EBR, and FENR are 8 bit registers. Only byte access is enabled which are two-state access. These registers are dedicated to the product in which flash memory is included. The product in which mask ROM is included does not have these registers. When the corresponding address is read in these products, the value is undefined. A write is disabled.

# 6.3 Descriptions of Registers of the Flash Memory

### 6.3.1 Flash Memory Control Register 1 (FLMCR1)

Bit	7	6	5	4	3	2	1	0
	_	SWE	ESU	PSU	EV	PV	E	Р
Initial value	0	0	0	0	0	0	0	0
Read/Write	_	R/W						

FLMCR1 is a register that makes the flash memory change to program mode, program-verify mode, erase mode, or erase-verify mode. For details on register setting, refer to section 6.5, Flash Memory Programming/Erasing. By setting this register, the flash memory enters program mode, erase mode, program-verify mode, or erase-verify mode. Read the data in the state that bits 6 to 0 of this register are cleared when using flash memory as normal built-in ROM.

### Bit 7—Reserved

This bit is always read as 0 and cannot be modified.

### Bit 6—Software Write Enable (SWE)

This bit is to set enabling/disabling of programming/enabling of flash memory (set when bits 5 to 0 and the EBR register are to be set).

Bit 6 SWE	Description
0	Programming/erasing is disabled. Other FLMCR1 register bits and all EBR bits cannot be set. (initial value)
1	Flash memory programming/erasing is enabled.

### Bit 5—Erase Setup (ESU)

This bit is to prepare for changing to erase mode. Set this bit to 1 before setting the E bit to 1 in FLMCR1 (do not set SWE, PSU, EV, PV, E, and P bits at the same time).

Bit 5 ESU	Description	
0	The erase setup state is cancelled (initial va	lue)
1	The flash memory changes to the erase setup state. Set this bit to 1 before sett the E bit to 1 in FLMCR1.	ing

### Bit 4—Program Setup (PSU)

This bit is to prepare for changing to program mode. Set this bit to 1 before setting the P bit to 1 in FLMCR1 (do not set SWE, ESU, EV, PV, E, and P bits at the same time).

Bit 4 PSU	Description	
0	The program setup state is cancelled	(initial value)
1	The flash memory changes to the program setup state. Set this bit to setting the P bit to 1 in FLMCR1.	o 1 before



### Bit 3—Erase-Verify (EV)

This bit is to set changing to or canceling erase-verify mode (do not set SWE, ESU, PSU, PV, E, and P bits at the same time).

Bit 3 EV	Description	
0	Erase-verify mode is cancelled	(initial value)
1	The flash memory changes to erase-verify mode	

### Bit 2—Program-Verify (PV)

This bit is to set changing to or canceling program-verify mode (do not set SWE, ESU, PSU, EV, E, and P bits at the same time).

Bit 2	Description	
PV	Description	
0	Program-verify mode is cancelled	(initial value)
1	The flash memory changes to program-verify mode	

### Bit 1—Erase (E)

This bit is to set changing to or canceling erase mode (do not set SWE, ESU, PSU, EV, PV, and P bits at the same time).

Bit 1 E	Description	
0	Erase mode is cancelled	(initial value)
1	When this bit is set to 1, while the SWE = 1 and ESU = 1, the flash changes to erase mode.	n memory



### Bit 0—Program (P)

This bit is to set changing to or canceling program mode (do not set SWE, ESU, PSU, EV, PV, and E bits at the same time).

Bit 0 P	Description	
0	Program mode is cancelled (ir	nitial value)
1	When this bit is set to 1, while the SWE = 1 and $PSU = 1$ , the flash mem changes to program mode.	ory

#### 6.3.2 Flash Memory Control Register 2 (FLMCR2)

Bit	7	6	5	4	3	2	1	0
	FLER	_	_	_	_	—	—	—
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	—	—	—	—	—	—	—

FLMCR2 is a register that displays the state of flash memory programming/erasing. FLMCR2 is a read-only register, and should not be written to.

### Bit 7—Flash Memory Error (FLER)

This bit is set when the flash memory detects an error and goes to the error-protection state during programming or erasing to the flash memory. See section 6.6.3, Error Protection, for details.

Bit 7 FLER	Description	
0	The flash memory operates normally.	(initial value)
1	Indicates that an error has occurred during an operation on flash m (programming or erasing).	nemory

### Bits 6 to 0—Reserved

These bits are always read as 0 and cannot be modified.



Bit	7	6	5	4	3	2	1	0
	_	—	—	EB4	EB3	EB2	EB1	EB0
Initial value	0	0	0	0	0	0	0	0
Read/Write	—		—	R/W	R/W	R/W	R/W	R/W

### 6.3.3 Erase Block Register (EBR)

EBR specifies the flash memory erase area block. EBR is initialized to H'00 when the SWE bit in FLMCR1 is 0. Do not set more than one bit at a time, as this will cause all the bits in EBR to be automatically cleared to 0. When each bit is set to 1 in EBR, the corresponding block can be erased. Other blocks change to the erase-protection state. See table 6.2 for the method of dividing blocks of the flash memory. When the whole bits are to be erased, erase them in turn in unit of a block.

EBR	Bit Name	Block (Size)	Address
0	EB0	EB0 (1 Kbyte)	H'0000 to H'03FF
1	EB1	EB1 (1 Kbyte)	H'0400 to H'07FF
2	EB2	EB2 (1 Kbyte)	H'0800 to H'0BFF
3	EB3	EB3 (1 Kbyte)	H'0C00 to H'0FFF
4	EB4	EB4 (12 Kbytes)	H'1000 to H'3FFF (HD64F38522)
		EB4 (28 Kbytes)	H'1000 to H'7FFF (HD64F38524)

### Table 6.2 Division of Blocks to Be Erased

### 6.3.4 Flash Memory Power Control Register (FLPWCR)

Bit	7	6	5	4	3	2	1	0
	PDWND		_	_				—
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	_	_	_	_	_	_	_

FLPWCR enables or disables a transition to the flash memory power-down mode when the LSI switches to subactive mode. The power supply circuit can be read in the subactive mode, although it is partly halted in the power-down mode.

### Bit 7—Power-down Disable (PDWND)

This bit selects the power-down mode of the flash memory when a transition to the subactive mode is made.

Bit 7 PDWND	Description
0	When this bit is 0 and a transition is made to the subactive mode, the flash memory enters the power-down mode. (initial value)
1	When this bit is 1, the flash memory remains in the normal mode even after a transition is made to the subactive mode.

### Bits 6 to 0—Reserved

These bits are always read as 0 and cannot be modified.

### 6.3.5 Flash Memory Enable Register (FENR)

Bit	7	6	5	4	3	2	1	0	
	FLSHE	_		—	—	—	—	_	]
Initial value	0	0	0	0	0	0	0	0	-
Read/Write	R/W	_						_	

FENR controls CPU access to the flash memory control registers, FLMCR1, FLMCR2, EBR, and FLPWCR.

### Bit 7—Flash Memory Control Register Enable (FLSHE)

This bit controls access to the flash memory control registers.

Bit 7 FLSHE	Description	
0	Flash memory control registers cannot be accessed	(initial value)
1	Flash memory control registers can be accessed	

### Bits 6 to 0—Reserved

These bits are always read as 0 and cannot be modified.



# 6.4 On-Board Programming Modes

There are two modes for programming/erasing of the flash memory; boot mode, which enables onboard programming/erasing, and programmer mode, in which programming/erasing is performed with a PROM programmer. On-board programming/erasing can also be performed in user program mode. At reset-start in reset mode, the series of HD64F38524 and HD64F38522 changes to a mode depending on the TEST pin settings, P95 pin settings, and input level of each port, as shown in table 6.3. The input level of each pin must be defined four states before the reset ends.

When changing to boot mode, the boot program built into this LSI is initiated. The boot program transfers the programming control program from the externally-connected host to on-chip RAM via SCI3. After erasing the entire flash memory, the programming control program is executed. This can be used for programming initial values in the on-board state or for a forcible return when programming/erasing can no longer be done in user program mode. In user program mode, individual blocks can be erased and programmed by branching to the user program/erase control program prepared by the user.

TEST	P95	P34	PB0	PB1	PB2	LSI State after Reset End
0	1	Х	Х	Х	Х	User Mode
0	0	1	Х	Х	Х	Boot Mode
1	Х	Х	0	0	0	Programmer Mode

X: Don't care

### 6.4.1 Boot Mode

Table 6.4 shows the boot mode operations between reset end and branching to the programming control program.

- 1. When boot mode is used, the flash memory programming control program must be prepared in the host beforehand. Prepare a programming control program in accordance with the description in section 6.5, Flash Memory Programming/Erasing.
- 2. SCI3 should be set to asynchronous mode, and the transfer format as follows: 8-bit data, 1 stop bit, and no parity. The inversion function of TXD and RXD pins by the SPCR register is set to "Not to be inverted," so do not put the circuit for inverting a value between the host and this LSI.

- 3. When the boot program is initiated, the chip measures the low-level period of asynchronous SCI communication data (H'00) transmitted continuously from the host. The chip then calculates the bit rate of transmission from the host, and adjusts the SCI3 bit rate to match that of the host. The reset should end with the RXD pin high. The RXD and TXD pins should be pulled up on the board if necessary. After the reset is complete, it takes approximately 100 states before the chip is ready to measure the low-level period.
- 4. After matching the bit rates, the chip transmits one H'00 byte to the host to indicate the completion of bit rate adjustment. The host should confirm that this adjustment end indication (H'00) has been received normally, and transmit one H'55 byte to the chip. If reception could not be performed normally, initiate boot mode again by a reset. Depending on the host's transfer bit rate and system clock frequency of this LSI, there will be a discrepancy between the bit rates of the host and the chip. To operate the SCI properly, set the host's transfer bit rate and system clock frequency of this LSI within the ranges listed in table 6.5.
- 5. In boot mode, a part of the on-chip RAM area is used by the boot program. The area H'F780 to H'FEEF is the area to which the programming control program is transferred from the host. The boot program area cannot be used until the execution state in boot mode switches to the programming control program.
- 6. Before branching to the programming control program, the chip terminates transfer operations by SCI3 (by clearing the RE and TE bits in SCR to 0), however the adjusted bit rate value remains set in BRR. Therefore, the programming control program can still use it for transfer of write data or verify data with the host. The TXD pin is high (PCR42 = 1, P42 = 1). The contents of the CPU general registers are undefined immediately after branching to the programming control program. These registers must be initialized at the beginning of the programming control program, as the stack pointer (SP), in particular, is used implicitly in subroutine calls, etc.
- Boot mode can be cleared by a reset. End the reset after driving the reset pin low, waiting at least 20 states, and then setting the TEST pin and P95 pin. Boot mode is also cleared when a WDT overflow occurs.
- 8. Do not change the TEST pin and P95 pin input levels in boot mode.



	Host Operation	LSI Operation			
ltem	Processing Contents	Processing Contents			
		Branches to boot program at reset-start.			
Bit rate adjustment	Continuously transmits data H'00 at specified bit rate.	<ul> <li>Measures low-level period of receive data H'00.</li> <li>Calculates bit rate and sets it in BRR of SCI3.</li> <li>Transmits data H'00 to the host to indicate that the adjustment has ended.</li> </ul>			
Flash memory erase	↓ Transmits data H'55 when data H'00 is received and no error occurs.	•			
		Checks flash memory data, erases all flash memory blocks in case of written data existing, and transmits data H'AA to host. (If erase could not be done, transmits data H'FF to host and aborts operation.)			
Transfer of programming control program	Transmits number of bytes (N) of programming control program to be transferred as 2-byte data (low-order byte following high-order byte)	Echobacks the 2-byte received data to host.			
+	•	+			
Transfer of programming control program (repeated for N times)	Transmits 1-byte of programming control program	Echobacks received data to host and also transfers it to RAM.			
Execution of		Transmits 1-byte data H'AA to host.			
Programming control program		Branches to programming control program transferred to on-chip RAM and starts execution.			

# Table 6.4 Boot Mode Operation

# Table 6.5 Oscillating Frequencies (f<sub>osc</sub>) for which Automatic Adjustment of LSI Bit Rate is Possible

Host Bit Rate	Oscillating Frequencies (f <sub>osc</sub> ) Range of LSI
19,200 bps	16 to 20 MHz
9,600 bps	8 to 20 MHz
4,800 bps	6 to 20 MHz
2,400 bps	2 to 20 MHz
1,200 bps	2 to 20 MHz

### 6.4.2 Programming/Erasing in User Program Mode

The term user mode refers to the status when a user program is being executed. On-board programming/erasing of an individual flash memory block can also be performed in user program mode by branching to a user program/erase control program. The user must set branching conditions and provide on-board means of supplying programming data. The flash memory must contain the user program/erase control program or a program that provides the user program/erase control program from external memory. As the flash memory itself cannot be read during programming/erasing, transfer the user program/erase control program to on-chip RAM, as in boot mode. Figure 6.3 shows a sample procedure for programming/erasing in user program mode. Prepare a user program/erase control program in accordance with the description in section 6.5, Flash Memory Programming/Erasing.



Figure 6.3 Programming/Erasing Flowchart Example in User Program Mode



### 6.4.3 Notes on On-Board Programming

- You must use the system clock oscillator when programming or erasing flash memory. The onchip oscillator should not be used for programming or erasing flash memory. See section 4.2 (5), On-Chip Oscillator Selection Method, for information on switching between the system clock oscillator and the on-chip oscillator.
- 2. The watchdog timer operates after a reset is canceled. When executing a program prepared by the user that performs programming and erasing in the user mode, the watchdog timer's overflow cycle should be set to an appropriate value. Refer to section 6.5.1, Program/Program-Verify, for information on the appropriate watchdog timer overflow cycle for programming, and refer to section 6.5.2, Erase/Erase-Verify, for information on the appropriate watchdog timer overflow cycle for erasing.

# 6.5 Flash Memory Programming/Erasing

A software method using the CPU is employed to program and erase flash memory in the onboard programming modes. Depending on the FLMCR1 setting, the flash memory operates in one of the following four modes: Program mode, program-verify mode, erase mode, and erase-verify mode. The programming control program in boot mode and the user program/erase control program in user program mode use these operating modes in combination to perform programming/erasing. Flash memory programming and erasing should be performed in accordance with the descriptions in section 6.5.1, Program/Program-Verify and section 6.5.2, Erase/Erase-Verify, respectively.

### 6.5.1 Program/Program-Verify

When writing data or programs to the flash memory, the program/program-verify flowchart shown in figure 6.4 should be followed. Performing programming operations according to this flowchart will enable data or programs to be written to the flash memory without subjecting the chip to voltage stress or sacrificing program data reliability.

- 1. Programming must be done to an empty address. Do not reprogram an address to which programming has already been performed.
- 2. Programming should be carried out 128 bytes at a time. A 128-byte data transfer must be performed even if writing fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
- 3. Prepare the following data storage areas in RAM: A 128-byte programming data area, a 128byte reprogramming data area, and a 128-byte additional-programming data area. Perform reprogramming data computation according to table 6.6, and additional programming data computation according to table 6.7.
- 4. Consecutively transfer 128 bytes of data in byte units from the reprogramming data area or additional-programming data area to the flash memory. The program address and 128-byte data are latched in the flash memory. The lower 8 bits of the start address in the flash memory destination area must be H'00 or H'80.
- 5. The time during which the P bit is set to 1 is the programming time. Table 6.8 shows the allowable programming times.
- 6. The watchdog timer (WDT) is set to prevent over-programming due to program runaway, etc. An overflow cycle of approximately 6.6 ms is allowed.
- 7. For a dummy write to a verify address, write 1-byte data H'FF to an address whose lower 1 bit is b'0. Verify data can be read in word size from the address to which a dummy write was performed.
- 8. The maximum number of repetitions of the program/program-verify sequence of the same bit is 1,000.





Figure 6.4 Program/Program-Verify Flowchart

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Program Data	Verify Data	Reprogram Data	Comments
0	0	1	Programming completed
0	1	0	Reprogram bit
1	0	1	_
1	1	1	Remains in erased state

# Table 6.6 Reprogram Data Computation Table

### Table 6.7 Additional-Program Data Computation Table

Reprogram Data	Verify Data	Additional-Program Data	Comments
0	0	0	Additional-program bit
0	1	1	No additional programming
1	0	1	No additional programming
1	1	1	No additional programming

### Table 6.8Programming Time

n (Number of Writes)	Programming Time	In Additional Programming	Comments
1 to 6	30	10	
7 to 1,000	200	—	

Note: Time shown in  $\mu$ s.



### 6.5.2 Erase/Erase-Verify

When erasing flash memory, the erase/erase-verify flowchart shown in figure 6.5 should be followed.

- 1. Prewriting (setting erase block data to all 0s) is not necessary.
- 2. Erasing is performed in block units. Make only a single-bit specification in the erase block register (EBR). To erase multiple blocks, each block must be erased in turn.
- 3. The time during which the E bit is set to 1 is the flash memory erase time.
- 4. The watchdog timer (WDT) is set to prevent over-erasing due to program runaway, etc. An overflow cycle of approximately 19.8 ms is allowed.
- 5. For a dummy write to a verify address, write 1-byte data H'FF to an address whose lower 1 bit is b'0. Verify data can be read in word size from the address to which a dummy write was performed.
- 6. If the read data is not erased successfully, set erase mode again, and repeat the erase/erase-verify sequence as before. The maximum number of repetitions of the erase/erase-verify sequence is 100.

### 6.5.3 Interrupt Handling when Programming/Erasing Flash Memory

All interrupts, are disabled while flash memory is being programmed or erased, or while the boot program is executing, for the following three reasons:

- 1. Interrupt during programming/erasing may cause a violation of the programming or erasing algorithm, with the result that normal operation cannot be assured.
- 2. If interrupt exception handling starts before the vector address is written or during programming/erasing, a correct vector cannot be fetched and the CPU malfunctions.
- 3. If an interrupt occurs during boot program execution, normal boot mode sequence cannot be carried out.









# 6.6 Program/Erase Protection

There are three kinds of flash memory program/erase protection; hardware protection, software protection, and error protection.

# 6.6.1 Hardware Protection

Hardware protection refers to a state in which programming/erasing of flash memory is forcibly disabled or aborted because of a transition to reset, subactive mode, subsleep mode, watch mode, or standby mode. Flash memory control register 1 (FLMCR1), flash memory control register 2 (FLMCR2), and erase block register (EBR) are initialized. In a reset via the  $\overline{\text{RES}}$  pin, the reset state is not entered unless the  $\overline{\text{RES}}$  pin is held low until oscillation stabilizes after powering on. In the case of a reset during operation, hold the  $\overline{\text{RES}}$  pin low for the  $\overline{\text{RES}}$  pulse width specified in the AC Characteristics section.

# 6.6.2 Software Protection

Software protection can be implemented against programming/erasing of all flash memory blocks by clearing the SWE bit in FLMCR1. When software protection is in effect, setting the P or E bit in FLMCR1 does not cause a transition to program mode or erase mode. By setting the erase block register (EBR), erase protection can be set for individual blocks. When EBR is set to H'00, erase protection is set for all blocks.

### 6.6.3 Error Protection

In error protection, an error is detected when CPU runaway occurs during flash memory programming/erasing, or operation is not performed in accordance with the program/erase algorithm, and the program/erase operation is aborted. Aborting the program/erase operation prevents damage to the flash memory due to over-programming or over-erasing.

When the following errors are detected during programming/erasing of flash memory, the FLER bit in FLMCR2 is set to 1, and the error protection state is entered.

- When the flash memory of the relevant address area is read during programming/erasing (including vector read and instruction fetch)
- Immediately after exception handling excluding a reset during programming/erasing
- When a SLEEP instruction is executed during programming/erasing

The FLMCR1, FLMCR2, and EBR settings are retained, however program mode or erase mode is aborted at the point at which the error occurred. Program mode or erase mode cannot be re-entered by re-setting the P or E bit. However, PV and EV bit setting is enabled, and a transition can be made to verify mode. Error protection can be cleared only by a power-on reset.

# 6.7 Programmer Mode

In programmer mode, a PROM programmer can be used to perform programming/erasing via a socket adapter, just as a discrete flash memory. Use a PROM programmer that supports the MCU device type with the on-chip Renesas Technology (former Hitachi Ltd.) 64-Kbyte flash memory (F-ZTAT64V3). A 10-MHz input clock is required. For the conditions for transition to programmer mode, see table 6.3.

### 6.7.1 Socket Adapter

The socket adapter converts the pin allocation of the HD64F38524 and HD64F38522 to that of the discrete flash memory HN28F101. The address of the on-chip flash memory is H'0000 to H'7FFF. Figure 6.6 shows a socket-adapter-pin correspondence diagram of the HD64F38524 and HD64F38522.



### 6.7.2 Programmer Mode Commands

The following commands are supported in programmer mode.

- Memory Read Mode
- Auto-Program Mode
- Auto-Erase Mode
- Status Read Mode

Status polling is used for auto-programming, auto-erasing, and status read modes. In status read mode, detailed internal information is output after the execution of auto-programming or auto-erasing. Table 6.9 shows the sequence of each command. In auto-programming mode, 129 cycles are required since 128 bytes are written at the same time. In memory read mode, the number of cycles depends on the number of address write cycles (n).

### Table 6.9 Command Sequence in Programmer Mode

	Number	1st Cycle			2nd Cycle		
Command Name	of Cycles	Mode	Address	Data	Mode	Address	Data
Memory read	1 + n	Write	Х	H'00	Read	RA	Dout
Auto-program	129	Write	Х	H'40	Write	WA	Din
Auto-erase	2	Write	Х	H'20	Write	Х	H'20
Status read	2	Write	Х	H'71	Write	Х	H'71

n: the number of address write cycles



HD64F385	24, HD64F38522	Socket Adapter	Ϊ.			
Pin No.		(Conversion to	:	HN28F107	1 (32 Pins)	
FP-80A TFP-80C	Pin Name	32-Pin Arrangement)	ł	Pin Name	Pin No.	
166-800		-1-1	Ł	FWE	1	
30	P71	1:	1	A9	26	
		41	L	A16	20	
36	P77	1:	:	A15	3	
56	P92			WE	31	
21	P60	1	÷	1/00	13	
22	P61	1:	:	I/01	14	
23	P62	1	Ŀ	1/02	15	
24	P63		:	I/O3	17	
25	P64		÷	1/04	18	
26	P65		:	I/O5	19	
27	P66	H	:	I/O6	20	
28	P67		-	I/07	21	
69	P40	1⊹	<u> </u>	A0	12	
70	P41		-	A1	11	
63	P32	1	÷	A2	10	
64	P33	+ ·	:	A3	9	
65	P34	1	÷	A4	8	
66	P35		!	A5	7	
67	P36	]-;	÷	A6	6	
68	P37		÷	A7	5	
29	P70	]:	÷	A8	27	
71	P42		÷	ŌĒ	24	
31	P72	<u>↓</u> ;	÷	A10	23	
32	P73		÷	A11	25	
33	P74		÷	A12	4	
34	P75	+	÷	A13	28	
35	P76	1:	:	A14	29	
72	P43		:	CE	22	
52	Vcc		÷	Vcc	32	
1	AVcc		÷	Vss	16	
6	X1		Į.			
11	TEST			[Legend]		
51	V1			FWE: I/O7 to I/O0:	Flash-write e	
52	Vcc				Address input/ou	
58	P94	4: 1	1	CE:	Chip enable	
4,59	CVcc, P95			OE: WE:	Output enabl Write enable	e
8	Vss		1			
53	Vss PB0	-[	÷	Note: The osc	cillation freque	ency
73			i.	of the o	scillator circu	
	PB1		1	should	be 10 MHz.	
75	PB2 OSC1,OSC2	Oscillator circuit	i.			
10,9	RES	Power-on	1			
Other than the a		reset circuit	i.			
	(0. 211)					
Element (				_	a Dia au	

Figure 6.6 Socket Adapter Pin Correspondence Diagram (HD64F38524, HD64F38522)

### 6.7.3 Memory Read Mode

- 1. After completion of auto-program/auto-erase/status read operations, a transition is made to the command wait state. When reading memory contents, a transition to memory read mode must first be made with a command write, after which the memory contents are read. Once memory read mode has been entered, consecutive reads can be performed.
- 2. In memory read mode, command writes can be performed in the same way as in the command wait state.
- 3. After powering on, memory read mode is entered.
- 4. Tables 6.10 to 6.12 show the AC characteristics.

### Table 6.10 AC Characteristics in Transition to Memory Read Mode

Conditions:  $V_{cc} = 3.3 \text{ V} \pm 0.3 \text{ V}$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	Figure 6.7
CE hold time	t <sub>ceh</sub>	0	—	ns	
CE setup time	t <sub>ces</sub>	0	—	ns	
Data hold time	t <sub>dh</sub>	50	—	ns	
Data setup time	t <sub>ds</sub>	50	—	ns	
Write pulse width	t <sub>wep</sub>	70	—	ns	
WE rise time	t,	—	30	ns	
WE fall time	t,		30	ns	



Figure 6.7 Timing Waveforms for Memory Read after Memory Write



# Table 6.11 AC Characteristics in Transition from Memory Read Mode to Another Mode

Conditions:	$V_{cc} = 3.3 \text{ V} \pm 0.2$	$3 V, V_{ss} = 0 V, T$	$f_a = 25^{\circ}C \pm 5^{\circ}C$
-------------	----------------------------------	------------------------	------------------------------------

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	Figure 6.8
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
WE rise time	t,	_	30	ns	
WE fall time	t <sub>r</sub>	_	30	ns	



Figure 6.8 Timing Waveforms in Transition from Memory Read Mode to Another Mode

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### Table 6.12 AC Characteristics in Memory Read Mode

Conditions:  $V_{cc} = 3.3 \text{ V} \pm 0.3 \text{ V}$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Access time	t <sub>acc</sub>	_	20	μs	Figure 6.9
CE output delay time	t <sub>ce</sub>	—	150	ns	Figure 6.10
OE output delay time	t <sub>oe</sub>	—	150	ns	
Output disable delay time	t <sub>df</sub>	—	100	ns	
Data output hold time	t <sub>oh</sub>	5		ns	



Figure 6.9  $\overline{CE}$  and  $\overline{OE}$  Enable State Read Timing Waveforms



Figure 6.10  $\overline{\text{CE}}$  and  $\overline{\text{OE}}$  Clock System Read Timing Waveforms

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### 6.7.4 Auto-Program Mode

- 1. When reprogramming previously programmed addresses, perform auto-erasing before autoprogramming.
- 2. Perform auto-programming once only on the same address block. It is not possible to program an address block that has already been programmed.
- 3. In auto-program mode, 128 bytes are programmed simultaneously. This should be carried out by executing 128 consecutive byte transfers. A 128-byte data transfer is necessary even when programming fewer than 128 bytes. In this case, H'FF data must be written to the extra addresses.
- 4. The lower 7 bits of the transfer address must be low. If a value other than an effective address is input, processing will switch to a memory write operation but a write error will be flagged.
- 5. Memory address transfer is performed in the second cycle (figure 6.11). Do not perform transfer after the third cycle.
- 6. Do not perform a command write during a programming operation.
- 7. Perform one auto-program operation for a 128-byte block for each address. Two or more additional programming operations cannot be performed on a previously programmed address block.
- 8. Confirm normal end of auto-programming by checking I/O6. Alternatively, status read mode can also be used for this purpose (I/O7 status polling uses the auto-program operation end decision pin).
- Status polling I/O6 and I/O7 pin information is retained until the next command write. As long as the next command write has not been performed, reading is possible by enabling CE and OE.
- 10. Table 6.13 shows the AC characteristics.



### Table 6.13 AC Characteristics in Auto-Program Mode

Conditions:  $V_{cc} = 3.3 \text{ V} \pm 0.3 \text{ V}$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	Figure 6.11
CE hold time	t <sub>ceh</sub>	0	—	ns	
CE setup time	t <sub>ces</sub>	0	—	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
Status polling start time	t <sub>wsts</sub>	1	_	ms	
Status polling access time	t <sub>spa</sub>	_	150	ns	
Address setup time	t <sub>as</sub>	0	_	ns	
Address hold time	t <sub>ah</sub>	60	_	ns	
Memory write time	t <sub>write</sub>	1	3000	ms	
WE rise time	t,	_	30	ns	
WE fall time	t,	_	30	ns	



Figure 6.11 Auto-Program Mode Timing Waveforms

### 6.7.5 Auto-Erase Mode

- 1. Auto-erase mode supports only entire memory erasing.
- 2. Do not perform a command write during auto-erasing.
- 3. Confirm normal end of auto-erasing by checking I/O6. Alternatively, status read mode can also be used for this purpose (I/O7 status polling uses the auto-erase operation end decision pin).
- Status polling I/O6 and I/O7 pin information is retained until the next command write. As long as the next command write has not been performed, reading is possible by enabling CE and OE.
- 5. Table 6.14 shows the AC characteristics.

### Table 6.14 AC Characteristics in Auto-Erase Mode

Item	Symbol	Min	Max	Unit	Notes
Command write cycle	t <sub>nxtc</sub>	20	_	μs	Figure 6.12
CE hold time	t <sub>ceh</sub>	0	_	ns	_
CE setup time	t <sub>ces</sub>	0	_	ns	_
Data hold time	t <sub>dh</sub>	50	_	ns	_
Data setup time	t <sub>ds</sub>	50	_	ns	_
Write pulse width	t <sub>wep</sub>	70	_	ns	_
Status polling start time	t <sub>ests</sub>	1	_	ms	_
Status polling access time	t <sub>spa</sub>	_	150	ns	_
Memory erase time	t <sub>erase</sub>	100	40000	ms	_
WE rise time	t,	_	30	ns	_
WE fall time	t <sub>r</sub>	_	30	ns	_

Conditions:  $V_{cc} = 3.3 \text{ V} \pm 0.3 \text{ V}$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 





Figure 6.12 Auto-Erase Mode Timing Waveforms

### 6.7.6 Status Read Mode

- 1. Status read mode is provided to identify the kind of abnormal end. Use this mode when an abnormal end occurs in auto-program mode or auto-erase mode.
- 2. The return code is retained until a command write other than a status read mode command write is executed.
- 3. Table 6.15 shows the AC characteristics and 6.16 shows the return codes.

### Table 6.15 AC Characteristics in Status Read Mode

Conditions:  $V_{cc} = 3.3 \text{ V} \pm 0.3 \text{ V}$ ,  $V_{ss} = 0 \text{ V}$ ,  $T_a = 25^{\circ}\text{C} \pm 5^{\circ}\text{C}$ 

Item	Symbol	Min	Max	Unit	Notes
Read time after command write	t <sub>nxtc</sub>	20	_	μs	Figure 6.13
CE hold time	t <sub>ceh</sub>	0	_	ns	
CE setup time	t <sub>ces</sub>	0	_	ns	
Data hold time	t <sub>dh</sub>	50	_	ns	
Data setup time	t <sub>ds</sub>	50	_	ns	
Write pulse width	t <sub>wep</sub>	70	_	ns	
OE output delay time	t <sub>oe</sub>	—	150	ns	
Disable delay time	t <sub>df</sub>	_	100	ns	
CE output delay time	t <sub>ce</sub>	_	150	ns	
WE rise time	t,	_	30	ns	
WE fall time	t,	—	30	ns	



Figure 6.13 Status Read Mode Timing Waveforms





Pin Name	Initial Value	Indications	
I/07	0	1: Abnormal end	
		0: Normal end	
I/O6	0	1: Command error	
		0: Otherwise	
I/O5	0	1: Programming error	
		0: Otherwise	
I/O4	0	1: Erasing error	
		0: Otherwise	
I/O3	0	_	
I/O2	0	_	
I/O1	0	1: Over counting of writing or erasing	
		0: Otherwise	
I/O0	0	1: Effective address error	
		0: Otherwise	

### Table 6.16 Status Read Mode Return Codes

### 6.7.7 Status Polling

- 1. The I/O7 status polling flag indicates the operating status in auto-program/auto-erase mode.
- 2. The I/O6 status polling flag indicates a normal or abnormal end in auto-program/auto-erase mode.

 Table 6.17
 Status Polling Output Truth Table

I/07	I/O6	I/O0 to 5	Status
0	0	0	During internal operation
1	0	0	Abnormal end
1	1	0	Normal end
0	1	0	_
#### 6.7.8 Programmer Mode Transition Time

Commands cannot be accepted during the oscillation stabilization period or the programmer mode setup period. After the programmer mode setup time, a transition is made to memory read mode.

Item	Symbol	Min	Max	Unit	Notes
Oscillation stabilization time(crystal oscillator)	T <sub>osc1</sub>	10	_	ms	Figure 6.14
Oscillation stabilization time(ceramic oscillator)	T <sub>osc1</sub>	5	_	ms	
Programmer mode setup time	T bmv	10	_	ms	
Vcc hold time	$T_{dwn}$	0		ms	

#### Table 6.18 Stipulated Transition Times to Command Wait State



# Figure 6.14 Oscillation Stabilization Time, Boot Program Transfer Time, and Power-Down Sequence

#### 6.7.9 Notes on Memory Programming

- 1. When performing programming using programmer mode on a chip that has been programmed/erased in an on-board programming mode, auto-erasing is recommended before carrying out auto-programming.
- 2. The flash memory is initially in the erased state when the device is shipped by Renesas Technology. For other chips for which the erasure history is unknown, it is recommended that auto-erasing be executed to check and supplement the initialization (erase) level.



# 6.8 Power-Down States for Flash Memory

In user mode, the flash memory will operate in either of the following states:

• Normal operating mode

The flash memory can be read and written to at high speed.

• Power-down operating mode

The power supply circuit of the flash memory is partly halted and can be read under low power consumption.

• Standby mode

All flash memory circuits are halted.

Table 6.19 shows the correspondence between the operating modes of this LSI and the flash memory. In subactive mode, the flash memory can be set to operate in power-down mode with the PDWND bit in FLPWCR. When the flash memory returns to its normal operating state from power-down mode or standby mode, a period to stabilize the power supply circuits that were stopped is needed. When the flash memory returns to its normal operating state, bits STS2 to STS0 in SYSCR1 must be set to provide a wait time of at least 20  $\mu$ s, even when the external clock is being used.

<b>Table 6.19</b>	Flash Memory Operating States
-------------------	-------------------------------

	Flash Memory Operating State						
LSI Operating State	PDWND = 0 (Initial value)	PDWND = 1					
Active mode	Normal operating mode	Normal operating mode					
Subactive mode	Power-down mode	Normal operating mode					
Sleep mode	Normal operating mode	Normal operating mode					
Subsleep mode	Standby mode	Standby mode					
Standby mode	Standby mode	Standby mode					
Watch mode	Standby mode	Standby mode					

# Section 7 RAM

# 7.1 Overview

The H8/38524, H8/38523, and H8/38522 have 1 Kbyte of high-speed static RAM on-chip, and the H8/38521 and H8/38520 have 512 bytes. The RAM is connected to the CPU by a 16-bit data bus, allowing high-speed 2-state access for both byte data and word data.

# 7.1.1 Block Diagram

Figure 7.1 shows a block diagram of the on-chip RAM.



Figure 7.1 RAM Block Diagram





# Section 8 I/O Ports

# 8.1 Overview

The LSI is provided with five 8-bit I/O ports, one 4-bit I/O port, two 3-bit I/O ports, one 8-bit input-only port, one 1-bit input-only port, and one 6-bit output-only port. Table 8.1 indicates the functions of each port.

Each port has of a port control register (PCR) that controls input and output, and a port data register (PDR) for storing output data. Input or output can be assigned to individual bits. See section 2.8.3, Bit-Manipulation Instruction, for information on executing bit-manipulation instructions to write data in PCR or PDR.

Ports 5, 6, 7, 8, and A are also used as liquid crystal display segment and common pins, selectable in 4-bit units.

Block diagrams of each port are given in appendix B, I/O Port Block Diagrams.

Port	Description	Pins	Other Functions	Function Switching Registers
Port 1	<ul><li> 3-bit I/O port</li><li> MOS input pull-up</li></ul>	P1 <sub>-</sub> ∕IRQ₃∕TMIF	External interrupt 3, timer event input pin TMIF	PMR1 TCRF
	option	P1₄/IRQ₄/ADTRG	External interrupt 4, A/D converter external trigger	PMR1 AMR
		P1 <sub>3</sub> /TMIG	Timer G input capture	PMR1 PMR2
Port 3	<ul><li> 8-bit I/O port</li><li> MOS input pull-up</li></ul>	P3 <sub>7</sub> /AEVL P3 <sub>6</sub> /AEVH	Asynchronous counter event input pins AEVL, AEVH	PMR3 ECCR
	option	P3 <sub>5</sub> to P3 <sub>3</sub>	None	PMR2
<ul><li>Large-current port</li><li>MOS open drain</li></ul>		P3 <sub>2</sub> , TMOFH P3 <sub>1</sub> , TMOFL	Timer F output compare output	PMR3
	output selectable (only P3 <sub>5</sub> )	P3 <sub>0</sub> /UD	Timer C count up/down selection input	PMR3

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### Table 8.1Port Functions

Port	Description	Pins	Other Functions	Function Switching Registers
Port 4	1-bit input port	P4 <sub>3</sub> /IRQ <sub>0</sub>	External interrupt 0	PMR2
	• 3-bit I/O port	P4 <sub>2</sub> /TXD <sub>32</sub> P4 <sub>1</sub> /RXD <sub>32</sub> P4 <sub>0</sub> /SCK <sub>32</sub>	SCI3 data output $(TXD_{32})$ , data input $(RXD_{32})$ , clock input/output $(SCK_{32})$	SCR3 SMR3 SPCR
Port 5	<ul><li> 8-bit I/O port</li><li> MOS input pull-up option</li></ul>	$\frac{P5_{7} \text{ to } P5_{6}}{WKP_{7} \text{ to } WKP_{0}}/$ $SEG_{8} \text{ to } SEG_{1}$	Wakeup input ( $\overline{WKP}_7$ to $\overline{WKP}_0$ ), segment output (SEG <sub>8</sub> to SEG <sub>1</sub> )	PMR5 LPCR
Port 6	<ul><li> 8-bit I/O port</li><li> MOS input pull-up option</li></ul>	$P6_7$ to $P6_7$ SEG <sub>16</sub> to SEG <sub>9</sub>	Segment output (SEG <sub>16</sub> to SEG <sub>9</sub> )	LPCR
Port 7	8-bit I/O port	$P7_{7}$ to $P7_{0}$ / SEG <sub>24</sub> to SEG <sub>17</sub>	Segment output (SEG <sub>24</sub> to SEG <sub>17</sub> )	LPCR
Port 8	8-bit I/O port	$P8_7$ to $P8_0$ / SEG <sub>32</sub> to SEG <sub>25</sub>	Segment output (SEG $_{32}$ to SEG $_{25}$ )	LPCR
Port 9	Dedicated 6-bit     output port	P9 <sub>5</sub> , P9 <sub>4</sub> , P9 <sub>2</sub> , P9 <sub>3</sub> /V <sub>ref</sub>	LVD reference voltage external input pin	LVDSR
		P9₁, P9₀/ PWM2, PWM1	10-bit PWM output	PMR9
	Input port	IRQAEC	None	
Port A	4-bit I/O port	$PA_{3}$ to $PA_{0}$ / COM <sub>4</sub> to COM <sub>1</sub>	Common output (COM₄ to COM₁)	LPCR
Port B	Dedicated 8-bit     input port	$PB_7$ to $PB_4$ / AN <sub>7</sub> to AN <sub>4</sub>	A/D converter analog input $(AN_7 \text{ to } AN_4)$	AMR
		PB <sub>3</sub> /AN <sub>3</sub> /IRQ <sub>1</sub>	A/D converter analog input (AN <sub>3</sub> ), external interrupt 1, timer event input (TMIC)	AMR PMRB TMC
		PB <sub>2</sub> /AN <sub>2</sub>	A/D converter analog input	AMR
		PB <sub>1</sub> /AN <sub>1</sub> /(extU) PB <sub>0</sub> /AN <sub>0</sub> /(extD)	A/D converter analog input (LVD detect voltage external input pin)	AMR (LVDCR)

# 8.2 Port 1

#### 8.2.1 Overview

Port 1 is a 3-bit I/O port. Figure 8.1 shows its pin configuration.





#### 8.2.2 Register Configuration and Description

Table 8.2 shows the port 1 register configuration.

# Table 8.2 Port 1 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 1	PDR1	R/W	—	H'FFD4
Port control register 1	PCR1	W	—	H'FFE4
Port pull-up control register 1	PUCR1	R/W	_	H'FFE0
Port mode register 1	PMR1	R/W	—	H'FFC8
Port mode register 2	PMR2	R/W	H'D8	H'FFC9

# (1) Port Data Register 1 (PDR1)

Bit	7	6	5	4	3	2	1	0
	P1,	_		P1 <sub>4</sub>	P1 <sub>3</sub>	—		—
Initial value	0	—	_	0	0	—	_	_
Read/Write	R/W	—		R/W	R/W	—		—

PDR1 is an 8-bit register that stores data for port 1 pins  $P1_7$ ,  $P1_4$ , and  $P1_3$ . If port 1 is read while PCR1 bits are set to 1, the values stored in PDR1 are read, regardless of the actual pin states. If port 1 is read while PCR1 bits are cleared to 0, the pin states are read.

### (2) Port Control Register 1 (PCR1)

Bit	7	6	5	4	3	2	1	0
	PCR1 <sub>7</sub>	_	_	PCR1 <sub>4</sub>	PCR1 <sub>3</sub>	_	—	—
Initial value	0	—	_	0	0	—	—	_
Read/Write	W	—	W	W	W	W	W	W

PCR1 is an 8-bit register for controlling whether each of the port 1 pins  $P1_7$ ,  $P1_4$ , and  $P1_3$  functions as an input pin or output pin. Setting a PCR1 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. The settings in PCR1 and in PDR1 are valid only when the corresponding pin is designated in PMR1 as a general I/O pin.

PCR1 is a write-only register, which is always read as all 1s.

# (3) Port Pull-Up Control Register 1 (PUCR1)

Bit	7	6	5	4	3	2	1	0
	PUCR1 <sub>7</sub>			$PUCR1_4$	PUCR1 <sub>3</sub>	_	_	—
Initial value	0	_	—	0	0	—	—	_
Read/Write	R/W	—	W	R/W	R/W	W	W	W

PUCR1 controls whether the MOS pull-up of each of the port 1 pins  $P1_7$ ,  $P1_4$ , and  $P1_3$  is on or off. When a PCR1 bit is cleared to 0, setting the corresponding PUCR1 bit to 1 turns on the MOS pullup for the corresponding pin, while clearing the bit to 0 turns off the MOS pull-up.

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#### (4) Port Mode Register 1 (PMR1)

Bit	7	6	5	4	3	2	1	0
	IRQ3	_	_	IRQ4	TMIG		—	—
Initial value	0	1	_	0	0	—	1	
Read/Write	R/W	—	W	R/W	R/W	W	—	W

PMR1 is an 8-bit read/write register, controlling the selection of pin functions for port 1 pins.

# Bit 7—P1<sub>7</sub>/IRQ<sub>3</sub>/TMIF Pin Function Switch (IRQ3)

This bit selects whether pin P1<sub>7</sub>/ $\overline{IRQ}_3$ /TMIF is used as P1<sub>7</sub> or as  $\overline{IRQ}_3$ /TMIF.

Bit 7 IRQ3	Description	
0	Functions as $P1_7 I/O$ pin	(initial value)
1	Functions as IRQ <sub>3</sub> /TMIF input pin	

Note: Rising or falling edge sensing can be designated for IRQ<sub>3</sub>, TMIF. For details on TMIF settings, see (3) Timer Control Register F (TCRF) in section 9.4.2, Register Descriptions.

#### Bit 6—Reserved

This bit is reserved; it is always read as 1 and cannot be modified.

#### Bit 5—Reserved

This bit is reserved; it can only be written with 0.

# Bit 4—P1<sub>4</sub>/IRQ<sub>4</sub>/ADTRG Pin Function Switch (IRQ4)

This bit selects whether pin  $P1_4/\overline{IRQ_4}/\overline{ADTRG}$  is used as  $P1_4$  or as  $\overline{IRQ_4}/\overline{ADTRG}$ .

Bit 4 IRQ4	Description
0	Functions as P1, I/O pin (initial value)
1	Functions as IRQ <sub>4</sub> /ADTRG input pin
Note:	For details of ADTRG pin setting, see section 12.3.2, Start of A/D Conversion by External Trigger Input.



# Bit 3—P1<sub>3</sub>/TMIG Pin Function Switch (TMIG)

This bit selects whether pin P1<sub>3</sub>/TMIG is used as P1<sub>3</sub> or as TMIG.

Bit 3 TMIG	Description	
0	Functions as P1 <sub>3</sub> I/O pin	(initial value)
1	Functions as TMIG input pin	

#### Bits 2 and 0—Reserved

These bits are reserved; they can only be written with 0.

### Bit 1—Reserved

This bit is reserved; it is always read as 1 and cannot be modified.

# (5) Port Mode Register 2 (PMR2)

Bit	7	6	5	4	3	2	1	0
		_	POF1	_	_	WDCKS	NCS	IRQ0
Initial value	1	1	0	1	1	0	0	0
Read/Write	—	—	R/W	—		R/W	R/W	R/W

PMR2 is an 8-bit read/write register. It controls whether the PMOS transistor internal to P3<sub>5</sub> is on or off, the selection of the watchdog timer clock, the selection of TMIG noise cancellation, and switching of the P4<sub>3</sub>/ $\overline{IRQ_0}$  pin functions.

Upon reset, PMR2 is initialized to H'D8.

This section only deals with the bits related to timer G and the watchdog timer. For the functions of the bits, see the descriptions of port 3 (POF1) and port 4 (IRQ0).



# Bit 2—Watchdog Timer Source Clock (WDCKS)

This bit selects the watchdog timer source clock.

Bit 2 WDCKS	Description	
0	Selects clock based on timer mode register W (TMW) setting $\!\!\!\!\!\!*$	(Initial value)
1	Selects $\phi_w/32$	

Note: \* See section 9.6, Watchdog Timer, for details.

# Bit 1—TMIG Noise Canceller Select (NCS)

This bit selects controls the noise cancellation circuit of the input capture input signal (TMIG).

Bit 1 NCS	Description	
0	No noise cancellation circuit	(Initial value)
1	Noise cancellation circuit	



# 8.2.3 Pin Functions

Table 8.3 shows the port 1 pin functions.

## Table 8.3Port 1 Pin Functions

Pin	Pin Functions and	d Selection Met	thod					
P1 <sub>7</sub> /IRQ <sub>3</sub> /TMIF	The pin function de and bit $PCR1_7$ in P		Q3 in PMR1, bit	s CKSL2 to CK	SL0 in TCRF,			
	IRQ <sub>3</sub> 0 1							
	PCR1 <sub>7</sub>	0	1	*	:			
	CKSL2 to CKSL0	3	k	Not 0**	0**			
	Pin function	P1, input pin	P1, output pin	$\overline{IRQ}_{3}$ input pin	IRQ <sub>3</sub> /TMIF input pin			
	Note: When this p to disable th	in is used as the e IRQ <sub>3</sub> interrupt		, clear bit IEN3 t	o 0 in IENR1			
P1₄/ĪRQ₄ ADTRG	-'							
	IRQ4	(	)	1				
	PCR1₄	0	1	*	:			
	TRGE	3	k	0	1			
	Pin function	$P1_4$ input pin	P1 <sub>4</sub> output pin	IRQ₄ input pin	IRQ₄/ADTRG input pin			
	Note: When this pin is used as the $\overline{\text{ADTRG}}$ input pin, clear bit IEN4 to 0 in IENR1 to disable the IRQ <sub>4</sub> interrupt.							
P1 <sub>3</sub> /TMIG	The pin function de	pends on bit TM	/IG in PMR1 an	d bit PCR1 <sub>3</sub> in F	PCR1.			
	TMIG	(	)	1				
	PCR1 <sub>3</sub>	0	1	ł	•			
	Pin function	$P1_{3}$ input pin	$P1_{3}$ output pin	TMIG ir	iput pin			

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\*: Don't care

#### 8.2.4 Pin States

Table 8.4 shows the port 1 pin states in each operating mode.

#### Table 8.4Port 1 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P1,∕IRQ,∕TMIF P1,∕IRQ,∕ADTRG P1,∕TMIG	High- impedance	Retains previous state	Retains previous state	High- impedance <sup>*</sup>		Functional	Functional

Note: \* A high-level signal is output when the MOS pull-up is in the on state.

# 8.2.5 MOS Input Pull-Up

Port 1 has a built-in MOS input pull-up function that can be controlled by software. When a PCR1 bit is cleared to 0, setting the corresponding PUCR1 bit to 1 turns on the MOS input pull-up for that pin. The MOS input pull-up function is in the off state after a reset.

PCR1 <sub>n</sub>	0	0	1	
PUCR1 <sub>n</sub>	0	1	*	
MOS input pull-up	Off	On	Off	

(n = 7, 4, 3) \*: Don't care



Section 8 I/O Ports

# 8.3 Port 3

#### 8.3.1 Overview

Port 3 is an 8-bit I/O port, configured as shown in figure 8.2.





# 8.3.2 Register Configuration and Description

Table 8.5 shows the port 3 register configuration.

### Table 8.5Port 3 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 3	PDR3	R/W	H'00	H'FFD6
Port control register 3	PCR3	W	H'00	H'FFE6
Port pull-up control register 3	PUCR3	R/W	H'00	H'FFE1
Port mode register 2	PMR2	R/W	H'D8	H'FFC9
Port mode register 3	PMR3	R/W	_	H'FFCA

#### (1) Port Data Register 3 (PDR3)

Bit	7	6	5	4	3	2	1	0
	P37	P36	P35	P34	P33	P3 <sub>2</sub>	P3 <sub>1</sub>	P30
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PDR3 is an 8-bit register that stores data for port 3 pins  $P3_7$  to  $P3_0$ . If port 3 is read while PCR3 bits are set to 1, the values stored in PDR3 are read, regardless of the actual pin states. If port 3 is read while PCR3 bits are cleared to 0, the pin states are read.

Upon reset, PDR3 is initialized to H'00.

# (2) Port Control Register 3 (PCR3)

Bit	7	6	5	4	3	2	1	0
	PCR37	PCR3 <sub>6</sub>	PCR3 <sub>5</sub>	PCR3 <sub>4</sub>	PCR3 <sub>3</sub>	PCR3 <sub>2</sub>	PCR3 <sub>1</sub>	PCR30
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PCR3 is an 8-bit register for controlling whether each of the port 3 pins  $P3_7$  to  $P3_0$  functions as an input pin or output pin. Setting a PCR3 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. The settings in PCR3 and in PDR3 are valid only when the corresponding pin is designated in PMR3 as a general I/O pin.

Upon reset, PCR3 is initialized to H'00.

PCR3 is a write-only register, which is always read as all 1s.



# (3) Port Pull-Up Control Register 3 (PUCR3)

Bit	7	6	5	4	3	2	1	0
	PUCR37	PUCR36	PUCR35	PUCR3 <sub>4</sub>	PUCR33	PUCR3 <sub>2</sub>	PUCR31	PUCR30
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PUCR3 controls whether the MOS pull-up of each of the port 3 pins  $P3_7$  to  $P3_0$  is on or off. When a PCR3 bit is cleared to 0, setting the corresponding PUCR3 bit to 1 turns on the MOS pull-up for the corresponding pin, while clearing the bit to 0 turns off the MOS pull-up.

Upon reset, PUCR3 is initialized to H'00.

# (4) Port Mode Register 2 (PMR2)

Bit	7	6	5	4	3	2	1	0
	_	—	POF1	_	—	WDCKS	NCS	IRQ0
Initial value	1	1	0	1	1	0	0	0
Read/Write	—	—	R/W	—	—	R/W	R/W	R/W

PMR2 is an 8-bit read/write register. It controls whether the PMOS transistor internal to P3<sub>5</sub> is on or off, the selection of the watchdog timer clock, the selection of TMIG noise cancellation, and switching of the P4<sub>3</sub>/ $\overline{IRQ_0}$  pin functions.

Upon reset, PMR2 is initialized to H'D8.

This section only deals with the bit that controls whether the PMOS transistor internal to pin  $P3_5$  is on or off. For the functions of the other bits, see the descriptions of port 1 (WDCKS and NCS) and port 4 (IRQ0).

# Bit 5—Pin P3<sub>5</sub> PMOS Transistor Control (POF1)

This bit selects whether the PMOS transistor of the output buffer for pin  $P3_5$  is on or off.

Bit 5 POF1	Description
0	CMOS output (initial value)
1	NMOS open-drain output
Noto	The pip is an NMOS apon drain output when this hit is not to 1 and P2 is an output

Note: The pin is an NMOS open-drain output when this bit is set to 1 and  $P3_5$  is an output.

#### (5) Port Mode Register 3 (PMR3)

Bit	7	6	5	4	3	2	1	0
	AEVL	AEVH	_			TMOFH	TMOFL	UD
Initial value	0	0				0	0	0
Read/Write	R/W	R/W	W	W	W	R/W	R/W	R/W

PMR3 is an 8-bit read/write register, controlling the selection of pin functions for port 3 pins.

#### Bit 7—P3,/AEVL Pin Function Switch (AEVL)

This bit selects whether pin  $P3_7$ /AEVL is used as  $P3_7$  or as AEVL.

Bit 7 AEVL	Description	
0	Functions as P3, I/O pin	(initial value)
1	Functions as AEVL input pin	

#### Bit 6—P3<sub>6</sub>/AEVH Pin Function Switch (AEVH)

This bit selects whether pin P3<sub>6</sub>/AEVH is used as P3<sub>6</sub> or as AEVH.

Bit 6 AEVH	Description	
0	Functions as P3 <sub>6</sub> I/O pin	(initial value)
1	Functions as AEVH input pin	

#### Bits 5 to 3—Reserved

These bits are reserved; they can only be written with 0.

### Bit 2—P3<sub>2</sub>/TMOFH Pin Function Switch (TMOFH)

This bit selects whether pin P3<sub>2</sub>/TMOFH is used as P3<sub>2</sub> or as TMOFH.

Bit 2 TMOFH	Description	
0	Functions as $P3_2$ I/O pin	(initial value)
1	Functions as TMOFH output pin	

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# Bit 1—P3<sub>1</sub>/TMOFL Pin Function Switch (TMOFL)

This bit selects whether pin  $P3_1/TMOFL$  is used as  $P3_1$  or as TMOFL.

Bit 1 TMOFL	Description	
0	Functions as P3, I/O pin	(initial value)
1	Functions as TMOFL output pin	

# Bit 0—P3<sub>0</sub>/UD Pin Function Switch (UD)

This bit selects whether pin  $P3_0/UD$  is used as  $P3_0$  or as UD.

Bit 0 UD	Description	
0	Functions as $P3_0 I/O pin$	(initial value)
1	Functions as UD input pin	



# 8.3.3 Pin Functions

Table 8.6 shows the port 3 pin functions.

## Table 8.6Port 3 Pin Functions

Pin	Pin Functions and	Selection Method					
P3 <sub>7</sub> /AEVL	The pin function depends on bit AEVL in PMR3 and bit PCR3, in PCR3.						
	AEVL	(	0		1		
	PCR3 <sub>7</sub>	0	1		*		
	Pin function	P3, input pin	P3, out	put pin	AEVL input pin		
P3 <sub>6</sub> /AEVH	The pin function dep	ends on bit AEVH ir	n PMR3 an	d bit PCR	3₀ in PCR3.		
	AEVH	(	0		1		
	PCR3 <sub>6</sub>	0	1		*		
	Pin function	P3 <sub>6</sub> input pin	P3 <sub>6</sub> out	put pin	AEVH input pin		
P3₅ to P3₃	The pin function dep	ends on the corresp	onding bit	in PCR3.			
	PCR3n	0		1			
	Pin function	P3, input pin P		3, output pin			
	(n = 5 to 3)						
P3 <sub>2</sub> /TMOFH	The pin function depends on bit TMOFH in PMR3 and bit PCR3 <sub>2</sub> in PCR3.						
	TMOFH	0		1			
	PCR3 <sub>2</sub>	0	1		*		
	Pin function	P3 <sub>2</sub> input pin	P3 <sub>2</sub> output pin		TMOFH output pin		
P3,/TMOFL	The pin function depends on bit TMOFL in PMR3 and bit PCR3, in PCR3.						
	TMOFL	0			1		
	PCR3,	0	1		*		
	Pin function	P3, input pin	P3, output pin		THOFL output pin		
P3,/UD	The pin function dep	ends on bit UD in P	MR3 and b	it PCR30	in PCR3.		
	UD	(	0		1		
	PCR3 <sub>0</sub>	0	1		*		
	Pin function	P3 <sub>0</sub> input pin	P3 <sub>o</sub> output pin		UD input pin		
					*: Dop't coro		

\*: Don't care



#### 8.3.4 Pin States

Table 8.7 shows the port 3 pin states in each operating mode.

Table 8.7	Port 3 Pir	<b>States</b>
-----------	------------	---------------

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P3 <sub>7</sub> /AEVL P3 <sub>6</sub> /AEVH P3 <sub>5</sub> P3 <sub>4</sub> P3 <sub>3</sub> P3 <sub>2</sub> /TMOFH P3 <sub>7</sub> /TMOFL P3 <sub>6</sub> /UD	High- impedance	Retains previous state	Retains previous state	High- impedance <sup>*</sup>		Functional	Functional

Note: \* A high-level signal is output when the MOS pull-up is in the on state.

# 8.3.5 MOS Input Pull-Up

Port 3 has a built-in MOS input pull-up function that can be controlled by software. When a PCR3 bit is cleared to 0, setting the corresponding PUCR3 bit to 1 turns on the MOS pull-up for that pin. The MOS pull-up function is in the off state after a reset.

PCR3 <sub>n</sub>	0	0	1	
PUCR3 <sub>n</sub>	0	1	*	
MOS input pull-up	Off	On	Off	

(n = 7 to 0) \*: Don't care

# 8.4 Port 4

## 8.4.1 Overview

Port 4 is a 3-bit I/O port and 1-bit input port, configured as shown in figure 8.3.





# 8.4.2 Register Configuration and Description

Table 8.8 shows the port 4 register configuration.

## Table 8.8Port 4 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 4	PDR4	R/W	H'F8	H'FFD7
Port control register 4	PCR4	W	H'F8	H'FFE7
Port mode register 2	PMR2	R/W	H'D8	H'FFC9



# (1) Port Data Register 4 (PDR4)

Bit	7	6	5	4	3	2	1	0
	—	_		—	P43	P4 2	P4 <sub>1</sub>	P4 0
Initial value	1	1	1	1	1	0	0	0
Read/Write	_	_	_	_	R	R/W	R/W	R/W

PDR4 is an 8-bit register that stores data for port 4 pins  $P4_2$  to  $P4_0$ . If port 4 is read while PCR4 bits are set to 1, the values stored in PDR4 are read, regardless of the actual pin states. If port 4 is read while PCR4 bits are cleared to 0, the pin states are read.

Upon reset, PDR4 is initialized to H'F8.

# (2) Port Control Register 4 (PCR4)



PCR4 is an 8-bit register for controlling whether each of port 4 pins  $P4_2$  to  $P4_0$  functions as an input pin or output pin. Setting a PCR4 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. PCR4 and PDR4 settings are valid when the corresponding pins are designated for general-purpose input/output by SCR3.

Upon reset, PCR4 is initialized to H'F8.

PCR4 is a write-only register, which is always read as all 1s.

#### (3) Port Mode Register 2 (PMR2)

Bit	7	6	5	4	3	2	1	0
			POF1			WDCKS	NCS	IRQ <sub>0</sub>
Initial value	1	1	0	1	1	0	0	0
Read/Write	_		R/W			R/W	R/W	R/W

PMR2 is an 8-bit read/write register. It controls whether the PMOS transistor internal to  $P3_5$  is on or off, the selection of the watchdog timer clock, the selection of TMIG noise cancellation, and switching of the  $P4_3/\overline{IRQ_0}$  pin functions.

Upon reset, PMR2 is initialized to H'D8.

This section only deals with the bit that controls switching of the  $P4_3/\overline{IRQ}_0$  pin functions. For the functions of the other bits, see the descriptions of port 1 (WDCKS and NCS) and port 3 (POF1).

# Bit 0—P4<sub>3</sub>/IRQ<sub>0</sub> Pin Function Switch (IRQ<sub>0</sub>)

This bit selects whether pin P4<sub>3</sub>/ $\overline{IRQ}_0$  is used as P4<sub>3</sub> or as  $\overline{IRQ}_0$ .

Bit 0 IRQ₀	Description	
0	Functions as P4 <sub>3</sub> input pin	(initial value)
1	Functions as $\overline{IRQ}_0$ input pin	



# 8.4.3 Pin Functions

Table 8.9 shows the port 4 pin functions.

## Table 8.9Port 4 Pin Functions

Pin	Pin Functions and	Pin Functions and Selection Method							
P4 <sub>3</sub> /IRQ <sub>0</sub>	The pin function dep	pends on bit IR	Q0 in PMR2.						
	IRQ0	(	0		1				
	Pin function	P4 <sub>3</sub> in	put pin		$\overline{IRQ}_{_0}$ in	put pin			
P4 <sub>2</sub> /TXD <sub>32</sub>	The pin function dep PCR4 <sub>2</sub> in PCR4.	pends on bit TE	in SCR3, bit SF	PC32 in	SPCR,	and bit			
	SPC32		0			1			
	TE		0			1			
	PCR4 <sub>2</sub>	0 1		1		*			
	Pin function	P4 <sub>2</sub> input pin P4 <sub>2</sub> outp		put pin TXI		KD <sub>32</sub> output pin			
P4 <sub>1</sub> /RXD <sub>32</sub>	The pin function dep	The pin function depends on bit RE in SCR3 and bit PCR4, in PCR4.							
1 52	RE	0				1			
	PCR4,	0	-	1		*			
	Pin function	P4, input pi	in P4 <sub>1</sub> out	P4, output pin		RXD <sub>32</sub> input pin			
P4 <sub>0</sub> /SCK <sub>32</sub>	The pin function dep and bit PCR4₀ in PC		E1 and CKE0 in	n SCR3	, bit CO	M in SMR3,			
	CKE1		0			1			
	CKE0		0		1	*			
	COM		0	1	*	*			
	PCR4 <sub>0</sub> 0 1		1	1 *		*			
	Pin function $P4_0$ input pin $P4_0$ output pin				output in	SCK <sub>32</sub> input pin			

RENESAS

\*: Don't care

#### 8.4.4 Pin States

Table 8.10 shows the port 4 pin states in each operating mode.

#### Table 8.10Port 4 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P4 <sub>3</sub> /IRQ <sub>0</sub> P4 <sub>2</sub> /TXD <sub>32</sub> P4 <sub>1</sub> /RXD <sub>32</sub> P4 <sub>0</sub> /SCK <sub>32</sub>	High- impedance	Retains previous state		High- impedance	Retains previous state	Functional	Functional

# 8.5 Port 5

#### 8.5.1 Overview

Port 5 is an 8-bit I/O port, configured as shown in figure 8.4.



Figure 8.4 Port 5 Pin Configuration



# 8.5.2 Register Configuration and Description

Table 8.11 shows the port 5 register configuration.

#### Table 8.11Port 5 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 5	PDR5	R/W	H'00	H'FFD8
Port control register 5	PCR5	W	H'00	H'FFE8
Port pull-up control register 5	PUCR5	R/W	H'00	H'FFE2
Port mode register 5	PMR5	R/W	H'00	H'FFCC

# (1) Port Data Register 5 (PDR5)

Bit	7	6	5	4	3	2	1	0
	P5 <sub>7</sub>	P5 <sub>6</sub>	P5₅	P54	P5₃	P5 <sub>2</sub>	P5,	P5₀
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PDR5 is an 8-bit register that stores data for port 5 pins  $P5_7$  to  $P5_0$ . If port 5 is read while PCR5 bits are set to 1, the values stored in PDR5 are read, regardless of the actual pin states. If port 5 is read while PCR5 bits are cleared to 0, the pin states are read.

Upon reset, PDR5 is initialized to H'00.

# (2) Port Control Register 5 (PCR5)

Bit	7	6	5	4	3	2	1	0
	PCR57	PCR5 <sub>6</sub>	PCR55	PCR5 <sub>4</sub>	PCR53	PCR52	PCR51	PCR50
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PCR5 is an 8-bit register for controlling whether each of the port 5 pins  $P5_7$  to  $P5_0$  functions as an input pin or output pin. Setting a PCR5 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. PCR5 and PDR5 settings are valid when the corresponding pins are designated for general-purpose input/output by PMR5 and bits SGS3 to SGS0 in LPCR.

Upon reset, PCR5 is initialized to H'00.

PCR5 is a write-only register, which is always read as all 1s.

# (3) Port Pull-Up Control Register 5 (PUCR5)

Bit	7	6	5	4	3	2	1	0
	PUCR57	PUCR5 <sub>6</sub>	PUCR55	PUCR54	PUCR53	PUCR52	PUCR51	PUCR50
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PUCR5 controls whether the MOS pull-up of each of port 5 pins  $P5_7$  to  $P5_0$  is on or off. When a PCR5 bit is cleared to 0, setting the corresponding PUCR5 bit to 1 turns on the MOS pull-up for the corresponding pin, while clearing the bit to 0 turns off the MOS pull-up.

Upon reset, PUCR5 is initialized to H'00.

# (4) Port Mode Register 5 (PMR5)

Bit	7	6	5	4	3	2	1	0
	WKP <sub>7</sub>	WKP <sub>6</sub>	$WKP_{5}$	$WKP_4$	$WKP_{3}$	WKP <sub>2</sub>	WKP <sub>1</sub>	WKP₀
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PMR5 is an 8-bit read/write register, controlling the selection of pin functions for port 5 pins.

Upon reset, PMR5 is initialized to H'00.

# Bit n—P5, $\overline{WKP}$ /SEG<sub>n+1</sub> Pin Function Switch (WKPn)

When pin P5n/WKPn/SEGn+1 is not used as  $SEG_{n+1}$ , these bits select whether the pin is used as P5n or  $\overline{WKP}_n$ .

Bit n WKPn	Description	
0	Functions as P5n I/O pin	(initial value)
1	Functions as $\overline{WKP}_n$ input pin	

RENESAS

(n = 7 to 0)

Note: For use as SEG<sub>nu1</sub>, see section 13.2.1, LCD Port Control Register (LPCR).

#### 8.5.3 Pin Functions

Table 8.12 shows the port 5 pin functions.

# Table 8.12Port 5 Pin Functions

Pin	Pin Functions and Selection Method								
P5 <sub>7</sub> /WKP <sub>7</sub> / SEG <sub>8</sub> to P5 <sub>8</sub> /WKP <sub>8</sub> /	The pin function dep in PCR5, and bits S			PMR5, bits PC	$R5_7$ to $PCR5_0$				
SEG,	$P5_7$ to $P5_4$				(n = 7 to 4)				
	SGS3 to SGS0		Other than 0010, 0011, 0100, 0101, 0110, 0111, 1000, 1001						
	WKP <sub>n</sub>	(	0	1	*				
	PCR5	0	1	*	*				
	Pin function	$P5_n$ input pin	P5, output pin	WKPn input pin	SEGn+1 output pin				
	P5 <sub>3</sub> to P5 <sub>0</sub>				(m= 3 to 0)				
	SGS3 to SGS0		001, 0010, 0011 0110, 0111, 100		0001, 0010, 0011, 0100, 0101, 0110, 0111, 1000				
	WKP <sub>m</sub>	(	1	*					
	PCR5 <sub>m</sub>	0	1	*	*				
	Pin function	P5 <sub>m</sub> input pin	P5 <sub>m</sub> output pin	WKPm output	SEGm+1				

RENESAS

\*: Don't care

output pin

pin

#### 8.5.4 Pin States

Table 8.13 shows the port 5 pin states in each operating mode.

#### Table 8.13Port 5 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P5,/WKP,/ SEG <sub>8</sub> to P5,/ WKP,/SEG <sub>1</sub>	High- impedance	Retains previous state		High- impedance*		Functional	Functional

Note: \* A high-level signal is output when the MOS pull-up is in the on state.

# 8.5.5 MOS Input Pull-Up

Port 5 has a built-in MOS input pull-up function that can be controlled by software. When a PCR5 bit is cleared to 0, setting the corresponding PUCR5 bit to 1 turns on the MOS pull-up for that pin. The MOS pull-up function is in the off state after a reset.

PCR5 <sub>n</sub>	0	0	1
PUCR5	0	1	*
MOS input pull-up	Off	On	Off

(n = 7 to 0) \*: Don't care



# 8.6 Port 6

#### 8.6.1 Overview

Port 6 is an 8-bit I/O port. The port 6 pin configuration is shown in figure 8.5.





### 8.6.2 Register Configuration and Description

Table 8.14 shows the port 6 register configuration.

## Table 8.14 Port 6 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 6	PDR6	R/W	H'00	H'FFD9
Port control register 6	PCR6	W	H'00	H'FFE9
Port pull-up control register 6	PUCR6	R/W	H'00	H'FFE3

#### (1) Port Data Register 6 (PDR6)

Bit	7	6	5	4	3	2	1	0
	P67	P6 <sub>6</sub>	P6 <sub>5</sub>	P64	P63	P6 <sub>2</sub>	P61	P60
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PDR6 is an 8-bit register that stores data for port 6 pins  $P6_7$  to  $P6_0$ .

If port 6 is read while PCR6 bits are set to 1, the values stored in PDR6 are read, regardless of the actual pin states. If port 6 is read while PCR6 bits are cleared to 0, the pin states are read.

Upon reset, PDR6 is initialized to H'00.

#### (2) Port Control Register 6 (PCR6)

Bit	7	6	5	4	3	2	1	0
	PCR67	PCR6 <sub>6</sub>	PCR6 <sub>5</sub>	PCR6 <sub>4</sub>	PCR6 <sub>3</sub>	PCR6 <sub>2</sub>	PCR6 <sub>1</sub>	PCR60
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PCR6 is an 8-bit register for controlling whether each of the port 6 pins  $P6_7$  to  $P6_0$  functions as an input pin or output pin.

Setting a PCR6 bit to 1 makes the corresponding pin  $(P6_7 \text{ to } P6_0)$  an output pin, while clearing the bit to 0 makes the pin an input pin. PCR6 and PDR6 settings are valid when the corresponding pins are designated for general-purpose input/output by bits SGS3 to SGS0 in LPCR.

Upon reset, PCR6 is initialized to H'00.

PCR6 is a write-only register, which is always read as all 1s.

### (3) Port Pull-Up Control Register 6 (PUCR6)

Bit	7	6	5	4	3	2	1	0
	PUCR67	PUCR66	PUCR65	PUCR64	PUCR63	PUCR6 <sub>2</sub>	PUCR61	PUCR60
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PUCR6 controls whether the MOS pull-up of each of the port 6 pins  $P6_7$  to  $P6_0$  is on or off. When a PCR6 bit is cleared to 0, setting the corresponding PUCR6 bit to 1 turns on the MOS pull-up for the corresponding pin, while clearing the bit to 0 turns off the MOS pull-up.

Upon reset, PUCR6 is initialized to H'00.

#### 8.6.3 Pin Functions

Table 8.15 shows the port 6 pin functions.

#### Table 8.15 Port 6 Pin Functions

#### Pin Pin Functions and Selection Method

 $\begin{array}{lll} \mathsf{P6}_7\!/\mathsf{SEG}_{_{16}} \text{ to } & \mathsf{The \ pin \ function \ depends \ on \ bits \ } \mathsf{PCR6}_7 \ to \ \mathsf{PCR6}_0 \ in \ \mathsf{PCR6} \ and \ bits \ \mathsf{SGS3} \ to \\ \mathsf{P6}_0\!/\mathsf{SEG}_9 & \mathsf{SGS0 \ in \ } \mathsf{LPCR}. \end{array}$ 

 $P6_7$  to  $P6_4$ 

(n = 7 to 4)

SGS3 to SGS0	Other than 0100, 1000, 1001,	0100, 0101, 0110, 0111, 1000, 1001, 1010, 1011	
PCR6 <sub>n</sub>	0	1	*
Pin function	P6, input pin	P6, output pin	$SEG_{n+9}$ output pin

#### P6, to P6

(m = 3 to 0)

$10_{3}1010_{0}$			(11 = 0.10.0)
SGS3 to SGS0	Other than 0011, 0111, 1000,	0011, 0100, 0101, 0110, 0111, 1000, 1001, 1010	
PCR6 <sub>m</sub>	0	1	*
Pin function	P6 <sub>m</sub> input pin	P6 <sub>m</sub> output pin	$SEG_{{}_{m+9}}  output  pin$

\*: Don't care

#### 8.6.4 Pin States

Table 8.16 shows the port 6 pin states in each operating mode.

#### Table 8.16Port 6 Pin States

Pin	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
$P6_7/SEG_{16}$ to $P6_0/SEG_9$	High- impedance	Retains previous state		High- impedance <sup>*</sup>		Functional	Functional

Note: \* A high-level signal is output when the MOS pull-up is in the on state.

# 8.6.5 MOS Input Pull-Up

Port 6 has a built-in MOS pull-up function that can be controlled by software. When a PCR6 bit is cleared to 0, setting the corresponding PUCR6 bit to 1 turns on the MOS pull-up for that pin. The MOS pull-up function is in the off state after a reset.

PCR6 <sub>n</sub>	0	0	1	
PUCR6 <sub>n</sub>	0	1	*	
MOS input pull-up	Off	On	Off	

(n = 7 to 0) \*: Don't care



Section 8 I/O Ports

# 8.7 Port 7

#### 8.7.1 Overview

Port 7 is an 8-bit I/O port, configured as shown in figure 8.6.





## 8.7.2 Register Configuration and Description

Table 8.17 shows the port 7 register configuration.

## Table 8.17 Port 7 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 7	PDR7	R/W	H'00	H'FFDA
Port control register 7	PCR7	W	H'00	H'FFEA

#### (1) Port Data Register 7 (PDR7)

Bit	7	6	5	4	3	2	1	0
	P77	P76	P75	P74	P73	P72	P7 <sub>1</sub>	P70
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W						

PDR7 is an 8-bit register that stores data for port 7 pins  $P7_7$  to  $P7_0$ . If port 7 is read while PCR7 bits are set to 1, the values stored in PDR7 are read, regardless of the actual pin states. If port 7 is read while PCR7 bits are cleared to 0, the pin states are read.

Upon reset, PDR7 is initialized to H'00.

### (2) Port Control Register 7 (PCR7)

Bit	7	6	5	4	3	2	1	0
	PCR77	PCR7 <sub>6</sub>	PCR75	PCR7 <sub>4</sub>	PCR7 <sub>3</sub>	PCR7 <sub>2</sub>	PCR71	PCR70
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PCR7 is an 8-bit register for controlling whether each of the port 7 pins  $P7_7$  to  $P7_0$  functions as an input pin or output pin. Setting a PCR7 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. PCR7 and PDR7 settings are valid when the corresponding pins are designated for general-purpose input/output by bits SGS3 to SGS0 in LPCR.

Upon reset, PCR7 is initialized to H'00.

PCR7 is a write-only register, which is always read as all 1s.

#### 8.7.3 Pin Functions

Table 8.18 shows the port 7 pin functions.

## Table 8.18Port 7 Pin Functions

Pin	Pin Functions and Selection Method						
P7 <sub>7</sub> /SEG <sub>24</sub> to P7 <sub>0</sub> /SEG <sub>17</sub>	The pin function depends on bits $PCR7_{7}$ to $PCR7_{0}$ in PCR7 and bits SGS3 to SGS0 in LPCR.						
	$P7_7$ to $P7_4$	(n = 7 to 4)					
	SGS3 to SGS0	Other than 0110, 1010, 1011,	0110, 0111, 1000, 1001, 1010, 1011, 1100, 1101				
	PCR7 <sub>n</sub>	0	1	*			
	Pin function	P7, input pin	P7 <sub>n</sub> output pin	SEG <sub>n+17</sub> output pin			
	$P7_{3}$ to $P7_{0}$	(m = 3 to 0)					
	SGS3 to SGS0	Other than 0101, 1001, 1010,	0101, 0110, 0111, 1000, 1001, 1010, 1011, 1100				
	PCR7 <sub>m</sub>	0	1	*			
	Pin function	P7 <sub>m</sub> input pin	P7 <sub>m</sub> output pin	SEG <sub>m+17</sub> output pin			

\*: Don't care

# 8.7.4 Pin States

Table 8.19 shows the port 7 pin states in each operating mode.

# Table 8.19Port 7 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
$P7_7/SEG_{24}$ to $P7_0/SEG_{17}$	High- impedance	Retains previous state		High- impedance		Functional	Functional
# 8.8 Port 8

### 8.8.1 Overview

Port 8 is an 8-bit I/O port configured as shown in figure 8.7.





# 8.8.2 Register Configuration and Description

Table 8.20 shows the port 8 register configuration.

### Table 8.20 Port 8 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 8	PDR8	R/W	H'00	H'FFDB
Port control register 8	PCR8	W	H'00	H'FFEB



### (1) Port Data Register 8 (PDR8)

Bit	7	6	5	4	3	2	1	0
	P87	P8 <sub>6</sub>	P8 <sub>5</sub>	P8 <sub>4</sub>	P8 <sub>3</sub>	P82	P8 <sub>1</sub>	P80
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

PDR8 is an 8-bit register that stores data for port 8 pins  $P8_7$  to  $P8_0$ . If port 8 is read while PCR8 bits are set to 1, the values stored in PDR8 are read, regardless of the actual pin states. If port 8 is read while PCR8 bits are cleared to 0, the pin states are read.

Upon reset, PDR8 is initialized to H'00.

# (2) Port Control Register 8 (PCR8)

Bit	7	6	5	4	3	2	1	0
	PCR87	PCR8 <sub>6</sub>	PCR85	PCR8 <sub>4</sub>	PCR83	PCR8 <sub>2</sub>	PCR8 <sub>1</sub>	PCR80
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

PCR8 is an 8-bit register for controlling whether the port 8 pins  $P8_7$  to  $P8_0$  functions as an input or output pin. Setting a PCR8 bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. PCR8 and PDR8 settings are valid when the corresponding pins are designated for general-purpose input/output by bits SGS3 to SGS0 in LPCR.

Upon reset, PCR8 is initialized to H'00.

PCR8 is a write-only register, which is always read as all 1s.

### 8.8.3 **Pin Functions**

Table 8.21 shows the port 8 pin functions.

### Table 8.21Port 8 Pin Functions

Pin	Pin Functions and	d Selection Method		
P8 <sub>7</sub> /SEG <sub>32</sub> to	The pin function de in LPCR.	epends on bits PCR8,	to PCR8₀ in PCR8 and	I bits SGS3 to SGS0
P8 <sub>0</sub> /SEG <sub>25</sub>	$P8_7$ to $P8_4$			(n = 7 to 4)
	SGS3 to SGS0	Other than 1000, 100 1101, 11	1000, 1001, 1010, 1011, 1100, 1101, 1110, 1111	
	PCR8 <sub>n</sub>	0 1		*
	Pin function	P8, input pin P8, output pin		SEG <sub>n+25</sub> output pin
	P8₃ to P8₀			(m = 3 to 0)
	SGS3 to SGS0	Other than 0111, 100 1100, 11	0111, 1000, 1001, 1010, 1011, 1100, 1101, 1110	
	PCR8 <sub>m</sub>	0	1	*

P8<sub>m</sub> input pin

P8<sub>m</sub> output pin

\*: Don't care

 $\mathsf{SEG}_{{}_{\mathsf{m+25}}}$  output pin

## 8.8.4 Pin States

Table 8.22 shows the port 8 pin states in each operating mode.

Pin function

### Table 8.22Port 8 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
$P8_7/SEG_{32}$ to $P8_0/SEG_{25}$	High- impedance	Retains previous state		High- impedance	Retains previous state	Functional	Functional

Section 8 I/O Ports

# 8.9 Port 9

### 8.9.1 Overview

Port 9 is a 6-bit output port, configured as shown in figure 8.8.





# 8.9.2 Register Configuration and Description

Table 8.23 shows the port 9 register configuration.

# Table 8.23 Port 9 Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register 9	PDR9	R/W	H'FF	H'FFDC
Port mode register 9	PMR9	R/W	_	H'FFEC

### (1) Port Data Register 9 (PDR9)

Bit	7	6	5	4	3	2	1	0
	_	—	P9 <sub>5</sub>	P94	P93	P9 <sub>2</sub>	P9 <sub>1</sub>	P90
Initial value	1	1	1	1	1	1	1	1
Read/Write	—	—	R/W	R/W	R/W	R/W	R/W	R/W

PDR9 is an 8-bit register that stores data for port 9 pins P9<sub>5</sub> to P9<sub>0</sub>.

Upon reset, PDR9 is initialized to H'FF.

# (2) Port Mode Register 9 (PMR9)

Bit	7	6	5	4	3	2	1	0
							PWM <sub>2</sub>	PWM <sub>1</sub>
Initial value	1	1	1	1	0		0	0
Read/Write	_				R/W	W	R/W	R/W

PMR9 is an 8-bit read/write register controlling the selection of the  $P9_0$  and  $P9_1$  pin functions.

### Bit 3—Reserved

This bit is reserved; it is readable/writable.

### Bit 2—Reserved

This bit is reserved; it can only be written with 0.

### Bits 1 and 0—P9<sub>n</sub>/PWM Pin Function Switches

These pins select whether pin P9n/PWMn+1 is used as P9n or as PWMn+1.

Bit n PWMn+1	Description	
0	Functions as P9, output pin	(initial value)
1	Functions as PWM <sub>n+1</sub> output pin	

(n = 0 or 1)



### 8.9.3 Pin Functions

Table 8.24 shows the port 9 pin functions.

### Table 8.24Port 9 Pin Functions

### Pin

### Pin Functions and Selection Method

# P9<sub>3</sub>/V<sub>ref</sub>\*

VREFSEL	0	1
Pin function	P9 <sub>3</sub> output pin	V <sub>ref</sub> input pin

P9 <sub>1</sub> /PWM <sub>n+1</sub> to			(n = 1 or 0)
P9/PWM	PMR9 <sub>n</sub>	0	1
	Pin function	P9 <sup>,</sup> output pin	PWM <sub>n+1</sub> output pin

Note: \* The V<sub>ref</sub> pin is the input pin for the LVD's external reference voltage.

### 8.9.4 Pin States

Table 8.25 shows the port 9 pin states in each operating mode.

### Table 8.25Port 9 Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P9₅ to P9₂ P9 <sub>n</sub> /PWM <sub>n+1</sub> to P9 <sub>n</sub> /PWM <sub>n+1</sub>	High- impedance	Retains previous state		High- impedance	Retains previous state	Functional	Functional

(n = 1 or 0)

# 8.10 Port A

### 8.10.1 Overview

Port A is a 4-bit I/O port, configured as shown in figure 8.9.





### 8.10.2 Register Configuration and Description

Table 8.26 shows the port A register configuration.

### Table 8.26 Port A Registers

Name	Abbr.	R/W	Initial Value	Address
Port data register A	PDRA	R/W	H'F0	H'FFDD
Port control register A	PCRA	W	H'F0	H'FFED

### (1) Port Data Register A (PDRA)

Bit	7	6	5	4	3	2	1	0
			_		PA <sub>3</sub>	PA <sub>2</sub>	PA <sub>1</sub>	PA <sub>0</sub>
Initial value	1	1	1	1	0	0	0	0
Read/Write		_			R/W	R/W	R/W	R/W

PDRA is an 8-bit register that stores data for port A pins  $PA_3$  to  $PA_0$ . If port A is read while PCRA bits are set to 1, the values stored in PDRA are read, regardless of the actual pin states. If port A is read while PCRA bits are cleared to 0, the pin states are read.

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Upon reset, PDRA is initialized to H'FO.

### (2) Port Control Register A (PCRA)

Bit	7	6	5	4	3	2	1	0
	—			—	PCRA <sub>3</sub>	$PCRA_2$	PCRA <sub>1</sub>	PCRA <sub>0</sub>
Initial value	1	1	1	1	0	0	0	0
Read/Write	_	_			W	W	W	W

PCRA controls whether each of port A pins  $PA_3$  to  $PA_0$  functions as an input pin or output pin. Setting a PCRA bit to 1 makes the corresponding pin an output pin, while clearing the bit to 0 makes the pin an input pin. PCRA and PDRA settings are valid when the corresponding pins are designated for general-purpose input/output by LPCR.

Upon reset, PCRA is initialized to H'F0.

PCRA is a write-only register, which is always read as all 1s.



### 8.10.3 **Pin Functions**

Table 8.27 shows the port A pin functions.

### Table 8.27Port A Pin Functions

Pin	Pin Functions and	Selection Method						
PA <sub>3</sub> /COM <sub>4</sub>	The pin function dep	The pin function depends on bit $PCRA_3$ in PCRA and bits SGS3 to SGS0.						
	SGS3 to SGS0	0000	0000	Not 0000				
	PCRA <sub>3</sub>	0	1	*				
	Pin function	$PA_{_3}$ input pin	PA <sub>3</sub> output pin	COM <sub>4</sub> output pin				
PA <sub>2</sub> /COM <sub>3</sub>	The pin function dep	ends on bit PCRA <sub>2</sub>	in PCRA and bits SG	GS3 to SGS0.				
	SGS3 to SGS0	0000	0000	Not 0000				
	PCRA <sub>2</sub>	0 1		*				
	Pin function	PA <sub>2</sub> input pin	PA <sub>2</sub> output pin	COM <sub>3</sub> output pin				
PA <sub>1</sub> /COM <sub>2</sub>	The pin function dep	The pin function depends on bit PCRA, in PCRA and bits SGS3 to SGS0.						
	SGS3 to SGS0	0000 0000		Not 0000				
	PCRA,	0	1	*				
	Pin function	PA, input pin	PA <sub>1</sub> output pin	COM <sub>2</sub> output pin				
PA <sub>0</sub> /COM <sub>1</sub>	The pin function dep	ends on bit PCRA	in PCRA and bits SG	S3 to SGS0.				
	SGS3 to SGS0	00	000	Not 0000				
	PCRA <sub>0</sub>	0	1	*				
	Pin function	$PA_0$ input pin	PA₀ input pin PA₀ output pin					
	L							

\*: Don't care



### 8.10.4 Pin States

Table 8.28 shows the port A pin states in each operating mode.

Table 8.28Port A Pin States

Pins	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
PA <sub>3</sub> /COM <sub>4</sub> PA <sub>2</sub> /COM <sub>3</sub> PA <sub>1</sub> /COM <sub>2</sub> PA <sub>0</sub> /COM <sub>1</sub>	High- impedance	Retains previous state		High- impedance	Retains previous state	Functional	Functional

# 8.11 Port B

### 8.11.1 Overview

Port B is an 8-bit input-only port, configured as shown in figure 8.10.



Figure 8.10 Port B Pin Configuration

### 8.11.2 Register Configuration and Description

Table 8.29 shows the port B register configuration.

### Table 8.29Port B Register

Name	Abbr.	R/W	Initial Value	Address
Port data register B	PDRB	R	—	H'FFDE
Port mode register B	PMRB	R/W	H'F7	H'FFEE

### (1) Port Data Register B (PDRB)

Bit	7	6	5	4	3	2	1	0
	PB <sub>7</sub>	$PB_6$	$PB_5$	PB <sub>4</sub>	PB <sub>3</sub>	PB <sub>2</sub>	PB <sub>1</sub>	PB <sub>0</sub>
	,							
Read/Write	R	R	R	R	R	R	R	R

Reading PDRB always gives the pin states. However, if a port B pin is selected as an analog input channel for the A/D converter by AMR bits CH3 to CH0, that pin reads 0 regardless of the input voltage.

### (2) Port Mode Register B (PMRB)

Bit	7	6	5	4	3	2	1	0	_
					IRQ1				
Initial value	1	1	1	1	0	1	1	1	
Read/Write					R/W				

PMRB is an 8-bit read/write register controlling the selection of the  $PB_3$  pin function. Upon reset, PMRB is initialized to H'F7.

### Bits 7 to 4 and 2 to 0—Reserved

Bits 7 to 4 and 2 to 0 are reserved; they are always read as 1 and cannot be modified.



# Bit 3—PB<sub>3</sub>/AN<sub>3</sub>/IRQ<sub>1</sub> Pin Function Switch (IRQ1)

These bits select whether pin PB<sub>3</sub>/AN<sub>3</sub>/ $\overline{IRQ}_1$  is used as PB<sub>3</sub>/AN<sub>3</sub> or as  $\overline{IRQ}_1$ /TMIC.

Bit 3 IRQ1	Description	
0	Functions as $PB_g/AN_s$ input pin	(initial value)
1	Functions as IRQ,/TMIC input pin	
Note:	Rising or falling edge sensing can be selected for the IRQ,/TMIC pin.	

For TMIC pin setting information, see the Timer Mode Register C (TMC) description in section 9.3.2, Register Descriptions.



# 8.11.3 Pin Functions

Table 8.30 shows the port B pin functions.

# Table 8.30 Port B Pin Functions

The pin function dep					
The pin function depends on bits CH3 to CH0 in AMR.					
CH3 to CH0	Not	1011		10	11
Pin function	PB <sub>7</sub> in	PB <sub>7</sub> input pin			out pin
The pin function dep	ends on bits C	H3 to CH0 in Al	MR.		
CH3 to CH0	Not	1010		10	10
Pin function	PB <sub>6</sub> in	put pin	А	N <sub>6</sub> in	out pin
The pin function dep	ends on bits C	H3 to CH0 in Al	MR.		
CH3 to CH0	Not	1001		10	01
Pin function	PB₅ in	put pin	AN₅ input pin		
The pin function depends on bits CH3 to CH0 in AMR.					
CH3 to CH0	Not		1000		
Pin function	$PB_4$ in	put pin	AN₄ input pin		
		H3 to CH0 in Al	VR and bit	IRQ	1 in PMRB and
IRQ,	(	)		1	
CH3 to CH0	Not 0111	0111		*	¢
TMC2 to TMC0	:	k	Not 11	1	111
Pin function	$PB_{_3}$ input pin	$AN_3$ input pin	IRQ₁ inpu	t pin	TMIC input pin
•		TMIC input pir	n, clear IEN	N1 to	0 in IENR1 to
The pin function dep	ends on bits C	H3 to CH0 in Al	MR.		
CH3 to CH0		Not 0110			0110
Pin function	P	B <sub>2</sub> input pin		A	N <sub>2</sub> input pin
	Pin function The pin function dep CH3 to CH0 Pin function The pin function dep CH3 to CH0 Pin function The pin function dep CH3 to CH0 Pin function The pin function dep bits TMC2 to TMC0 IRQ, CH3 to CH0 TMC2 to TMC0 Pin function Note: When this pin disable the I The pin function dep CH3 to CH0	Pin functionPB, inThe pin function depends on bits CCH3 to CH0NotPin functionPB, inThe pin function depends on bits CCH3 to CH0NotPin functionPB, inThe pin function depends on bits CCH3 to CH0NotPin function depends on bits CCH3 to CH0NotPin function depends on bits CCH3 to CH0NotPin function depends on bits Cbits TMC2 to TMC0 in TMC.IRQ,CCH3 to CH0Not 0111TMC2 to TMC02Pin functionPB, input pinNote:When this pin is used as the disable the IRQ1 interrupt.The pin function depends on bits CCH3 to CH0O	Pin functionPB, input pinThe pin function depends on bits CH3 to CH0 in AlCH3 to CH0Not 1010Pin functionPB, input pinThe pin function depends on bits CH3 to CH0 in AlCH3 to CH0Not 1001Pin functionPB, input pinThe pin function depends on bits CH3 to CH0 in AlCH3 to CH0Not 1001Pin functionPB, input pinThe pin function depends on bits CH3 to CH0 in AlCH3 to CH0Not 1000Pin functionPB, input pinThe pin function depends on bits CH3 to CH0 in AlDits TMC2 to TMC0 in TMC.IRQ,0CH3 to CH0Not 0111OCH3 to CH0Note:When this pin is used as the TMIC input pinNote:When this pin is used as the TMIC input pinNote:When this pin is used as the CH3 to CH0 in AlCH3 to CH0Not 0110Not 0110Not 0110	Pin functionPB, input pinAThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 1010Pin functionPB, input pinAThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 1001Pin functionPB, input pinAThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 1001Pin functionPB, input pinAThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 1000Pin functionPB, input pinAThe pin function depends on bits CH3 to CH0 in AMR and bitsbits TMC2 to TMC0IRQ,0CH3 to CH0Not 01110CH3 to CH0CH3 to CH0Not 01110IRQ,0Not 01110Not 01110RQ, input pinNote:When this pin is used as the TMIC input pin, clear IEI disable the IRQ1 interrupt.The pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 0110	Pin functionPB, input pinAN, inputThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 101010Pin functionPB, input pinAN, inputThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 100110Pin functionPB, input pinAN, inputThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 100110Pin functionPB, input pinAN, inputThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 100010Pin functionPB, input pinAN, inputThe pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 100010Pin function depends on bits CH3 to CH0 in AMR and bit IRQ: bits TMC2 to TMC0 in TMC.IRQ,IRQ,011Pin functionPB, input pinAN, input pinTMC2 to TMC0*Not 111Pin functionPB, input pinIRQ, input pinNote:When this pin is used as the TMIC input pin, clear IEN1 to disable the IRQ1 interrupt.IRQ input pin, clear IEN1 to disable the IRQ1 interrupt.The pin function depends on bits CH3 to CH0 in AMR.CH3 to CH0Not 0110



Pin	Pin Functions and Selection Method						
PB <sub>1</sub> /AN <sub>1</sub> /extU	Switching is accomplished by combining CH3 to CH0 in AMR and VINTUSEL in LVDCR as shown below.						
	VINTUSEL	C	)	1			
	CH3 to CH0	Not B'0101	B'0101	*			
	Pin function	PB, input pin	AN, input pin	extU input pin			
PB <sub>0</sub> /AN <sub>0</sub> /extD	Switching is accomp LVDCR as shown be	, ,	CH3 to CH0 in AMF	R and VINTDSEL in			
	VINTDSEL	C	)	1			
	CH3 to CH0	Not B'0100	B'0100	*			
	Pin function	PB <sub>o</sub> input pin AN <sub>o</sub> input pin extD input pin					
. <u></u>				*: Don't caro			

### \*: Don't care

# 8.12 Input/Output Data Inversion Function

### 8.12.1 Overview

With input pin  $RXD_{32}$  and output pin  $TXD_{32}$ , the data can be handled in inverted form.



Figure 8.11 Input/Output Data Inversion Function

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### 8.12.2 **Register Configuration and Descriptions**

Table 8.31 shows the registers used by the input/output data inversion function.

### Table 8.31 Register Configuration

Name	Abbr.	R/W	Address
Serial port control register	SPCR	R/W	H'FF91

### (1) Serial Port Control Register (SPCR)

Bit	7	6	5	4	3	2	1	0
			SPC32		SCINV3	SCINV2		
Initial value	1	1	0		0	0		_
Read/Write	_	_	R/W	W	R/W	R/W	W	W

SPCR is an 8-bit readable/writable register that performs RXD<sub>32</sub> and TXD<sub>32</sub> pin input/output data inversion switching.

### Bits 7 and 6—Reserved

Bits 7 and 6 are reserved; they are always read as 1 and cannot be modified.

### Bit 5—P4<sub>2</sub>/TXD<sub>32</sub> Pin Function Switch (SPC32)

This bit selects whether pin P4<sub>2</sub>/TXD<sub>32</sub> is used as P4<sub>2</sub> or as TXD<sub>32</sub>.

Bit 5 SPC32	Description	
0	Functions as $P4_{_2}$ I/O pin	(initial value)
1	Functions as $TXD_{_{32}}$ output pin*	
Note: * S	Set the TE bit in SCB3 after setting this bit to 1	

Set the TE bit in SCR3 after setting this bit to 1. inole.

# Bit 4—Reserved

Bit 4 is reserved; it can only be written with 0.



# Bit 3—TXD<sub>32</sub> Pin Output Data Inversion Switch

Bit 3 specifies whether or not  $TXD_{32}$  pin output data is to be inverted.

Bit 3 SCINV3	Description	
0	$TXD_{_{32}}$ output data is not inverted	(initial value)
1	TXD <sub>32</sub> output data is inverted	

### Bit 2—RXD<sub>32</sub> Pin Input Data Inversion Switch

Bit 2 specifies whether or not  $RXD_{32}$  pin input data is to be inverted.

Bit 2 SCINV2	Description	
0	RXD <sub>32</sub> input data is not inverted	(initial value)
1	RXD <sub>32</sub> input data is inverted	

### Bits 1 and 0—Reserved

Bits 1 and 0 are reserved; they can only be written with 0.

# 8.12.3 Note on Modification of Serial Port Control Register

When a serial port control register is modified, the data being input or output up to that point is inverted immediately after the modification, and an invalid data change is input or output. When modifying a serial port control register, do so in a state in which data changes are invalidated.



# 8.13 Application Note

# 8.13.1 The Management of the Un-Use Terminal

If an I/O pin not used by the user system is floating, pull it up or down.

- If an unused pin is an input pin, handle it in one of the following ways:
  - Pull it up to  $V_{\rm cc}$  with an on-chip pull-up MOS.
  - Pull it up to  $V_{\rm cc}$  with an external resistor of approximately 100 kΩ.
  - Pull it down to  $V_{ss}$  with an external resistor of approximately 100 k $\Omega$ .
  - For a pin also used by the A/D converter, pull it up to  $\mathrm{AV}_{\mathrm{cc}}$
- If an unused pin is an output pin, handle it in one of the following ways:
  - Set the output of the unused pin to high and pull it up to  $V_{cc}$  with an on-chip pull-up MOS.
  - Set the output of the unused pin to high and pull it up to  $V_{\rm cc}$  with an external resistor of approximately 100 kΩ.
  - Set the output of the unused pin to low and pull it down to GND with an external resistor of approximately 100 k $\Omega$ .





# Section 9 Timers

# 9.1 Overview

This LSI provides six timers: timers A, C, F, G, and a watchdog timer, and an asynchronous event counter. The functions of these timers are outlined in table 9.1.

Table 9.1	Timer	Functions
-----------	-------	-----------

Name	Fu	Inctions	Internal Clock	Event Input Pin	Waveform Output Pin	Remarks
Timer A	•	8-bit timer	φ/8 to φ/8192	_	_	
	•	Interval function	(8 choices)			
	•	Time base	$\phi_W$ /128 (choice of 4 overflow periods)	_		
Timer C	•	8-bit timer	$\varphi/4$ to $\varphi/8192, \varphi_w/4$	TMIC	_	Up-count/
	•	Interval function	(7 choices)			down-count controllable by
	•	Event counting function				software or
	•	Up-count/down-count selectable				hardware
Timer F	٠	16-bit timer	φ/4 to φ/32, φ <sub>W</sub> /4	TMIF	TMOFL	
	•	Event counting function	(4 choices)		TMOFH	
	•	Also usable as two independent 8-bit timers				
	•	Output compare output function				
Timer G	٠	8-bit timer	$\phi/2$ to $\phi/64$ , $\phi_w/4$	TMIG	_	Counter
	٠	Input capture function	(4 choices)			clearing option
	•	Interval function				Built-in capture input signal noise canceler
Watchdog timer	•	Reset signal generated when 8-bit counter overflows	φ/64 to φ/8192 φ <sub>w</sub> /32 On-chip oscillator	_	_	



Name	Fu	nctions	Internal Clock	Event Input Pin	Waveform Output Pin	Remarks
Asynchro-	•	16-bit counter	φ/2 to φ/8	AEVL	_	
nous event counter	•	Also usable as two independent 8-bit counters	(3 choices)	AEVH IRQAEC		
	•	Counts events asynchronous to $\phi$ and $\phi w$				
	•	Can count asynchronous events (rising/falling/both edges) independ-ently of the MCU's internal clock				

# 9.2 Timer A

### 9.2.1 Overview

Timer A is an 8-bit timer with interval timing and real-time clock time-base functions. The clock time-base function is available when a 32.768 kHz crystal resonator is connected as the subclock.

### (1) Features

Features of timer A are given below.

- Choice of eight internal clock sources (φ/8192, φ/4096, φ/2048, φ/512, φ/256, φ/128, φ/32, φ/8).
- Choice of four overflow periods (1 s, 0.5 s, 0.25 s, 31.25 ms) when timer A is used as a clock time base (using a 32.768 kHz crystal resonator is connected as the subclock).
- An interrupt is requested when the counter overflows.
- Use of module standby mode enables this module to be placed in standby mode independently when not used.

# (2) Block Diagram

Figure 9.1 shows a block diagram of timer A.



Figure 9.1 Block Diagram of Timer A



# (3) Register Configuration

Table 9.2 shows the register configuration of timer A.

### Table 9.2Timer A Registers

Name	Abbr.	R/W	Initial Value	Address
Timer mode register A	TMA	R/W	—	H'FFB0
Timer counter A	TCA	R	H'00	H'FFB1
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA

### 9.2.2 Register Descriptions

### (1) Timer Mode Register A (TMA)

Bit	7	6	5	4	3	2	1	0
	_	—			TMA3	TMA2	TMA1	TMA0
Initial value				1	0	0	0	0
Read/Write	W	W	W		R/W	R/W	R/W	R/W

TMA is an 8-bit read/write register for selecting the prescaler, and input clock.

### Bits 7 to 5—Reserved

Bits 7 to 5 are reserved; only 0 can be written to these bits.

### Bit 4—Reserved

Bit 4 is reserved; it is always read as 1, and cannot be modified.



# Bits 3 to 0—Internal Clock Select (TMA3 to TMA0)

Bits 3 to 0 select the clock input to TCA. The selection is made as follows.

				Description	
Bit 3 TMA3	Bit 2 TMA2	Bit 1 TMA1	Bit 0 TMA0	Prescaler and Divider Ratio or Overflow Period	Function
0	0	0	0	PSS, φ/8192 (initial value)	Interval timer
			1	PSS,	_
		1	0	PSS,	_
			1	PSS, φ/512	_
	1	0	0	PSS,	_
			1	PSS,	_
		1	0	PSS,	_
			1	PSS, φ/8	_
1	0	0	0	PSW, 1 s	Clock time
			1	PSW, 0.5 s	base
		1	0	PSW, 0.25 s	(when using
			1	PSW, 0.03125 s	32.768 kHz)
	1	0	0	PSW and TCA are reset	_
			1	_	
		1	0	—	
			1		



### (2) Timer Counter A (TCA)

Bit	7	6	5	4	3	2	1	0
	TCA7	TCA6	TCA5	TCA4	TCA3	TCA2	TCA1	TCA0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

TCA is an 8-bit read-only up-counter, which is incremented by internal clock input. The clock source for input to this counter is selected by bits TMA3 to TMA0 in timer mode register A (TMA). TCA values can be read by the CPU in active mode, but cannot be read in subactive mode. When TCA overflows, the IRRTA bit in interrupt request register 1 (IRR1) is set to 1.

TCA is cleared by setting bits TMA3 and TMA2 of TMA to 11.

Upon reset, TCA is initialized to H'00.

## (3) Clock Stop Register 1 (CKSTPR1)

Bit:	7	6	5	4	3	2	1	0
	—		S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value:	1	1	1	1	1	1	1	1
Read/Write:			R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to timer A is described here. For details of the other bits, see the sections on the relevant modules.

### Bit 0—Timer A Module Standby Mode Control (TACKSTP)

Bit 0 controls setting and clearing of module standby mode for timer A.

TACKSTP	Description	
0	Timer A is set to module standby mode	
1	Timer A module standby mode is cleared	(initial value)

# 9.2.3 Timer Operation

# (1) Interval Timer Operation

When bit TMA3 in timer mode register A (TMA) is cleared to 0, timer A functions as an 8-bit interval timer.

Upon reset, TCA is cleared to H'00 and bit TMA3 is cleared to 0, so up-counting and interval timing resume immediately. The clock input to timer A is selected by bits TMA2 to TMA0 in TMA; any of eight internal clock signals output by prescaler S can be selected.

After the count value in TCA reaches H'FF, the next clock signal input causes timer A to overflow, setting bit IRRTA to 1 in interrupt request register 1 (IRR1). If IENTA = 1 in interrupt enable register 1 (IENR1), a CPU interrupt is requested.\*

At overflow, TCA returns to H'00 and starts counting up again. In this mode timer A functions as an interval timer that generates an overflow output at intervals of 256 input clock pulses.

Note: \* For details on interrupts, see section 3.3, Interrupts.

### (2) Real-Time Clock Time Base Operation

When bit TMA3 in TMA is set to 1, timer A functions as a real-time clock time base by counting clock signals output by prescaler W. The overflow period of timer A is set by bits TMA1 and TMA0 in TMA. A choice of four periods is available. In time base operation (TMA3 = 1), setting bit TMA2 to 1 clears both TCA and prescaler W to their initial values of H'00.



# 9.2.4 Timer A Operation States

Table 9.3 summarizes the timer A operation states.

### Table 9.3 Timer A Operation States

Oper	ation Mode	Reset	Active	Sleep	Watch	Sub- active	Sub- sleep	Standby	Module Standby
TCA	Interval	Reset	Functions	Functions	Halted	Halted	Halted	Halted	Halted
	Clock time base	Reset	Functions	Functions	Functions	Functions	Functions	Halted	Halted
ТМА		Reset	Functions	Retained	Retained	Functions	Retained	Retained	Retained

Note: When the real-time clock time base function is selected as the internal clock of TCA in active mode or sleep mode, the internal clock is not synchronous with the system clock, so it is synchronized by a synchronizing circuit. This may result in a maximum error of  $1/\phi$  (s) in the count cycle.

### 9.2.5 Application Note

When bit 0 (TACKSTP) of the clock stop register 1 (CKSTPR1) is cleared to 0, bit 3 (TMA3) of the timer mode register A (TMA) cannot be rewritten.

Set bit 0 (TACKSTP) of the clock stop register 1 (CKSTPR1) to 1 before rewriting bit 3 (TMA3) of the timer mode register A (TMA).



# 9.3 Timer C

### 9.3.1 Overview

Timer C is an 8-bit timer that increments or decrements each time a clock pulse is input. This timer has two operation modes, interval and auto reload.

# (1) Features

Features of timer C are given below.

- Choice of seven internal clock sources (φ/8192, φ/2048, φ/512, φ/64, φ/16, φ/4, φ<sub>w</sub>/4) or an external clock (can be used to count external events).
- An interrupt is requested when the counter overflows.
- Up/down-counter switching is possible by hardware or software.
- Subactive mode or subsleep mode operation is possible when  $\phi_w/4$  is selected as the internal clock, or when an external clock is selected.
- Use of module standby mode enables this module to be placed in standby mode independently when not used.



### (2) Block Diagram

Figure 9.2 shows a block diagram of timer C.





# (3) Pin Configuration

Table 9.4 shows the timer C pin configuration.

### Table 9.4Pin Configuration

Name	Abbr.	I/O	Function
Timer C event input	TMIC	Input	Input pin for event input to TCC
Timer C up/down select	UD	Input	Timer C up/down-count selection

# (4) Register Configuration

Table 9.5 shows the register configuration of timer C.

### Table 9.5Timer C Registers

Name	Abbr.	R/W	Initial Value	Address
Timer mode register C	TMC	R/W	H'18	H'FFB4
Timer counter C	TCC	R	H'00	H'FFB5
Timer load register C	TLC	W	H'00	H'FFB5
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA

### 9.3.2 Register Descriptions

### (1) Timer Mode Register C (TMC)

Bit	7	6	5	4	3	2	1	0
	TMC7	TMC6	TMC5			TMC2	TMC1	TMC0
Initial value	0	0	0	1	1	0	0	0
Read/Write	R/W	R/W	R/W			R/W	R/W	R/W

TMC is an 8-bit read/write register for selecting the auto-reload function and input clock, and performing up/down-counter control.

Upon reset, TMC is initialized to H'18.

### Bit 7—Auto-Reload Function Select (TMC7)

Bit 7 selects whether timer C is used as an interval timer or auto-reload timer.

Bit 7 TMC7	Description	
0	Interval timer function selected	(initial value)
1	Auto-reload function selected	



# Bits 6 and 5—Counter Up/Down Control (TMC6, TMC5)

Selects whether TCC up/down control is performed by hardware using UD pin input, or whether TCC functions as an up-counter or a down-counter.

Bit 6 TMC6	Bit 5 TMC5	Description	
0	0	TCC is an up-counter	(initial value)
0	1	TCC is a down-counter	
1	*	Hardware control by UD pin input UD pin input high: Down-counter UD pin input Iow: Up-counter	
			* Dault anna

\*: Don't care

### Bits 4 and 3—Reserved

Bits 4 and 3 are reserved; they are always read as 1 and cannot be modified.

### Bits 2 to 0—Clock Select (TMC2 to TMC0)

Bits 2 to 0 select the clock input to TCC. For external event counting, either the rising or falling edge can be selected.

Bit 1 TMC1	Bit 0 TMC0	Description	
0	0	Internal clock: ø/8192	(initial value)
0	1	Internal clock: ø/2048	
1	0	Internal clock: ¢/512	
1	1	Internal clock: ø/64	
0	0	Internal clock: ø/16	
0	1	Internal clock:	
1	0	Internal clock: $\phi_w/4$	
1	1	External event (TMIC): rising or falling edge $^{*}$	
	TMC1           0           1           0           0	TMC1         TMC0           0         0           1         0           1         1           0         0           0         1           1         1           0         0           0         1           0         1	TMC1         TMC0         Description           0         0         Internal clock: φ/8192           0         1         Internal clock: φ/2048           1         0         Internal clock: φ/512           1         1         Internal clock: φ/64           0         0         Internal clock: φ/64           0         0         Internal clock: φ/4           1         Internal clock: φ/4           1         0         Internal clock: φ/4

Note: \* The edge of the external event signal is selected by bit IEG1 in the IRQ edge select register (IEGR). See IRQ Edge Select Register (IEGR) in section 3.3.2, Interrupt Control Registers, for details. IRQ1 in port mode register B (PMRB) must be set to 1 before setting 111 in bits TMC2 to TMC0.

### (2) Timer Counter C (TCC)

Bit	7	6	5	4	3	2	1	0
	TCC7	TCC6	TCC5	TCC4	тсс3	TCC2	TCC1	TCC0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

TCC is an 8-bit read-only up/down-counter, which is incremented or decremented by internal clock or external event input. The clock source for input to this counter is selected by bits TMC2 to TMC0 in timer mode register C (TMC). TCC values can be read by the CPU at any time.

When TCC overflows from H'FF to H'00 or to the value set in TLC, or underflows from H'00 to H'FF or to the value set in TLC, the IRRTC bit in IRR2 is set to 1.

TCC is allocated to the same address as TLC.

Upon reset, TCC is initialized to H'00.

### (3) Timer Load Register C (TLC)

Bit	7	6	5	4	3	2	1	0
	TLC7	TLC6	TLC5	TLC4	TLC3	TLC2	TLC1	TLC0
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

TLC is an 8-bit write-only register for setting the reload value of timer counter C (TCC).

When a reload value is set in TLC, the same value is loaded into timer counter C as well, and TCC starts counting up/down from that value. When TCC overflows or underflows during operation in auto-reload mode, the TLC value is loaded into TCC. Accordingly, overflow/underflow period can be set within the range of 1 to 256 input clocks.

The same address is allocated to TLC as to TCC.

Upon reset, TLC is initialized to H'00.



# (4) Clock Stop Register 1 (CKSTPR1)

Bit:	7	6	5	4	3	2	1	0
			S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value:	1	1	1	1	1	1	1	1
Read/Write:	_	_	R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to timer C is described here. For details of the other bits, see the sections on the relevant modules.

### Bit 1—Timer C Module Standby Mode Control (TCCKSTP)

Bit 1 controls setting and clearing of module standby mode for timer C.

TCCKSTP	Description	
0	Timer C is set to module standby mode	
1	Timer C module standby mode is cleared	(initial value)

### 9.3.3 Timer Operation

### (1) Interval Timer Operation

When bit TMC7 in timer mode register C (TMC) is cleared to 0, timer C functions as an 8-bit interval timer.

Upon reset, TCC is initialized to H'00 and TMC to H'18, so TCC continues up-counting as an interval up-counter without halting immediately after a reset. The timer C operating clock is selected from seven internal clock signals output by prescalers S and W, or an external clock input at pin TMIC. The selection is made by bits TMC2 to TMC0 in TMC.

TCC up/down-count control can be performed either by software or hardware. The selection is made by bits TMC6 and TMC5 in TMC.

After the count value in TCC reaches H'FF (H'00), the next clock input causes timer C to overflow (underflow), setting bit IRRTC in IRR2 to 1. If IENTC = 1 in interrupt enable register 2 (IENR2), a CPU interrupt is requested.

At overflow (underflow), TCC returns to H'00 (H'FF) and starts counting up (down) again.

During interval timer operation (TMC7 = 0), when a value is set in timer load register C (TLC), the same value is set in TCC.

Note: For details on interrupts, see section 3.3, Interrupts.

### (2) Auto-Reload Timer Operation

Setting bit TMC7 in TMC to 1 causes timer C to function as an 8-bit auto-reload timer. When a reload value is set in TLC, the same value is loaded into TCC, becoming the value from which TCC starts its count.

After the count value in TCC reaches H'FF (H'00), the next clock signal input causes timer C to overflow/underflow. The TLC value is then loaded into TCC, and the count continues from that value. The overflow/underflow period can be set within a range from 1 to 256 input clocks, depending on the TLC value.

The clock sources, up/down control, and interrupts in auto-reload mode are the same as in interval mode.

In auto-reload mode (TMC7 = 1), when a new value is set in TLC, the TLC value is also set in TCC.

### (3) Event Counter Operation

Timer C can operate as an event counter, counting rising or falling edges of an external event signal input at pin TMIC. External event counting is selected by setting bits TMC2 to TMC0 in timer mode register C (TMC) to all 1s (111). TCC counts up/down at the rising/falling edge of an external event signal input at pin TMIC.

When timer C is used to count external event input, bit IRQ1 in PMRB should be set to 1 and bit IEN1 in IENR1 cleared to 0 to disable interrupt IRQ1 requests.

# (4) TCC Up/Down Control by Hardware

With timer C, TCC up/down control can be performed by UD pin input. When bit TMC6 in TMC is set to 1, TCC functions as an up-counter when UD pin input is low, and as a down-counter when high.

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When using UD pin input, set bit UD in PMR3 to 1.

### 9.3.4 Timer C Operation States

Table 9.6 summarizes the timer C operation states.

### Table 9.6 Timer C Operation States

						Sub-	Sub-		Module
Operation Mode		Reset	Active	Sleep	Watch	active	sleep	Standby	Standby
TCC	Interval	Reset	Functions	Functions	Halted		Functions/ Halted <sup>*</sup>	Halted	Halted
	Auto reload	Reset	Functions	Functions	Halted		Functions/ Halted <sup>*</sup>	Halted	Halted
TMC		Reset	Functions	Retained	Retained	Functions	Retained	Retained	Retained

Note: \* When  $\phi$ w/4 is selected as the TCC internal clock in active mode or sleep mode, since the system clock and internal clock are mutually asynchronous, synchronization is maintained by a synchronization circuit. This results in a maximum count cycle error of  $1/\phi$  (s). When the counter is operated in subactive mode or subsleep mode, either select  $\phi$ w/4 as the internal clock or select an external clock. The counter will not operate on any other internal clock. If  $\phi$ w/4 is selected as the internal clock for the counter when  $\phi$ w/8 has been selected as subclock  $\phi_{sue}$ , the lower 2 bits of the counter operate on the same cycle, and the operation of the least significant bit is unrelated to the operation of the counter.



# 9.4 Timer F

### 9.4.1 Overview

Timer F is a 16-bit timer with a built-in output compare function. As well as counting external events, timer F also provides for counter resetting, interrupt request generation, toggle output, etc., using compare match signals. Timer F can also be used as two independent 8-bit timers (timer FH and timer FL).

### (1) Features

Features of timer F are given below.

- Choice of four internal clock sources (φ/32, φ/16, φ/4, φw/4) or an external clock (can be used as an external event counter)
- TMOFH/TMOFL pin toggle output provided using a single compare match signal (toggle output initial value can be set)
- Counter resetting by a compare match signal
- Two interrupt sources: one compare match, one overflow
- Can operate as two independent 8-bit timers (timer FH and timer FL) (in 8-bit mode).

	Timer FH 8-Bit Timer*	Timer FL 8-Bit Timer/Event Counter				
Internal clock	Choice of 4 (\$\phi/32, \$\phi/16, \$\phi/4, \$\phiw/4\$)					
Event input	<u> </u>	TMIF pin				
Toggle output	One compare match signal, output to TMOFH pin(initial value settable)	One compare match signal, output to TMOFL pin (initial value settable)				
Counter reset	Counter can be reset by compare match signal					
Interrupt sources	One compare match One overflow					

Note: \* When timer F operates as a 16-bit timer, it operates on the timer FL overflow signal.



- Operation in watch mode, subactive mode, and subsleep mode
   When \u03c6w/4 is selected as the internal clock, timer F can operate in watch mode, subactive mode, and subsleep mode.
- Use of module standby mode enables this module to be placed in standby mode independently when not used.


### (2) Block Diagram

Figure 9.3 shows a block diagram of timer F.





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## (3) Pin Configuration

Table 9.7 shows the timer F pin configuration.

## Table 9.7Pin Configuration

Name	Abbr.	I/O	Function
Timer F event input	TMIF	Input	Event input pin for input to TCFL
Timer FH output	TMOFH	Output	Timer FH toggle output pin
Timer FL output	TMOFL	Output	Timer FL toggle output pin

## (4) Register Configuration

Table 9.8 shows the register configuration of timer F.

## Table 9.8Timer F Registers

Name	Abbr.	R/W	Initial Value	Address
Timer control register F	TCRF	W	H'00	H'FFB6
Timer control/status register F	TCSRF	R/W	H'00	H'FFB7
8-bit timer counter FH	TCFH	R/W	H'00	H'FFB8
8-bit timer counter FL	TCFL	R/W	H'00	H'FFB9
Output compare register FH	OCRFH	R/W	H'FF	H'FFBA
Output compare register FL	OCRFL	R/W	H'FF	H'FFBB
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA

### 9.4.2 Register Descriptions

16-bit Timer Counter (TCF)
 8-bit Timer Counter (TCFH)
 8-bit Timer Counter (TCFL)



TCF is a 16-bit read/write up-counter configured by cascaded connection of 8-bit timer counters TCFH and TCFL. In addition to the use of TCF as a 16-bit counter with TCFH as the upper 8 bits and TCFL as the lower 8 bits, TCFH and TCFL can also be used as independent 8-bit counters.

TCFH and TCFL can be read and written by the CPU, but when they are used in 16-bit mode, data transfer to and from the CPU is performed via a temporary register (TEMP). For details of TEMP, see section 9.4.3, CPU Interface.

TCFH and TCFL are each initialized to H'00 upon reset.

a. 16-bit mode (TCF)

When CKSH2 is cleared to 0 in TCRF, TCF operates as a 16-bit counter. The TCF input clock is selected by bits CKSL2 to CKSL0 in TCRF.

TCF can be cleared in the event of a compare match by means of CCLRH in TCSRF. When TCF overflows from H'FFFF to H'0000, OVFH is set to 1 in TCSRF. If OVIEH in TCSRF is 1 at this time, IRRTFH is set to 1 in IRR2, and if IENTFH in IENR2 is 1, an interrupt request is sent to the CPU.



b. 8-bit mode (TCFL/TCFH)

When CKSH2 is set to 1 in TCRF, TCFH, and TCFL operate as two independent 8-bit counters. The TCFH (TCFL) input clock is selected by bits CKSH2 to CKSH0 (CKSL2 to CKSL0) in TCRF.

TCFH (TCFL) can be cleared in the event of a compare match by means of CCLRH (CCLRL) in TCSRF.

When TCFH (TCFL) overflows from H'FF to H'00, OVFH (OVFL) is set to 1 in TCSRF. If OVIEH (OVIEL) in TCSRF is 1 at this time, IRRTFH (IRRTFL) is set to 1 in IRR2, and if IENTFH (IENTFL) in IENR2 is 1, an interrupt request is sent to the CPU.

(2) 16-bit Output Compare Register (OCRF)
8-bit Output Compare Register (OCRFH)
8-bit Output Compare Register (OCRFL)



OCRF is a 16-bit read/write register composed of the two registers OCRFH and OCRFL. In addition to the use of OCRF as a 16-bit register with OCRFH as the upper 8 bits and OCRFL as the lower 8 bits, OCRFH and OCRFL can also be used as independent 8-bit registers.

OCRFH and OCRFL can be read and written by the CPU, but when they are used in 16-bit mode, data transfer to and from the CPU is performed via a temporary register (TEMP). For details of TEMP, see section 9.4.3, CPU Interface.

OCRFH and OCRFL are each initialized to H'FF upon reset.

a. 16-bit mode (OCRF)

When CKSH2 is cleared to 0 in TCRF, OCRF operates as a 16-bit register. OCRF contents are constantly compared with TCF, and when both values match, CMFH is set to 1 in TCSRF. At the same time, IRRTFH is set to 1 in IRR2. If IENTFH in IENR2 is 1 at this time, an interrupt request is sent to the CPU.

Toggle output can be provided from the TMOFH pin by means of compare matches, and the output level can be set (high or low) by means of TOLH in TCRF.

## b. 8-bit mode (OCRFH/OCRFL)

When CKSH2 is set to 1 in TCRF, OCRFH, and OCRFL operate as two independent 8-bit registers. OCRFH contents are compared with TCFH, and OCRFL contents are with TCFL. When the OCRFH (OCRFL) and TCFH (TCFL) values match, CMFH (CMFL) is set to 1 in TCSRF. At the same time, IRRTFH (IRRTFL) is set to 1 in IRR2. If IENTFH (IENTFL) in IENR2 is 1 at this time, an interrupt request is sent to the CPU.

Toggle output can be provided from the TMOFH pin (TMOFL pin) by means of compare matches, and the output level can be set (high or low) by means of TOLH (TOLL) in TCRF.

## (3) Timer Control Register F (TCRF)

Bit:	7	6	5	4	3	2	1	0
	TOLH	CKSH2	CKSH1	CKSH0	TOLL	CKSL2	CKSL1	CKSL0
Initial value:	0	0	0	0	0	0	0	0
Read/Write:	W	W	W	W	W	W	W	W

TCRF is an 8-bit write-only register that switches between 16-bit mode and 8-bit mode, selects the input clock from among four internal clock sources or external event input, and sets the output level of the TMOFH and TMOFL pins.

TCRF is initialized to H'00 upon reset.

# Bit 7—Toggle Output Level H (TOLH)

Bit 7 sets the TMOFH pin output level. The output level is effective immediately after this bit is written.

Bit 7 TOLH	Description	
0	Low level	(initial value)
1	High level	



### Bits 6 to 4—Clock Select H (CKSH2 to CKSH0)

Bits 6 to 4 select the clock input to TCFH from among four internal clock sources or TCFL overflow.

Bit 6 CKSH2	Bit 5 CKSH1	Bit 4 CKSH0	Description	
0	0	0	16-bit mode, counting on TCFL overflow signal	(initial value)
0	0	1	_	
0	1	0	_	
0	1	1	Use prohibited	
1	0	0	Internal clock: counting on	
1	0	1	Internal clock: counting on	
1	1	0	Internal clock: counting on	
1	1	1	Internal clock: counting on $\phi w/4$	

## Bit 3—Toggle Output Level L (TOLL)

Bit 3 sets the TMOFL pin output level. The output level is effective immediately after this bit is written.

Bit 3 TOLL	Description	
0	Low level	(initial value)
1	High level	



### Bits 2 to 0—Clock Select L (CKSL2 to CKSL0)

Bits 2 to 0 select the clock input to TCFL from among four internal clock sources or external event input.

Bit 2 CKSL2	Bit 1 CKSL1	Bit 0 CKSL0	Description
0	0	0	Counting on external event (TMIF) rising/falling edge*
0	0	1	(initial value)
0	1	0	
0	1	1	Use prohibited
1	0	0	Internal clock: counting on $\phi/32$
1	0	1	Internal clock: counting on $\phi/16$
1	1	0	Internal clock: counting on $\phi/4$
1	1	1	Internal clock: counting on $\phi w/4$

Note: \* External event edge selection is set by IEG3 in the IRQ edge select register (IEGR). For details, see IRQ Edge Select Register (IEGR) in section 3.3.2, Interrupt Control Registers.

Note that the timer F counter may increment if the setting of IRQ3 in port mode register 1 (PMR1) is changed from 0 to 1 or from 1 to 0 while the TMIF pin is low in order to change the TMIF pin function.

#### (4) Timer Control/Status Register F (TCSRF)

Bit:	7	6	5	4	3	2	1	0
	OVFH	CMFH	OVIEH	CCLRH	OVFL	CMFL	OVIEL	CCLRL
Initial value:	0	0	0	0	0	0	0	0

Note: \* Bits 7, 6, 3, and 2 can only be written with 0, for flag clearing.

TCSRF is an 8-bit read/write register that performs counter clear selection, overflow flag setting, and compare match flag setting, and controls enabling of overflow interrupt requests.

TCSRF is initialized to H'00 upon reset.



## Bit 7—Timer Overflow Flag H (OVFH)

Bit 7 is a status flag indicating that TCFH has overflowed from H'FF to H'00. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 7 OVFH	Description	
0	Clearing condition: After reading OVFH = 1, cleared by writing 0 to OVFH	(initial value)
1	Setting condition: Set when TCFH overflows from H'FF to H'00	

#### Bit 6—Compare Match Flag H (CMFH)

Bit 6 is a status flag indicating that TCFH has matched OCRFH. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 6 CMFH	Description	
0	Clearing condition: After reading CMFH = 1, cleared by writing 0 to CMFH	(initial value)
1	Setting condition: Set when the TCFH value matches the OCRFH value	

### Bit 5—Timer Overflow Interrupt Enable H (OVIEH)

Bit 5 selects enabling or disabling of interrupt generation when TCFH overflows.

Bit 5 OVIEH	Description	
0	TCFH overflow interrupt request is disabled	(initial value)
1	TCFH overflow interrupt request is enabled	

### Bit 4—Counter Clear H (CCLRH)

In 16-bit mode, bit 4 selects whether TCF is cleared when TCF and OCRF match.

In 8-bit mode, bit 4 selects whether TCFH is cleared when TCFH and OCRFH match.

Bit 4 CCLRH	Description	
0	16-bit mode: TCF clearing by compare match is disabled 8-bit mode: TCFH clearing by compare match is disabled	(initial value)
1	16-bit mode: TCF clearing by compare match is enabled 8-bit mode: TCFH clearing by compare match is enabled	

### Bit 3—Timer Overflow Flag L (OVFL)

Bit 3 is a status flag indicating that TCFL has overflowed from H'FF to H'00. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 3 OVFL	Description	
0	Clearing condition: After reading OVFL = 1, cleared by writing 0 to OVFL	(initial value)
1	Setting condition: Set when TCFL overflows from H'FF to H'00	

## Bit 2—Compare Match Flag L (CMFL)

Bit 2 is a status flag indicating that TCFL has matched OCRFL. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 2 CMFL	Description	
0	Clearing condition: After reading CMFL = 1, cleared by writing 0 to CMFL	(initial value)
1	Setting condition: Set when the TCFL value matches the OCRFL value	



## Bit 1—Timer Overflow Interrupt Enable L (OVIEL)

Bit 1 selects enabling or disabling of interrupt generation when TCFL overflows.

Bit 1 OVIEL	Description	
0	TCFL overflow interrupt request is disabled	(initial value)
1	TCFL overflow interrupt request is enabled	

### Bit 0—Counter Clear L (CCLRL)

Bit 0 selects whether TCFL is cleared when TCFL and OCRFL match.

Bit 0 CCLRL	Description	
0	TCFL clearing by compare match is disabled	(initial value)
1	TCFL clearing by compare match is enabled	

### (5) Clock Stop Register 1 (CKSTPR1)

Bit:	7	6	5	4	3	2	1	0
		_	S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value:	1	1	1	1	1	1	1	1
Read/Write:	_	_	R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to timer F is described here. For details of the other bits, see the sections on the relevant modules.

### Bit 2—Timer F Module Standby Mode Control (TFCKSTP)

Bit 2 controls setting and clearing of module standby mode for timer F.

TFCKSTP	Description	
0	Timer F is set to module standby mode	
1	Timer F module standby mode is cleared	(initial value)

## 9.4.3 CPU Interface

TCF and OCRF are 16-bit read/write registers, but the CPU is connected to the on-chip peripheral modules by an 8-bit data bus. When the CPU accesses these registers, it therefore uses an 8-bit temporary register (TEMP).

When performing TCF read/write access or OCRF write access in 16-bit mode, data will not be transferred correctly if only the upper byte or only the lower byte is accessed. Access must be performed for all 16 bits (using two consecutive byte-size MOV instructions), and the upper byte must be accessed before the lower byte.

In 8-bit mode, there are no restrictions on the order of access.

## (1) Write Access

Write access to the upper byte results in transfer of the upper-byte write data to TEMP. Next, write access to the lower byte results in transfer of the data in TEMP to the upper register byte, and direct transfer of the lower-byte write data to the lower register byte.



Figure 9.4 shows an example in which H'AA55 is written to TCF.



Figure 9.4 Write Access to TCF (CPU  $\rightarrow$  TCF)

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### (2) Read Access

In access to TCF, when the upper byte is read the upper-byte data is transferred directly to the CPU and the lower-byte data is transferred to TEMP. Next, when the lower byte is read, the lower-byte data in TEMP is transferred to the CPU.

In access to OCRF, when the upper byte is read the upper-byte data is transferred directly to the CPU. When the lower byte is read, the lower-byte data is transferred directly to the CPU.

Figure 9.5 shows an example in which TCF is read when it contains H'AAFF.



Figure 9.5 Read Access to TCF (TCF  $\rightarrow$  CPU)

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## 9.4.4 Operation

Timer F is a 16-bit counter that increments on each input clock pulse. The timer F value is constantly compared with the value set in output compare register F, and the counter can be cleared, an interrupt requested, or port output toggled, when the two values match. Timer F can also function as two independent 8-bit timers.

## (1) Timer F Operation

Timer F has two operating modes, 16-bit timer mode and 8-bit timer mode. The operation in each of these modes is described below.

a. Operation in 16-bit timer mode

When CKSH2 is cleared to 0 in timer control register F (TCRF), timer F operates as a 16-bit timer.

Following a reset, timer counter F (TCF) is initialized to H'0000, output compare register F (OCRF) to H'FFFF, and timer control register F (TCRF) and timer control/status register F (TCSRF) to H'00. The counter starts incrementing on external event (TMIF) input. The external event edge selection is set by IEG3 in the IRQ edge select register (IEGR).

The timer F operating clock can be selected from three internal clocks output by prescaler S or an external clock by means of bits CKSL2 to CKSL0 in TCRF.

OCRF contents are constantly compared with TCF, and when both values match, CMFH is set to 1 in TCSRF. If IENTFH in IENR2 is 1 at this time, an interrupt request is sent to the CPU, and at the same time, TMOFH pin output is toggled. If CCLRH in TCSRF is 1, TCF is cleared. TMOFH pin output can also be set by TOLH in TCRF.

When TCF overflows from H'FFFF to H'0000, OVFH is set to 1 in TCSRF. If OVIEH in TCSRF and IENTFH in IENR2 are both 1, an interrupt request is sent to the CPU.

b. Operation in 8-bit timer mode

When CKSH2 is set to 1 in TCRF, TCF operates as two independent 8-bit timers, TCFH and TCFL. The TCFH/TCFL input clock is selected by CKSH2 to CKSH0/CKSL2 to CKSL0 in TCRF.

When the OCRFH/OCRFL and TCFH/TCFL values match, CMFH/CMFL is set to 1 in TCSRF. If IENTFH/IENTFL in IENR2 is 1, an interrupt request is sent to the CPU, and at the same time, TMOFH pin/TMOFL pin output is toggled. If CCLRH/CCLRL in TCSRF is 1, TCFH/TCFL is cleared. TMOFH pin/TMOFL pin output can also be set by TOLH/TOLL in TCRF.

When TCFH/TCFL overflows from H'FF to H'00, OVFH/OVFL is set to 1 in TCSRF. If OVIEH/OVIEL in TCSRF and IENTFH/IENTFL in IENR2 are both 1, an interrupt request is sent to the CPU.

## (2) TCF Increment Timing

TCF is incremented by clock input (internal clock or external event input).

a. Internal clock operation

Bits CKSH2 to CKSH0 or CKSL2 to CKSL0 in TCRF select one of four internal clock sources ( $\phi/32$ ,  $\phi/16$ ,  $\phi/4$ , or  $\phi$ w/4) created by dividing the system clock ( $\phi$  or  $\phi$ w).

b. External event operation

External event input is selected by clearing CKSL2 to 0 in TCRF. TCF can increment on either the rising or falling edge of external event input. External event edge selection is set by IEG3 in the interrupt controller's IEGR register. An external event pulse width of at least 2 system clocks ( $\phi$ ) is necessary. Shorter pulses will not be counted correctly.

## (3) TMOFH/TMOFL Output Timing

In TMOFH/TMOFL output, the value set in TOLH/TOLL in TCRF is output. The output is toggled by the occurrence of a compare match. Figure 9.6 shows the output timing.



# Figure 9.6 TMOFH/TMOFL Output Timing



## (4) TCF Clear Timing

TCF can be cleared by a compare match with OCRF.

### (5) Timer Overflow Flag (OVF) Set Timing

OVF is set to 1 when TCF overflows from H'FFFF to H'0000.

## (6) Compare Match Flag Set Timing

The compare match flag (CMFH or CMFL) is set to 1 when the TCF and OCRF values match. The compare match signal is generated in the last state during which the values match (when TCF is updated from the matching value to a new value). When TCF matches OCRF, the compare match signal is not generated until the next counter clock.

#### (7) Timer F Operation Modes

Timer F operation modes are shown in table 9.9.

Table 9.9	Timer F Operation Modes
-----------	-------------------------

Operation Mode	Reset	Active	Sleep	Watch	Sub- active	Sub- sleep	Standby	Module Standby
TCF	Reset	Functions	Functions	Functions/ Halted <sup>*</sup>	Functions/ Halted <sup>*</sup>	Functions/ Halted <sup>*</sup>	Halted	Halted
OCRF	Reset	Functions	Held	Held	Functions	Held	Held	Held
TCRF	Reset	Functions	Held	Held	Functions	Held	Held	Held
TCSRF	Reset	Functions	Held	Held	Functions	Held	Held	Held

Note: \* When  $\phi_w/4$  is selected as the TCF internal clock in active mode or sleep mode, since the system clock and internal clock are mutually asynchronous, synchronization is maintained by a synchronization circuit. This results in a maximum count cycle error of  $1/\phi$  (s). When the counter is operated in subactive mode, watch mode, or subsleep mode,  $\phi_w/4$  must be selected as the internal clock. The counter will not operate if any other internal clock is selected.



## 9.4.5 Application Notes

The following types of contention and operation can occur when timer F is used.

## (1) 16-bit Timer Mode

In toggle output, TMOFH pin output is toggled when all 16 bits match and a compare match signal is generated. If a TCRF write by a MOV instruction and generation of the compare match signal occur simultaneously, TOLH data is output to the TMOFH pin as a result of the TCRF write. TMOFL pin output is unstable in 16-bit mode, and should not be used; the TMOFL pin should be used as a port pin.

If an OCRFL write and compare match signal generation occur simultaneously, the compare match signal is invalid. However, if the written data and the counter value match, a compare match signal will be generated at that point. As the compare match signal is output in synchronization with the TCFL clock, a compare match will not result in compare match signal generation if the clock is stopped.

Compare match flag CMFH is set when all 16 bits match and a compare match signal is generated. Compare match flag CMFL is set if the setting conditions for the lower 8 bits are satisfied.

When TCF overflows, OVFH is set. OVFL is set if the setting conditions are satisfied when the lower 8 bits overflow. If a TCFL write and overflow signal output occur simultaneously, the overflow signal is not output.

### (2) 8-bit Timer Mode

### a. TCFH, OCRFH

In toggle output, TMOFH pin output is toggled when a compare match occurs. If a TCRF write by a MOV instruction and generation of the compare match signal occur simultaneously, TOLH data is output to the TMOFH pin as a result of the TCRF write.

If an OCRFH write and compare match signal generation occur simultaneously, the compare match signal is invalid. However, if the written data and the counter value match, a compare match signal will be generated at that point. The compare match signal is output in synchronization with the TCFH clock.

If a TCFH write and overflow signal output occur simultaneously, the overflow signal is not output.



## b. TCFL, OCRFL

In toggle output, TMOFL pin output is toggled when a compare match occurs. If a TCRF write by a MOV instruction and generation of the compare match signal occur simultaneously, TOLL data is output to the TMOFL pin as a result of the TCRF write.

If an OCRFL write and compare match signal generation occur simultaneously, the compare match signal is invalid. However, if the written data and the counter value match, a compare match signal will be generated at that point. As the compare match signal is output in synchronization with the TCFL clock, a compare match will not result in compare match signal generation if the clock is stopped.

If a TCFL write and overflow signal output occur simultaneously, the overflow signal is not output.

# (3) Clear Timer FH, Timer FL Interrupt Request Flags (IRRTFH, IRRTFL), Timer Overflow Flags H, L (OVFH, OVFL) and Compare Match Flags H, L (CMFH, CMFL)

When  $\phi$ w/4 is selected as the internal clock, "Interrupt factor generation signal" will be operated with  $\phi$ w and the signal will be outputted with  $\phi$ w width. And, "Overflow signal" and "Compare match signal" are controlled with 2 cycles of  $\phi$ w signals. Those signals are outputted with 2 cycles width of  $\phi$ w (figure 9.7)

In active (high-speed, medium-speed) mode, even if you cleared interrupt request flag during the term of validity of "Interrupt factor generation signal", same interrupt request flag is set. (figure 9.7 (1)) And, you cannot be cleared timer overflow flag and compare match flag during the term of validity of "Overflow signal" and "Compare match signal".

For interrupt request flag is set right after interrupt request is cleared, interrupt process to one time timer FH, timer FL interrupt might be repeated. (figure 9.7 (2)) Therefore, to definitely clear interrupt request flag in active (high-speed, medium-speed) mode, clear should be processed after the time that calculated with below (1) formula. And, to definitely clear timer overflow flag and compare match flag, clear should be processed after read timer control status register F (TCSRF) after the time that calculated with below (1) formula. For ST of (1) formula, please substitute the longest number of execution states in used instruction. (10 states of RTE instruction when MULXU, DIVXU instruction is not used, 14 states when MULXU, DIVXU instruction is used) In subactive mode, there are not limitation for interrupt request flag, timer overflow flag, and compare match flag clear.



The term of validity of "Interrupt factor generation signal"

= 1 cycle of  $\phi w$  + waiting time for completion of executing instruction

+ interrupt time synchronized with  $\phi = 1/\phi w + ST \times (1/\phi) + (2/\phi)$  (second).....(1)

ST: Executing number of execution states

Method 1 is recommended to operate for time efficiency.

Method 1

- 1. Prohibit interrupt in interrupt handling routine (set IENFH, IENFL to 0).
- 2. After program process returned normal handling, clear interrupt request flags (IRRTFH, IRRTFL) after more than that calculated with (1) formula.
- 3. After read timer control status register F (TCSRF), clear timer overflow flags (OVFH, OVFL) and compare match flags (CMFH, CMFL).
- 4. Operate interrupt permission (set IENFH, IENFL to 1).

Method 2

- 1. Set interrupt handling routine time to more than time that calculated with (1) formula.
- 2. Clear interrupt request flags (IRRTFH, IRRTFL) at the end of interrupt handling routine.
- 3. After read timer control status register F (TCSRF), clear timer overflow flags (OVFH, OVFL) and compare match flags (CMFH, CMFL).

All above attentions are also applied in 16-bit mode and 8-bit mode.





Figure 9.7 Clear Interrupt Request Flag when Interrupt Factor Generation Signal is Valid

## (4) Timer Counter (TCF) Read/Write

When  $\phi$ w/4 is selected as the internal clock in active (high-speed, medium-speed) mode, write on TCF is impossible. And, when read TCF, as the system clock and internal clock are mutually asynchronous, TCF synchronizes with synchronization circuit. This results in a maximum TCF read value error of ±1.

When read/write TCF in active (high-speed, medium-speed) mode is needed, please select internal clock except for  $\phi$ w/4 before read/write.

In subactive mode, even  $\phi$ w/4 is selected as the internal clock, normal read/write TCF is possible.

# 9.5 Timer G

### 9.5.1 Overview

Timer G is an 8-bit timer with dedicated input capture functions for the rising/falling edges of pulses input from the input capture input pin (input capture input signal). High-frequency component noise in the input capture input signal can be eliminated by a noise canceler, enabling accurate measurement of the input capture input signal duty cycle. If input capture input is not set, timer G functions as an 8-bit interval timer.

### (1) Features

Features of timer G are given below.

- Choice of four internal clock sources ( $\phi/64$ ,  $\phi/32$ ,  $\phi/2$ ,  $\phi w/4$ )
- Dedicated input capture functions for rising and falling edges
- Level detection at counter overflow It is possible to detect whether overflow occurred when the input capture input signal was high or when it was low.
- Selection of whether or not the counter value is to be cleared at the input capture input signal rising edge, falling edge, or both edges
- Two interrupt sources: one input capture, one overflow. The input capture input signal rising or falling edge can be selected as the interrupt source.
- A built-in noise canceler eliminates high-frequency component noise in the input capture input signal.
- Watch mode, subactive mode, or subsleep mode operation is possible when  $\phi w/4$  is selected as the internal clock.
- Use of module standby mode enables this module to be placed in standby mode independently when not used.



### (2) Block Diagram

Figure 9.8 shows a block diagram of timer G.





### (3) Pin Configuration

Table 9.10 shows the timer G pin configuration.

### Table 9.10Pin Configuration

Name	Abbr.	I/O	Function
Input capture input	TMIG	Input	Input capture input pin

### (4) Register Configuration

Table 9.11 shows the register configuration of timer G.

### Table 9.11 Timer G Registers

Name	Abbr.	R/W	Initial Value	Address
Timer mode register G	TMG	R/W	H'00	H'FFBC
Timer counter G	TCG	_	H'00	_
Input capture register GF	ICRGF	R	H'00	H'FFBD
Input capture register GR	ICRGR	R	H'00	H'FFBE
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA

### 9.5.2 Register Descriptions

### (1) Timer Counter G (TCG)

Bit:	7	6	5	4	3	2	1	0
	TCG7	TCG6	TCG5	TCG4	TCG3	TCG2	TCG1	TCG0
Initial value:	0	0	0	0	0	0	0	0
Read/Write:	_	_	_	_	_	_		_

TCG is an 8-bit up-counter, which is incremented by clock input. The input clock is selected by bits CKS1 and CKS0 in TMG.

TMIG in PMR1 is set to 1 to operate TCG as an input capture timer, or cleared to 0 to operate TCG as an interval timer<sup>\*</sup>. In input capture timer operation, the TCG value can be cleared by the rising edge, falling edge, or both edges of the input capture input signal, according to the setting made in TMG.

When TCG overflows from H'FF to H'00, if OVIE in TMG is 1, IRRTG in IRR2 is set to 1, and if IENTG in IENR2 is 1, an interrupt request is sent to the CPU.

For details of the interrupt, see section 3.3, Interrupts.

TCG cannot be read or written by the CPU. It is initialized to H'00 upon reset.

Note: \* An input capture signal may be generated when TMIG is modified.





## (2) Input Capture Register GF (ICRGF)

Bit:	7	6	5	4	3	2	1	0
	ICRGF7	ICRGF6	ICRGF5	ICRGF4	ICRGF3	ICRGF2	ICRGF1	ICRGF0
Initial value:	0	0	0	0	0	0	0	0
Read/Write:	R	R	R	R	R	R	R	R

ICRGF is an 8-bit read-only register. When a falling edge of the input capture input signal is detected, the current TCG value is transferred to ICRGF. If IIEGS in TMG is 1 at this time, IRRTG in IRR2 is set to 1, and if IENTG in IENR2 is 1, an interrupt request is sent to the CPU.

For details of the interrupt, see section 3.3, Interrupts.

To ensure dependable input capture operation, the pulse width of the input capture input signal must be at least  $2\phi$  or  $2\phi_{streak}$  (when the noise canceler is not used).

ICRGF is initialized to H'00 upon reset.

## (3) Input Capture Register GR (ICRGR)

Bit:	7	6	5	4	3	2	1	0
	ICRGR7	ICRGR6	ICRGR5	ICRGR4	ICRGR3	ICRGR2	ICRGR1	ICRGR0
Initial value:	0	0	0	0	0	0	0	0
Read/Write:	R	R	R	R	R	R	R	R

ICRGR is an 8-bit read-only register. When a rising edge of the input capture input signal is detected, the current TCG value is transferred to ICRGR. If IIEGS in TMG is 0 at this time, IRRTG in IRR2 is set to 1, and if IENTG in IENR2 is 1, an interrupt request is sent to the CPU.

For details of the interrupt, see section 3.3, Interrupts.

To ensure dependable input capture operation, the pulse width of the input capture input signal must be at least  $2\phi$  or  $2\phi_{SUB}$  (when the noise canceler is not used).

ICRGR is initialized to H'00 upon reset.



### (4) Timer Mode Register G (TMG)

Bit:	7	6	5	4	3	2	1	0
	OVFH	OVFL	OVIE	IIEGS	CCLR1	CCLR0	CKS1	CKS0
Initial value:	0		0	0	0	0	0	0
initial value.	0	0	0	0	0	0	0	0

Note: \* Bits 7 and 6 can only be written with 0, for flag clearing.

TMG is an 8-bit read/write register that performs TCG clock selection from four internal clock sources, counter clear selection, and edge selection for the input capture input signal interrupt request, controls enabling of overflow interrupt requests, and also contains the overflow flags.

TMG is initialized to H'00 upon reset.

### Bit 7—Timer Overflow Flag H (OVFH)

Bit 7 is a status flag indicating that TCG has overflowed from H'FF to H'00 when the input capture input signal is high. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 7 OVFH	Description	
0	Clearing condition: After reading OVFH = 1, cleared by writing 0 to OVFH (initial value	¥)
1	Setting condition: Set when input capture input signal is high level and TCG overflows from H'FF to H'00	)



## Bit 6—Timer Overflow Flag L (OVFL)

Bit 6 is a status flag indicating that TCG has overflowed from H'FF to H'00 when the input capture input signal is low, or in interval operation. This flag is set by hardware and cleared by software. It cannot be set by software.

Bit 6 OVFL	Description	
0	Clearing condition: After reading OVFL = 1, cleared by writing 0 to OVFL	(initial value)
1	Setting condition: Set when TCG overflows from H'FF to H'00 while input capture input sig level or during interval operation	gnal is high

## Bit 5—Timer Overflow Interrupt Enable (OVIE)

Bit 5 selects enabling or disabling of interrupt generation when TCG overflows.

Bit 5 OVIE	Description	
0	TCG overflow interrupt request is disabled	(initial value)
1	TCG overflow interrupt request is enabled	

## Bit 4—Input Capture Interrupt Edge Select (IIEGS)

Bit 4 selects the input capture input signal edge that generates an interrupt request.

Bit 4		
IIEGS	Description	
0	Interrupt generated on rising edge of input capture input signal	(initial value)
1	Interrupt generated on falling edge of input capture input signal	



### Bits 3 and 2—Counter Clear 1 and 0 (CCLR1, CCLR0)

Bits 3 and 2 specify whether or not TCG is cleared by the rising edge, falling edge, or both edges of the input capture input signal.

Bit 3 CCLR1	Bit 2 CCLR0	Description	
0	0	TCG clearing is disabled	(initial value)
0	1	TCG cleared by falling edge of input capture input signal	
1	0	TCG cleared by rising edge of input capture input signal	
1	1	TCG cleared by both edges of input capture input signal	

#### Bits 1 and 0—Clock Select (CKS1, CKS0)

Bits 1 and 0 select the clock input to TCG from among four internal clock sources.

Bit 1 CKS1	Bit 0 CKS0	Description	
0	0	Internal clock: counting on $\phi/64$	(initial value)
0	1	Internal clock: counting on $\phi/32$	
1	0	Internal clock: counting on $\phi/2$	
1	1	Internal clock: counting on $\phi$ w/4	

## (5) Clock Stop Register 1 (CKSTPR1)

Bit:	7	6	5	4	3	2	1	0
	—		S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value:	1	1	1	1	1	1	1	1
Read/Write:			R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to timer G is described here. For details of the other bits, see the sections on the relevant modules.



## Bit 3—Timer G Module Standby Mode Control (TGCKSTP)

Bit 3 controls setting and clearing of module standby mode for timer G.

TGCKSTP	Description	
0	Timer G is set to module standby mode	
1	Timer G module standby mode is cleared	(initial value)

#### 9.5.3 Noise Canceler

The noise canceler consists of a digital low-pass filter that eliminates high-frequency component noise from the pulses input from the input capture input pin. The noise canceler is set by NCS<sup>\*</sup> in PMR2.

Figure 9.9 shows a block diagram of the noise canceler.



Figure 9.9 Noise Canceler Block Diagram

The noise canceler consists of five latch circuits connected in series and a match detector circuit. When the noise cancellation function is not used (NCS = 0), the system clock is selected as the sampling clock. When the noise cancellation function is used (NCS = 1), the sampling clock is the internal clock selected by CKS1 and CKS0 in TMG, the input capture input is sampled on the rising edge of this clock, and the data is judged to be correct when all the latch outputs match. If all the outputs do not match, the previous value is retained. After a reset, the noise canceler output is initialized when the falling edge of the input capture input signal has been sampled five times.



Therefore, after making a setting for use of the noise cancellation function, a pulse with at least five times the width of the sampling clock is a dependable input capture signal. Even if noise cancellation is not used, an input capture input signal pulse width of at least  $2\phi$  or  $2\phi_{SUB}$  is necessary to ensure that input capture operations are performed properly

Note: \* An input capture signal may be generated when the NCS bit is modified.

Figure 9.10 shows an example of noise canceler timing.

In this example, high-level input of less than five times the width of the sampling clock at the input capture input pin is eliminated as noise.



Figure 9.10 Noise Canceler Timing (Example)



## 9.5.4 Operation

Timer G is an 8-bit timer with built-in input capture and interval functions.

### (1) Timer G Functions

Timer G is an 8-bit up-counter with two functions, an input capture timer function and an interval timer function.

The operation of these two functions is described below.

a. Input capture timer operation

When the TMIG bit in port mode register 1 (PMR1) is set to 1, timer G functions as an input capture timer<sup>\*</sup>.

In a reset, timer mode register G (TMG), timer counter G (TCG), input capture register GF (ICRGF), and input capture register GR (ICRGR) are all initialized to H'00.

Following a reset, TCG starts counting on the  $\phi/64$  internal clock.

The input clock can be selected from four internal clock sources by bits CKS1 and CKS0 in TMG.

When a rising edge/falling edge is detected in the input capture signal input from the TMIG pin, the TCG value at that time is transferred to ICRGR/ICRGF. When the edge selected by IIEGS in TMG is input, IRRTG in IRR2 is set to 1, and if the IENTG bit in IENR2 is 1 at this time, an interrupt request is sent to the CPU. For details of the interrupt, see section 3.3, Interrupts.

TCG can be cleared by a rising edge, falling edge, or both edges of the input capture signal, according to the setting of bits CCLR1 and CCLR0 in TMG. If TCG overflows when the input capture signal is high, the OVFH bit in TMG is set; if TCG overflows when the input capture signal is low, the OVFL bit in TMG is set. If the OVIE bit in TMG is 1 when these bits are set, IRRTG in IRR2 is set to 1, and if the IENTG bit in IENR2 is 1, timer G sends an interrupt request to the CPU. For details of the interrupt, see section 3.3, Interrupts.

Timer G has a built-in noise canceler that enables high-frequency component noise to be eliminated from pulses input from the TMIG pin. For details, see section 9.5.3, Noise Canceler.

Note: \* An input capture signal may be generated when TMIG is modified.



b. Interval timer operation

When the TMIG bit in PMR1 is cleared to 0, timer G functions as an interval timer. Following a reset, TCG starts counting on the  $\phi/64$  internal clock. The input clock can be selected from four internal clock sources by bits CKS1 and CKS0 in TMG. TCG increments on the selected clock, and when it overflows from H'FF to H'00, the OVFL bit in TMG is set to 1. If the OVIE bit in TMG is 1 at this time, IRRTG in IRR2 is set to 1, and if the IENTG bit in IENR2 is 1, timer G sends an interrupt request to the CPU. For details of the interrupt, see section 3.3, Interrupts.

## (2) Count Timing

TCG is incremented by internal clock input. Bits CKS1 and CKS0 in TMG select one of four internal clock sources ( $\phi/64$ ,  $\phi/32$ ,  $\phi/2$ , or  $\phi w/4$ ) created by dividing the system clock ( $\phi$ ) or watch clock ( $\phi w$ ).

# (3) Input Capture Input Timing

a. Without noise cancellation function

For input capture input, dedicated input capture functions are provided for rising and falling edges.

Figure 9.11 shows the timing for rising/falling edge input capture input.



Figure 9.11 Input Capture Input Timing (without Noise Cancellation Function)

RENESAS



b. With noise cancellation function

When noise cancellation is performed on the input capture input, the passage of the input capture signal through the noise canceler results in a delay of five sampling clock cycles from the input capture input signal edge.

Figure 9.12 shows the timing in this case.



Figure 9.12 Input Capture Input Timing (with Noise Cancellation Function)

## (4) Timing of Input Capture by Input Capture Input

Figure 9.13 shows the timing of input capture by input capture input



Figure 9.13 Timing of Input Capture by Input Capture Input

## (5) TCG Clear Timing

TCG can be cleared by the rising edge, falling edge, or both edges of the input capture input signal.

Figure 9.14 shows the timing for clearing by both edges.



Figure 9.14 TCG Clear Timing



# (6) Timer G Operation Modes

Timer G operation modes are shown in table 9.12.

# Table 9.12 Timer G Operation Modes

Operat Mode	tion	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module Standby
TCG	Input capture	Reset	Functions*	Functions*	Functions/ halted*	Functions/ halted*	Functions/ halted*	Halted	Halted
	Interval	Reset	Functions*	Functions*	Functions/ halted*	Functions/ halted*	Functions/ halted*	Halted	Halted
ICRGF		Reset	Functions*	Functions*	Functions/ halted*	Functions/ halted*	Functions/ halted*	Retained	Retained
ICRGR		Reset	Functions*	Functions*	Functions/ halted*	Functions/ halted*	Functions/ halted*	Retained	Retained
TMG		Reset	Functions	Retained	Retained	Functions	Retained	Retained	Retained
Note:	the mair 1/φ( nois (φw/	system ntained s). Whe e cance 8, øw/4	is selected clock and ir by a synch n $\phi$ w/4 is se eler operate , $\phi$ w/2). Not eler do not c	nternal cloc ronization c elected as t on the φw/ e that wher	k are mutua sircuit. This he TCG int 4 internal c n another ir	ally asynch results in a ernal clock lock withou nternal cloc	ronous, sy maximum in watch n It regard to k is selecte	nchronizat count cyc node, TCG the $\phi_{SUB}$ si ed, TCG ar	ion is le error of and the ubclock

in input capture.

To operate the timer G in subactive mode or subsleep mode, select  $\phi w/4$  as the TCG internal clock and  $\phi w/2$  as the subclock  $\varphi_{\text{sub}}$ . Note that when other internal clock is selected, or when  $\phi w/8$  or  $\phi w/4$  is selected as the subclock  $\varphi_{\text{sub}}$ , TCG and the noise canceler do not operate.

### 9.5.5 Application Notes

## (1) Internal Clock Switching and TCG Operation

Depending on the timing, TCG may be incremented by a switch between different internal clock sources. Table 9.13 shows the relation between internal clock switchover timing (by write to bits CKS1 and CKS0) and TCG operation.

When TCG is internally clocked, an increment pulse is generated on detection of the falling edge of an internal clock signal, which is divided from the system clock ( $\phi$ ) or subclock ( $\phi$ w). For this reason, in a case like No. 3 in table 9.13 where the switch is from a high clock signal to a low clock signal, the switchover is seen as a falling edge, causing TCG to increment.

## Table 9.13 Internal Clock Switching and TCG Operation









### (2) Notes on Port Mode Register Modification

The following points should be noted when a port mode register is modified to switch the input capture function or the input capture input noise canceler function.

• Switching input capture input pin function

Note that when the pin function is switched by modifying TMIG in port mode register 1 (PMR1), which performs input capture input pin control, an edge will be regarded as having been input at the pin even though no valid edge has actually been input. Input capture input signal input edges, and the conditions for their occurrence, are summarized in table 9.14.
Input Capture Input Signal Input Edge	Conditions
Generation of rising edge	When TMIG is modified from 0 to 1 while the TMIG pin is high
	When NCS is modified from 0 to 1 while the TMIG pin is high, then TMIG is modified from 0 to 1 before the signal is sampled five times by the noise canceler
Generation of falling edge	When TMIG is modified from 1 to 0 while the TMIG pin is high
	When NCS is modified from 0 to 1 while the TMIG pin is low, then TMIG is modified from 0 to 1 before the signal is sampled five times by the noise canceler
	When NCS is modified from 0 to 1 while the TMIG pin is high, then TMIG is modified from 1 to 0 after the signal is sampled five times by the noise canceler

# Table 9.14Input Capture Input Signal Input Edges Due to Input Capture Input Pin<br/>Switching, and Conditions for Their Occurrence

Note: When the P1<sub>3</sub> pin is not set as an input capture input pin, the timer G input capture input signal is low.

• Switching input capture input noise canceler function

When performing noise canceler function switching by modifying NCS in port mode register 2 (PMR2), which controls the input capture input noise canceler, TMIG should first be cleared to 0. Note that if NCS is modified without first clearing TMIG, an edge will be regarded as having been input at the pin even though no valid edge has actually been input. Input capture input signal input edges, and the conditions for their occurrence, are summarized in table 9.15.

# Table 9.15Input Capture Input Signal Input Edges Due to Noise Canceler FunctionSwitching, and Conditions for Their Occurrence

Input Capture Input Signal Input Edge	Conditions
Generation of rising edge	When the TMIG pin is modified from 0 to 1 while TMIG is 1, then NCS is modified from 0 to 1 before the signal is sampled five times by the noise canceler
Generation of falling edge	When the TMIG pin is modified from 1 to 0 while TMIG is 1, then NCS is modified from 1 to 0 before the signal is sampled five times by the noise canceler



When the pin function is switched and an edge is generated in the input capture input signal, if this edge matches the edge selected by the input capture interrupt select (IIEGS) bit, the interrupt request flag will be set to 1. The interrupt request flag should therefore be cleared to 0 before use. Figure 9.15 shows the procedure for port mode register manipulation and interrupt request flag clearing. When switching the pin function, set the interrupt-disabled state before manipulating the port mode register, then, after the port mode register operation has been performed, wait for the time required to confirm the input capture input signal as an input capture signal (at least two system clocks when the noise canceler is not used; at least five sampling clocks when the noise canceler is used), before clearing the interrupt enable flag to 0. There are two ways of preventing interrupt request flag setting when the pin function is switched: by controlling the pin level so that the conditions shown in tables 9.14 and 9.15 are not satisfied, or by setting the opposite of the generated edge in the IIEGS bit in TMG.



Figure 9.15 Port Mode Register Manipulation and Interrupt Enable Flag Clearing Procedure



#### 9.5.6 Timer G Application Example

Using timer G, it is possible to measure the high and low widths of the input capture input signal as absolute values. For this purpose, CCLR1 and CCLR0 in TMG should both be set to 1.

Figure 9.16 shows an example of the operation in this case.



Figure 9.16 Timer G Application Example



# 9.6 Watchdog Timer

#### 9.6.1 Overview

The watchdog timer has an 8-bit counter that is incremented by an input clock. If a system runaway allows the counter value to overflow before being rewritten, the watchdog timer can reset the chip internally.

#### (1) Features

Features of the watchdog timer are given below.

- Ten internal clocks (φ/64, φ/128, φ/256, φ/512, φ/1024, φ/2048, φ/4096, φ/8192, φw/32, or watchdog on-chip oscillator) are available for selection for use by the counter.
- A reset signal is generated when the counter overflows. The overflow period can be set from 1 to 256 times the selected clock (from approximately 4 ms to 1,000 ms when  $\phi = 2.00$  MHz).
- Use of module standby mode enables this module to be placed in standby mode independently when not used. See section 5.9, Module Standby Mode, for details.

#### (2) Block Diagram



Figure 9.17 shows a block diagram of the watchdog timer.

Figure 9.17 Block Diagram of Watchdog Timer

#### (3) Register Configuration

Table 9.16 shows the register configuration of the watchdog timer.

#### Table 9.16 Watchdog Timer Registers

Name	Abbr.	R/W	Initial Value	Address
Timer control/status register W	TCSRW	R/W	H'AA	H'FFB2
Timer counter W	TCW	R/W	H'00	H'FFB3
Timer mode register W	TMW	R/W	H'FF	H'FFF8
Clock stop register 2	CKSTPR2	R/W	H'FF	H'FFFB
Port mode register 2	PMR2	R/W	H'D8	H'FFC9



#### 9.6.2 Register Descriptions

#### (1) Timer Control/Status Register W (TCSRW)

Bit	7	6	5	4	3	2	1	0
	B6WI	TCWE	B4WI	TCSRWE	B2WI	WDON	B0WI	WRST
Initial value	1	0	1	0	1	1	1	0
Read/Write	R	(R/W)*	R	(R/W)*	R	(R/W)*	R	(R/W)*

Note: \* Write is enabled only under certain conditions, which are given in the descriptions of the individual bits.

TCSRW is an 8-bit read/write register that controls write access to TCW and TCSRW itself, controls watchdog timer operations, and indicates operating status.

#### Bit 7—Bit 6 Write Disable (B6WI)

Bit 7 controls the writing of data to bit 6 in TCSRW.

Bit 7 B6WI	Description	
0	Bit 6 is write-enabled	
1	Bit 6 is write-protected	(initial value)

This bit is always read as 1. Data written to this bit is not stored.

#### Bit 6—Timer Counter W Write Enable (TCWE)

Bit 6 controls the writing of data to TCW.

Bit 6 TCWE	Description	
0	Data cannot be written to TCW	(initial value)
1	Data can be written to TCW	

#### Bit 5—Bit 4 Write Disable (B4WI)

Bit 5 controls the writing of data to bit 4 in TCSRW.

Bit 5 B4WI	Description	
0	Bit 4 is write-enabled	
1	Bit 4 is write-protected	(initial value)

This bit is always read as 1. Data written to this bit is not stored.

#### Bit 4—Timer Control/Status Register W Write Enable (TCSRWE)

Bit 4 controls the writing of data to bits 2 and 0 in TCSRW.

Bit 4 TCSRWE	Description	
0	Data cannot be written to bits 2 and 0	(initial value)
1	Data can be written to bits 2 and 0	

#### Bit 3—Bit 2 Write Inhibit (B2WI)

Bit 3 controls the writing of data to bit 2 in TCSRW.

Bit 3 B2WI	Description	
0	Bit 2 is write-enabled	
1	Bit 2 is write-protected	(initial value)

This bit is always read as 1. Data written to this bit is not stored.



# Bit 2—Watchdog Timer On (WDON)

Bit 2 enables watchdog timer operation.

Bit 2 WDON	Description	
0	Watchdog timer operation is disabled	
	Clearing condition: When TCSRWE is set to 1 and 0 is written to B2WI and WDON. Note that a reset sets WDON to 1.	
1	Watchdog timer operation is enabled	(initial value)
	Setting condition: Reset, or when TCSRWE is set to 1 and 0 is written to B2WI and 1 is written to WDON	

Counting starts when this bit is set to 1, and stops when this bit is cleared to 0.

#### Bit 1—Bit 0 Write Inhibit (B0WI)

Bit 1 controls the writing of data to bit 0 in TCSRW.

Bit 1 B0WI	Description	
0	Bit 0 is write-enabled	
1	Bit 0 is write-protected	(initial value)

This bit is always read as 1. Data written to this bit is not stored.



#### Bit 0—Watchdog Timer Reset (WRST)

Bit 0 indicates that TCW has overflowed, generating an internal reset signal. The internal reset signal generated by the overflow resets the entire chip. WRST is cleared to 0 by a reset from the  $\overline{\text{RES}}$  pin, or when software writes 0.

Bit 0 WRST	Description
0	Clearing conditions: Reset by RES pin When TCSRWE = 1, and 0 is written in both B0WI and WRST
1	Setting condition: When TCW overflows and an internal reset signal is generated

#### (2) Timer Counter W (TCW)

Bit	7	6	5	4	3	2	1	0
	TCW7	TCW6	TCW5	TCW4	TCW3	TCW2	TCW1	TCW0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W							

TCW is an 8-bit read/write up-counter that counts up by the internal clock. The clock source is selected based on the timer mode register (TMW) setting if WDCKS is 0 and is  $\phi$ w/32 if WDCKS is 1. TCW is always read or written to by the CPU.

When TCW overflows from H'FF to H'00, an internal reset signal is generated and WRST is set to 1 in TCSRW. Upon reset, TCW is initialized to H'00.



#### (3) Timer Mode Register (TMW)

Bit	7	6	5	4	3	2	1	0
	_		—	—	CKS3	CKS2	CKS1	CKS0
Initial value	1	1	1	1	1	1	1	1
Read/Write	—	—	—	—	R/W	R/W	R/W	R/W

The input clock is selected using combinations of CKS3 to CKS0.

#### Bits 7 to 4-Reserved

These bits are always read as 1.

#### Bits 3 to 0—Clock Select (CKS3 to CKS0)

These bits are used to select the clock input to TCW from among 10 internal options. Clock source selection using this register is enabled when WDCKS in port mode register 2 (PMR2) is cleared to 0. If WDCKS is set to 1 the  $\phi$ w/32 clock source is selected, regardless of the settings of the bits in this register.

Bit 3 CKS3	Bit 2 CKS2	Bit 1 CKS1	Bit 0 CKS0	Description	
1	0	0	0	Internal clock: <a>[</a> /64 count	
			1	Internal clock: \$\phi/128 count	
		1	0	Internal clock: φ/256 count	
			1	Internal clock: <a>[1] fill provide the second seco</a>	
	1	0	0	Internal clock: \$\phi/1024 count	
			1	Internal clock: φ/2048 count	
		1	0	Internal clock: <a>phi/4096</a> count	
			1	Internal clock: ø/8192 count	(initial value)
0	Х	Х	Х	Watchdog on-chip oscillator	

Note: X: Don't care

#### (4) Clock Stop Register 2 (CKSTPR2)

Bit	7	6	5	4	3	2	1	0
	LVDCKSTP	—		PW2CKSTP	AECKSTP	WDCKSTP	PW1CKSTP	LDCKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	_		R/W	R/W	R/W	R/W	R/W

CKSTPR2 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to the watchdog timer is described here. For details of the other bits, see the sections on the relevant modules.

#### Bit 2—Watchdog Timer Module Standby Mode Control (WDCKSTP)

Bit 2 controls setting and clearing of module standby mode for the watchdog timer.

0	Watchdog timer is set to module standby mode	
1	Watchdog timer module standby mode is cleared	(initial value)
Note:	WDCKSTP is valid when the WDON bit is cleared to 0 in timer control/status (TCSRW). If WDCKSTP is set to 0 while WDON is set to 1 (during watchdog operation), 0 will be set in WDCKSTP but the watchdog timer will continue it function and will not enter module standby mode. When the watchdog functi WDON is cleared to 0 by software, the WDCKSTP setting will become valid watchdog timer will enter module standby mode.	g timer s watchdog on ends and



#### (5) Port Mode Register 2 (PMR2)

Bit	7	6	5	4	3	2	1	0
	—	_	POF1			WDCKS	NCS	IRQ0
Initial value	1	1	0	1	1	0	0	0
Read/Write	—	—	R/W	—	—	R/W	R/W	R/W

PMR2 is an 8-bit read/write register, mainly controlling the selection of pin functions for port 2. Only the bit relating to the watchdog timer is described here. For details of the other bits, see section 8, I/O Ports.

#### Bit 2—Watchdog Timer Source Clock Select (WDCKS)

This bit selects the watchdog timer source clock.

WDCKS	Description	
0	Selects clock based on timer mode register W (TMW) setting	(initial value)
1	φw/32 selected	

#### 9.6.3 Timer Operation

The watchdog timer has an 8-bit counter (TCW) that is incremented by clock input. The input clock is selected by the WDCKS in port mode register 2 (PMR2). If WDCKS is cleared to 0 the clock selection is specified by the setting of timer mode register W (TMW), and if WDCKS is set to 1 the  $\phi$ w/32 clock source is selected. When TCSRWE = 1 in TCSRW, if 0 is written in B2WI and 1 is simultaneously written in WDON, TCW starts counting up. (Write access to TCSRW is required twice to turn on the watchdog timer. However, WDON is set to 1 after a reset is cancelled, TCW starts to be incremented even without gaining write access to TCSRW.) When the TCW count value reaches H'FF, the next clock input causes the watchdog timer to overflow, and an internal reset signal is generated one base clock ( $\phi$  or  $\phi_{SUB}$ ) cycle later. The internal reset signal is output for 512 clock cycles of the  $\phi_{osc}$  clock. It is possible to write to TCW, causing TCW to count up from the written value. The overflow period can be set in the range from 1 to 256 input clocks, depending on the value written in TCW.

Figure 9.18 shows an example of watchdog timer operations.







#### 9.6.4 Watchdog Timer Operation States

Table 9.17 summarizes the watchdog timer operation states for the H8/38524 Group.

 Table 9.17
 Watchdog Timer Operation States

Operation Mode	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module Standby
TCW	Reset	Functions	Functions		Functions/ Halted <sup>*1</sup>		Functions/ Halted <sup>*2</sup>	Halted
TCSRW	Reset	Functions	Functions		Functions/ Halted <sup>*1</sup>			
TMW	Reset	Functions	Functions		Functions/ Halted <sup>*1</sup>			

RENESAS

Notes: 1. Operates when  $\frac{\phi w}{32}$  or the on-chip oscillator is selected as the internal clock.

2. Operates only when the on-chip oscillator is selected.



# 9.7 Asynchronous Event Counter (AEC)

#### 9.7.1 Overview

The asynchronous event counter is incremented by external event clock or internal clock input.

#### (1) Features

Features of the asynchronous event counter are given below.

• Can count asynchronous events

Can count external events input asynchronously without regard to the operation of base clocks  $\varphi$  and  $\varphi_{_{SUB}}.$ 

The counter has a 16-bit configuration, enabling it to count up to  $65536 (2^{16})$  events.

- Can also be used as two independent 8-bit event counter channels.
- Can be used as single-channel independent 16-bit event counter.
- Event/clock input is enabled only when IRQAEC is high or event counter PWM output (IECPWM) is high.
- Both edge sensing can be used for IRQAEC or event counter PWM output (IECPWM) interrupts. When the asynchronous counter is not used, independent interrupt function use is possible.
- When an event counter PWM is used, event clock input enabling/disabling can be performed automatically in a fixed cycle.
- External event input or a prescaler output clock can be selected by software for the ECH and ECL clock sources.  $\phi/2$ ,  $\phi/4$ , or  $\phi/8$  can be selected as the prescaler output clock.
- Both edge counting is possible for AEVL and AEVH.
- Counter resetting and halting of the count-up function controllable by software
- Automatic interrupt generation on detection of event counter overflow
- Use of module standby mode enables this module to be placed in standby mode independently when not used.



#### (2) Block Diagram



Figure 9.19 shows a block diagram of the asynchronous event counter.

Figure 9.19 Block Diagram of Asynchronous Event Counter



## (3) Pin Configuration

Table 9.18 shows the asynchronous event counter pin configuration.

### Table 9.18Pin Configuration

Name	Abbr.	I/O	Function
Asynchronous event input H	AEVH	Input	Event input pin for input to event counter H
Asynchronous event input L	AEVL	Input	Event input pin for input to event counter L
Event input enable interrupt input	IRQAEC	Input	Input pin for interrupt enabling event input

#### (4) Register Configuration

Table 9.19 shows the register configuration of the asynchronous event counter.

Table 9.19         Asynchronous Event Counter Register;	<b>Table 9.19</b>	Asynchronous Ev	vent Counter	Registers
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Name	Abbr.	R/W	Initial Value	Address
Event counter PWM compare register H	ECPWCRH	R/W	H'FF	H'FF8C
Event counter PWM compare register L	ECPWCRL	R/W	H'FF	H'FF8D
Event counter PWM data register H	ECPWDRH	W	H'00	H'FF8E
Event counter PWM data register L	ECPWDRL	W	H'00	H'FF8F
Input pin edge select register	AEGSR	R/W	H'00	H'FF92
Event counter control register	ECCR	R/W	H'00	H'FF94
Event counter control/status register	ECCSR	R/W	H'00	H'FF95
Event counter H	ECH	R	H'00	H'FF96
Event counter L	ECL	R	H'00	H'FF97
Clock stop register 2	CKSTPR2	R/W	H'FF	H'FFFB

#### 9.7.2 Register Configurations

#### (1) Event Counter PWM Compare Register H (ECPWCRH)

Bit	7	6	5	4	3	2	1	0
	ECPWCRH7	ECPWCRH6	ECPWCRH5	ECPWCRH4	ECPWCRH3	ECPWCRH2	ECPWCRH1	ECPWCRH0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

Note: When ECPWME in AEGSR is 1, event counter PWM is operating and therefore ECPWCRH should not be modified.

When changing the conversion period, event counter PWM must be halted by clearing ECPWME to 0 in AEGSR before modifying ECPWCRH.

ECPWCRH is an 8-bit read/write register that sets the event counter PWM waveform conversion period.

#### (2) Event Counter PWM Compare Register L (ECPWCRL)

Bit	7	6	5	4	3	2	1	0
	ECPWCRL7	ECPWCRL6	ECPWCRL5	ECPWCRL4	ECPWCRL3	ECPWCRL2	ECPWCRL1	ECPWCRL0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

Note: When ECPWME in AEGSR is 1, event counter PWM is operating and therefore ECPWCRL should not be modified.

When changing the conversion period, event counter PWM must be halted by clearing ECPWME to 0 in AEGSR before modifying ECPWCRL.

ECPWCRL is an 8-bit read/write register that sets the event counter PWM waveform conversion period.



#### (3) Event Counter PWM Data Register H (ECPWDRH)

Bit	7	6	5	4	3	2	1	0
	ECPWDRH7	ECPWDRH6	ECPWDRH5	ECPWDRH4	ECPWDRH3	ECPWDRH2	ECPWDRH1	ECPWDRH0
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

Note: When ECPWME in AEGSR is 1, event counter PWM is operating and therefore ECPWDRH should not be modified.

When changing the data, event counter PWM must be halted by clearing ECPWME to 0 in AEGSR before modifying ECPWDRH.

ECPWDRH is an 8-bit write-only register that controls event counter PWM waveform generator data.

#### (4) Event Counter PWM Data Register L (ECPWDRL)

Bit	7	6	5	4	3	2	1	0
	ECPWDRL7	ECPWDRL6	ECPWDRL5	ECPWDRL4	ECPWDRL3	ECPWDRL2	ECPWDRL1	ECPWDRL0
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

Note: When ECPWME in AEGSR is 1, event counter PWM is operating and therefore ECPWDRL should not be modified.

When changing the data, event counter PWM must be halted by clearing ECPWME to 0 in AEGSR before modifying ECPWDRL.

ECPWDRL is an 8-bit write-only register that controls event counter PWM waveform generator data.

#### (5) Input Pin Edge Selection Register (AEGSR)

Bit	7	6	5	4	3	2	1	0
	AHEGS1	AHEGS0	ALEGS1	ALEGS0	AIEGS1	AIEGS0	ECPWME	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W						

AEGSR is an 8-bit read/write register that selects rising, falling, or both edge sensing for the AEVH, AEVL, and IRQAEC pins.

#### Bits 7 and 6—AEC Edge Select H

Bits 7 and 6 select rising, falling, or both edge sensing for the AEVH pin.

Bit 7 AHEGS1	Bit 6 AHEGS0	Description	
0	0	Falling edge on AEVH pin is sensed	(initial value)
	1	Rising edge on AEVH pin is sensed	
1	0	Both edges on AEVH pin are sensed	
	1	Use prohibited	

#### Bits 5 and 4—AEC Edge Select L

Bits 5 and 4 select rising, falling, or both edge sensing for the AEVL pin.

Bit 5 ALEGS1	Bit 4 ALEGS0	Description	
0	0	Falling edge on AEVL pin is sensed	(initial value)
	1	Rising edge on AEVL pin is sensed	
1	0	Both edges on AEVL pin are sensed	
	1	Use prohibited	



#### Bits 3 and 2—IRQAEC Edge Select

Bits 3 and 2 select rising, falling, or both edge sensing for the IRQAEC pin.Bit 3Bit 2

AIEGS1	AIEGS0	Description	
0	0	Falling edge on IRQAEC pin is sensed	(initial value)
	1	Rising edge on IRQAEC pin is sensed	
1	0	Both edges on IRQAEC pin are sensed	
	1	Use prohibited	

#### Bit 1—Event Counter PWM Enable

Bit 1 controls enabling/disabling of event counter PWM and selection/deselection of IRQAEC.

Bit 1 ECPWME	Description	
0	AEC PWM halted, IRQAEC selected	(initial value)
1	AEC PWM operation enabled, IRQAEC deselected	

#### Bit 0—Reserved

Bit 0 is a readable/writable reserved bit. It is initialized to 0 by a reset.

Note: Do not set this bit to 1.



#### (6) Event Counter Control Register (ECCR)

Bit	7	6	5	4	3	2	1	0
	ACKH1	ACKH0	ACKL1	ACKL0	PWCK2	PWCK1	PWCK0	_
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W						

ECCR performs counter input clock and IRQAEC/IECPWM control.

#### Bits 7 and 6—AEC Clock Select H (ACKH1, ACKH0)

Bits 7 and 6 select the clock used by ECH.

Bit 7 ACKH1	Bit 6 ACKH0	Description	
0	0	AEVH pin input	(initial value)
	1	ф/2	
1	0	ф/4	
	1	ф/8	

#### Bits 5 and 4—AEC Clock Select L (ACKL1, ACKL0)

Bits 5 and 4 select the clock used by ECL.

Bit 5 ACKL1	Bit 4 ACKL0	Description	
0	0	AEVL pin input	(initial value)
	1	φ/2	
1	0	φ/4	
	1	φ/8	

#### Bits 3 to 1—Event Counter PWM Clock Select (PWCK2, PWCK1, PWCK0)

Bit 3 PWCK2	Bit 2 PWCK1	Bit 1 PWCK0	Description	
0	0	0	φ/2	(initial value)
		1	φ/4	
	1	0	φ/8	
		1	φ/16	
1	*	0	ф/32	
		1	ф/64	

Bits 3 to 1 select the event counter PWM clock.

\*: Don't care

#### Bit 0—Reserved

Bit 0 is a readable/writable reserved bit. It is initialized to 0 by a reset.

Note: Do not set this bit to 1.

#### (7) Event Counter Control/Status Register (ECCSR)

Bit	7	6	5	4	3	2	1	0
	OVH	OVL		CH2	CUEH	CUEL	CRCH	CRCL
Initial Value	0	0	0	0	0	0	0	0
Read/Write	R/W*	R/W*	R/W	R/W	R/W	R/W	R/W	R/W

Note: \* Bits 7 and 6 can only be written with 0, for flag clearing.

ECCSR is an 8-bit read/write register that controls counter overflow detection, counter resetting, and halting of the count-up function.

ECCSR is initialized to H'00 upon reset.

#### Bit 7—Counter Overflow H (OVH)

Bit 7 is a status flag indicating that ECH has overflowed from H'FF to H'00. This flag is set when ECH overflows. It is cleared by software but cannot be set by software. OVH is cleared by reading it when set to 1, then writing 0.

When ECH and ECL are used as a 16-bit event counter with CH2 cleared to 0, OVH functions as a status flag indicating that the 16-bit event counter has overflowed from H'FFFF to H'0000.

Bit 7 OVH	Description	
0	ECH has not overflowed Clearing condition: After reading OVH = 1, cleared by writing 0 to OVH	(initial value)
1	ECH has overflowed Setting condition: Set when ECH overflows from H'FF to H'00	

#### Bit 6—Counter Overflow L (OVL)

Bit 6 is a status flag indicating that ECL has overflowed from H'FF to H'00. This flag is set when ECL overflows. It is cleared by software but cannot be set by software. OVL is cleared by reading it when set to 1, then writing 0.

Bit 6 OVL	Description	
0	ECL has not overflowed Clearing condition: After reading OVL = 1, cleared by writing 0 to OVL	(initial value)
1	ECL has overflowed Setting condition: Set when ECL overflows from H'FF to H'00	

#### Bit 5—Reserved

Bit 5 is a readable/writable reserved bit. It is initialized to 0 by a reset.



## Bit 4—Channel Select (CH2)

Bit 4 selects whether ECH and ECL are used as a single-channel 16-bit event counter or as two independent 8-bit event counter channels. When CH2 is cleared to 0, ECH and ECL function as a 16-bit event counter which is incremented each time an event clock is input to the AEVL pin. In this case, the overflow signal from ECL is selected as the ECH input clock. When CH2 is set to 1, ECH and ECL function as independent 8-bit event counters which are incremented each time an event clock is input to the AEVH or AEVL pin, respectively.

Bit 4 CH2	Description				
0	ECH and ECL are used together as a single-channel 16-bit event counter				
	(initial value)				
1	ECH and ECL are used as two independent 8-bit event counter channels				

#### Bit 3—Count-up Enable H (CUEH)

Bit 3 enables event clock input to ECH. When 1 is written to this bit, event clock input is enabled and increments the counter. When 0 is written to this bit, event clock input is disabled and the ECH value is held. The AEVH pin or the ECL overflow signal can be selected as the event clock source by bit CH2.

Bit 3 CUEH	Description	
0	ECH event clock input is disabled ECH value is held	(initial value)
1	ECH event clock input is enabled	

#### Bit 2—Count-up Enable L (CUEL)

Bit 2 enables event clock input to ECL. When 1 is written to this bit, event clock input is enabled and increments the counter. When 0 is written to this bit, event clock input is disabled and the ECL value is held.

Bit 2 CUEL	Description	
0	ECL event clock input is disabled ECL value is held	(initial value)
1	ECL event clock input is enabled	

#### Bit 1—Counter Reset Control H (CRCH)

Bit 1 controls resetting of ECH. When this bit is cleared to 0, ECH is reset. When 1 is written to this bit, the counter reset is cleared and the ECH count-up function is enabled.

Bit 1 CRCH	Description	
0	ECH is reset	(initial value)
1	ECH reset is cleared and count-up function is enabled	

#### Bit 0—Counter Reset Control L (CRCL)

Bit 0 controls resetting of ECL. When this bit is cleared to 0, ECL is reset. When 1 is written to this bit, the counter reset is cleared and the ECL count-up function is enabled.

Bit 0 CRCL	Description	
0	ECL is reset	(initial value)
1	ECL reset is cleared and count-up function is enabled	

#### (8) Event Counter H (ECH)

Bit	7	6	5	4	3	2	1	0
	ECH7	ECH6	ECH5	ECH4	ECH3	ECH2	ECH1	ECH0
Initial Value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

ECH is an 8-bit read-only up-counter that operates either as an independent 8-bit event counter or as the upper 8-bit up-counter of a 16-bit event counter configured in combination with ECL. The external asynchronous event AEVH pin,  $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ , or the overflow signal from lower 8-bit counter ECL can be selected as the input clock source. ECH can be cleared to H'00 by software, and is also initialized to H'00 upon reset.



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#### (9) Event Counter L (ECL)

Bit	7	6	5	4	3	2	1	0
	ECL7	ECL6	ECL5	ECL4	ECL3	ECL2	ECL1	ECL0
Initial Value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

ECL is an 8-bit read-only up-counter that operates either as an independent 8-bit event counter or as the lower 8-bit up-counter of a 16-bit event counter configured in combination with ECH. The event clock from the external asynchronous event AEVL pin,  $\phi/2$ ,  $\phi/4$ , or  $\phi/8$  is used as the input clock source. ECL can be cleared to H'00 by software, and is also initialized to H'00 upon reset.

#### (10) Clock Stop Register 2 (CKSTPR2)

Bit	7	6	5	4	3	2	1	0
	LVDCKSTP	_	_	PW2CKSTP	AECKSTP	WDCKSTP	PW1CKSTP	LDCKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	_	_	R/W	R/W	R/W	R/W	R/W

CKSTPR2 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to the asynchronous event counter is described here. For details of the other bits, see the sections on the relevant modules.

#### Bit 3—Asynchronous Event Counter Module Standby Mode Control (AECKSTP)

Bit 3 controls setting and clearing of module standby mode for the asynchronous event counter.

AECKSTP	Description	
0	Asynchronous event counter is set to module standby mode	
1	Asynchronous event counter module standby mode is cleared	(initial value)



#### 9.7.3 Operation

#### (1) 16-bit Event Counter Operation

When bit CH2 is cleared to 0 in ECCSR, ECH and ECL operate as a 16-bit event counter.

Any of four input clock sources— $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ , or AEVL pin input—can be selected by means of bits ACKL1 and ACKL0 in ECCR.

When AEVL pin input is selected, input sensing is selected with bits ALEGS1 and ALEGS0.

The input clock is enabled only when IRQAEC is high or IECPWM is high. When IRQAEC is low or IECPWM is low, the input clock is not input to the counter, which therefore does not operate. Figure 9.20 shows an example of the software processing when ECH and ECL are used as a 16-bit event counter.



Figure 9.20 Example of Software Processing when Using ECH and ECL as 16-Bit Event Counter

As CH2 is cleared to 0 by a reset, ECH and ECL operate as a 16-bit event counter after a reset, and as ACKL1 and ACKL0 are cleared to 00, the operating clock is asynchronous event input from the AEVL pin (using falling edge sensing). When the next clock is input after the count value reaches H'FF in both ECH and ECL, ECH and ECL overflow from H'FFFF to H'0000, the OVH flag is set to 1 in ECCSR, the ECH and ECL count values each return to H'00, and counting up is restarted. When overflow occurs, the IRREC bit is set to 1 in IRR2. If the IENEC bit in IENR2 is 1 at this time, an interrupt request is sent to the CPU.

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# (2) 8-bit Event Counter Operation

When bit CH2 is set to 1 in ECCSR, ECH and ECL operate as independent 8-bit event counters.

 $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ , or AEVH pin input can be selected as the input clock source for ECH by means of bits ACKH1 and ACKH0 in ECCR, and  $\phi/2$ ,  $\phi/4$ ,  $\phi/8$ , or AEVL pin input can be selected as the input clock source for ECL by means of bits ACKL1 and ACKL0 in ECCR.

Input sensing is selected with bits AHEGS1 and AHEGS0 when AEVH pin input is selected, and with bits ALEGS1 and ALEGS0 when AEVL pin input is selected.

The input clock is enabled only when IRQAEC is high or IECPWM is high. When IRQAEC is low or IECPWM is low, the input clock is not input to the counter, which therefore does not operate. Figure 9.21 shows an example of the software processing when ECH and ECL are used as 8-bit event counters.



Figure 9.21 Example of Software Processing when Using ECH and ECL as 8-Bit Event Counters

ECH and ECL can be used as 8-bit event counters by carrying out the software processing shown in the example in figure 9.21. When the next clock is input after the ECH count value reaches H'FF, ECH overflows, the OVH flag is set to 1 in ECCSR, the ECH count value returns to H'00, and counting up is restarted. Similarly, when the next clock is input after the ECL count value reaches H'FF, ECL overflows, the OVL flag is set to 1 in ECCSR, the ECL count value returns to H'00, and counting up is restarted. When overflow occurs, the IRREC bit is set to 1 in IRR2. If the IENEC bit in IENR2 is 1 at this time, an interrupt request is sent to the CPU.



#### (3) IRQAEC Operation

When ECPWME in AEGSR is 0, the ECH and ECL input clocks are enabled only when IRQAEC is high. When IRQAEC is low, the input clocks are not input to the counters, and so ECH and ECL do not count. ECH and ECL count operations can therefore be controlled from outside by controlling IRQAEC. In this case, ECH and ECL cannot be controlled individually.

IRQAEC can also operate as an interrupt source. In this case the vector number is 6 and the vector addresses are H'000C and H'000D.

Interrupt enabling is controlled by IENEC2 in IENR1. When an IRQAEC interrupt is generated, IRR1 interrupt request flag IRREC2 is set to 1. If IENEC2 in IENR1 is set to 1 at this time, an interrupt request is sent to the CPU.

Rising, falling, or both edge sensing can be selected for the IRQAEC input pin, with bits AIAGS1 and AIAGS0 in AEGSR.

Note: The control of switching between the system clock oscillator and the on-chip oscillator during resets should be performed by setting the IRQAEC input level. Refer to section 4, Clock Pulse Generators, for details.

#### (4) Event Counter PWM Operation

When ECPWME in AEGSR is 1, the ECH and ECL input clocks are enabled only when event counter PWM output (IECPWM) is high. When IECPWM is low, the input clocks are not input to the counters, and so ECH and ECL do not count. ECH and ECL count operations can therefore be controlled cyclically from outside by controlling event counter PWM. In this case, ECH and ECL cannot be controlled individually.

IECPWM can also operate as an interrupt source. In this case the vector number is 6 and the vector addresses are H'000C and H'000D.

Interrupt enabling is controlled by IENEC2 in IENR1. When an IECPWM interrupt is generated, IRR1 interrupt request flag IRREC2 is set to 1. If IENEC2 in IENR1 is set to 1 at this time, an interrupt request is sent to the CPU.

Rising, falling, or both edge detection can be selected for IECPWM interrupt sensing with bits AIAGS1 and AIAGS0 in AEGSR.



Figure 9.22 and table 9.20 show examples of event counter PWM operation.



Figure 9.22 Event Counter Operation Waveform

Note:  $N_{dr}$  and  $N_{cm}$  above must be set so that  $N_{dr} < N_{cm}$ . If the settings do not satisfy this condition, do not set ECPWME in AEGSR to 1.

#### Table 9.20 Examples of Event Counter PWM Operation

Conditions:  $f_{osc} = 4 \text{ MHz}$ ,  $f_{\phi} = 2 \text{ MHz}$ , high-speed active mode, ECPWCR value ( $N_{cm}$ ) = H'7A11, ECPWDR value ( $N_{dr}$ ) = H'16E3

Clock Source Selection	Clock Source Cycle (T) <sup>*</sup>	ECPWCR Value (N <sub>cm</sub> )	ECPWDR Value (N <sub>dr</sub> )	$t_{_{off}} = T \times (N_{_{dr}} + 1)$	$t_{cm} = T \times (N_{cm} + 1)$	$\mathbf{t}_{_{\mathrm{on}}} = \mathbf{t}_{_{\mathrm{cm}}} - \mathbf{t}_{_{\mathrm{off}}}$
ф/2	1 µs	H'7A11 -D'31249 - -		5.86 ms	31.25 ms	25.39 ms
ф/4	2 µs			11.72 ms	62.5 ms	50.78 ms
φ/8	4 µs			23.44 ms	125.0 ms	101.56 ms
ф/16	8 µs			46.88 ms	250.0 ms	203.12 ms
ф/32	16 µs	_		93.76 ms	500.0 ms	406.24 ms
ф/64	32 µs	_		187.52 ms	1000.0 ms	812.48 ms

Note: \* t<sub>off</sub> minimum width

#### (5) Clock Input Enable/Disable Function Operation

The clock input to the event counter can be controlled by the IRQAEC pin when ECPWME in AEGSR is 0, and by event counter PWM output IECPWM when ECPWME in AEGSR is 1. As this function forcibly terminates the clock input by each signal, a maximum error of one count will occur depending the IRQAEC or IECPWM timing.

Figure 9.23 shows an example of the operation of this function.



Figure 9.23 Example of Clock Control Operation



#### 9.7.4 Asynchronous Event Counter Operation Modes

Asynchronous event counter operation modes are shown in table 9.21.

#### Table 9.21 Asynchronous Event Counter Operation Modes

Operation Mode	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module Standby
AEGSR	Reset	Functions	Functions	Retained*1	Functions	Functions	Retained*1	Retained
ECCR	Reset	Functions	Functions	Retained*1	Functions	Functions	Retained*1	Retained
ECCSR	Reset	Functions	Functions	Retained*1	Functions	Functions	Retained <sup>*1</sup>	Retained
ECH	Reset	Functions	Functions	Functions <sup>*1*2</sup>	Functions*2	Functions*2	Functions <sup>*1*2</sup>	Halted
ECL	Reset	Functions	Functions	Functions <sup>*1*2</sup>	Functions*2	Functions*2	Functions <sup>*1*2</sup>	Halted
IRQAEC	Reset	Functions	Functions	Retained*3	Functions	Functions	Retained <sup>*3</sup>	Retained*4
Event counter PWM	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained

Notes: 1. When an asynchronous external event is input, the counter increments but the counter overflow H/L flags are not affected.

2. Operates when asynchronous external events are selected; halted and retained otherwise.

3. Clock control by IRQAEC operates, but interrupts do not.

4. As the clock is stopped in module standby mode, IRQAEC has no effect.



#### 9.7.5 Application Notes

- 1. When reading the values in ECH and ECL, the correct value will not be returned if the event counter increments during the read operation. Therefore, if the counter is being used in the 8-bit mode, clear bits CUEH and CUEL in ECCSR to 0 before reading ECH or ECL. If the counter is being used in the 16-bit mode, clear CUEL only to 0 before reading ECH or ECL.
- 2. Use a clock with a frequency of up to 16 MHz for input to the AEVH and AEVL pins, and ensure that the high and low widths of the clock are at least half the OSC clock cycle duration. The duty cycle is immaterial.

Mode		Maximum AEVH/AEVL Pin Input Clock Frequency
Active (high-speed), sleep (high-speed)		16 MHz
Active (medium-speed), sleep (medium-speed	) (ф/16)	2 • f <sub>osc</sub>
	(¢/32)	f <sub>osc</sub>
	(¢/64)	1/2 • f <sub>osc</sub>
$f_{osc} = 1 \text{ MHz to 4 MHz}$	(¢/128)	1/4 • f <sub>osc</sub>
Watch, subactive, subsleep, standby	(¢w/2)	1000 kHz
	(¢w/4)	500 kHz
φw = 32.768 kHz	(¢w/8)	250 kHz

- 3. When using the clock in the 16-bit mode, set CUEH to 1 first, then set CRCH to 1 in ECCSR. Or, set CUEH and CRCH simultaneously before inputting the clock. After that, do not change the CUEH value while using in the 16-bit mode. Otherwise, an error counter increment may occur. Also, to reset the counter, clear CRCH and CRCL to 0 simultaneously or clear CRCL and CRCH to 0 sequentially, in that order.
- 4. When ECPWME in AEGSR is 1, event counter PWM is operating and therefore ECPWCRH, ECPWCRL, ECPWDRH, and ECPWDRL should not be modified. When changing the data, event counter PWM must be halted by clearing ECPWME to 0 in AEGSR before modifying these registers.
- 5. The event counter PWM data register and event counter PWM compare register must be set so that event counter PWM data register < event counter PWM compare register. If the settings do not satisfy this condition, do not set ECPWME to 1 in AEGSR.
- 6. As synchronization is established internally when an IRQAEC interrupt is generated, a maximum error of 1  $t_{cvc}$  will occur between clock halting and interrupt acceptance.





# Section 10 Serial Communication Interface

# 10.1 Overview

This LSI is provided with one serial communication interface, SCI3.

Serial communication interface 3 (SCI3) can carry out serial data communication in either asynchronous or synchronous mode.

#### 10.1.1 Features

Features of SCI3 are listed below.

- Choice of asynchronous or synchronous mode for serial data communication
  - Asynchronous mode

Serial data communication is performed asynchronously, with synchronization provided character by character. In this mode, serial data can be exchanged with standard asynchronous communication LSIs such as a Universal Asynchronous Receiver/Transmitter (UART) or Asynchronous Communication Interface Adapter (ACIA).

There is a choice of 16 data transfer formats.

Data length	7, 8, 5 bits
Stop bit length	1 or 2 bits
Parity	Even, odd, or none
Receive error detection	Parity, overrun, and framing errors
Break detection	Break detected by reading the $RXD_{_{32}}$ pin level directly when a framing error occurs

- Synchronous mode

Serial data communication is synchronized with a clock. In this mode, serial data can be exchanged with another LSI that has a synchronous communication function.

Data length	8 bits
Receive error detection	Overrun errors



• Full-duplex communication

Separate transmission and reception units are provided, enabling transmission and reception to be carried out simultaneously. The transmission and reception units are both double-buffered, allowing continuous transmission and reception.

- On-chip baud rate generator, allowing any desired bit rate to be selected
- Choice of an internal or external clock as the transmit/receive clock source
- Six interrupt sources: transmit end, transmit data empty, receive data full, overrun error, framing error, and parity error

Note: The system clock generator must be used when carrying out this function.
# 10.1.2 Block Diagram



Figure 10.1 shows a block diagram of SCI3.



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# 10.1.3 Pin Configuration

Table 10.1 shows the SCI3 pin configuration.

# Table 10.1Pin Configuration

Name	Abbr.	I/O	Function
SCI3 clock	SCK <sub>32</sub>	I/O	SCI3 clock input/output
SCI3 receive data input	RXD <sub>32</sub>	Input	SCI3 receive data input
SCI3 transmit data output	TXD <sub>32</sub>	Output	SCI3 transmit data output

# 10.1.4 Register Configuration

Table 10.2 shows the SCI3 register configuration.

#### Table 10.2 Registers

Name	Abbr.	R/W	Initial Value	Address
Serial mode register	SMR	R/W	H'00	H'FFA8
Bit rate register	BRR	R/W	H'FF	H'FFA9
Serial control register 3	SCR3	R/W	H'00	H'FFAA
Transmit data register	TDR	R/W	H'FF	H'FFAB
Serial status register	SSR	R/W	H'84	H'FFAC
Receive data register	RDR	R	H'00	H'FFAD
Transmit shift register	TSR	Protected	—	_
Receive shift register	RSR	Protected	—	_
Bit rate counter	BRC	Protected	—	_
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA
Serial port control register	SPCR	R/W		H'FF91

# **10.2** Register Descriptions

# 10.2.1 Receive Shift Register (RSR)

Bit	7	6	5	4	3	2	1	0
			<u> </u>					
Read/Write								_

RSR is a register used to receive serial data. Serial data input to RSR from the  $RXD_{32}$  pin is set in the order in which it is received, starting from the LSB (bit 0), and converted to parallel data. When one byte of data is received, it is transferred to RDR automatically.

RSR cannot be read or written directly by the CPU.

### 10.2.2 Receive Data Register (RDR)

Bit	7	6	5	4	3	2	1	0
	RDR7	RDR6	RDR5	RDR4	RDR3	RDR2	RDR1	RDR0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

RDR is an 8-bit register that stores received serial data.

When reception of one byte of data is finished, the received data is transferred from RSR to RDR, and the receive operation is completed. RSR is then able to receive data. RSR and RDR are double-buffered, allowing consecutive receive operations.

RDR is a read-only register, and cannot be written by the CPU.

RDR is initialized to H'00 upon reset, and in standby, module standby or watch mode.



# 10.2.3 Transmit Shift Register (TSR)



TSR is a register used to transmit serial data. Transmit data is first transferred from TDR to TSR, and serial data transmission is carried out by sending the data to the  $TXD_{32}$  pin in order, starting from the LSB (bit 0). When one byte of data is transmitted, the next byte of transmit data is transferred to TDR, and transmission started, automatically. Data transfer from TDR to TSR is not performed if no data has been written to TDR (if bit TDRE is set to 1 in the serial status register (SSR)).

TSR cannot be read or written directly by the CPU.

# 10.2.4 Transmit Data Register (TDR)

Bit	7	6	5	4	3	2	1	0
	TDR7	TDR6	TDR5	TDR4	TDR3	TDR2	TDR1	TDR0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

TDR is an 8-bit register that stores transmit data. When TSR is found to be empty, the transmit data written in TDR is transferred to TSR, and serial data transmission is started. Continuous transmission is possible by writing the next transmit data to TDR during TSR serial data transmission.

TDR can be read or written by the CPU at any time.

TDR is initialized to H'FF upon reset, and in standby, module standby, or watch mode.



Bit	7	6	5	4	3	2	1	0
	COM	CHR	PE	PM	STOP	MP	CKS1	CKS0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

# 10.2.5 Serial Mode Register (SMR)

SMR is an 8-bit register used to set the serial data transfer format and to select the clock source for the baud rate generator.

SMR can be read or written by the CPU at any time.

SMR is initialized to H'00 upon reset, and in standby, module standby, or watch mode.

# Bit 7—Communication Mode (COM)

Bit 7 selects whether SCI3 operates in asynchronous mode or synchronous mode.

Bit 7 COM	Description	
0	Asynchronous mode	(initial value)
1	Synchronous mode	

### Bit 6—Character Length (CHR)

Bit 6 selects either 7 or 8 bits as the data length to be used in asynchronous mode. In synchronous mode the data length is always 8 bits, irrespective of the bit 6 setting.

Bit 6 CHR		Description	
0		8-bit data/5-bit data <sup>*2</sup>	(initial value)
1		7-bit data <sup>*1</sup> /5-bit data <sup>*2</sup>	
Notes:	1.	When 7-bit data is selected, the MSB (bit 7) of TDR is not transmitted.	
	2	When 5-bit data is selected, set both PE and MP to 1. The three most s	ignificant hits

 When 5-bit data is selected, set both PE and MP to 1. The three most significant bits (bits 7, 6, and 5) of TDR are not transmitted.



# Bit 5—Parity Enable (PE)

Bit 5 selects whether a parity bit is to be added during transmission and checked during reception in asynchronous mode. In synchronous mode parity bit addition and checking is not performed, irrespective of the bit 5 setting.

Bit 5 PE	Description	
0	Parity bit addition and checking disabled <sup>*2</sup> (	initial value)
1	Parity bit addition and checking enabled <sup>*1/*2</sup>	
Notes:	When PE is set to 1, even or odd parity, as designated by bit PM, is added data before it is sent, and the received parity bit is checked against the parit designated by bit PM.	
	For the case where 5-bit data is selected, see table 10.11.	

### Bit 4—Parity Mode (PM)

Bit 4 selects whether even or odd parity is to be used for parity addition and checking. The PM bit setting is only valid in asynchronous mode when bit PE is set to 1, enabling parity bit addition and checking. The PM bit setting is invalid in synchronous mode, and in asynchronous mode if parity bit addition and checking is disabled.

Bit 4 PM	Description
0	Even parity*1 (initial value
1	Odd parity*2
Notes: 1.	When even parity is selected, a parity bit is added in transmission so that the total number of 1 bits in the transmit data plus the parity bit is an even number; in reception, a check is carried out to confirm that the number of 1 bits in the receive data plus the

parity bit is an even number.When odd parity is selected, a parity bit is added in transmission so that the total number of 1 bits in the transmit data plus the parity bit is an odd number; in reception, a check is carried out to confirm that the number of 1 bits in the receive data plus the parity bit is an odd number.



# Bit 3—Stop Bit Length (STOP)

Bit 3 selects 1 bit or 2 bits as the stop bit length in asynchronous mode. The STOP bit setting is only valid in asynchronous mode. When synchronous mode is selected the STOP bit setting is invalid since stop bits are not added.

Bit 3 STOP	Description	
0	1 stop bit <sup>*1</sup> (init	tial value)
1	2 stop bits <sup>*2</sup>	
Notes:	1. In transmission, a single 1 bit (stop bit) is added at the end of a transmit chara	acter.
	2. In transmission, two 1 bits (stop bits) are added at the end of a transmit chara	cter.

In reception, only the first of the received stop bits is checked, irrespective of the STOP bit setting. If the second stop bit is 1 it is treated as a stop bit, but if 0, it is treated as the start bit of the next transmit character.

### Bit 2— 5-Bit Communication (MP)

When this bit is set to 1, the 5-bit communication format is enabled. When writing 1 to this bit, always write 1 to bit 5 (RE) at the same time.

### Bits 1 and 0—Clock Select 1, 0 (CKS1, CKS0)

Bits 1 and 0 choose  $\phi/64$ ,  $\phi/16$ ,  $\phi w/2$ , or  $\phi$  as the clock source for the baud rate generator.

For the relation between the clock source, bit rate register setting, and baud rate, see section 10.2.8, Bit rate register (BRR).

Bit 1 CKS1	Bit 0 CKS0	Description	
0	0	φ clock	(initial value)
0	1	$\phi$ w/2 clock <sup>*1</sup> / $\phi$ w clock <sup>*2</sup>	
1	0	φ/16 clock	
1	1	φ/64 clock	

Notes: 1.  $\phi$  w/2 clock in active (medium-speed/high-speed) mode and sleep mode

2.  $\phi$  w clock in subactive mode and subsleep mode. In subactive or subsleep mode, SCI3 can be operated when CPU clock is  $\phi$ w/2 only.



### 10.2.6 Serial Control Register 3 (SCR3)

Bit	7	6	5	4	3	2	1	0
	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

SCR3 is an 8-bit register for selecting transmit or receive operation, the asynchronous mode clock output, interrupt request enabling or disabling, and the transmit/receive clock source.

SCR3 can be read or written by the CPU at any time.

SCR3 is initialized to H'00 upon reset, and in standby, module standby or watch mode.

# Bit 7—Transmit Interrupt Enable (TIE)

Bit 7 selects enabling or disabling of the transmit data empty interrupt request (TXI) when transmit data is transferred from the transmit data register (TDR) to the transmit shift register (TSR), and bit TDRE in the serial status register (SSR) is set to 1.

TXI can be released by clearing bit TDRE or bit TIE to 0.

Bit 7		
TIE	Description	
0	Transmit data empty interrupt request (TXI) disabled	(initial value)
1	Transmit data empty interrupt request (TXI) enabled	

# Bit 6—Receive Interrupt Enable (RIE)

Bit 6 selects enabling or disabling of the receive data full interrupt request (RXI) and the receive error interrupt request (ERI) when receive data is transferred from the receive shift register (RSR) to the receive data register (RDR), and bit RDRF in the serial status register (SSR) is set to 1. There are three kinds of receive errors: overrun, framing, and parity.

RXI and ERI can be released by clearing bit RDRF or the FER, PER, or OER error flag to 0, or by clearing bit RIE to 0.

Bit 6 RIE	Description	
0	Receive data full interrupt request (RXI) and receive error interrupt request (ERI) disabled	(initial value)
1	Receive data full interrupt request (RXI) and receive error interrupt request (ERI) enabled	

### Bit 5—Transmit Enable (TE)

Bit 5 selects enabling or disabling of the start of transmit operation.

Bit	5
	5

Descri

TE	Description	
0	Transmit operation disabled <sup>*1</sup> (TXD32 pin is I/O port)	(initial value)
1	Transmit operation enabled*2 (TXD32 pin is transmit data pin)	
Notes: 1	. Bit TDRE in SSR is fixed at 1.	

 When transmit data is written to TDR in this state, bit TDRE in SSR is cleared to 0 and serial data transmission is started. Be sure to carry out serial mode register (SMR) settings, and setting of bit SPC32 in SPCR, to decide the transmission format before setting bit TE to 1.



# Bit 4—Receive Enable (RE)

Bit 4 selects enabling or disabling of the start of receive operation.

Bit 4 RE		Description	
0		Receive operation disabled <sup>*1</sup> (RXD32 pin is I/O port)	(initial value)
1		Receive operation enabled <sup>*2</sup> (RXD32 pin is receive data pin)	
Notes:		Note that the RDRF, FER, PER, and OER flags in SSR are not affecte cleared to 0, and retain their previous state.	d when bit RE is
	2.	In this state, serial data reception is started when a start bit is detected	l in asynchronous

mode or serial clock input is detected in synchronous mode. Be sure to carry out serial mode register (SMR) settings to decide the reception format before setting bit RE to 1.

# Bit 3— Reserved (MPIE)

Bit 3 is reserved.

### Bit 2—Transmit End Interrupt Enable (TEIE)

Bit 2 selects enabling or disabling of the transmit end interrupt request (TEI) if there is no valid transmit data in TDR when MSB data is to be sent.

Bit 2 TEIE		Description	
0		Transmit end interrupt request (TEI) disabled (initial value	e)
1		Transmit end interrupt request (TEI) enabled*	
Note:	*	TEI can be released by clearing bit TDRE to 0 and clearing bit TEND to 0 in SSR, or b clearing bit TEIE to 0.	зу



# Bits 1 and 0—Clock Enable 1 and 0 (CKE1, CKE0)

Bits 1 and 0 select the clock source and enabling or disabling of clock output from the  $SCK_{32}$  pin. The combination of CKE1 and CKE0 determines whether the  $SCK_{32}$  pin functions as an I/O port, a clock output pin, or a clock input pin.

The CKE0 bit setting is only valid in case of internal clock operation (CKE1 = 0) in asynchronous mode. In synchronous mode, or when external clock operation is used (CKE1 = 1), bit CKE0 should be cleared to 0.

After setting bits CKE1 and CKE0, set the operating mode in the serial mode register (SMR).

Description Bit 1 Bit 0 CKE1 CKE0 Communication Mode **Clock Source** SCK,, Pin Function I/O port<sup>\*1</sup> 0 0 Asynchronous Internal clock Serial clock output\*1 Internal clock Synchronous Clock output\*2 0 1 Asynchronous Internal clock Synchronous Reserved Clock input\*3 1 0 Asynchronous External clock Synchronous External clock Serial clock input 1 1 Asynchronous Reserved

Reserved

For details on clock source selection, see table 10.9 in section 10.3.1, Overview.

Notes: 1. Initial value

2. A clock with the same frequency as the bit rate is output.

3. Input a clock with a frequency 16 times the bit rate.

Synchronous



### 10.2.7 Serial Status Register (SSR)

Bit	7	6	5	4	3	2	1	0
	TDRE	RDRF	OER	FER	PER	TEND	MPBR	MPBT
Initial value	1	0	0	0	0	1	0	0
Read/Write	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R/(W)*	R	R	R/W

Note: \* Only a write of 0 for flag clearing is possible.

SSR is an 8-bit register containing status flags that indicate the operational status of SCI3, and multiprocessor bits.

SSR can be read or written to by the CPU at any time, but 1 cannot be written to bits TDRE, RDRF, OER, PER, and FER.

Bits TEND and MPBR are read-only bits, and cannot be modified.

SSR is initialized to H'84 upon reset, and in standby, module standby, or watch mode.

### Bit 7—Transmit Data Register Empty (TDRE)

Bit 7 indicates that transmit data has been transferred from TDR to TSR.

Bit 7 TDRE	Description	
0	Transmit data written in TDR has not been transferred to TSR Clearing conditions: After reading TDRE = 1, cleared by writing 0 to TDRE When data is written to TDR by an instruction	
1	Transmit data has not been written to TDR, or transmit data written in TDR has been transferred to TSR Setting conditions: When bit TE in SCR3 is cleared to 0 When data is transferred from TDR to TSR	(initial value)

#### Bit 6—Receive Data Register Full (RDRF)

Bit 6 indicates that received data is stored in RDR.

Bit 6 RDRI	Description
0	There is no receive data in RDR(initial value)Clearing conditions:After reading RDRF = 1, cleared by writing 0 to RDRFWhen RDR data is read by an instruction
1	There is receive data in RDR Setting condition: When reception ends normally and receive data is transferred from RSR to RDR
Note:	If an error is detected in the receive data, or if the RE bit in SCR3 has been cleared to 0, RDR and bit RDRF are not affected and retain their previous state. Note that if data reception is completed while bit RDRF is still set to 1, an overrun error (OER) will result and the receive data will be lost.

### Bit 5—Overrun Error (OER)

Bit 5 indicates that an overrun error has occurred during reception.

Bit 5 OER	Description
0	Reception in progress or completed <sup>*1</sup> (initial value
	Clearing condition:
	After reading OER = 1, cleared by writing 0 to OER
1	An overrun error has occurred during reception <sup>*2</sup>
	Setting condition:
	When reception is completed with RDRF set to 1
Notes: 1	. When bit RE in SCR3 is cleared to 0, bit OER is not affected and retains its previous
	state.

2. RDR retains the receive data it held before the overrun error occurred, and data received after the error is lost. Reception cannot be continued with bit OER set to 1, and in synchronous mode, transmission cannot be continued either.



# Bit 4—Framing Error (FER)

Bit 4 indicates that a framing error has occurred during reception in asynchronous mode.

Bit 4 FER		Description	
0		Reception in progress or completed <sup>*1</sup> Clearing condition: After reading FER = 1, cleared by writing 0 to FER	(initial value)
1		A framing error has occurred during reception Setting condition: When the stop bit at the end of the receive data is checked for a value of 1 at the end of reception, and the stop bit is 0 <sup>*2</sup>	
Notes:	1.	When bit RE in SCR3 is cleared to 0, bit FER is not affected and retains it state.	s previous
	2.	Note that, in 2-stop-bit mode, only the first stop bit is checked for a value of second stop bit is not checked. When a framing error occurs the receive of the section of the performance of the section	lata is

transferred to RDR but bit RDRF is not set. Reception cannot be continued with bit FER set to 1. In synchronous mode, neither transmission nor reception is possible when bit FER is set to 1.

# Bit 3—Parity Error (PER)

Bit 3 indicates that a parity error has occurred during reception with parity added in asynchronous mode.

Bit 3 PER		Description	
0		Reception in progress or completed <sup>*1</sup> (initial value) Clearing condition: After reading PER = 1, cleared by writing 0 to PER	ne)
1		A parity error has occurred during reception <sup>*2</sup> Setting condition: When the number of 1 bits in the receive data plus parity bit does not match the parity designated by bit PM in the serial mode register (SMR)	
Notes:	1.	When bit RE in SCR3 is cleared to 0, bit PER is not affected and retains its previous state.	;
	2.	Receive data in which a parity error has occurred is still transferred to RDR, but bit RDRF is not set. Reception cannot be continued with bit PER set to 1. In synchronomode, neither transmission nor reception is possible when bit FER is set to 1.	us

#### Bit 2—Transmit End (TEND)

Bit 2 indicates that bit TDRE is set to 1 when the last bit of a transmit character is sent.

Bit 2 is a read-only bit and cannot be modified.

Bit 2 TEND	Description	
0	Transmission in progress Clearing conditions: After reading TDRE = 1, cleared by writing 0 to TDRE When data is written to TDR by an instruction	
1	Transmission ended Setting conditions: When bit TE in SCR3 is cleared to 0 When bit TDRE is set to 1 when the last bit of a transmit character is s	(initial value) ent

#### Bit 1—Reserved (MPBR)

Bit 1 is read-only and reserved. It cannot be written to.

#### Bit 0— Reserved (MPBT)

Bit 0 is reserved. The write value should always be 0.

### 10.2.8 Bit Rate Register (BRR)

Bit	7	6	5	4	3	2	1	0
	BRR7	BRR6	BRR5	BRR4	BRR3	BRR2	BRR1	BRR0
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W							

BRR is an 8-bit register that designates the transmit/receive bit rate in accordance with the baud rate generator operating clock selected by bits CKS1 and CKS0 of the serial mode register (SMR).

BRR can be read or written by the CPU at any time.

BRR is initialized to H'FF upon reset, and in standby, module standby, or watch mode.

Table 10.3 shows examples of BRR settings in asynchronous mode. The values shown are for active (high-speed) mode.

RENESAS

	φ														
	16.4 kHz				19.2 kHz			1 MHz		1.2288 MHz			2 MHz		
Bit Rate (bit/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)
110	—	—	_	—	—	_	2	17	-1.36	2	21	-0.83	3	8	-1.36
150	—	—	_	0	3	0	2	12	0.16	3	3	0	2	25	0.16
200		_	—	0	2	0	2	9	-2.34	3	2	0	3	4	-2.34
250	0	1	2.5	—	—	_	3	1	-2.34	0	153	-0.26	2	15	-2.34
300	—	—	—	0	1	0	0	103	0.16	3	1	0	2	12	0.16
600	—	—	_	0	0	0	0	51	0.16	3	0	0	0	103	0.16
1200				—	—	_	0	25	0.16	2	1	0	0	51	0.16
2400				_	_	_	0	12	0.16	2	0	0	0	25	0.16
4800				_	_	_	_	_	_	0	7	0	0	12	0.16
9600				_	_	_	_	_	_	0	3	0	_	_	_
19200				_		—	—		—	0	1	0	—		_
31250				_	—	—	0	0	0	_	—	—	0	1	0
38400					_	_	_		_	0	0	0	_		_

<b>Table 10.3</b>	Examples of BRR	Settings for Variou	s Bit Rates (Asynchron	nous Mode) (1)

	φ									
		5 MHz			8 MHz			10 MHz		
Bit Rate (bit/s)	n	N	Error (%)	n	N	Error (%)	n	N	Error (%)	
110	3	21	0.88	3	35	-1.36	3	43	0.88	
150	3	15	1.73	3	25	0.16	3	32	-1.36	
200	3	11	1.73	3	19	-2.34	3	23	1.73	
250	3	9	-2.34	3	15	-2.34	3	19	-2.34	
300	3	7	1.73	3	12	0.16	3	15	1.73	
600	3	3	1.73	2	25	0.16	3	7	1.73	
1200	3	1	1.73	2	12	0.16	3	3	1.73	
2400	3	0	1.73	0	103	0.16	3	1	1.73	
4800	2	1	1.73	0	51	0.16	3	0	1.73	
9600	2	0	173	0	25	0.16	2	1	1.73	
19200	0	7	1.73	0	12	0.16	2	0	1.73	
31250	0	4	0	0	7	0	0	9	0	
38400	0	3	1.73	_	_	_	0	7	1.73	

 Table 10.3
 Examples of BRR Settings for Various Bit Rates (Asynchronous Mode) (2)

Notes: No indication: Setting not possible.

-: Setting possible, but errors may result.

1. The value set in BRR is given by the following equation:

$$N = \frac{\varphi}{(32 \times 2^{2n} \times B)} - 1$$

where B: Bit rate (bit/s)

- N: Baud rate generator BRR setting (0  $\leq$  N  $\leq$  255)
- φ: System clock frequency
- n: Baud rate generator input clock number (n = 0, 2, or 3) (The relation between n and the clock is shown in table 10.4.)
- 2. The error in table 10.3 is the value obtained from the following equation, rounded to two decimal places.

 $Error (\%) = \frac{B \text{ (rate obtained from n, N, OSC)} - R(bit rate in left-hand column in table 10.3.)}{R \text{ (bit rate in left-hand column in table 10.3.)}} \times 100$ 



### Table 10.4 Relation between n and Clock

		SMR Setting				
n	Clock	CKS1	CKS0			
0	φ	0	0			
0	φw/2 <sup>*1</sup> /φw <sup>*2</sup>	0	1			
2	ф/16	1	0			
3	ф/64	1	1			

Notes: 1.  $\phi$  w/2 clock in active (medium-speed/high-speed) mode and sleep mode

2.  $\phi$  w clock in subactive mode and subsleep mode

In subactive or subsleep mode, SCI3 can be operated when CPU clock is  $\phi w/2$  only.

Table 10.5 shows the maximum bit rate for each frequency. The values shown are for active (high-speed) mode.

Table 10.5         Maximum Bit Rate for Each Frequency (Asynchronou)	s Mode)
--	---------

		Maximum Bit Rate	Setting		
OSC (MHz)	φ (MHz)	(bit/s)	n	Ν	
0.0384*	0.0192	600	0	0	
2	1	31250	0	0	
2.4576	1.2288	38400	0	0	
4	2	62500	0	0	
10	5	156250	0	0	
16	8	250000	0	0	
20	10	312500	0	0	

Note: \* When SMR is set up to CKS1 = 0, CKS0 = 1.

Table 10.6 shows examples of BRR settings in synchronous mode. The values shown are for active (high-speed) mode.

	φ									
Bit Rate	19.2 kHz				1 MHz	:	2 MHz			
(bit/s)	n	Ν	Error	n	Ν	Error	n	Ν	Error	
200	0	23	0	_	_	_	_			
250	—	—		—	_	—	2	124	0	
300	2	0	0		—	_				
500				—	_	—	—	—	_	
1K				0	249	0				
2.5K				0	99	0	0	199	0	
5K				0	49	0	0	99	0	
10K				0	24	0	0	49	0	
25K				0	9	0	0	19	0	
50K				0	4	0	0	9	0	
100K				_	—	_	0	4	0	
250K				0	0	0	0	1	0	
500K							0	0	0	
1M										

Table 10.6	Examples of BRR Settings for	Various Bit Rates (Synchronous Mode) (1)
		(1)



	φ										
Bit Rate		5 MHz	2		8 MHz	2		10 MHz			
(bit/s)	n	Ν	Error	n	Ν	Error	n	Ν	Error		
200	_	—	_	_		—	0	12499	0		
250	_	_	_	3	124	0	2	624	0		
300	_	_	_	_		_	0	8332	0		
500	—	—	_	2	249	0	0	4999	0		
1K	_	_	_	2	124	0	0	2499	0		
2.5K	—	—	_	2	49	0	0	999	0		
5K	0	249	0	2	24	0	0	499	0		
10K	0	124	0	0	199	0	0	249	0		
25K	0	49	0	0	79	0	0	99	0		
50K	0	24	0	0	39	0	0	49	0		
100K	—	—	_	0	19	0	0	24	0		
250K	0	4	0	0	7	0	0	9	0		
500K	—	—	_	0	3	0	0	4	0		
1M	—	_	_	0	1	0	—	_	_		

### Table 10.6 Examples of BRR Settings for Various Bit Rates (Synchronous Mode) (2)

Blank: Cannot be set.

— : A setting can be made, but an error will result.

Notes: The value set in BRR is given by the following equation:

$$N = \frac{\phi}{(4 \times 2^{2n} \times B)} - 1$$

where B: Bit rate (bit/s)

- N: Baud rate generator BRR setting ( $0 \le N \le 255$ )
- φ: System clock frequency
- n: Baud rate generator input clock number (n = 0, 2, or 3)

(The relation between n and the clock is shown in table 10.7.)

		SMR Setting					
n	Clock	CKS1	CKS0				
0	φ	0	0				
0	$\phi_w/2^{*1}/\phi w^{*2}$	0	1				
2	ф/16	1	0				
3	ф/64	1	1				

#### Table 10.7 Relation between n and Clock

Notes: 1.  $\varphi w/2$  clock in active (medium-speed/high-speed) mode and sleep mode

φw clock in subactive mode and subsleep mode
 In subactive or subsleep mode, SCI3 can be operated when CPU clock is φw/2 only.

# 10.2.9 Clock stop register 1 (CKSTPR1)

Bit	7	6	5	4	3	2	1	0
		_	S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	—	—	R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bits relating to SCI3 are described here. For details of the other bits, see the sections on the relevant modules.

### Bit 5—SCI3 Module Standby Mode Control (S32CKSTP)

Bit 5 controls setting and clearing of module standby mode for SCI3.

#### S32CKSTP Description

0	SCI3 is set to module standby mode <sup>*</sup>	
1	SCI3 module standby mode is cleared	(initial value)
Note:	* All SCI3 register is initialized in module standby mode	

Note: \* All SCI3 register is initialized in module standby mode.



# 10.2.10 Serial Port Control Register (SPCR)

Bit	7	6	5	4	3	2	1	0
	_	_	SPC32		SCINV3	SCINV2		
Initial value	1	1	0		0	0		_
Read/Write			R/W	W	R/W	R/W	W	W

SPCR is an 8-bit readable/writable register that performs  $RXD_{32}$  and  $TXD_{32}$  pin input/output data inversion switching.

#### Bits 7 and 6—Reserved

Bits 7 and 6 are reserved; they are always read as 1 and cannot be modified.

# Bit 5—P4<sub>2</sub>/TXD<sub>32</sub> Pin Function Switch (SPC32)

This bit selects whether pin P42/TXD32 is used as P42 or as TXD32.

Bit 5 SPC32	Description	
0	Functions as $P4_2$ I/O pin	(initial value)
1	Functions as TXD <sub>32</sub> output pin*	
Noto: *	Set the TE hit in SCR3 after setting this hit to 1	

Note: \* Set the TE bit in SCR3 after setting this bit to 1.

### Bit 4—Reserved

Bit 4 is reserved; only 0 can be written to this bit.

### Bit 3—TXD<sub>32</sub> Pin Output Data Inversion Switch

Bit 3 specifies whether or not  $TXD_{32}$  pin output data is to be inverted.

Bit 3 SCINV3	Description	
0	TXD <sub>32</sub> output data is not inverted	(initial value)
1	TXD <sub>32</sub> output data is inverted	

# Bit 2—RXD<sub>32</sub> Pin Input Data Inversion Switch

Bit 2 specifies whether or not RXD<sub>32</sub> pin input data is to be inverted.

Bit 2 SCINV2	Description	
0	RXD <sub>32</sub> input data is not inverted	(initial value)
1	RXD <sub>32</sub> input data is inverted	

#### Bits 1 and 0—Reserved

Bits 1 and 0 are reserved; only 0 can written to these bits.

#### Operation 10.3

#### 10.3.1 **Overview**

SCI3 can perform serial communication in two modes: asynchronous mode in which synchronization is provided character by character, and synchronous mode in which synchronization is provided by clock pulses. The serial mode register (SMR) is used to select asynchronous or synchronous mode and the data transfer format, as shown in table 10.8.

The clock source for SCI3 is determined by bit COM in SMR and bits CKE1 and CKE0 in SCR3, as shown in table 10.9.

#### (1) Asynchronous Mode

- Choice of 5-, 7-, or 8-bit data length
- Choice of parity addition and addition of 1 or 2 stop bits. (The combination of these parameters determines the data transfer format and the character length.)
- Framing error (FER), parity error (PER), overrun error (OER), and break detection during reception
- Choice of internal or external clock as the clock source . When internal clock is selected: SCI3 operates on the baud rate generator clock, and a clock with the same frequency as the bit rate can be output.

When external clock is selected: A clock with a frequency 16 times the bit rate must be input. (The on-chip baud rate generator is not used.)



# (2) Synchronous Mode

- Data transfer format: Fixed 8-bit data length
- Overrun error (OER) detection during reception
- Choice of internal or external clock as the clock source

When internal clock is selected: SCI3 operates on the baud rate generator clock, and a serial clock is output.

When external clock is selected: The on-chip baud rate generator is not used, and SCI3 operates on the input serial clock.

#### Table 10.8 SMR Settings and Corresponding Data Transfer Formats

SMR				Data	Transfer Form	nat		
Bit 7 COM	Bit 6 CHR	Bit 2 MP	Bit 5 PE	Bit 3 STOP	Mode	Data Length	Parity Bit	Stop Bit Length
0	0	0	0	0	Asynchronous	8-bit data	No	1 bit
				1	mode			2 bits
			1	0	-		Yes	1 bit
				1	-			2 bits
	1	_	0	0	-	7-bit data	No	1 bit
				1	-			2 bits
			1	0	-		Yes	1 bit
				1	-			2 bits
	0	1	0	0	Setting			
				1	prohibited			
			1	0	Asynchronous	5-bit data	No	1 bit
				1	mode			2 bits
	1	_	0	0	Setting			
				1	prohibited			
			1	0	Asynchronous	5-bit data	Yes	1 bit
				1	mode			2 bits
	*	0	*	*	Synchronous mode	8-bit data	No	No
								*: Don't ca

RENESAS

Bit 7 B	Bit 1								
		Bit 0		Transmit/Receive Clock					
COM C	CKE1	CKE0	Mode	Clock Source	SCK <sub>32</sub> Pin Function				
0 0	)	0	Asynchronous	Internal	I/O port (SCK <sub>32</sub> pin not used)				
	-	1	mode		Outputs clock with same frequency as bit rate				
1		0	-	External	Inputs clock with frequency 16 times bit rate				
1 0	)	0	Synchronous	Internal	Outputs serial clock				
1		0	mode	External	Inputs serial clock				
0 1		1	Reserved (Do r	not specify these	combinations)				
1 0	)	1	-						
1 1		1	-						

#### Table 10.9 SMR and SCR3 Settings and Clock Source Selection

# (3) Interrupts and Continuous Transmission/Reception

SCI3 can carry out continuous reception using RXI and continuous transmission using TXI. These interrupts are shown in table 10.10.

#### Table 10.10 Transmit/Receive Interrupts

Interrupt	Flags	Interrupt Request Conditions	Notes
RXI	RDRF RIE	When serial reception is performed normally and receive data is transferred from RSR to RDR, bit RDRF is set to 1, and if bit RIE is set to 1 at this time, RXI is enabled and an interrupt is requested. (See figure 10.2(a).)	The RXI interrupt routine reads the receive data transferred to RDR and clears bit RDRF to 0. Continuous reception can be performed by repeating the above operations until reception of the next RSR data is completed.
ΤΧΙ	TDRE TIE	When TSR is found to be empty (on completion of the previous transmission) and the transmit data placed in TDR is transferred to TSR, bit TDRE is set to 1. If bit TIE is set to 1 at this time, TXI is enabled and an interrupt is requested. (See figure 10.2(b).)	The TXI interrupt routine writes the next transmit data to TDR and clears bit TDRE to 0. Continuous transmission can be performed by repeating the above operations until the data transferred to TSR has been transmitted.
TEI	TEND TEIE	When the last bit of the character in TSR is transmitted, if bit TDRE is set to 1, bit TEND is set to 1. If bit TEIE is set to 1 at this time, TEI is enabled and an interrupt is requested. (See figure 10.2(c).)	TEI indicates that the next transmit data has not been written to TDR when the last bit of the transmit character in TSR is sent.







Figure 10.2(b) TDRE Setting and TXI Interrupt



Figure 10.2(c) TEND Setting and TEI Interrupt

#### **Operation in Asynchronous Mode** 10.3.2

In asynchronous mode, serial communication is performed with synchronization provided character by character. A start bit indicating the start of communication and one or two stop bits indicating the end of communication are added to each character before it is sent.

SCI3 has separate transmission and reception units, allowing full-duplex communication. As the transmission and reception units are both double-buffered, data can be written during transmission and read during reception, making possible continuous transmission and reception.

#### (1) **Data Transfer Format**

The general data transfer format in asynchronous communication is shown in figure 10.3.



# Figure 10.3 Data Format in Asynchronous Communication

In asynchronous communication, the communication line is normally in the mark state (high level). SCI3 monitors the communication line and when it detects a space (low level), identifies this as a start bit and begins serial data communication.

One transfer data character consists of a start bit (low level), followed by transmit/receive data (LSB-first format, starting from the least significant bit), a parity bit (high or low level), and finally one or two stop bits (high level).

In asynchronous mode, synchronization is performed by the falling edge of the start bit during reception. The data is sampled on the 8th pulse of a clock with a frequency 16 times the bit period, so that the transfer data is latched at the center of each bit.



Table 10.11 shows the 16 data transfer formats that can be set in asynchronous mode. The format is selected by the settings in the serial mode register (SMR).

	s	MR		Serial Data Transfer Format and Frame Length						gth					
CHR	PE	MP	STOP	1	2	3	4	5	6	7	8	9	10	11	12
0	0	0	0	s			8-k	oit da	ita				STOP		
0	0	0	1	s			8-k	oit da	ita				STOP	STOP	2
0	0	1	0			Se	etting	prol	hibit	ed					
0	0	1	1			Se	etting	prol	hibit	ed					
0	1	0	0	S			8-k	oit da	ita				Р	STOP	2
0	1	0	1	s			8-k	oit da	ita				Р	STOP	STOP
0	1	1	0	s			5-k	oit da	ita	STOP					
0	1	1	1	S			5-k	oit da	ita	STOP	STOP				
1	0	0	0	s			7-k	oit da	ita			STOF			
1	0	0	1	s			7-k	oit da	ita			STOP	STOP		
1	0	1	0			Se	etting	prol	hibit	ed					
1	0	1	1			Se	etting	prol	hibit	ed					
1	1	0	0	s			7-k	oit da	ita			Ρ	STOP		
1	1	0	1	s			7-k	oit da	ita			Ρ	STOP	STOP	
1	1	1	0	S			5-k	oit da	ita	Ρ	STOP				
1	1	1	1	s			5-k	oit da	ita	Р	STOP	STOP			

 Table 10.11 Data Transfer Formats (Asynchronous Mode)

[Legend]

S: Start bit

STOP: Stop bit

P: Parity bit

# (2) Clock

Either an internal clock generated by the baud rate generator or an external clock input at the  $SCK_{32}$  pin can be selected as the SCI3 transmit/receive clock. The selection is made by means of bit COM in SMR and bits SCE1 and CKE0 in SCR3. See table 10.9 for details on clock source selection.

When an external clock is input at the  $SCK_{32}$  pin, the clock frequency should be 16 times the bit rate.

When SCI3 operates on an internal clock, the clock can be output at the  $SCK_{32}$  pin. In this case the frequency of the output clock is the same as the bit rate, and the phase is such that the clock rises at the center of each bit of transmit/receive data, as shown in figure 10.4.



# Figure 10.4 Phase Relationship between Output Clock and Transfer Data (Asynchronous Mode) (8-Bit Data, Parity, 2 Stop Bits)

# (3) Data Transfer Operations

### (a) SCI3 Initialization

Before data is transferred on SCI3, bits TE and RE in SCR3 must first be cleared to 0, and then SCI3 must be initialized as follows.

Note: If the operation mode or data transfer format is changed, bits TE and RE must first be cleared to 0.
When bit TE is cleared to 0, bit TDRE is set to 1.
Note that the RDRF, PER, FER, and OER flags and the contents of RDR are retained when RE is cleared to 0.
When an external clock is used in asynchronous mode, the clock should not be stopped during operation, including initialization. When an external clock is used in synchronous mode, the clock should not be supplied during operation, including initialization.



Figure 10.5 shows an example of a flowchart for initializing SCI3.



Figure 10.5 Example of SCI3 Initialization Flowchart

Renesas

# (b) Transmitting

Figure 10.6 shows an example of a flowchart for data transmission. This procedure should be followed for data transmission after initializing SCI3.



Figure 10.6 Example of Data Transmission Flowchart (Asynchronous Mode)

RENESAS

SCI3 operates as follows when transmitting data.

SCI3 monitors bit TDRE in SSR, and when it is cleared to 0, recognizes that data has been written to TDR and transfers data from TDR to TSR. It then sets bit TDRE to 1 and starts transmitting. If bit TIE in SCR3 is set to 1 at this time, a TXI request is made.

Serial data is transmitted from the  $TXD_{32}$  pin using the relevant data transfer format in table 10.11. When the stop bit is sent, SCI3 checks bit TDRE. If bit TDRE is cleared to 0, SCI3 transfers data from TDR to TSR, and when the stop bit has been sent, starts transmission of the next frame. If bit TDRE is set to 1, bit TEND in SSR bit is set to 1the mark state, in which 1s are transmitted, is established after the stop bit has been sent. If bit TEIE in SCR3 is set to 1 at this time, a TEI request is made.

Figure 10.7 shows an example of the operation when transmitting in asynchronous mode.



Figure 10.7 Example of Operation when Transmitting in Asynchronous Mode (8-Bit Data, Parity, 1 Stop Bit)

# (c) Receiving

Figure 10.8 shows an example of a flowchart for data reception. This procedure should be followed for data reception after initializing SCI3.





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Figure 10.8 Example of Data Reception Flowchart (Asynchronous Mode) (cont)



SCI3 operates as follows when receiving data.

SCI3 monitors the communication line, and when it detects a 0 start bit, performs internal synchronization and begins reception. Reception is carried out in accordance with the relevant data transfer format in table 10.11. The received data is first placed in RSR in LSB-to-MSB order, and then the parity bit and stop bit(s) are received. SCI3 then carries out the following checks.

• Parity check

SCI3 checks that the number of 1 bits in the receive data conforms to the parity (odd or even) set in bit PM in the serial mode register (SMR).

- Stop bit check SCI3 checks that the stop bit is 1. If two stop bits are used, only the first is checked.
- Status check

SCI3 checks that bit RDRF is set to 0, indicating that the receive data can be transferred from RSR to RDR.

If no receive error is found in the above checks, bit RDRF is set to 1, and the receive data is stored in RDR. If bit RIE is set to 1 in SCR3, an RXI interrupt is requested. If the error checks identify a receive error, bit OER, PER, or FER is set to 1 depending on the kind of error. Bit RDRF retains its state prior to receiving the data. If bit RIE is set to 1 in SCR3, an ERI interrupt is requested.

Table 10.12 shows the conditions for detecting a receive error, and receive data processing.

Note: No further receive operations are possible while a receive error flag is set. Bits OER, FER, PER, and RDRF must therefore be cleared to 0 before resuming reception.

Receive Error	Abbr.	<b>Detection Conditions</b>	Receive Data Processing
Overrun error	OER	When the next date receive operation is completed while bit RDRF is still set to 1 in SSR	Receive data is not transferred from RSR to RDR
Framing error	FER	When the stop bit is 0	Receive data is transferred from RSR to RDR
Parity error	PER	When the parity (odd or even) set in SMR is different from that of the received data	Receive data is transferred from RSR to RDR

Table 10.12 Receive Error Detection Conditions and Receive Data Processing





Figure 10.9 shows an example of the operation when receiving in asynchronous mode.

Figure 10.9 Example of Operation when Receiving in Asynchronous Mode (8-Bit Data, Parity, 1 Stop Bit)

# 10.3.3 Operation in Synchronous Mode

In synchronous mode, SCI3 transmits and receives data in synchronization with clock pulses. This mode is suitable for high-speed serial communication.

SCI3 has separate transmission and reception units, allowing full-duplex communication with a shared clock.

As the transmission and reception units are both double-buffered, data can be written during transmission and read during reception, making possible continuous transmission and reception.
#### (1) Data Transfer Format



The general data transfer format in asynchronous communication is shown in figure 10.10.

#### Figure 10.10 Data Format in Synchronous Communication

In synchronous communication, data on the communication line is output from one falling edge of the serial clock until the next falling edge. Data confirmation is guaranteed at the rising edge of the serial clock.

One transfer data character begins with the LSB and ends with the MSB. After output of the MSB, the communication line retains the MSB state.

When receiving in synchronous mode, SCI3 latches receive data at the rising edge of the serial clock.

The data transfer format uses a fixed 8-bit data length.

Parity bits cannot be added.

#### (2) Clock

Either an internal clock generated by the baud rate generator or an external clock input at the  $SCK_{32}$  pin can be selected as the SCI3 serial clock. The selection is made by means of bit COM in SMR and bits CKE1 and CKE0 in SCR3. See table 10.9 for details on clock source selection.

When SCI3 operates on an internal clock, the serial clock is output at the  $SCK_{32}$  pin. Eight pulses of the serial clock are output in transmission or reception of one character, and when SCI3 is not transmitting or receiving, the clock is fixed at the high level.



### (3) Data Transfer Operations

## (a) SCI3 Initialization

Data transfer on SCI3 first of all requires that SCI3 be initialized as described in section 10.3.2 (3), (a) SCI3 Initialization, and shown in figure 10.5.

#### (b) Transmitting

Figure 10.11 shows an example of a flowchart for data transmission. This procedure should be followed for data transmission after initializing SCI3.





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SCI3 operates as follows when transmitting data.

SCI3 monitors bit TDRE in SSR, and when it is cleared to 0, recognizes that data has been written to TDR and transfers data from TDR to TSR. It then sets bit TDRE to 1 and starts transmitting. If bit TIE in SCR3 is set to 1 at this time, a TXI request is made.

When clock output mode is selected, SCI3 outputs 8 serial clock pulses. When an external clock is selected, data is output in synchronization with the input clock.

Serial data is transmitted from the TXD32 pin in order from the LSB (bit 0) to the MSB (bit 7). When the MSB (bit 7) is sent, checks bit TDRE. If bit TDRE is cleared to 0, SCI3 transfers data from TDR to TSR, and starts transmission of the next frame. If bit TDRE is set to 1, SCI3 sets bit TEND to 1 in SSR, and after sending the MSB (bit 7), retains the MSB state. If bit TEIE in SCR3 is set to 1 at this time, a TEI request is made.

After transmission ends, the SCK pin is fixed at the high level.

Note: Transmission is not possible if an error flag (OER, FER, or PER) that indicates the data reception status is set to 1. Check that these error flags are all cleared to 0 before a transmit operation.

Figure 10.12 shows an example of the operation when transmitting in synchronous mode.



Figure 10.12 Example of Operation when Transmitting in Synchronous Mode

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#### (c) Receiving

Figure 10.13 shows an example of a flowchart for data reception. This procedure should be followed for data reception after initializing SCI3.



Figure 10.13 Example of Data Reception Flowchart (Synchronous Mode)



SCI3 operates as follows when receiving data.

SCI3 performs internal synchronization and begins reception in synchronization with the serial clock input or output.

The received data is placed in RSR in LSB-to-MSB order.

After the data has been received, SCI3 checks that bit RDRF is set to 0, indicating that the receive data can be transferred from RSR to RDR.

If this check shows that there is no overrun error, bit RDRF is set to 1, and the receive data is stored in RDR. If bit RIE is set to 1 in SCR3, an RXI interrupt is requested. If the check identifies an overrun error, bit OER is set to 1.

Bit RDRF remains set to 1. If bit RIE is set to 1 in SCR3, an ERI interrupt is requested.

See table 10.12 for the conditions for detecting a receive error, and receive data processing.

Note: No further receive operations are possible while a receive error flag is set. Bits OER, FER, PER, and RDRF must therefore be cleared to 0 before resuming reception.

Figure 10.14 shows an example of the operation when receiving in synchronous mode.



Figure 10.14 Example of Operation when Receiving in Synchronous Mode

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#### (d) Simultaneous transmit/receive

Figure 10.15 shows an example of a flowchart for a simultaneous transmit/receive operation. This procedure should be followed for simultaneous transmission/reception after initializing SCI3.







- Notes: 1. When switching from transmission to simultaneous transmission/reception, check that SCI3 has finished transmitting and that bits TDRE and TEND are set to 1, clear bit TE to 0, and then set bits TE and RE to 1 simultaneously.
  - 2. When switching from reception to simultaneous transmission/reception, check that SCI3 has finished receiving, clear bit RE to 0, then check that bit RDRF and the error flags (OER, FER, and PER) are cleared to 0, and finally set bits TE and RE to 1 simultaneously.

## 10.4 Interrupts

SCI3 can generate six kinds of interrupts: transmit end, transmit data empty, receive data full, and three receive error interrupts (overrun error, framing error, and parity error). These interrupts have the same vector address.

The various interrupt requests are shown in table 10.13.

Interrupt Abbr.	Interrupt Request	Vector Address
RXI	Interrupt request initiated by receive data full flag (RDRF)	H'0024
TXI	Interrupt request initiated by transmit data empty flag (TDRE)	-
TEI	Interrupt request initiated by transmit end flag (TEND)	-
ERI	Interrupt request initiated by receive error flag (OER, FER, PER)	_

#### **Table 10.13 SCI3 Interrupt Requests**

Each interrupt request can be enabled or disabled by means of bits TIE and RIE in SCR3.

When bit TDRE is set to 1 in SSR, a TXI interrupt is requested. When bit TEND is set to 1 in SSR, a TEI interrupt is requested. These two interrupts are generated during transmission.

The initial value of bit TDRE in SSR is 1. Therefore, if the transmit data empty interrupt request (TXI) is enabled by setting bit TIE to 1 in SCR3 before transmit data is transferred to TDR, a TXI interrupt will be requested even if the transmit data is not ready.

Also, the initial value of bit TEND in SSR is 1. Therefore, if the transmit end interrupt request (TEI) is enabled by setting bit TEIE to 1 in SCR3 before transmit data is transferred to TDR, a TEI interrupt will be requested even if the transmit data has not been sent.

Effective use of these interrupt requests can be made by having processing that transfers transmit data to TDR carried out in the interrupt service routine.

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To prevent the generation of these interrupt requests (TXI and TEI), on the other hand, the enable bits for these interrupt requests (bits TIE and TEIE) should be set to 1 after transmit data has been transferred to TDR.

When bit RDRF is set to 1 in SSR, an RXI interrupt is requested, and if any of bits OER, PER, and FER is set to 1, an ERI interrupt is requested. These two interrupt requests are generated during reception.

For further details, see section 3.3, Interrupts.

# **10.5** Application Notes

The following points should be noted when using SCI3.

#### 1. Relation between writes to TDR and bit TDRE

Bit TDRE in the serial status register (SSR) is a status flag that indicates that data for serial transmission has not been prepared in TDR. When data is written to TDR, bit TDRE is cleared to 0 automatically. When SCI3 transfers data from TDR to TSR, bit TDRE is set to 1.

Data can be written to TDR irrespective of the state of bit TDRE, but if new data is written to TDR while bit TDRE is cleared to 0, the data previously stored in TDR will be lost of it has not yet been transferred to TSR. Accordingly, to ensure that serial transmission is performed dependably, you should first check that bit TDRE is set to 1, then write the transmit data to TDR once only (not two or more times).

#### 2. Operation when a number of receive errors occur simultaneously

If a number of receive errors are detected simultaneously, the status flags in SSR will be set to the states shown in table 10.14. If an overrun error is detected, data transfer from RSR to RDR will not be performed, and the receive data will be lost.

SSR Status Flags			gs	Receive Data Transfer				
RDRF*	OER	FER	PER	$RSR \rightarrow RDR$	Receive Error Status			
1	1	0	0	Х	Overrun error			
0	0	1	0	0	Framing error			
0	0	0	1	0	Parity error			
1	1	1	0	Х	Overrun error + framing error			
1	1	0	1	Х	Overrun error + parity error			
0	0	1	1	0	Framing error + parity error			
1	1	1	1	Х	Overrun error + framing error + parity error			

## Table 10.14 SSR Status Flag States and Receive Data Transfer

O: Receive data is transferred from RSR to RDR.

X: Receive data is not transferred from RSR to RDR.

Note: \* Bit RDRF retains its state prior to data reception. However, note that if RDR is read after an overrun error has occurred in a frame because reading of the receive data in the previous frame was delayed, RDRF will be cleared to 0.

## 3. Break detection and processing

When a framing error is detected, a break can be detected by reading the value of the  $RXD_{32}$  pin directly. In a break, the input from the  $RXD_{32}$  pin becomes all 0s, with the result that bit FER is set and bit PER may also be set.

SCI3 continues the receive operation even after receiving a break. Note, therefore, that even though bit FER is cleared to 0 it will be set to 1 again.

## 4. Mark state and break detection

When bit TE is cleared to 0, the  $TXD_{32}$  pin functions as an I/O port whose input/output direction and level are determined by PDR and PCR. This fact can be used to set the  $TXD_{32}$  pin to the mark state, or to detect a break during transmission.

To keep the communication line in the mark state (1 state) until bit TE is set to 1, set PCR = 1 and PDR = 1. Since bit TE is cleared to 0 at this time, the  $TXD_{32}$  pin functions as an I/O port and 1 is output.

To detect a break, clear bit TE to 0 after setting PCR = 1 and PDR = 0.

When bit TE is cleared to 0, the transmission unit is initialized regardless of the current transmission state, the  $TXD_{32}$  pin functions as an I/O port, and 0 is output from the  $TXD_{32}$  pin.

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### 5. Receive error flags and transmit operation (synchronous mode only)

When a receive error flag (OER, PER, or FER) is set to 1, transmission cannot be started even if bit TDRE is cleared to 0. The receive error flags must be cleared to 0 before starting transmission.

Note also that receive error flags cannot be cleared to 0 even if bit RE is cleared to 0.

## 6. Receive data sampling timing and receive margin in asynchronous mode

In asynchronous mode, SCI3 operates on a basic clock with a frequency 16 times the transfer rate. When receiving, SCI3 performs internal synchronization by sampling the falling edge of the start bit with the basic clock. Receive data is latched internally at the 8th rising edge of the basic clock. This is illustrated in figure 10.16.



Figure 10.16 Receive Data Sampling Timing in Asynchronous Mode

Consequently, the receive margin in asynchronous mode can be expressed as shown in equation (1).

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$$M = \{(0.5 - \frac{1}{2N}) - \frac{D - 0.5}{N} - (L - 0.5) F\} \times 100 [\%] \qquad \dots Equation (1)$$

where M: Receive margin (%)
N: Ratio of bit rate to clock (N = 16)
D: Clock duty (D = 0.5 to 1.0)
L: Frame length (L = 9 to 12)
F: Absolute value of clock frequency deviation

Substituting 0 for F (absolute value of clock frequency deviation) and 0.5 for D (clock duty) in equation (1), a receive margin of 46.875% is given by equation (2).

When D = 0.5 and F = 0,  $M = \{0.5 - 1/(2 \times 16)\} \times 100 [\%]$ = 46.875% .... Equation (2)

However, this is only a computed value, and a margin of 20% to 30% should be allowed when carrying out system design.

#### 7. Relation between RDR reads and bit RDRF

In a receive operation, SCI3 continually checks the RDRF flag. If bit RDRF is cleared to 0 when reception of one frame ends, normal data reception is completed. If bit RDRF is set to 1, this indicates that an overrun error has occurred.

When the contents of RDR are read, bit RDRF is cleared to 0 automatically. Therefore, if bit RDR is read more than once, the second and subsequent read operations will be performed while bit RDRF is cleared to 0. Note that, when an RDR read is performed while bit RDRF is cleared to 0, if the read operation coincides with completion of reception of a frame, the next frame of data may be read. This is illustrated in figure 10.17.



Figure 10.17 Relation between RDR Read Timing and Data

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In this case, only a single RDR read operation (not two or more) should be performed after first checking that bit RDRF is set to 1. If two or more reads are performed, the data read the first time should be transferred to RAM, etc., and the RAM contents used. Also, ensure that there is sufficient margin in an RDR read operation before reception of the next frame is completed. To be precise in terms of timing, the RDR read should be completed before bit 7 is transferred in synchronous mode, or before the STOP bit is transferred in asynchronous mode.

### 8. Transmit and receive operations when making a state transition

Make sure that transmit and receive operations have completely finished before carrying out state transition processing.

## 9. Switching SCK<sub>32</sub> function

If pin SCK<sub>32</sub> is used as a clock output pin by SCI3 in synchronous mode and is then switched to a general input/output pin (a pin with a different function), the pin outputs a low level signal for half a system clock ( $\phi$ ) cycle immediately after it is switched.

This can be prevented by either of the following methods according to the situation.

a. When an  $SCK_{32}$  function is switched from clock output to non clock-output

When stopping data transfer, issue one instruction to clear bits TE and RE to 0 and to set bits CKE1 and CKE0 in SCR3 to 1 and 0, respectively. In this case, bit COM in SMR should be left 1. The above prevents  $SCK_{32}$  from being used as a general input/output pin. To avoid an intermediate level of voltage from being applied to  $SCK_{32}$ , the line connected to  $SCK_{32}$  should be pulled up to the  $V_{cc}$  level via a resistor, or supplied with output from an external device.

- b. When an  $SCK_{32}$  function is switched from clock output to general input/output When stopping data transfer,
  - (i) Issue one instruction to clear bits TE and RE to 0 and to set bits CKE1 and CKE0 in SCR3 to 1 and 0, respectively.
  - (ii) Clear bit COM in SMR to 0
  - (iii) Clear bits CKE1 and CKE0 in SCR3 to 0

Note that special care is also needed here to avoid an intermediate level of voltage from being applied to  $SCK_{32}$ .

## 10. Set up at subactive or subsleep mode

At subactive or subsleep mode, SCI3 becomes possible use only at CPU clock is  $\phi w/2$ .

#### 11. Oscillator use with serial communications interface

When implementing the serial communications interface, the system clock oscillator must be used. The on-chip oscillator should not be used in this case. See section 4.2 (5), On-Chip Oscillator Selection Method, for information on switching between the system clock oscillator and the on-chip oscillator.





# Section 11 10-Bit PWM

## 11.1 Overview

This LSI is provided with two on-chip 10-bit PWMs (pulse width modulators), designated PWM1 and PWM2, with identical functions. The PWMs can be used as D/A converters by connecting a low-pass filter. In this section the suffix m (m = 1 or 2) is used with register names, etc., as in PWDRLm, which denotes the PWDRL registers for each PWM.

#### 11.1.1 Features

Features of the 10-bit PWMs are as follows.

- Choice of four conversion periods
   Any of the following conversion periods can be chosen:
   4,096/\$\op\$, with a minimum modulation width of 4/\$\op\$
   2,048/\$\op\$, with a minimum modulation width of 2/\$\op\$
   1,024/\$\op\$, with a minimum modulation width of 1/\$\op\$
   512/\$\op\$, with a minimum modulation width of 1/\$\op\$
- Pulse division method for less ripple
- Use of module standby mode enables this module to be placed in standby mode independently when not used.

It is possible to select between two types of PWM output: pulse-division PWM and event counter PWM (PWM incorporating AEC). Refer to section 9.7, Asynchronous Event Counter (AEC), for information on event counter PWM.



#### 11.1.2 Block Diagram

Figure 11.1 shows a block diagram of the 10-bit PWM.



Figure 11.1 Block Diagram of the 10-bit PWM (1-Channel Configuration)

#### **11.1.3 Pin Configuration**

Table 11.1 shows the output pin assigned to the 10-bit PWM.

#### Table 11.1 Pin Configuration

Name	Abbr.	I/O	Function
PWM1 output pin	PWM1	Output	Pulse-division PWM waveform output (PWM1)/ event counter PWM output (IECPWM)
PWM2 output pin	PWM2	Output	Pulse-division PWM waveform output (PWM2)/ event counter PWM output (IECPWM)

#### 11.1.4 Register Configuration

Table 11.2 shows the register configuration of the 10-bit PWM.

#### Table 11.2 Register Configuration

Name	Abbr.	R/W	Initial Value	Address
PWM1 control register	PWCR1	W	H'F8	H'FFD0
PWM1 data register U	PWDRU1	W	H'FC	H'FFD1
PWM1 data register L	PWDRL1	W	H'00	H'FFD2
PWM2 control register	PWCR2	W	H'F8	H'FFCD
PWM2 data register U	PWDRU2	W	H'FC	H'FFCE
PWM2 data register L	PWDRL2	W	H'00	H'FFCF
Clock stop register 2	CKSTPR2	R/W	H'FF	H'FFFB

## **11.2 Register Descriptions**

#### 11.2.1 PWM Control Register (PWCRm)

Bit	7	6	5	4	3	2	1	0
	—		—		—	PWCRm2	PWCRm1	PWCRm0
Initial value	1	1	1	1	1	0	0	0
Read/Write	—				_	W	W	W

On the H8/38524 Group, PWCRm is an 8-bit write-only register used to select the input clock and PWM output type. At reset PWCRm is initialized to H'F8.

#### Bits 7 to 3—Reserved

Bits 7 to 3 are reserved; they are always read as 1, and cannot be modified.



#### Bit 2—Output Format Select (PWCRm2)

This bit selects the format of the output from the PWMm output pin.

This bit is write-only. Reading it always returns 1.

Bit 2 PWCRm2	Description	
0	Pulse-division PWM	(initial value)
1	Event counter PWM	

#### Bits 1 and 0—Clock Select 1 and 0 (PWCRm1, PWCRm0)

Bits 1 and 0 select the clock supplied to the 10-bit PWM. These bits are write-only bits; they are always read as 1.

Bit 0 PWCRm0	Description	
0	The input clock is $\phi$ (t $\phi^* = 1/\phi$ ) The conversion period is 512/ $\phi$ , with a minimum modulation width of 1/2 $\phi$	(initial value)
1	The input clock is $\phi/2$ (t $\phi^* = 2/\phi$ ) The conversion period is 1,024/ $\phi$ , with a minimum modulation width of 1/ $\phi$	
0	The input clock is $\phi/4$ (t $\phi^* = 4/\phi$ ) The conversion period is 2,048/ $\phi$ , with a minimum modulation width of 2/ $\phi$	
1	The input clock is $\phi/8$ (t $\phi^* = 8/\phi$ ) The conversion period is 4,096/ $\phi$ , with a minimum modulation width of 4/ $\phi$	
	<b>PWCRm0</b> 0 1	PWCRm0       Description         0       The input clock is φ (tφ* = 1/φ) The conversion period is 512/φ, with a minimum modulation width of 1/2φ         1       The input clock is φ/2 (tφ* = 2/φ) The conversion period is 1,024/φ, with a minimum modulation width of 1/φ         0       The input clock is φ/4 (tφ* = 4/φ) The conversion period is 2,048/φ, with a minimum modulation width of 2/φ         1       The input clock is φ/8 (tφ* = 8/φ) The conversion period is 4,096/φ, with a minimum

Note: \* Period of PWM input clock.

PWDRUm								
Bit	7	6	5	4	3	2	1	0
							PWDRUm1	PWDRUm0
Initial value	1	1	1	1	1	1	0	0
Read/Write							W	W
PWDRLm								
Bit	7	6	5	4	3	2	1	0
	PWDRLm7	PWDRLm6	PWDRLm5	PWDRLm4	PWDRLm3	PWDRLm2	PWDRLm1	PWDRLm0
Initial value	0	0	0	0	0	0	0	0
Read/Write	W	W	W	W	W	W	W	W

### 11.2.2 PWM Data Registers U and L (PWDRUm, PWDRLm)

PWDRUm and PWDRLm form a 10-bit write-only register, with the upper 2 bits assigned to PWDRUm and the lower 8 bits to PWDRLm. The value written to PWDRUm and PWDRLm gives the total high-level width of one PWM waveform cycle.

When 10-bit data is written to PWDRUm and PWDRLm, the register contents are latched in the PWM waveform generator, updating the PWM waveform generation data. The 10-bit data should always be written in the following sequence:

- 1. Write the lower 8 bits to PWDRLm.
- 2. Write the upper 2 bits to PWDRUm for the same channel.

PWDRUm and PWDRLm are write-only registers. If they are read, all bits are read as 1.

Upon reset, PWDRUm is initialized to H'FC, and PWDRLm to H'00.



### 11.2.3 Clock Stop Register 2 (CKSTPR2)

Bit	7	6	5	4	3	2	1	0
	LVDCKSTP	_	—	PW2CKSTP	AECKSTP	WDCKSTP	PW1CKSTP	LDCKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	_	_	R/W	R/W	R/W	R/W	R/W

CKSTPR2 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to the PWM is described here. For details of the other bits, see the sections on the relevant modules.

#### Bits 4 and 1—PWM Module Standby Mode Control (PWmCKSTP)

Bits 4 and 1 control setting and clearing of module standby mode for the PWMm.

0	PWMm is set to module standby mode	
1	PWMm module standby mode is cleared	(initial value)



## 11.3 Operation

#### 11.3.1 Operation

When using the 10-bit PWM, set the registers in the following sequence.

- 1. Set PWM1 or PWM2 in PMR9 to 1 for the PWM channel to be used, so that pin P9<sub>0</sub>/PWM1 or P9<sub>1</sub>/PWM2 is designated as the PWM output pin, or both are designated as PWM output pins.
- 3. Set the output waveform data in PWDRUm and PWDRLm. Be sure to write in the correct sequence, first PWDRLm then PWDRUm for the same channel. When data is written to PWDRUm, the data will be latched in the PWM waveform generator, updating the PWM waveform generation in synchronization with internal signals.

One conversion period consists of 4 pulses, as shown in figure 11.2. The total of the high-level pulse widths during this period  $(T_{H})$  corresponds to the data in PWDRUm and PWDRLm. This relation can be represented as follows.

 $T_{H} = (data value in PWDRUm and PWDRLm + 4) \cdot t_{\phi}/2$ 

where t $\phi$  is the PWM input clock period:  $1/\phi$  (PWCRm = H'0),  $2/\phi$  (PWCRm = H'1),  $4/\phi$  (PWCRm = H'2), or  $8/\phi$  (PWCRm = H'3).



Example: Settings in order to obtain a conversion period of 1,024  $\mu$ s: When PWCRm1 = 0 and PWCRm0 = 0, the conversion period is 512/ $\phi$ , so  $\phi$  must be 0.5 MHz. In this case, tfn = 256  $\mu$ s, with 1/2 $\phi$  (resolution) = 1.0  $\mu$ s. When PWCRm1 = 0 and PWCRm0 = 1, the conversion period is 1,024/ $\phi$ , so  $\phi$  must be 1 MHz. In this case, tfn = 256  $\mu$ s, with 1/ $\phi$  (resolution) = 1.0  $\mu$ s. When PWCRm1 = 1 and PWCRm0 = 0, the conversion period is 2,048/ $\phi$ , so  $\phi$  must be 2 MHz. In this case, tfn = 256  $\mu$ s, with 2/ $\phi$  (resolution) = 1.0  $\mu$ s. When PWCRm1 = 1 and PWCRm0 = 1, the conversion period is 4,096/ $\phi$ , so  $\phi$  must be 4 MHz. In this case, tfn = 256  $\mu$ s, with 4/ $\phi$  (resolution) = 1.0  $\mu$ s. Accordingly, for a conversion period of 1,024  $\mu$ s, the system clock frequency ( $\phi$ ) must be 0.5 MHz, 1 MHz, 2 MHz, or 4 MHz.



Figure 11.2 PWM Output Waveform

#### 11.3.2 PWM Operation Modes

PWM operation modes are shown in table 11.3.

#### Table 11.3 PWM Operation Modes

Operation Mode	Reset	Active	Sleep	Watch	Sub- active	Sub- sleep	Standby	Module Standby
PWCRm	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained
PWDRUm	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained
PWDRLm	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained

# Section 12 A/D Converter

## 12.1 Overview

This LSI includes on-chip a resistance-ladder-based successive-approximation analog-to-digital converter, and can convert up to 8 channels of analog input.

## 12.1.1 Features

The A/D converter has the following features.

- 10-bit resolution
- Eight input channels
- Conversion time: approx. 12.4 μs per channel (at 5 MHz operation)/6.2 μs (at 10 MHz operation)
- Built-in sample-and-hold function
- Interrupt requested on completion of A/D conversion
- A/D conversion can be started by external trigger input
- Use of module standby mode enables this module to be placed in standby mode independently when not used.



#### 12.1.2 Block Diagram





Figure 12.1 Block Diagram of the A/D Converter

#### 12.1.3 **Pin Configuration**

Table 12.1 shows the A/D converter pin configuration.

#### Table 12.1 Pin Configuration

Name	Abbr.	I/O	Function
Analog power supply	$AV_{cc}$	Input	Power supply and reference voltage of analog part
Analog ground	AV <sub>ss</sub>	Input	Ground and reference voltage of analog part
Analog input 0	AN <sub>o</sub>	Input	Analog input channel 0
Analog input 1	AN <sub>1</sub>	Input	Analog input channel 1
Analog input 2	AN <sub>2</sub>	Input	Analog input channel 2
Analog input 3	AN <sub>3</sub>	Input	Analog input channel 3
Analog input 4	AN <sub>4</sub>	Input	Analog input channel 4
Analog input 5	AN <sub>5</sub>	Input	Analog input channel 5
Analog input 6	AN <sub>6</sub>	Input	Analog input channel 6
Analog input 7	AN <sub>7</sub>	Input	Analog input channel 7
External trigger input	ADTRG	Input	External trigger input for starting A/D conversion

#### 12.1.4 **Register Configuration**

Table 12.2 shows the A/D converter register configuration.

#### Table 12.2 Register Configuration

Name	Abbr.	R/W	Initial Value	Address
A/D mode register	AMR	R/W	H'30	H'FFC6
A/D start register	ADSR	R/W	H'7F	H'FFC7
A/D result register H	ADRRH	R	Not fixed	H'FFC4
A/D result register L	ADRRL	R	Not fixed	H'FFC5
Clock stop register 1	CKSTPR1	R/W	H'FF	H'FFFA

## **12.2 Register Descriptions**

#### 12.2.1 A/D Result Registers (ADRRH, ADRRL)

Bit	7	6	5	4	3	2	1	0	7	6	5	4	3	2	1	0
	ADR9	ADR8	ADR7	ADR6	ADR5	ADR4	ADR3	ADR2	ADR1	ADR0	_	—	—	—	_	—
Initial value	Unde-	_		_												
	fined															
Read/Write	R	R	R	R	R	R	R	R	R	R	—	_	_	_	_	_
	·												~			
	ADRRH											AD	RRL			

ADRRH and ADRRL together comprise a 16-bit read-only register for holding the results of analog-to-digital conversion. The upper 8 bits of the data are held in ADRRH, and the lower 2 bits in ADRRL.

ADRRH and ADRRL can be read by the CPU at any time, but the ADRRH and ADRRL values during A/D conversion are not fixed. After A/D conversion is complete, the conversion result is stored as 10-bit data, and this data is held until the next conversion operation starts.

ADRRH and ADRRL are not cleared on reset.

#### 12.2.2 A/D Mode Register (AMR)

Bit	7	6	5	4	3	2	1	0
	CKS	TRGE	_		CH3	CH2	CH1	CH0
Initial value	0	0	1	1	0	0	0	0
Read/Write	R/W	R/W			R/W	R/W	R/W	R/W

AMR is an 8-bit read/write register for specifying the A/D conversion speed, external trigger option, and the analog input pins.

Upon reset, AMR is initialized to H'30.

#### Bit 7—Clock Select (CKS)

Bit 7 sets the A/D conversion speed.

Bit 7		Conversion Time					
CKS	<b>Conversion Period</b>	φ = 1 MHz	φ = 5 MHz	φ = 10 MHz			
0	62/	62 µs	12.4 µs	6.2 µs			
1	31/ <b></b>	31 µs	*	*			

Note: \* The operation cannot be guaranteed if the conversion time is less than 6.2  $\mu$ s. Make sure to select a setting that gives a conversion time of 6.2  $\mu$ s or more.

#### Bit 6—External Trigger Select (TRGE)

Bit 6 enables or disables the start of A/D conversion by external trigger input.

Bit 6 TRGE	Description
0	Disables start of A/D conversion by external trigger (initial value)
1	Enables start of A/D conversion by rising or falling edge of external trigger at pin ADTRG*
Note: *	The external trigger (ADTRG) edge is selected by bit IEG4 of IEGR. See (1) IRQ Edge Select Register (IEGR) in section 3.3.2, Interrupt Control Registers, for details.

#### Bits 5 and 4—Reserved

Bits 5 and 4 are reserved; they are always read as 1, and cannot be modified.



#### Bits 3 to 0—Channel Select (CH3 to CH0)

Bits 3 to 0 select the analog input channel.

The channel selection should be made while bit ADSF is cleared to 0.

Bit 3 CH3	Bit 2 CH2	Bit 1 CH1	Bit 0 CH0	Analog Input Channel	
0	0	*	*	No channel selected	(initial value)
0	1	0	0	AN <sub>o</sub>	
0	1	0	1	AN <sub>1</sub>	
0	1	1	0	AN <sub>2</sub>	
0	1	1	1	AN <sub>3</sub>	
1	0	0	0	AN₄	
1	0	0	1	AN <sub>5</sub>	
1	0	1	0	AN <sub>6</sub>	
1	0	1	1	AN <sub>7</sub>	
1	1	*	*	Setting prohibited	

\*: Don't care

#### 12.2.3 A/D Start Register (ADSR)

Bit	7	6	5	4	3	2	1	0	
	ADSF	_						_	
Initial value	0	1	1	1	1	1	1	1	_
Read/Write	R/W	_	_	_	_	_	_	_	

The A/D start register (ADSR) is an 8-bit read/write register for starting and stopping A/D conversion.

A/D conversion is started by writing 1 to the A/D start flag (ADSF) or by input of the designated edge of the external trigger signal, which also sets ADSF to 1. When conversion is complete, the converted data is set in ADRRH and ADRRL, and at the same time ADSF is cleared to 0.

#### Bit 7—A/D Start Flag (ADSF)

Bit 7 controls and indicates the start and end of A/D conversion.

Bit 7 ADSF	Description					
0	Read: Indicates the completion of A/D conversion (initia					
	Write: Stops A/D conversion					
1	Read: Indicates A/D conversion in progress					
	Write: Starts A/D conversion					

#### Bits 6 to 0—Reserved

Bits 6 to 0 are reserved; they are always read as 1, and cannot be modified.

#### 12.2.4 Clock Stop Register 1 (CKSTPR1)

Bit	7	6	5	4	3	2	1	0
	_		S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTP	TACKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	_	_	R/W	R/W	R/W	R/W	R/W	R/W

CKSTPR1 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to the A/D converter is described here. For details of the other bits, see the sections on the relevant modules.

#### Bit 4—A/D Converter Module Standby Mode Control (ADCKSTP)

Bit 4 controls setting and clearing of module standby mode for the A/D converter.

ADCKSTP	Description	
0	A/D converter is set to module standby mode	
1	A/D converter module standby mode is cleared	(initial value)



## 12.3 Operation

#### 12.3.1 A/D Conversion Operation

The A/D converter operates by successive approximations, and yields its conversion result as 10bit data.

A/D conversion begins when software sets the A/D start flag (bit ADSF) to 1. Bit ADSF keeps a value of 1 during A/D conversion, and is cleared to 0 automatically when conversion is complete.

The completion of conversion also sets bit IRRAD in interrupt request register 2 (IRR2) to 1. An A/D conversion end interrupt is requested if bit IENAD in interrupt enable register 2 (IENR2) is set to 1.

If the conversion time or input channel needs to be changed in the A/D mode register (AMR) during A/D conversion, bit ADSF should first be cleared to 0, stopping the conversion operation, in order to avoid malfunction.

## 12.3.2 Start of A/D Conversion by External Trigger Input

The A/D converter can be made to start A/D conversion by input of an external trigger signal. External trigger input is enabled at pin ADTRG when bit IRQ4 in PMR1 is set to 1 and bit TRGE in AMR is set to 1. Then when the input signal edge designated in bit IEG4 of interrupt edge select register (IEGR) is detected at pin ADTRG, bit ADSF in ADSR will be set to 1, starting A/D conversion.

Figure 12.2 shows the timing.



Figure 12.2 External Trigger Input Timing



#### 12.3.3 A/D Converter Operation Modes

A/D converter operation modes are shown in table 12.3.

#### Table 12.3 A/D Converter Operation Modes

Operation Mode	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module Standby		
AMR	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained		
ADSR	Reset	Functions	Functions	Retained	Retained	Retained	Retained	Retained		
ADRRH	$Retained^*$	Functions	Functions	Retained	Retained	Retained	Retained	Retained		
ADRRL	$Retained^*$	Functions	Functions	Retained	Retained	Retained	Retained	Retained		
Note: *	Undefined	Undefined in a power-on reset.								

## 12.4 Interrupts

When A/D conversion ends (ADSF changes from 1 to 0), bit IRRAD in interrupt request register 2 (IRR2) is set to 1.

A/D conversion end interrupts can be enabled or disabled by means of bit IENAD in interrupt enable register 2 (IENR2).

For further details see section 3.3, Interrupts.



# 12.5 Typical Use

An example of how the A/D converter can be used is given below, using channel 1 (pin AN1) as the analog input channel. Figure 12.3 shows the operation timing.

- 1. Bits CH3 to CH0 of the A/D mode register (AMR) are set to 0101, making pin AN<sub>1</sub> the analog input channel. A/D interrupts are enabled by setting bit IENAD to 1, and A/D conversion is started by setting bit ADSF to 1.
- 2. When A/D conversion is complete, bit IRRAD is set to 1, and the A/D conversion result is stored in ADRRH and ADRRL. At the same time ADSF is cleared to 0, and the A/D converter goes to the idle state.
- 3. Bit IENAD = 1, so an A/D conversion end interrupt is requested.
- 4. The A/D interrupt handling routine starts.
- 5. The A/D conversion result is read and processed.
- 6. The A/D interrupt handling routine ends.

If ADSF is set to 1 again afterward, A/D conversion starts and steps 2 through 6 take place.

Figures 12.4 and 12.5 show flow charts of procedures for using the A/D converter.





Figure 12.3 Typical A/D Converter Operation Timing







Figure 12.4 Flow Chart of Procedure for Using A/D Converter (Polling by Software)





Figure 12.5 Flow Chart of Procedure for Using A/D Converter (Interrupts Used)



# 12.6 A/D Conversion Accuracy Definitions

This LSI's A/D conversion accuracy definitions are given below.

Resolution

The number of A/D converter digital output codes

• Quantization error

The deviation inherent in the A/D converter, given by 1/2 LSB (see figure 12.6).

Offset error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from the minimum voltage value 0000000000 to 0000000001 (see figure 12.7).

• Full-scale error

The deviation of the analog input voltage value from the ideal A/D conversion characteristic when the digital output changes from 1111111110 to 111111111 (see figure 12.7).

Nonlinearity error

The error with respect to the ideal A/D conversion characteristics between zero voltage and full-scale voltage. Does not include offset error, full-scale error, or quantization error.

• Absolute accuracy

The deviation between the digital value and the analog input value. Includes offset error, fullscale error, quantization error, and nonlinearity error.






Figure 12.7 A/D Conversion Accuracy Definitions (2)

RENESAS

# 12.7 Application Notes

# 12.7.1 Permissible Signal Source Impedance

This LSI's analog input is designed such that conversion precision is guaranteed for an input signal for which the signal source impedance is 10 k $\Omega$  or less. This specification is provided to enable the A/D converter's sample-and-hold circuit input capacitance to be charged within the sampling time; if the sensor output impedance exceeds 10 k $\Omega$ , charging may be insufficient and it may not be possible to guarantee A/D conversion precision. However, a large capacitance provided externally, the input load will essentially comprise only the internal input resistance of 10 k $\Omega$ , and the signal source impedance is ignored. However, as a low-pass filter effect is obtained in this case, it may not be possible to follow an analog signal with a large differential coefficient (e.g., 5 mV/µs or greater) (see figure 12.8). When converting a high-speed analog signal, a low-impedance buffer should be inserted.

# 12.7.2 Influences on Absolute Precision

Adding capacitance results in coupling with GND, and therefore noise in GND may adversely affect absolute precision. Be sure to make the connection to an electrically stable GND.

Care is also required to ensure that filter circuits do not interfere with digital signals or act as antennas on the mounting board.



Figure 12.8 Analog Input Circuit Example

# 12.7.3 Additional Usage Notes

- Data in ADRRH and ADRRL should be read only when the A/D start flag (ADSF) in the A/D start register (ADSR) is cleared to 0.
- Changing the digital input signal at an adjacent pin during A/D conversion may adversely affect conversion accuracy.
- When A/D conversion is started after clearing module standby mode, wait for  $10 \phi$  clock cycles before starting.
- In active mode or sleep mode, analog power supply current (AI<sub>stopl</sub>) flows into the ladder resistance even when the A/D converter is not operating. Therefore, if the A/D converter is not used, it is recommended that AV<sub>cc</sub> be connected to the system power supply and the ADCKSTP (A/D converter module standby mode control) bit be cleared to 0 in clock stop register 1 (CKSTPR1).





# Section 13 LCD Controller/Driver

# 13.1 Overview

This LSI has an on-chip segment type LCD control circuit, LCD driver, and power supply circuit, enabling it to directly drive an LCD panel.

# 13.1.1 Features

Features of the LCD controller/driver are given below.

• Display capacity

Duty Cycle	Internal Driver
Static	32 seg
1/2	32 seg
1/3	32 seg
1/4	32 seg

• LCD RAM capacity

8 bits  $\times$  16 bytes (128 bits)

- Word access to LCD RAM
- All four segment output pins can be used individually as port pins.
- Common output pins not used because of the duty cycle can be used for common doublebuffering (parallel connection).
- Display possible in operating modes other than standby mode
- Choice of 11 frame frequencies
- Built-in power supply split-resistance, supplying LCD drive power
- Use of module standby mode enables this module to be placed in standby mode independently when not used.
- A or B waveform selectable by software
- Removal of split-resistance can be controlled in software.





# 13.1.2 Block Diagram

Figures 13.1 shows a block diagram of the LCD controller/driver.



Figure 13.1 Block Diagram of LCD Controller/Driver

# 13.1.3 Pin Configuration

Table 13.1 shows the LCD controller/driver pin configuration.

# Table 13.1 Pin Configuration

Name	Abbr.	I/O	Function
Segment output pins	$SEG_{32}$ to $SEG_1$	Output	LCD segment drive pins All pins are multiplexed as port pins (setting programmable)
Common output pins	$COM_4$ to $COM_1$	Output	LCD common drive pins Pins can be used in parallel with static or 1/2 duty
LCD power supply pins	V <sub>1</sub> , V <sub>2</sub> , V <sub>3</sub>		Used when a bypass capacitor is connected externally, and when an external power supply circuit is used

# 13.1.4 Register Configuration

Table 13.2 shows the register configuration of the LCD controller/driver.

# Table 13.2 LCD Controller/Driver Registers

Name	Abbr.	R/W	Initial Value	Address
LCD port control register	LPCR	R/W	—	H'FFC0
LCD control register	LCR	R/W	H'80	H'FFC1
LCD control register 2	LCR2	R/W	—	H'FFC2
LCD RAM	—	R/W	Undefined	H'F740 to H'F74F
Clock stop register 2	CKSTPR2	R/W	H'FF	H'FFFB

# **13.2** Register Descriptions

# 13.2.1 LCD Port Control Register (LPCR)

Bit	7	6	5	4	3	2	1	0
	DTS1	DTS0	CMX		SGS3	SGS2	SGS1	SGS0
Initial value	0	0	0		0	0	0	0
Read/Write	R/W	R/W	R/W	W	R/W	R/W	R/W	R/W

LPCR is an 8-bit read/write register which selects the duty cycle and LCD driver pin functions.

#### Bits 7 to 5—Duty Cycle Select 1 and 0 (DTS1, DTS0), Common Function Select (CMX)

The combination of DTS1 and DTS0 selects static, 1/2, 1/3, or 1/4 duty. CMX specifies whether or not the same waveform is to be output from multiple pins to increase the common drive power when not all common pins are used because of the duty setting.

Bit 7 DTS1	Bit 6 DTS0	Bit 5 CMX	Duty Cycle	Common Drivers	Notes
0	0	0	Static	$COM_1$ (initial value)	Do not use $COM_4$ , $COM_3$ , and $COM_2$ .
		1	_	COM <sub>4</sub> to COM <sub>1</sub>	$COM_4$ , $COM_3$ , and $COM_2$ output the same waveform as $COM_1$ .
0	1	0	1/2 duty	$\text{COM}_{_2}$ and $\text{COM}_{_1}$	Do not use $COM_4$ and $COM_3$ .
		1	_	$COM_4$ to $COM_1$	$COM_4$ outputs the same waveform as $COM_3$ , and $COM_2$ outputs the same waveform as $COM_1$ .
1	0	0	1/3 duty	COM <sub>3</sub> to COM <sub>1</sub>	Do not use COM <sub>4</sub> .
		1		COM <sub>4</sub> to COM <sub>1</sub>	Do not use COM <sub>4</sub> .
1	1	0	1/4 duty	$COM_4$ to $COM_1$	_
		1			

# Bit 4—Reserved

Bit 4 is reserved. It can only be written with 0.

# Bits 3 to 0—Segment Driver Select 3 to 0 (SGS3 to SGS0)

Bits 3 to 0 select the segment drivers to be used.

					Function of Pins SEG <sub>32</sub> to SEG <sub>1</sub>							_
Bit 3 SGS3	Bit 2 SGS2	Bit 1 SGS1	Bit 0 SGS0		$SEG_{28}$ to $SEG_{25}$	$SEG_{24}$ to $SEG_{21}$	SEG <sub>20</sub> to SEG <sub>17</sub>	SEG <sub>16</sub> to SEG <sub>13</sub>	SEG <sub>12</sub> to SEG <sub>9</sub>	SEG <sub>s</sub> to SEG <sub>5</sub>	SEG₄ to SEG₁	Notes
0	0	0	0	Port	Port	Port	Port	Port	Port	Port	Port	(Initial value)
			1	Port	Port	Port	Port	Port	Port	Port	SEG	
		1	0	Port	Port	Port	Port	Port	Port	SEG	SEG	_
			1	Port	Port	Port	Port	Port	SEG	SEG	SEG	_
	1	0	0	Port	Port	Port	Port	SEG	SEG	SEG	SEG	_
			1	Port	Port	Port	SEG	SEG	SEG	SEG	SEG	_
		1	0	Port	Port	SEG	SEG	SEG	SEG	SEG	SEG	
			1	Port	SEG	SEG	SEG	SEG	SEG	SEG	SEG	_
1	0	0	0	SEG	SEG	SEG	SEG	SEG	SEG	SEG	SEG	_
			1	SEG	SEG	SEG	SEG	SEG	SEG	SEG	Port	
		1	0	SEG	SEG	SEG	SEG	SEG	SEG	Port	Port	_
			1	SEG	SEG	SEG	SEG	SEG	Port	Port	Port	_
	1	0	0	SEG	SEG	SEG	SEG	Port	Port	Port	Port	_
			1	SEG	SEG	SEG	Port	Port	Port	Port	Port	_
		1	0	SEG	SEG	Port	Port	Port	Port	Port	Port	_
			1	SEG	Port	Port	Port	Port	Port	Port	Port	

Function of Pins SEG<sub>32</sub> to SEG



# 13.2.2 LCD Control Register (LCR)

Bit	7	6	5	4	3	2	1	0
		PSW	ACT	DISP	CKS3	CKS2	CKS1	CKS0
Initial value	1	0	0	0	0	0	0	0
Read/Write		R/W	R/W	R/W	R/W	R/W	R/W	R/W

LCR is an 8-bit read/write register which performs LCD drive power supply on/off control and display data control, and selects the frame frequency.

LCR is initialized to H'80 upon reset.

#### Bit 7—Reserved

Bit 7 is reserved; it is always read as 1 and cannot be modified.

#### Bit 6—LCD Drive Power Supply On/Off Control (PSW)

Bit 6 can be used to turn the LCD drive power supply off when LCD display is not required in a power-down mode, or when an external power supply is used. When the ACT bit is cleared to 0, or in standby mode, the LCD drive power supply is turned off regardless of the setting of this bit.

Bit 6 PSW	Description	
0	LCD drive power supply off	(initial value)
1	LCD drive power supply on	

#### Bit 5—Display Function Activate (ACT)

Bit 5 specifies whether or not the LCD controller/driver is used. Clearing this bit to 0 halts operation of the LCD controller/driver. The LCD drive power supply is also turned off, regardless of the setting of the PSW bit. However, register contents are retained.

Bit 5 ACT	Description	
0	LCD controller/driver operation halted	(initial value)
1	LCD controller/driver operates	

# Bit 4—Display Data Control (DISP)

Bit 4 specifies whether the LCD RAM contents are displayed or blank data is displayed regardless of the LCD RAM contents.

Bit 4 DISP	Description	
0	Blank data is displayed	(initial value)
1	LCD RAM data is display	

# Bits 3 to 0—Frame Frequency Select 3 to 0 (CKS3 to CKS0)

Bits 3 to 0 select the operating clock and the frame frequency. In subactive mode, watch mode, and subsleep mode, the system clock ( $\phi$ ) is halted, and therefore display operations are not performed if one of the clocks from  $\phi/2$  to  $\phi/256$  is selected. If LCD display is required in these modes,  $\phi$ w,  $\phi$ w/2, or  $\phi$ w/4 must be selected as the operating clock.

Bit 3	Bit 2	Bit 1	Bit 0		Frame Frequency <sup>*2</sup>		
CKS3	CKS2	CKS1	CKS0	Operating Clock	φ = 2 MHz	φ = 250 kHz <sup>*1</sup>	
0	*	0	0	φw	128 Hz <sup>*3</sup> (initia	l value)	
0	*	0	1	φw/2	64 Hz <sup>*3</sup>		
0	*	1	*	φw/4	32 Hz <sup>*3</sup>		
1	0	0	0	ф/2	_	244 Hz	
1	0	0	1	ф/4	977 Hz	122 Hz	
1	0	1	0	ф/8	488 Hz	61 Hz	
1	0	1	1	ф/16	244 Hz	30.5 Hz	
1	1	0	0	ф/32	122 Hz	—	
1	1	0	1	ф/64	61 Hz	—	
1	1	1	0	ф/128	30.5 Hz	—	
1	1	1	1	ф/256		_	

\*: Don't care

Notes: 1. This is the frame frequency in active (medium-speed,  $\phi$ osc/16) mode when  $\phi$  = 2 MHz.

- 2. When 1/3 duty is selected, the frame frequency is 4/3 times the value shown.
- 3. This is the frame frequency when  $\phi w = 32.768$  kHz.



# 13.2.3 LCD Control Register 2 (LCR2)

Bit	7	6	5	4	3	2	1	0
	LCDAB	_	—	—	CDS3	CDS2	CDS1	CDS0
Initial value	0	1	1	—	0	0	0	0
Read/Write	R/W	_	—	R/W	R/W	R/W	R/W	R/W

LCR2 is an 8-bit read/write register which controls switching between the A waveform and B waveform and removal of split-resistance. LCR2 is initialized to H'7F upon a reset.

#### Bit 7—A Waveform/B Waveform Switching Control (LCDAB)

Bit 7 specifies whether the A waveform or B waveform is used as the LCD drive waveform.

Bit 7 LCDAB	Description	
0	Drive using A waveform	(initial value)
1	Drive using B waveform	

#### Bits 6 and 5—Reserved

Bits 6 and 5 are reserved; they are always read as 1 and cannot be modified.

#### Bit 4—Reserved

Bit 4 is reserved; this can only be written with 0.



# Bits 3 to 0-Removal of Split-Resistance Control

Bit 3 Bit 2 Bit 1 Bit 0 CDS3 CDS2 CDS1 CDS0 Description (initial value) Split-resistance connected Split-resistance removed Split-resistance connected 

These bits control whether the split-resistance is removed or connected.



# 13.2.4 Clock Stop Register 2 (CKSTPR2)

Bit	7	6	5	4	3	2	1	0
	LVDCKSTP			PW2CKSTP	AECKSTP	WDCKSTP	PW1CKSTP	LDCKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W			R/W	R/W	R/W	R/W	R/W

CKSTPR2 is an 8-bit read/write register that performs module standby mode control for peripheral modules. Only the bit relating to the LCD controller/driver is described here. For details of the other bits, see the sections on the relevant modules.

# Bit 0—LCD Controller/Driver Module Standby Mode Control (LDCKSTP)

Bit 0 controls setting and clearing of module standby mode for the LCD controller/driver.

Bit 0 LDCKSTP	Description	
0	LCD controller/driver is set to module standby mode	
1	LCD controller/driver module standby mode is cleared	(initial value)

# 13.3 Operation

# 13.3.1 Settings up to LCD Display

To perform LCD display, the hardware and software related items described below must first be determined.

# (1) Hardware Settings

a. Using 1/2 duty

When 1/2 duty is used, interconnect pins  $V_2$  and  $V_3$  as shown in figure 13.2.



Figure 13.2 Handling of LCD Drive Power Supply when Using 1/2 Duty

b. Large-panel display

As the impedance of the built-in power supply split-resistance is large, it may not be suitable for driving a large panel. If the display lacks sharpness when using a large panel, refer to section 13.3.4, Boosting the LCD Drive Power Supply. When static or 1/2 duty is selected, the common output drive capability can be increased. Set CMX to 1 when selecting the duty cycle. In this mode, with a static duty cycle pins COM<sub>4</sub> to COM<sub>1</sub> output the same waveform, and with 1/2 duty the COM<sub>1</sub> waveform is output from pins COM<sub>2</sub> and COM<sub>1</sub>, and the COM<sub>2</sub> waveform is output from pins COM<sub>4</sub> and COM<sub>3</sub>.



# (2) Software Settings

a. Duty selection

Any of four duty cycles—static, 1/2 duty, 1/3 duty, or 1/4 duty—can be selected with bits DTS1 and DTS0.

b. Segment selection

The segment drivers to be used can be selected with bits SGS<sub>3</sub> to SGS<sub>0</sub>.

c. Frame frequency selection

The frame frequency can be selected by setting bits  $CKS_3$  to  $CKS_0$ . The frame frequency should be selected in accordance with the LCD panel specification. For the clock selection method in watch mode, subactive mode, and subsleep mode, see section 13.3.3, Operation in Power-Down Modes.

d. A or B waveform selection

Either the A or B waveform can be selected as the LCD waveform to be used by means of LCDAB.



# 13.3.2 Relationship between LCD RAM and Display

The relationship between the LCD RAM and the display segments differs according to the duty cycle. LCD RAM maps for the different duty cycles are shown in figures 13.3 to 13.6.

After setting the registers required for display, data is written to the part corresponding to the duty using the same kind of instruction as for ordinary RAM, and display is started automatically when turned on. Word- or byte-access instructions can be used for RAM setting.

	Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0
H'F740	SEG2	SEG2	SEG2	SEG2	SEG1	SEG1	SEG1	SEG1
H'F74F	SEG32	SEG32	SEG32	SEG32	SEG31	SEG31	SEG31	SEG31
	↓ COM4	↓ COM3	∳ COM2	↓ COM1	↓ COM4	↓ COM3	↓ COM2	¢ COM1
	COIVI4	COIVIS	COIVIZ	CONT	COIVI4	COIVIS	COIVIZ	COMI

Figure 13.3 LCD RAM Map (1/4 Duty)





Figure 13.4 LCD RAM Map (1/3 Duty)



Figure 13.5 LCD RAM Map (1/2 Duty)





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Figure 13.7 Output Waveforms for Each Duty Cycle (A Waveform)

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Figure 13.8 Output Waveforms for Each Duty Cycle (B Waveform)



Data		0	0	1	1
М		0	1	0	1
Static	Common output	<b>V</b> <sub>1</sub>	V <sub>ss</sub>	V <sub>1</sub>	V <sub>ss</sub>
	Segment output	V,	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>1</sub>
1/2 duty	Common output	V <sub>2</sub> , V <sub>3</sub>	V <sub>2</sub> , V <sub>3</sub>	<b>V</b> <sub>1</sub>	V <sub>ss</sub>
	Segment output	V <sub>1</sub>	V <sub>ss</sub>	V <sub>ss</sub>	V <sub>1</sub>
1/3 duty	Common output	V <sub>3</sub>	V <sub>2</sub>	<b>V</b> <sub>1</sub>	V <sub>ss</sub>
	Segment output	V <sub>2</sub>	V <sub>3</sub>	V <sub>ss</sub>	V <sub>1</sub>
1/4 duty	Common output	V <sub>3</sub>	V <sub>2</sub>	<b>V</b> <sub>1</sub>	V <sub>ss</sub>
	Segment output	V <sub>2</sub>	V <sub>3</sub>	V <sub>ss</sub>	V <sub>1</sub>

M: LCD alternation signal

# 13.3.3 Operation in Power-Down Modes

This LSI the LCD controller/driver can be operated even in the power-down modes. The operating state of the LCD controller/driver in the power-down modes is summarized in table 13.4.

In subactive mode, watch mode, and subsleep mode, the system clock oscillator stops, and therefore, unless  $\phi w$ ,  $\phi w/2$ , or  $\phi w/4$  has been selected by bits CKS3 to CKS0, the clock will not be supplied and display will halt. Since there is a possibility that a direct current will be applied to the LCD panel in this case, it is essential to ensure that  $\phi w$ ,  $\phi w/2$ , or  $\phi w/4$  is selected. In active (medium-speed) mode, the system clock is switched, and therefore CKS3 to CKS0 must be modified to ensure that the frame frequency does not change.



Mode		Reset	Active	Sleep	Watch	Sub-active	Sub-sleep	Standby	Module Standby
Clock	φ	Runs	Runs	Runs	Stops	Stops	Stops	Stops	Stops*4
	φw	Runs	Runs	Runs	Runs	Runs	Runs	Stops <sup>*1</sup>	Stops*4
Display	ACT = 0	Stops	Stops	Stops	Stops	Stops	Stops	Stops <sup>*2</sup>	Stops
operation	ACT = 1	Stops	Functions	Functions	Functions*3	Functions*3	Functions*3	Stops*2	Stops

Table 13.4 Power-Down Modes and Display Operation

Notes: 1. The subclock oscillator does not stop, but clock supply is halted.

2. The LCD drive power supply is turned off regardless of the setting of the PSW bit.

3. Display operation is performed only if  $\phi w, \, \phi w/2, \, or \, \phi w/4$  is selected as the operating clock.

4. The clock supplied to the LCD stops.

#### 13.3.4 Boosting the LCD Drive Power Supply

When a large panel is driven, the on-chip power supply capacity may be insufficient. If the power supply capacity is insufficient when  $V_{cc}$  is used as the power supply, the power supply impedance must be reduced. This can be done by connecting bypass capacitors of around 0.1 to 0.3  $\mu$ F to pins  $V_1$  to  $V_3$ , as shown in figure 13.9, or by adding a split-resistance externally.



Figure 13.9 Connection of External Split-Resistance

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# Section 14 Power-On Reset and Low-Voltage Detection Circuits

# 14.1 Overview

This LSI can include a power-on reset circuit and low-voltage detection circuit.

The low-voltage detection circuit consists of two circuits: LVDI (interrupt by low voltage detect) and LVDR (reset by low voltage detect) circuits.

This circuit is used to prevent abnormal operation (runaway execution) from occurring due to the power supply voltage fall and to recreate the state before the power supply voltage fall when the power supply voltage rises again.

Even if the power supply voltage falls, the unstable state when the power supply voltage falls below the guaranteed operating voltage can be removed by entering standby mode<sup>\*</sup> when exceeding the guaranteed operating voltage and during normal operation. Thus, system stability can be improved. If the power supply voltage falls more, the reset state is automatically entered. If the power supply voltage rises again, the reset state is held for a specified period, then active mode is automatically entered.

Figure 14.1 is a block diagram of the power-on reset circuit and the low-voltage detection circuit.

Note: \* The voltage maintained in standby mode is the same as the RAM data retaining voltage  $(V_{RAM})$ . See section 17.2.2, DC Characteristics, for information on retaining voltage.



# 14.1.1 Features

The features of the power-on reset circuit and low-voltage detection circuit are described below.

• Power-on reset circuit

Uses an external capacitor to generate an internal reset signal when power is first supplied.

• Low-voltage detection circuit

LVDR: Monitors the power-supply voltage, and generates an internal reset signal when the voltage falls below a specified value.

LVDI: Monitors the power-supply voltage, and generates an interrupt when the voltage falls below or rises above respective specified values.

Two pairs of detection levels for reset generation voltage are available: when only the LVDR circuit is used, or when the LVDI and LVDR circuits are both used. In addition, power supply rise/drop detection voltages and a detection voltage reference voltage may be input from an external source, allowing the detection level to be set freely by the user.

# 14.1.2 Block Diagram

A block diagram of the power-on reset circuit and low-voltage detection circuit are shown in figure 14.1.



Figure 14.1 Diagram of Power-On Reset Circuit and Low-Voltage Detection Circuit



# 14.1.3 Pin Description

The pins of the power-on reset circuit and low-voltage detection circuit are listed in table 14.1.

## Table 14.1Pin Description

Pin	Symbol	I/O	Function
Low-voltage detection circuit reference voltage input pin	Vref	Input	Reference voltage input for low- voltage detection circuit
Low-voltage detection circuit power supply drop detection voltage input pin	extD	Input	Power supply drop detection voltage input pin for low-voltage detection circuit
Low-voltage detection circuit power supply rise detection voltage input pin	extU	Input	Power supply rise detection voltage input pin for low-voltage detection circuit

# 14.1.4 Register Descriptions

The registers of the power-on reset circuit and low-voltage detection circuit are listed in table 14.2.

#### Table 14.2 Register Descriptions

Name	Symbol	R/W	Initial Value	Address
Low-voltage detection control register	LVDCR	R/W	H'00	H'FF86
Low-voltage detection status register	LVDSR	R/W	H'00	H'FF87
Low-voltage detection counter	LVDCNT	R	H'00	H'FFC3

# 14.2 Individual Register Descriptions

# 14.2.1 Low-Voltage Detection Control Register (LVDCR)

Bit	7	6	5	4	3	2	1	0
	LVDE	_	VINTDSEL	VINTUSEL	LVDSEL	LVDRE	LVDDE	LVDUE
Initial value	0*	0	0	0	0*	0*	0	0
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: \* These bits are not initialized by resets trigged by LVDR. They are initialized by poweron resets and watchdog timer resets. LVDCR is an 8-bit read/write register. It is used to control whether or not the low-voltage detection circuit is used, settings for external input of power supply rise and drop detection voltages, the LVDR detection level setting, enabling or disabling of resets triggered by the low-voltage detection reset circuit (LVDR), and enabling or disabling of interrupts triggered by power supply voltage drops or rises.

#### Bit 7—LVD Enable (LVDE)

This bit is used to control whether or not the low-voltage detection circuit is used.

Bit 7 LVDE	Description	
0	Low-voltage detection circuit not used (standby status)	(initial value)
1	Low-voltage detection circuit used	

#### Bit 6—Reserved

This bit is a read/write enabled reserved bit.

#### Bit 5—Power Supply Drop (LVDD) Detection Level External Input Select (VINTDSEL)

This bit is used to select the power supply drop detection level.

Bit 5 VINTDSEL	Description	
0	LVDD detection level generated by on-chip ladder resistor	(initial value)
1	LVDD detection level input to extD pin	

## Bit 4—Power Supply Rise (LVDU) Detection Level External Input Select (VINTUSEL)

This bit is used to select the power supply rise detection level.

Bit 4 VINTUSEL	Description	
0	LVDU detection level generated by on-chip ladder resistor	(initial value)
1	LVDU detection level input to extU pin	



# Bit 3—LVDR Detection Level Select (LVDSEL)

This bit is used to select the LVDR detection level. Select 2.3 V (typical) reset if voltage rise and drop detection interrupts are to be used. For reset detection only, Select 3.3 V (typical) reset.

# Bit 3 LVDSEL Description 0 Reset detection voltage 2.3 V (typ.) 1 Reset detection voltage 3.3 V (typ.)

#### Bit 2—LVDR Enable (LVDRE)

This bit is used to control whether resets triggered by LVDR are enabled or disabled.

Bit 2 LVDRE	Description	
0	LVDR resets disabled	(initial value)
1	LVDR resets enabled	

# Bit 1—Voltage Drop Interrupt Enable (LVDDE)

This bit is used to control whether voltage drop interrupt requests are enabled or disabled.

Bit 1 LVDDE	Description	
0	Voltage drop interrupt requests disabled	(initial value)
1	Voltage drop interrupt requests enabled	

#### Bit 0—Voltage Rise Interrupt Enable (LVDUE)

This bit is used to control whether voltage rise interrupt requests are enabled or disabled.

Bit 0 LVDUE	Description	
0	Voltage rise interrupt requests disabled	(initial value)
1	Voltage rise interrupt requests enabled	

Table 14.3 shows the relationship between LVDCR settings and function selections. Refer to table 14.3 when making settings to LVDCR.

	LVDC	R Setting	y Value		Low-Voltage		Low-Voltage Detection	Low-Voltage Detection
LVDE	LVDSEL	LVDRE	LVDDE	LVDUE	Power-on Reset	Detection Reset	Voltage Drop Interrupt	Voltage Rise Interrupt
0	*	*	*	*	0	_	_	_
1	1	1	0	0	0	0	_	_
1	0	0	1	0	0	_	0	_
1	0	0	1	1	0	_	0	0
1	0	1	1	1	0	0	0	0

#### Table 14.3 LVDCR Settings and Function Selections

Note: Setting values marked with an asterisk (\*) are invalid.

#### 14.2.2 Low-Voltage Detection Status Register (LVDSR)

Bit	7	6	5	4	3	2	1	0
	OVF	—	—	—	VREFSEL	_	LVDDF	LVDUF
Initial value	0*	0	0	0	0	0	0*	0*
Read/Write	R/W	R/W	R/W	R/W	R/W	R/W	R/W	R/W

Note: \* These bits initialized by resets trigged by LVDR.

LVDSR is an 8-bit read/write register. It is used to control external input selection, indicates when the reference voltage is stable, and indicates if the power supply voltage goes below or above a specified range.

#### Bit 7—LVD Reference Voltage Stabilized Flag (OVF)

This bit indicates when the low-voltage detection counter (LVDCNT) overflows.

Bit 7 OVF	Description	
0	[Clearing condition] When 0 is written after reading 1	(initial value)
1	[Setting condition] When the low-voltage detection counter (LVDCNT) overflows	

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#### Bits 6 to 4—Reserved

These bits are read/write enabled reserved bits.

# Bit 3—Reference Voltage External Input Select (VREFSEL)

This bit is used to select the reference voltage.

Bit 3 VREFSEL	Description	
0	The on-chip circuit is used to generate the reference voltage	(initial value)
1	The reference voltage is input to the Vref pin from an external source	e

#### Bit 2—Reserved

This bit is reserved. It is always read as 0 and cannot be written to.

## Bit 1—LVD Power Supply Voltage Drop Flag (LVDDF)

This bit indicates when a power supply voltage drop has been detected.

Bit 1 LVDDF	Description	
0	[Clearing condition] When 0 is written after reading 1	(initial value)
1	[Setting condition] When the power supply voltage drops below Vint(D)	

#### Bit 0—LVD Power Supply Voltage Rise Flag (LVDUF)

This bit indicates when a power supply voltage rise has been detected.

Bit 0 LVDUF	Description	
0	[Clearing condition] When 0 is written after reading 1	(initial value)
1	[Setting condition] When the power supply voltage drops below Vint(D) while the LVD LVDCR is set to 1, and it rises above Vint(U) before dropping belov	

Bit	7	6	5	4	3	2	1	0
	CNT7	CNT6	CNT5	CNT4	CNT3	CNT2	CNT1	CNT0
Initial value	0	0	0	0	0	0	0	0
Read/Write	R	R	R	R	R	R	R	R

# 14.2.3 Low-Voltage Detection Counter (LVDCNT)

LVDCNT is a read-only 8-bit up-counter. Counting begins when 1 is written to LVDE. The counter increments using  $\phi/4$  as the clock source until it overflows by switching from H'FF to H'00, at which time the OVF bit in the LVDSR register is set to 1, indicating that the on-chip reference voltage generator has stabilized. If the LVD function is used, it is necessary to stand by until the counter has overflowed. The initial value of LVDCNT is H'00.

# 14.2.4 Clock Stop Register 2 (CKSTPR2)

Bit	7	6	5	4	3	2	1	0
	LVDCKSTP	_	—	PW2CKSTP	AECKSTP	WDCKSTP	PW1CKSTP	LDCKSTP
Initial value	1	1	1	1	1	1	1	1
Read/Write	R/W	_	—	R/W	R/W	R/W	R/W	R/W

CKSTPR2 is an 8-bit read/write register. It is used to control the module's module standby mode. Only the bits relevant to the LVD function are described in this section. Refer to the sections on the other modules for information about the other bits.

# Bit 7—LVD Module Standby Control (LVDCKSTP)

This bit is used to control setting of the LVD function to module standby status and cancellation of that status.

Bit 7 LVDCKSTP	Description	
0	Sets LVD to module standby status	
1	Cancels LVD module standby status	(initial value)



# 14.3 Operation

# 14.3.1 Power-On Reset Circuit

Figure 14.2 shows the timing of the operation of the power-on reset circuit. As the power-supply voltage rises, the capacitor which is externally connected to the  $\overline{\text{RES}}$  pin is gradually charged via the on-chip pull-up resistor (typ. 100 k $\Omega$ ). Since the state of the  $\overline{\text{RES}}$  pin is transmitted within the chip, the prescaler S and the entire chip are in their reset states. When the level on the  $\overline{\text{RES}}$  pin reaches the specified value, the prescaler S is released from its reset state and it starts counting. The OVF signal is generated to release the internal reset signal after the prescaler S has counted 131,072 clock ( $\phi$ ) cycles. The noise cancellation circuit of approximately 100 ns is incorporated to prevent the incorrect operation of the chip by noise on the  $\overline{\text{RES}}$  pin.

To achieve stable operation of this LSI, the power supply needs to rise to its full level and settles within the specified time. The maximum time required for the power supply to rise and settle after power has been supplied ( $t_{PWON}$ ) is determined by the oscillation frequency ( $f_{osc}$ ) and capacitance which is connected to  $\overline{\text{RES}}$  pin ( $C_{\overline{\text{RES}}}$ ). If  $t_{PWON}$  means the time required to reach 90 % of power supply voltage, the power supply circuit should be designed to satisfy the following formula.

$$\begin{split} t_{_{PWON}} \ (ms) &\leq 80 \times C_{\overline{\text{RES}}} \ (\mu\text{F}) \pm 10/\text{f}_{_{OSC}} \ (\text{MHz}) \\ (t_{_{PWON}} &\leq 3000 \ \text{ms}, \ C_{\overline{\text{RES}}} \geq 0.22 \ \mu\text{F}, \ \text{and} \ \text{f}_{_{OSC}} = 10 \ \text{in} \ 2\text{-MHz} \ \text{to} \ 10\text{-MHz} \ \text{operation}) \end{split}$$

Note that the power supply voltage (Vcc) must fall below Vpor = 100 mV and rise after charge on the  $\overline{\text{RES}}$  pin is removed. To remove charge on the  $\overline{\text{RES}}$  pin, it is recommended that the diode should be placed near Vcc. If the power supply voltage (Vcc) rises from the point above Vpor, a power-on reset may not occur.



Figure 14.2 Operational Timing of Power-On Reset Circuit

# 14.3.2 Low-Voltage Detection Circuit

#### (1) LVDR (Reset by Low Voltage Detect) Circuit

Figure 14.3 shows the timing of the LVDR function. The LVDR enters the module-standby state after a power-on reset is canceled. To operate the LVDR, set the LVDE bit in LVDCR to 1, wait for 150  $\mu$ s (t<sub>LVDON</sub>) until the reference voltage and the low-voltage-detection power supply have stabilized, based on overflow of LVDNT, etc., then set the LVDRE bit in LVDCR to 1. After that, the output settings of ports must be made. To cancel the low-voltage detection circuit, first the LVDRE bit should be cleared to 0 and then the LVDE bit should be cleared to 0. The LVDE and LVDRE bits must not be cleared to 0 simultaneously because incorrect operation may occur.

When the power-supply voltage falls below the Vreset voltage (typ. = 2.3 V or 3.3 V), the LVDR clears the  $\overline{LVDRES}$  signal to 0, and resets the prescaler S. The low-voltage detection reset state remains in place until a power-on reset is generated. When the power-supply voltage rises above the Vreset voltage again, the prescaler S starts counting. It counts 131,072 clock ( $\phi$ ) cycles, and then releases the internal reset signal. In this case, the LVDE, LVDSEL, and LVDRE bits in LVDCR are not initialized.

Note that if the power supply voltage (Vcc) falls below  $V_{LVDRmin} = 1.0$  V and then rises from that point, the low-voltage detection reset may not occur.

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If the power supply voltage (Vcc) falls below Vpor = 100 mV, a power-on reset occurs.



Figure 14.3 Operational Timing of LVDR Circuit

# (2) LVDI (Interrupt by Low Voltage Detect) Circuit

Figure 14.4 shows the timing of LVDI functions. The LVDI enters the module-standby state after a power-on reset is canceled. To operate the LVDI, set the LVDE bit in LVDCR to 1, wait for 150  $\mu$ s (t<sub>LVDON</sub>) until the reference voltage and the low-voltage-detection power supply have stabilized, based on overflow of LVDNT, etc., then set the LVDDE and LVDUE bits in LVDCR to 1. After that, the output settings of ports must be made. To cancel the low-voltage detection circuit, first the LVDDE and LVDUE bits should all be cleared to 0 and then the LVDE bit should be cleared to 0. The LVDE bit must not be cleared to 0 at the same timing as the LVDDE and LVDUE bits because incorrect operation may occur.

When the power-supply voltage falls below Vint (D) (typ. = 3.7 V) voltage, the LVDI clears the  $\overline{\text{LVDINT}}$  signal to 0 and the LVDDF bit in LVDSR is set to 1. If the LVDDE bit is 1 at this time, an IRQ0 interrupt request is simultaneously generated. In this case, the necessary data must be saved in the external EEPROM, etc, and a transition must be made to standby mode or watch mode. Until this processing is completed, the power supply voltage must be higher than the lower limit of the guaranteed operating voltage.
When the power-supply voltage does not fall below Vreset1 (typ. = 2.3 V) voltage but rises above Vint (U) (typ. = 4.0 V) voltage, the LVDI sets the  $\overline{LVDINT}$  signal to 1. If the LVDUE bit is 1 at this time, the LVDUF bit in LVDSR is set to 1 and an IRQ0 interrupt request is simultaneously generated.

If the power supply voltage (Vcc) falls below Vreset1 (typ. = 2.3 V) voltage, the LVDR function is performed.



Figure 14.4 Operational Timing of LVDI Circuit

The reference voltage, power supply voltage drop detection level, and power supply voltage rise detection level can be input to the LSI from external sources via the Vref, extD, and extU pins. Figure 14.5 shows the operational timing using input from the Vref, extD, and extU pins.

First, make sure that the voltages input to pins extD and extU are set to higher levels than the interrupt detection voltage Vexd. After initial settings are made, a power supply drop interrupt is generated if the extD input voltage drops below Vexd. After a power supply drop interrupt is generated, if the external power supply voltage rises and the extU input voltage rises higher than Vexd, a power supply rise interrupt is generated. As with the on-chip circuit, the above function should be used in conjunction with LVDR (Vreset1) when the LVDI function is used.





Figure 14.5 Operational Timing of Low-Voltage Detection Interrupt Circuit (Using Pins Vref, extD, and extU)





Figure 14.6 shows a usage example for the LVD function employing pins Vref, extD, and extU.

Figure 14.6 LVD Function Usage Example Employing Pins Vref, extD, and extU

Below is an explanation of the method for calculating the external resistor values when using the Vref, extD, and extU pins for input of reference and detection voltages from sources external to the LSI.

Procedure:

- 1. First, determine the overall resistance value, R. The current consumed by the resistor is determined by the value of R. A lower R will result in a greater current flow, and a higher R will result in a reduced current flow. The value of R is dependent on the configuration of the system in which the LSI is installed.
- 2. Determine the power supply drop detection voltage (Vint(D) and the power supply rise detection voltage (Vint(U).
- 3. Using a resistance value calculation table like the one shown below, plug in values for R, Vreset1, Vint(D), and Vint(U) to calculate the values of Vref, R1, R2, and R3.

Ex. No	Vref (V)	R (kΩ)	Vreset1	Vint(D)	Vint(U)	R1 (kΩ)	R2 (kΩ)	R3 (kΩ)
1	1.30	1000	2.5	2.7	2.9	517	33	450
2	1.41	1000	2.7	2.9	3	514	16	470
3	1.57	1000	3	3.2	3.5	511	42	447
4	2.09	1000	4	4.5	4.7	536	20	444

### **Resistance Value Calculation Table**

4. Using an error calculation table like the one shown below, plug in values for R1, R2, R3, and Vref to calculate the deviation of Vreset1, Vint(D), and Vint(U). Make sure to double check the maximum and minimum values for each value.

### **Error Calculation Table**

	R1	R2	R3	Resistance Value Error (%)	_Comparator	Vreset1	Vint(D)	Vint(U)
Vref (V)	(kΩ)	(kΩ)	(kΩ)	5	Error (V)	(V)	(V)	(V)
1.3	517	33	450	R1+Err, R2/R3-Err	0.1	2.59	2.94	3.15
					0	2.49	2.84	3.05
					-0.1	2.39	2.74	2.95
				R1-Err, R2/R3+Err	0.1	2.59	2.66	2.85
					0	2.49	2.56	2.75
					-0.1	2.39	2.46	2.65
				R1/R2/R3 No Err	0.1	2.59	2.79	2.99
					0	2.49	2.69	2.89
					-0.1	2.39	2.59	2.79
				R1/R2+Err, R3-Err	0.1	2.59	2.93	3.16
					0	2.49	2.83	3.06
					-0.1	2.39	2.73	2.96
				R1/R2-Err, R3+Err	0.1	2.59	2.67	2.84
					0	2.49	2.57	2.74
					-0.1	2.39	2.47	2.64

### (3) Operation and Cancellation Setting Procedure Using LVDR and LVDI

Settings should be made as indicated below in order to ensure proper operation of the low voltage detection circuit or to cancel operation. Figure 14.7 shows the setting timing for low voltage detection circuit operation and cancellation.

- 1. To turn on the low voltage detection circuit, first set the LVDE bit in LVDCR to 1.
- 2. After waiting for LVDCNT overflow, etc., to ensure that the stabilization time  $(t_{LVDON} = 150 \ \mu s)$  for the reference voltage and low voltage detection power supply has elapsed, clear bits LVDDF and LVDUF in LVDSR to 0. If necessary, set the LVDRE, LVDDE, and LVDUE bits in LVDCR to 1.
- 3. To cancel operation of the low voltage detection circuit, clear bits LVDRE, LVDDE, and LVDUE to 0, then clear bit LVDE to 0. Bit LVDE should not be cleared at the same time as bits LVDRE, LVDDE, and LVDUE to avoid malfunction.



Figure 14.7 Low Voltage Detection Circuit Operation and Cancellation Setting Timing





# Section 15 Power Supply Circuit

This LSI incorporates an internal power supply step-down circuit. Use of this circuit enables the internal power supply to be fixed at a constant level of approximately 3.0 V, independently of the voltage of the power supply connected to the external  $V_{cc}$  pin. As a result, the current consumed when an external power supply is used at 3.0 V or above can be held down to virtually the same low level as when used at approximately 3.0 V. If the external power supply is 3.0 V or below, the internal voltage will be practically the same as the external voltage. It is, of course, also possible to use the same level of external power supply voltage and internal power supply voltage without using the internal power supply step-down circuit.

## 15.1 When Using Internal Power Supply Step-Down Circuit

Connect the external power supply to the  $V_{cc}$  pin, and connect a capacitance of approximately 0.1  $\mu$ F between  $CV_{cc}$  and  $V_{ss}$ , as shown in figure 15.1. The internal step-down circuit is made effective simply by adding this external circuit. In the external circuit interface, the external power supply voltage connected to  $V_{cc}$  and the GND potential connected to  $V_{ss}$  are the reference levels. For example, for port input/output levels, the  $V_{cc}$  level is the reference for the high level, and the  $V_{ss}$  level is that for the low level. The A/D converter analog power supply is not affected by the internal step-down circuit.



Figure 15.1 Power Supply Connection when Internal Step-Down Circuit is Used



## 15.2 When Not Using Internal Power Supply Step-Down Circuit

When the internal power supply step-down circuit is not used, connect the external power supply to the  $CV_{cc}$  pin and  $V_{cc}$  pin, as shown in figure 15.2. The external power supply is then input directly to the internal power supply. The permissible range for the power supply voltage is 2.7 V to 3.6 V. Operation cannot be guaranteed if a voltage outside this range (less than 3.0 V or more than 3.6 V) is input.



Figure 15.2 Power Supply Connection when Internal Step-Down Circuit is Not Used



# Section 16 List of Registers

The register list gives information on the on-chip I/O register addresses, how the register bits are configured, and the register states in each operating mode. The information is given as shown below.

- 1. Register addresses (address order)
- Registers are listed from the lower allocation addresses.
- Registers are classified by functional modules.
- The data bus width is indicated.
- The number of access states is indicated.
- 2. Register bits
- Bit configurations of the registers are described in the same order as the register addresses.
- Reserved bits are indicated by in the bit name column.
- When registers consist of 16 bits, bits are described from the MSB side.
- 3. Register states in each operating mode
- Register states are described in the same order as the register addresses.
- The register states described here are for the basic operating modes. If there is a specific reset for an on-chip peripheral module, refer to the section on that on-chip peripheral module.



### 16.1 Register Addresses (Address Order)

The data bus width indicates the numbers of bits by which the register is accessed.

The number of access states indicates the number of states based on the specified reference clock.

Register Name	Abbre- viation	Bit No	Address	Module Name	Data Bus Width	Access State
Flash memory control register 1	FLMCR1	8	H'F020	ROM	8	2
Flash memory control register 2	FLMCR2	8	H'F021	ROM	8	2
Flash memory power control register	FLPWCR	8	H'F022	ROM	8	2
Erase block register	EBR	8	H'F023	ROM	8	2
Flash memory enable register	FENR	8	H'F02B	ROM	8	2
Low-voltage detection control register	LVDCR	8	H'FF86	LVD	8	2
Low-voltage detection status register	LVDSR	8	H'FF87	LVD	8	2
Event counter PWM compare register H	ECPWCRH	8	H'FF8C	AEC <sup>*1</sup>	8	2
Event counter PWM compare register L	ECPWCRL	8	H'FF8D	AEC <sup>*1</sup>	8	2
Event counter PWM data register H	ECPWDRH	8	H'FF8E	AEC*1	8	2
Event counter PWM data register L	ECPWDRL	8	H'FF8F	AEC <sup>*1</sup>	8	2
Wakeup edge select register	WEGR	8	H'FF90	Interrupts	8	2
Serial port control register	SPCR	8	H'FF91	SCI3	8	2
Input pin edge select register	AEGSR	8	H'FF92	AEC <sup>*1</sup>	8	2
Event counter control register	ECCR	8	H'FF94	AEC*1	8	2
Event counter control/status register	ECCSR	8	H'FF95	AEC <sup>*1</sup>	8	2
Event counter H	ECH	8	H'FF96	AEC*1	8	2
Event counter L	ECL	8	H'FF97	AEC*1	8	2
Serial mode register	SMR	8	H'FFA8	SCI3	8	3
Bit rate register	BRR	8	H'FFA9	SCI3	8	3
Serial control register 3	SCR3	8	H'FFAA	SCI3	8	3

Register Name	Abbre- viation	Bit No	Address	Module Name	Data Bus Width	Access State
Transmit data register	TDR	8	H'FFAB	SCI3	8	3
Serial status register	SSR	8	H'FFAC	SCI3	8	3
Receive data register	RDR	8	H'FFAD	SCI3	8	3
Timer mode register A	ТМА	8	H'FFB0	Timer A	8	2
Timer counter A	TCA	8	H'FFB1	Timer A	8	2
Timer control/status register W	TCSRW	8	H'FFB2	WDT <sup>*2</sup>	8	2
Timer counter W	TCW	8	H'FFB3	WDT <sup>*2</sup>	8	2
Timer mode register C	TMC	8	H'FFB4	Timer C	8	2
Timer counter C / Timer load register C	TCC/ TLC	8	H'FFB5	Timer C	8	2
Timer control register F	TCRF	8	H'FFB6	Timer F	8	2
Timer control status register F	TCSRF	8	H'FFB7	Timer F	8	2
8-bit timer counter FH	TCFH	8	H'FFB8	Timer F	8	2
8-bit timer counter FL	TCFL	8	H'FFB9	Timer F	8	2
Output compare register FH	OCRFH	8	H'FFBA	Timer F	8	2
Output compare register FL	OCRFL	8	H'FFBB	Timer F	8	2
Timer mode register G	TMG	8	H'FFBC	Timer G	8	2
Input capture register GF	ICRGF	8	H'FFBD	Timer G	8	2
Input capture register GR	ICRGR	8	H'FFBE	Timer G	8	2
LCD port control register	LPCR	8	H'FFC0	LCD <sup>*3</sup>	8	2
LCD control register	LCR	8	H'FFC1	LCD <sup>*3</sup>	8	2
LCD control register 2	LCR2	8	H'FFC2	LCD <sup>*3</sup>	8	2
Low-voltage detection counter	LVDCNT	8	H'FFC3	LVD	8	2
A/D result register H	ADRRH	8	H'FFC4	A/D converter	8	2
A/D result register L	ADRRL	8	H'FFC5	A/D converter	8	2
A/D mode register	AMR	8	H'FFC6	A/D converter	8	2
A/D start register	ADSR	8	H'FFC7	A/D converter	8	2
Port mode register 1	PMR1	8	H'FFC8	I/O port	8	2
Port mode register 2	PMR2	8	H'FFC9	I/O port	8	2
Port mode register 3	PMR3	8	H'FFCA	I/O port	8	2
Port mode register 5	PMR5	8	H'FFCC	I/O port	8	2



Register Name	Abbre- viation	Bit No	Address	Module Name	Data Bus Width	Access State
PWM2 control register	PWCR2	8	H'FFCD	10-bit PWM	8	2
PWM2 data register U	PWDRU2	8	H'FFCE	10-bit PWM	8	2
PWM2 data register L	PWDRL2	8	H'FFCF	10-bit PWM	8	2
PWM1 control register	PWCR1	8	H'FFD0	10-bit PWM	8	2
PWM1 data register U	PWDRU1	8	H'FFD1	10-bit PWM	8	2
PWM1 data register L	PWDRL1	8	H'FFD2	10-bit PWM	8	2
Port data register 1	PDR1	8	H'FFD4	I/O port	8	2
Port data register 3	PDR3	8	H'FFD6	I/O port	8	2
Port data register 4	PDR4	8	H'FFD7	I/O port	8	2
Port data register 5	PDR5	8	H'FFD8	I/O port	8	2
Port data register 6	PDR6	8	H'FFD9	I/O port	8	2
Port data register 7	PDR7	8	H'FFDA	I/O port	8	2
Port data register 8	PDR8	8	H'FFDB	I/O port	8	2
Port data register 9	PDR9	8	H'FFDC	I/O port	8	2
Port data register A	PDRA	8	H'FFDD	I/O port	8	2
Port data register B	PDRB	8	H'FFDE	I/O port	8	2
Port pull-up control register 1	PUCR1	8	H'FFE0	I/O port	8	2
Port pull-up control register 3	PUCR3	8	H'FFE1	I/O port	8	2
Port pull-up control register 5	PUCR5	8	H'FFE2	I/O port	8	2
Port pull-up control register 6	PUCR6	8	H'FFE3	I/O port	8	2
Port control register 1	PCR1	8	H'FFE4	I/O port	8	2
Port control register 3	PCR3	8	H'FFE6	I/O port	8	2
Port control register 4	PCR4	8	H'FFE7	I/O port	8	2
Port control register 5	PCR5	8	H'FFE8	I/O port	8	2
Port control register 6	PCR6	8	H'FFE9	I/O port	8	2
Port control register 7	PCR7	8	H'FFEA	I/O port	8	2
Port control register 8	PCR8	8	H'FFEB	I/O port	8	2
Port mode register 9	PMR9	8	H'FFEC	I/O port	8	2
Port control register A	PCRA	8	H'FFED	I/O port	8	2
Port mode register B	PMRB	8	H'FFEE	I/O port	8	2
System control register 1	SYSCR1	8	H'FFF0	SYSTEM	8	2

Register Name	Abbre- viation	Bit No	Address	Module Name	Data Bus Width	Access State
System control register 2	SYSCR2	8	H'FFF1	SYSTEM	8	2
IRQ edge select register	IEGR	8	H'FFF2	Interrupts	8	2
Interrupt enable register 1	IENR1	8	H'FFF3	Interrupts	8	2
Interrupt enable register 2	IENR2	8	H'FFF4	Interrupts	8	2
Oscillator control register	OSCCR	8	H'FFF5	CPG	8	2
Interrupt request register 1	IRR1	8	H'FFF6	Interrupts	8	2
Interrupt request register 2	IRR2	8	H'FFF7	Interrupts	8	2
Timer mode register W	TMW	8	H'FFF8	WDT <sup>*2</sup>	8	2
Wakeup interrupt request register	IWPR	8	H'FFF9	Interrupts	8	2
Clock stop register 1	CKSTPR1	8	H'FFFA	SYSTEM	8	2
Clock stop register 2	CKSTPR2	8	H'FFFB	SYSTEM	8	2

Notes: 1. AEC: Asynchronous event counter

2. WDT: Watchdog timer

3. LCD: LCD controller/driver



## 16.2 Register Bits

Register bit names of the on-chip peripheral modules are described below.

Register Abbreviatior	n Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
FLMCR1	_	SWE	ESU	PSU	EV	PV	E	Р	ROM
FLMCR2	FLER	_	_	_	_	_	_	_	-
FLPWCR	PDWND	_	_	_	_	_	_	_	-
EBR	_	_	_	EB4	EB3	EB2	EB1	EB0	-
FENR	FLSHE	_	_	_	_	_	_	_	-
LVDCR	LVDE	_	VINTDSEL	VINTUSEL	LVDSEL	LVDRE	LVDDE	LVDUE	Low-
LVDSR	OVF	—	_	_	VREFSEL	_	LVDDF	LVDUF	voltage detect circuit
ECPWCRH	ECPWCRH	ECPWCRH6	ECPWCRH	5 ECPWCRH4	ECPWCRH	BECPWCRH2	2 ECPWCRH1	ECPWCRH0	AEC <sup>*1</sup>
ECPWCRL	ECPWCRL7	ECPWCRL6	ECPWCRL5	ECPWCRL4	ECPWCRL3	ECPWCRL2	ECPWCRL1	ECPWCRL0	-
ECPWDRH	ECPWDRH	ECPWDRH6	ECPWDRH	5 ECPWDRH4	ECPWDRH	BECPWDRH2	2 ECPWDRH1	ECPWDRH0	-
ECPWDRL	ECPWDRL7	ECPWDRL6	ECPWDRL5	ECPWDRL4	ECPWDRL3	BECPWDRL2	ECPWDRL1	ECPWDRL0	-
WEGR	WKEGS7	WKEGS6	WKEGS5	WKEGS4	WKEGS3	WKEGS2	WKEGS1	WKEGS0	Interrupts
SPCR	_	_	SPC32	_	SCINV3	SCINV2	_	_	SCI3
AEGSR	AHEGS1	AHEGS0	ALEGS1	ALEGS0	AIEGS1	AIEGS0	ECPWME	—	AEC <sup>*1</sup>
ECCR	ACKH1	ACKH0	ACKL1	ACKL0	PWCK2	PWCK1	PWCK0	_	-
ECCSR	OVH	OVL	_	CH2	CUEH	CUEL	CRCH	CRCL	-
ECH	ECH7	ECH6	ECH5	ECH4	ECH3	ECH2	ECH1	ECH0	-
ECL	ECL7	ECL6	ECL5	ECL4	ECL3	ECL2	ECL1	ECL0	-
SMR	COM	CHR	PE	PM	STOP	MP	CKS1	CKS0	SCI3
BRR	BRR7	BRR6	BRR5	BRR4	BRR3	BRR2	BRR1	BRR0	-
SCR3	TIE	RIE	TE	RE	MPIE	TEIE	CKE1	CKE0	-
TDR	TDR7	TDR6	TDR5	TDR4	TDR3	TDR2	TDR1	TDR0	-
SSR	TDRE	RDRF	OER	FER	PER	TEND	MPBR	MPBT	-
RDR	RDR7	RDR6	RDR5	RDR4	RDR3	RDR2	RDR1	RDR0	-
ТМА	_	_	_	_	TMA3	TMA2	TMA1	TMA0	Timer A
ТСА	TCA7	TCA6	TCA5	TCA4	TCA3	TCA2	TCA1	TCA0	-
TCSRW	B6WI	TCWE	B4WI	TCSRWE	B2WI	WDON	B0WI	WRST	WDT <sup>*2</sup>
TCW	TCW7	TCW6	TCW5	TCW4	TCW3	TCW2	TCW1	TCW0	-

Register Abbreviatio	n Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
TMC	TMC7	TMC6	TMC5	_	_	TMC2	TMC1	TMC0	Timer C
TCC/ TLC	TCC7/ TLC7	TCC6/ TLC6	TCC5/ TLC5	TCC4/ TLC4	TCC3/ TLC3	TCC2/ TLC2	TCC1/ TLC1	TCC0/ TLC0	-
TCRF	TOLH	CKSH2	CKSH1	CKSH0	TOLL	CKSL2	CKSL1	CKSL0	Timer F
TCSRF	OVFH	CMFH	OVIEH	CCLRH	OVFL	CMFL	OVIEL	CCLRL	-
TCFH	TCFH7	TCFH6	TCFH5	TCFH4	TCFH3	TCFH2	TCFH1	TCFH0	-
TCFL	TCFL7	TCFL6	TCFL5	TCFL4	TCFL3	TCFL2	TCFL1	TCFL0	-
OCRFH	OCRFH7	OCRFH6	OCRFH5	OCRFH4	OCRFH3	OCRFH2	OCRFH1	OCRFH0	•
OCRFL	OCRFL7	OCRFL6	OCRFL5	OCRFL4	OCRFL3	OCRFL2	OCRFL1	OCRFL0	•
TMG	OVFH	OVFL	OVIE	IIEGS	CCLR1	CCLR0	CKS1	CKS0	Timer G
ICRGF	ICRGF7	ICRGF6	ICRGF5	ICRGF4	ICRGF3	ICRGF2	ICRGF1	ICRGF0	-
ICRGR	ICRGR7	ICRGR6	ICRGR5	ICRGR4	ICRGR3	ICRGR2	ICRGR1	ICRGR0	
LPCR	DTS1	DTS0	CMX	_	SGS3	SGS2	SGS1	SGS0	LCD <sup>*3</sup>
LCR	_	PSW	ACT	DISP	CKS3	CKS2	CKS1	CKS0	-
LCR2	LCDAB	_	_	_	CDS3	CDS2	CDS1	CDS0	
LVDCNT	CNT7	CNT6	CNT5	CNT4	CNT3	CNT2	CNT1	CNT0	Low- voltage detect circuit
ADRRH	ADR9	ADR8	ADR7	ADR6	ADR5	ADR4	ADR3	ADR2	A/D
ADRRL	ADR1	ADR0	_	_	_	_	_	_	converter
AMR	CKS	TRGE	_	_	CH3	CH2	CH1	CH0	-
ADSR	ADSF	_	_	_	_	_	_	_	-
PMR1	IRQ3	_	_	IRQ4	TMIG	_	_	_	I/O port
PMR2	_	_	POF1	_	_	WDCKS	NCS	IRQ0	-
PMR3	AEVL	AEVH	_	_	_	TMOFH	TMOFL	UD	-
PMR5								WKP0	-
1 101113	WKP7	WKP6	WKP5	WKP4	WKP3	WKP2	WKP1		
PWCR2			WKP5	WKP4	WКР3 —		PWCR21	-	
	— — —	— — —	WКР5 — —	WKP4 — —	WКР3 — —		PWCR21	-	
PWCR2	_	PWDRL26	_	_		PWCR22	PWCR21 PWDRU21	PWCR20 PWDRU20	
PWCR2 PWDRU2	_		_	_		PWCR22 — PWDRL22	PWCR21 PWDRU21	PWCR20 PWDRU20 PWDRL20	
PWCR2 PWDRU2 PWDRL2	_		_	_		PWCR22 — PWDRL22	PWCR21 PWDRU21 PWDRL21 PWCR11	PWCR20 PWDRU20 PWDRL20	PWM
PWCR2 PWDRU2 PWDRL2 PWCR1	— — PWDRL27 — —	— — — — —	 PWDRL25  	 PWDRL24 	 PWDRL23 	PWCR22  PWDRL22 PWCR12 	PWCR21 PWDRU21 PWDRL21 PWCR11 PWDRU11	PWCR20 PWDRU20 PWDRL20 PWCR10	PWM
PWCR2 PWDRU2 PWDRL2 PWCR1 PWDRU1	— — PWDRL27 — —	— — — — —	— — PWDRL25 — —	 PWDRL24 	 PWDRL23 	PWCR22  PWDRL22 PWCR12 	PWCR21 PWDRU21 PWDRL21 PWCR11 PWDRU11	PWCR20 PWDRU20 PWDRL20 PWCR10 PWDRU10	PWM



Register Abbreviatio	n Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
PDR4	_	—	_	_	P43	P42	P41	P40	I/O port
PDR5	P57	P56	P55	P54	P53	P52	P51	P50	_
PDR6	P67	P66	P65	P64	P63	P62	P61	P60	_
PDR7	P77	P76	P75	P74	P73	P72	P71	P70	_
PDR8	P87	P86	P85	P84	P83	P82	P81	P80	_
PDR9	_	_	P95	P94	P93	P92	P91	P90	_
PDRA	_	_	_	_	PA3	PA2	PA1	PA0	_
PDRB	PB7	PB6	PB5	PB4	PB3	PB2	PB1	PB0	_
PUCR1	PUCR17	_	_	PUCR14	PUCR13	_	_	_	_
PUCR3	PUCR37	PUCR36	PUCR35	PUCR34	PUCR33	PUCR32	PUCR31	PUCR30	_
PUCR5	PUCR57	PUCR56	PUCR55	PUCR54	PUCR53	PUCR52	PUCR51	PUCR50	_
PUCR6	PUCR67	PUCR66	PUCR65	PUCR64	PUCR63	PUCR62	PUCR61	PUCR60	_
PCR1	PCR17	_	_	PCR14	PCR13	_	_	_	_
PCR3	PCR37	PCR36	PCR35	PCR34	PCR33	PCR32	PCR31	PCR30	_
PCR4	_	_	_	_	_	PCR42	PCR41	PCR40	_
PCR5	PCR57	PCR56	PCR55	PCR54	PCR53	PCR52	PCR51	PCR50	_
PCR6	PCR67	PCR66	PCR65	PCR64	PCR63	PCR62	PCR61	PCR60	_
PCR7	PCR77	PCR76	PCR75	PCR74	PCR73	PCR72	PCR71	PCR70	_
PCR8	PCR87	PCR86	PCR85	PCR84	PCR83	PCR82	PCR81	PCR80	_
PMR9	_	_	_	_	_	_	PWM2	PWM1	_
PCRA	_	_	_	_	PCRA3	PCRA2	PCRA1	PCRA0	_
PMRB	_	_	_	_	IRQ1	_	_	_	_
SYSCR1	SSBY	STS2	STS1	STS0	LSON	_	MA1	MA0	SYSTEM
SYSCR2	_	_	_	NESEL	DTON	MSON	SA1	SA0	_
IEGR	_	_	_	IEG4	IEG3	_	IEG1	IEG0	Interrupts
IENR1	IENTA	_	IENWP	IEN4	IEN3	IENEC2	IEN1	IEN0	_
IENR2	IENDT	IENAD	_	IENTG	IENTFH	IENTFL	IENTC	IENEC	_
OSCCR	SUBSTP	_	_	—	_	IRQAECF	OSCF	_	CPG
IRR1	IRRTA	_	_	IRRI4	IRRI3	IRREC2	IRRI1	IRRI0	Interrupts
IRR2	IRRDT	IRRAD	_	IRRTG	IRRTFH	IRRTFL	IRRTC	IRREC	

Register Abbreviation	n Bit 7	Bit 6	Bit 5	Bit 4	Bit 3	Bit 2	Bit 1	Bit 0	Module Name
TMW	—	_	_	_	CKS3	CKS2	CKS1	CKS0	WDT <sup>*2</sup>
IWPR	IWPF7	IWPF6	IWPF5	IWPF4	IWPF3	IWPF2	IWPF1	IWPF0	Interrupts
CKSTPR1	_	_	S32CKSTP	ADCKSTP	TGCKSTP	TFCKSTP	TCCKSTF	TACKSTP	SYSTEM
CKSTPR2	LVDCKST	P	_	PW2CKSTP	AECKSTF	WDCKSTF	PW1CKSTP	LDCKSTP	-

Notes: 1. AEC: Asynchronous event counter

2. WDT: Watchdog timer

3. LCD: LCD controller/driver



### 16.3 Register States in Each Operating Mode

Register Abbreviation	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module
FLMCR1	Initialized	_	_	Initialized	Initialized	Initialized	Initialized	ROM
FLMCR2	Initialized	_	_		_	_	_	_
FLPWCR	Initialized	_	_		_	_	_	_
EBR	Initialized	_	—	Initialized	Initialized	Initialized	Initialized	_
FENR	Initialized	_	_		_	_	_	_
LVDCR	Initialized	_	_	_	_	_	_	Low- voltage
LVDSR	Initialized	—		—	—	—	—	detect circuit
ECPWCRH	Initialized	_	_	_	_	_	_	AEC <sup>*1</sup>
ECPWCRL	Initialized	_	_	_	_	_	_	_
ECPWDRH	Initialized	_	_	_	_	_	_	_
ECPWDRL	Initialized	_	_	_	_	_	_	_
WEGR	Initialized	_	_	_	_	_	_	Interrupts
SPCR	Initialized	_	_	_	_	_	_	SCI3
AEGSR	Initialized	_	_	_	_	_	_	AEC <sup>*1</sup>
ECCR	Initialized	_	_	_	_	_	_	_
ECCSR	Initialized	_	_	_	_	_	_	_
ECH	Initialized	_	_	_	_	_	_	_
ECL	Initialized	_	_		_	_	_	_
SMR	Initialized	_	_	Initialized	_	_	Initialized	SCI3
BRR	Initialized	_	_	Initialized	_	_	Initialized	_
SCR3	Initialized	_	_	Initialized	_	_	Initialized	_
TDR	Initialized	_	_	Initialized	_	_	Initialized	_
SSR	Initialized	_	_	Initialized	_	_	Initialized	_
RDR	Initialized	—	_	Initialized	—	_	Initialized	_
ТМА	Initialized	_	_	_	_	_	_	Timer A
ТСА	Initialized	_	_	_	_	_	_	_
TCSRW	Initialized	—	_		_	_	_	WDT <sup>*2</sup>
TCW	Initialized	_	_		_	_	_	_

Register Abbreviation	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module
ТМС	Initialized	_	_	_	_	_	_	Timer C
TCC	Initialized	—	—	_	_	_	_	_
TLC	Initialized	—	_	_	_	_	_	_
TCRF	Initialized	_	_	_	_	_	_	Timer F
TCSRF	Initialized	—	—	_	_	_	_	_
TCFH	Initialized	—	_	_	_	_	_	_
TCFL	Initialized	—	_	_	_	_	_	_
OCRFH	Initialized	_	_	_	_	—	_	_
OCRFL	Initialized	—	_	_	_	_	_	_
TMG	Initialized	_	_	_	_	_	_	Timer G
ICRGF	Initialized	—	—	_	_	_	_	_
ICRGR	Initialized	_	_	_	_	_	_	_
LPCR	Initialized	_	_	_	_	_	_	LCD <sup>*3</sup>
LCR	Initialized	—	—	—	_	_	_	_
LCR2	Initialized	_	_	_	_	_	_	_
LVDCNT	Initialized	_	_	_	_	_	_	Low- voltage detect circuit
ADRRH	_	_	_	_	_	_	_	A/D
ADRRL	_	_	_	—	_	_	_	converter
AMR	Initialized	_	_	_	_	_	_	_
ADSR	Initialized	_		Initialized	Initialized	Initialized	Initialized	_
PMR1	Initialized	_		_	_	_	_	I/O port
PMR2	Initialized	_	_	_	_	_	_	_
PMR3	Initialized	_	_	_	_	—	_	_
PMR5	Initialized	_	_	_	_	_	_	_
PWCR2	Initialized	_	_	_	_	_	_	10-bit
PWDRU2	Initialized	—	—	_	_	_	_	PWM
PWDRL2	Initialized	_	_	_	_	_	_	_
PWCR1	Initialized	_	_	_	_	_		_
PWDRU1	Initialized	_	_		_	_	_	_
PWDRL1	Initialized	_	_	_	_	_		_



Register Abbreviation	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module
PDR1	Initialized	—	_	_	_	_	_	I/O port
PDR3	Initialized	—	_	_	_	_	_	_
PDR4	Initialized	_	_	_	_	_	_	—
PDR5	Initialized	_	_	_	_	_	_	—
PDR6	Initialized	_	_		_		_	_
PDR7	Initialized	_	_	_	—	_	_	_
PDR8	Initialized	_	_	_	_	_	_	_
PDR9	Initialized	_			_	_	_	_
PDRA	Initialized	_	_	_	_	_	_	_
PDRB	Initialized	_	_	_	_	_	_	_
PUCR1	Initialized	_			_	_	_	_
PUCR3	Initialized	_	_	_	_	_	_	_
PUCR5	Initialized	_	_	_	_	_	_	_
PUCR6	Initialized	_	_		_		_	_
PCR1	Initialized	_	_	_	—	_	_	_
PCR3	Initialized	_	_	_	_	_	_	_
PCR4	Initialized	_	_		_		_	_
PCR5	Initialized	_	_	_	_	_	_	_
PCR6	Initialized	_	_	_	_	_	_	_
PCR7	Initialized	_			_	_	_	_
PCR8	Initialized	_	_	_	_	_	_	_
PMR9	Initialized	_	_	_	_	_	_	_
PCRA	Initialized	_			_	_	_	_
PMRB	Initialized	_	_	_	—	_	_	_
SYSCR1	Initialized	_	_	_	_	_	_	SYSTEM
SYSCR2	Initialized	_	_		_		_	_
IEGR	Initialized	_	_	_	_	_	_	Interrupts
IENR1	Initialized	_	_	_	_	_	_	_
IENR2	Initialized	—			_	_	_	_
OSCCR	Initialized	—			_	_	_	CPG
IRR1	Initialized	—			—	_	—	Interrupts
IRR2	Initialized	_	_	_		_	_	_

Register Abbreviation	Reset	Active	Sleep	Watch	Subactive	Subsleep	Standby	Module
TMW	Initialized	_	_	_	_	_	_	WDT <sup>*2</sup>
IWPR	Initialized	_		_	_	_	_	Interrupts
CKSTPR1	Initialized	_	_	_	_	_	_	SYSTEM
CKSTPR2	Initialized		_	_	_	_		_

Notes: — is not initialized

1. AEC: Asynchronous event counter

2. WDT: Watchdog timer

3. LCD: LCD controller/driver





# Section 17 Electrical Characteristics

# 17.1 Absolute Maximum Ratings (Flash Memory Version and Mask ROM Version)

Table 17.1 lists the absolute maximum ratings.

<b>Table 17.1</b>	Absolute Maximum Ratings
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Item		Symbol	Value	Unit	Note
Power supply volta	age	V <sub>cc</sub>	–0.3 to +7.0	V	*1
		CV <sub>cc</sub>	–0.3 to +4.3	V	
Analog power sup	ply voltage	$AV_{cc}$	–0.3 to +7.0	V	
Input voltage	Other than port B	V <sub>in</sub>	–0.3 to V $_{\rm cc}$ +0.3	V	
	Port B	$AV_{in}$	–0.3 to AV <sub>cc</sub> +0.3	V	
Port 9 pin voltage		$V_{_{P9}}$	–0.3 to V $_{\rm cc}$ +0.3	V	
Operating tempera	ature	T <sub>opr</sub>	-20 to +75 <sup>*2</sup> (regular specifications)	°C	
			-40 to +85 <sup>*2</sup> (wide-range temperature specifications)	_	
Storage temperatu	ire	T <sub>stg</sub>	–55 to +125	°C	_

Notes: 1. Permanent damage may result if maximum ratings are exceeded. Normal operation should be under the conditions specified in Electrical Characteristics. Exceeding these values can result in incorrect operation and reduced reliability.

2. The operating temperature ranges from –20°C to +75°C when programming or erasing the flash memory.



# 17.2 Electrical Characteristics (Flash Memory Version and Mask ROM Version)

- 17.2.1 Power Supply Voltage and Operating Ranges
- (1) Power Supply Voltage and Oscillation Frequency Range (System Clock Oscillator Selected)



### (2) Power Supply Voltage and Oscillation Frequency Range (On-Chip Oscillator Selected)





# (3) Power Supply Voltage and Operating Frequency Range (System Clock Oscillator Selected)



### (4) Power Supply Voltage and Operating Frequency Range (On-Chip Oscillator Selected)





# (5) Analog Power Supply Voltage and A/D Converter Operating Range (System Clock Oscillator Selected)



# (6) Analog Power Supply Voltage and A/D Converter Operating Range (On-Chip Oscillator Selected)





### 17.2.2 DC Characteristics

Table 17.2 lists the DC characteristics.

### Table 17.2DC Characteristics

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

				Valu	es			
Item	Symbol	Applicable Pins	Min	Тур	Max	Unit	Test Condition	Notes
Input high	V <sub>IH</sub>	RES,	$V_{cc}  imes 0.8$	—	V <sub>cc</sub> + 0.3	V	$V_{cc}$ = 4.0 V to 5.5 V	
voltage		WKP <sub>0</sub> to WKP <sub>7</sub> , IRQ <sub>0</sub> , IRQ <sub>3</sub> , IRQ <sub>4</sub> , AEVL, AEVH, TMIC, TMIF, TMIG, ADTRG, SCK <sub>32</sub>	$V_{cc} \times 0.9$		V <sub>cc</sub> + 0.3	_	Other than above	_
		RXD <sub>32</sub> , UD	$V_{cc}  imes 0.7$	—	$V_{cc}$ + 0.3	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	_
			$V_{cc}  imes 0.8$	—	$V_{cc}$ + 0.3		Other than above	
		OSC <sub>1</sub>	$V_{cc}  imes 0.8$	—	$V_{cc}$ + 0.3	V	$V_{cc}$ = 4.0 V to 5.5 V	
		P1 <sub>3</sub> , P1 <sub>4</sub> , P1 <sub>7</sub> ,	$V_{cc}  imes 0.9$	—	$V_{cc}$ + 0.3	_	Other than above	_
			$V_{cc} \times 0.7$	—	$V_{cc}$ + 0.3	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	
		P3° to P3°, P4° to P4°, P5° to P5°, P6° to P5°, P7° to P7°, P8° to P8°, P4° to P8°,	$V_{cc} \times 0.8$		V <sub>cc</sub> + 0.3		Other than above	
		PB <sub>0</sub> to PB <sub>7</sub>	$V_{cc}  imes 0.7$	—	$AV_{cc} + 0.3$	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	_
			$V_{cc}  imes 0.8$	_	$AV_{cc} + 0.3$	_	Other than above	-
		IRQAEC, P9 <sub>5</sub> *5	$V_{cc}  imes 0.8$	—	$V_{cc} + 0.3$	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	
			$V_{cc} \times 0.9$	_	$V_{cc}$ + 0.3		Other than above	
		<b>IRQ</b> ,	$V_{cc}  imes 0.8$	_	$AV_{cc} + 0.3$	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	_
			$V_{cc}  imes 0.9$	_	$AV_{cc} + 0.3$		Other than above	

				Valu	es			
ltem	Symbol	Applicable Pins	Min	Тур	Max	Unit	Test Condition	Notes
Input low voltage	IL	$\begin{tabular}{c} \hline \hline RES, \\ \hline WKP_0 to $ \overline{WKP}_7$, \\ \hline IRQ_0$, $ IRQ_1$, \\ \hline IRQ_3$, $ IRQ_4$, \\ $ IRQAEC$, $ P9_5^{*5}$, \\ \hline \end{tabular}$	- 0.3	—	$V_{cc} \times 0.2$	V	$V_{cc}$ = 4.0 V to 5.5 V	
		AEVL, AEVH, TMIC, TMIF, TMIG, ADTRG, SCK <sub>32</sub>	- 0.3	_	$V_{cc} \times 0.1$		Other than above	_
		RXD <sub>32</sub> , UD	- 0.3	—	$V_{cc}  imes 0.3$	V	$V_{cc}$ = 4.0 V to 5.5 V	
			- 0.3	_	$V_{cc}  imes 0.2$		Other than above	_
		OSC <sub>1</sub>	- 0.3	—	$V_{cc}  imes 0.2$	V	$V_{\rm cc}$ = 4.0 V to 5.5 V	
			- 0.3	—	$V_{cc}  imes 0.1$		Other than above	_
		P1 <sub>3</sub> , P1 <sub>4</sub> , P1 <sub>7</sub> , P3 <sub>0</sub> to P3 <sub>7</sub> , P4 <sub>0</sub> to P4 <sub>3</sub> , P5 <sub>0</sub> to P5 <sub>7</sub> , P6 <sub>0</sub> to P6 <sub>7</sub> ,	- 0.3	_	$V_{cc} \times 0.3$	V	$V_{cc}$ = 4.0 V to 5.5 V	
		$PO_{0}$ to $PO_{7}$ , $PO_{1}$ to $PO_{7}$ , $PO_{1}$ to $PO_{7}$ , $PO_{1}$ to $PO_{3}$ , $PO_{1}$ to $PO_{3}$ , $PO_{1}$ to $PO_{7}$	- 0.3	_	$V_{cc}  imes 0.2$		Other than above	_
Output high voltage	V <sub>OH</sub>	P1 <sub>3</sub> , P1 <sub>4</sub> , P1 <sub>7</sub> , P3 <sub>0</sub> to P3 <sub>7</sub> , P4 <sub>0</sub> to P4 <sub>2</sub> ,	V <sub>cc</sub> - 1.0	_	_	V	$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$ $-I_{OH} = 1.0 \text{ mA}$	
vollage		$P5_{0}$ to $P5_{7}$ , $P6_{0}$ to $P6_{7}$ ,	V <sub>cc</sub> - 0.5	—	_		$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$ $-I_{OH} = 0.5 \text{ mA}$	_
		$P7_{0}$ to $P7_{7}$ , $P8_{0}$ to $P8_{7}$ , $PA_{0}$ to $PA_{3}$	V <sub>cc</sub> - 0.3	—	—		-I <sub>OH</sub> = 0.1 mA	_



				Value	s			
Item	Symbol	Applicable Pins	Min	Тур	Max	Unit	Test Condition	Notes
Output low	v V <sub>ol</sub>	P1 <sub>3</sub> , P1 <sub>4</sub> , P1 <sub>7</sub> ,	—	—	0.6	V	$V_{cc}$ = 4.0 V to 5.5 V	
voltage		$P4_0$ to $P4_2$ , $P5_0$ to $P5_7$ , $P6_0$ to $P6_7$ ,					I <sub>oL</sub> = 1.6 mA	
		$P7_{0}$ to $P7_{7}$ , $P8_{0}$ to $P8_{7}$ , $PA_{0}$ to $PA_{3}$	_	_	0.5		I <sub>oL</sub> = 0.4 mA	_
		P3 <sub>0</sub> to P3 <sub>7</sub>	—	—	1.0		$V_{cc} = 4.0 \text{ V to } 5.5 \text{ V}$ $I_{oL} = 10 \text{ mA}$	_
					0.6		$V_{cc} = 4.0 \text{ V to 5.5 V}$	
							$I_{OL} = 1.6 \text{ mA}$	
			_	_	0.5		$I_{01} = 0.4 \text{ mA}$	_
		P9₀ to P9₅	_		1.5		$V_{cc}$ = 4.0 V to 5.5 V	_
							I <sub>oL</sub> = 15 mA	
			_	_	1.0		$V_{\rm cc}$ = 4.0 V to 5.5 V	
							I <sub>oL</sub> = 10 mA	
			_	—	0.8		$V_{\rm cc}$ = 4.0 V to 5.5 V	-
							$I_{OL} = 8 \text{ mA}$	
			_	_	1.0		$I_{OL} = 5 \text{ mA}$	_
			_	_	0.6		I <sub>oL</sub> = 1.6 mA	_
			_	_	0.5		$I_{OL} = 0.4 \text{ mA}$	
Input/ output leakage current	I <sub>IL</sub>	$\begin{array}{l} \overline{\text{RES}}, P4_{3}, \\ P1_{3}, P1_{4}, P1_{7}, \\ OSC_{1}, X_{1}, \\ P3_{0} \text{ to } P3_{7}, \\ P4_{0} \text{ to } P4_{2}, \\ P5_{0} \text{ to } P5_{7}, \\ P6_{0} \text{ to } P6_{7}, \\ P7_{0} \text{ to } P7_{7}, \\ P8_{0} \text{ to } P8_{7}, \\ IRQAEC, \\ PA_{0} \text{ to } P4_{3}, \\ P9_{0} \text{ to } P9_{5} \end{array}$	_	_	1.0	μ	$V_{\rm IN}$ = 0.5 V to $V_{\rm cc}$ – 0.5 V	_
		PB <sub>0</sub> to PB <sub>7</sub>	_	_	1.0	_	$V_{IN} = 0.5 \text{ V to AV}_{CC}$ - 0.5 V	

				Value	S			
ltem	Symbol	Applicable Pins	Min	Тур	Max	Unit	Test Condition	Notes
Pull-up MOS	-I <sub>p</sub>	P1 <sub>3</sub> , P1 <sub>4</sub> , P1 <sub>7</sub> , P3 <sub>0</sub> to P3 <sub>7</sub> ,	20	_	200	μA	$V_{cc} = 5.0 V,$ $V_{IN} = 0.0 V$	
current		$P5_{0}$ to $P5_{7}$ , $P6_{0}$ to $P6_{7}$	_	40	_	μA	$V_{\rm CC} = 2.7 \text{ V},$ $V_{\rm IN} = 0.0 \text{ V}$	Reference value
Input capaci- tance	C <sub>in</sub>	All input pins except power supply pin	_	_	15.0	pF	f = 1 MHz, $V_{IN} = 0.0 V,$ $T_a = 25^{\circ}C$	
Active mode supply current	I <sub>OPE1</sub>	V <sub>cc</sub>	_	0.6	—	mA	Active (high-speed) mode $V_{cc} = 2.7 V,$ $f_{osc} = 2 MHz$	*1 *3 *4 Approx. max. value = $1.1 \times$ Typ.
			_	1.0	—			*2 *3 *4
								Approx. max. value = 1.1 × Typ.
			_	0.8	_		Active (high-speed) mode $V_{cc} = 5 V$ , $f_{osc} = 2 MHz$	*1*3*4 Approx. max. value = 1.1 × Typ.
			_	1.5	_			*2 *3 *4
								Approx. max. value = $1.1 \times$ Typ.
			—	1.6	_		Active (high-speed) mode $V_{cc} = 5 V$ , $f_{osc} = 4 MHz$	*1 *3 *4 Approx. max. value = 1.1 × Typ.
			_	2.0	_			*2 *3 *4
				-				
			_	3.3	7.0		Active (high-speed)	*1 *3 *4
			_	4.0	7.0		mode V <sub>cc</sub> = 5 V, f <sub>osc</sub> = 10 MHz	*2 *3 *4

				Value	s			
Item	Symbol	Applicable Pins	Min	Тур	Мах	Unit	<b>Test Condition</b>	Notes
Active I <sub>OPE2</sub> mode supply current	I <sub>OPE2</sub>	V <sub>cc</sub>	_	0.2	_	mA	Active (medium- speed) mode $V_{cc} = 2.7 V$ , $f_{osc} = 2 MHz$ , $\phi_{osc}/128$	*1 *3 *4 Approx. max. value = $1.1 \times$ Typ.
			—	0.5	_			*2*3*4 Approx. max. value = $1.1 \times$ Typ.
			_	0.4	_		Active (medium- speed) mode $V_{cc} = 5 V$ , $f_{osc} = 2 MHz$ , $\phi_{osc}/128$	*1 *3 *4 Approx. max. value = $1.1 \times$ Typ.
			—	0.8	_			*2*3*4 Approx. max. value = 1.1 × Typ.
			_	0.6	_		Active (medium- speed) mode $V_{cc} = 5 V$ , $f_{osc} = 4 MHz$ , $\phi_{osc}/128$	*1*3*4 Approx. max. value = $1.1 \times$ Typ.
			_	0.9	—			*2 *3 *4
			_	0.9	3.0		Active (medium-	*1 *3 *4
			_	1.2	3.0		speed) mode $V_{cc} = 5 V,$ $f_{osc} = 10 MHz,$ $\phi_{osc}/128$	*2 *3 *4

				Value	S			
ltem	Symbol	Applicable Pins	Min	Тур	Max	Unit	<b>Test Condition</b>	Notes
Sleep	I <sub>sleep</sub>	V <sub>cc</sub>	_	0.3	_	mA	V <sub>cc</sub> = 2.7 V,	*1 *3 *4
mode supply current		ŭ					$f_{osc} = 2 \text{ MHz}$	Approx. max. value = 1.1 × Typ.
			—	0.8	_			*2 *3 *4
								Approx. max. value = 1.1 × Typ.
			_	0.5	_		V <sub>cc</sub> = 5 V,	*1 *3 *4
							f <sub>osc</sub> = 2 MHz	Approx. max. value = 1.1 × Typ.
			—	0.9	_			*2 *3 *4
								Approx. max. value = 1.1 × Typ.
			—	0.9			$V_{cc} = 5 V$ ,	*1 *3 *4
							f <sub>osc</sub> = 4 MHz	Approx. max. value = 1.1 × Typ.
			_	1.3	_			*2 *3 *4
			_	1.5	5.0		V <sub>cc</sub> = 5 V,	*1 *3 *4
			_	2.2	5.0		$f_{osc} = 10 \text{ MHz}$	*2 *3 *4
Subactive	I <sub>SUB</sub>	V <sub>cc</sub>	_	11.3	_	μA	V <sub>cc</sub> = 2.7 V,	*1 *3 *4
mode supply	'SUB	♥ cc		11.0		μΑ	LCD on, 32-kHz crystal	Reference value
current			_	12.7			resonator used	*2 *3 *4
							$(\phi_{SUB} = \phi_W / 8)$	Reference value
			_	16.3	50	_	$V_{cc} = 2.7 V,$	*1 *3 *4
			_	30	50		LCD on,	*2 *3 *4
							32-kHz crystal resonator used $(\phi_{SUB} = \phi_W/2)$	



#### Section 17 Electrical Characteristics

				Value	s			
ltem	Symbol	Applicable Pins	Min	Тур	Max	Unit	Test Condition	Notes
Subsleep mode supply current	I <sub>SUBSP</sub>	V <sub>cc</sub>	_	4.0	16	μΑ	$V_{cc} = 2.7 V,$ LCD on, 32-kHz crystal resonator used $(\phi_{SUB} = \phi_W/2)$	*3 *4
Watch mode supply	I <sub>watch</sub>	V <sub>cc</sub>	—	1.4		μA	$V_{cc} = 2.7 V,$ $T_a = 25^{\circ}C,$ 32-kHz crystal	*1 *3 *4 Reference value
current			_	1.8	_		resonator used, LCD not used	*2 *3 *4 Reference value
			_	1.8	6.0		V <sub>cc</sub> = 2.7 V, 32-kHz crystal resonator used, LCD not used	*3*4
Standby mode supply current	I <sub>stby</sub>	V <sub>cc</sub>	_	0.3	_	μA	$V_{cc} = 2.7 V,$ $T_a = 25^{\circ}C,$ 32-kHz crystal resonator not used	*1*3*4 Reference value
			_	0.5	_		$V_{cc} = 2.7 V,$ $T_a = 25^{\circ}C,$ 32-kHz crystal resonator not used	*2 *3 *4 Reference value
			_	0.05	_		$V_{cc} = 2.7 V,$ $T_a = 25^{\circ}C,$ SUBSTP (subclock oscillator control register) setting = 1	*2*4 Reference value
			_	0.6	_		$V_{cc} = 5.0 V,$ $T_a = 25^{\circ}C,$ 32-kHz crystal resonator not used	*2 *3 *4 Reference value
			_	0.16	_		$V_{cc} = 5.0 V,$ $T_a = 25^{\circ}C,$ SUBSTP (subclock oscillator control register) setting = 1	*2 *4 Reference value
			_	1.0	5.0		32-kHz crystal resonator not used	*3 *4
RAM data retaining voltage	V <sub>RAM</sub>	V <sub>cc</sub>	2.0	_		V		*6

		Applicable		Value	s		Test	
ltem	Symbol	Pins	Min	Тур	Max	Unit	Condition	Notes
Allowable output low current (per pin)	I <sub>ol</sub>	Output pins except ports 3 and 9	_	_	2.0	mA	V <sub>cc</sub> = 4.0 V to 5.5 V	
		Port 3	_	—	10.0		V <sub>cc</sub> = 4.0 V to 5.5 V	_
		Output pins except port 9	—	—	0.5			_
		Port 9	—	—	15.0		V <sub>cc</sub> = 4.0 V to 5.5 V	
			—	—	5.0		Other than above	
Allowable output low current (total)	$\Sigma I_{\rm OL}$	Output pins except ports 3 and 9	_	—	40.0	mA	$V_{cc} = 4.0 V$ to 5.5 V	
		Port 3	—	—	80.0		V <sub>cc</sub> = 4.0 V to 5.5 V	_
		Output pins except port 9	—	—	20.0			_
		Port 9	—	—	80.0			
Allowable output high current (per pin)	— <b>I</b> <sub>он</sub>	All output pins	_	—	2.0	mA	V <sub>cc</sub> = 4.0 V to 5.5 V	
			_	_	0.2		Other than above	
Allowable output high current (total)	$\Sigma - I_{OH}$	All output pins	_	_	15.0	mA	V <sub>cc</sub> = 4.0 V to 5.5 V	
			_	_	10.0		Other than above	_
V <sub>cc</sub> start voltage	$V_{\text{ccstart}}$	V <sub>cc</sub>	0	_	0.1	V		*2
$V_{\rm cc}$ rising gradient	$\mathrm{SV}_{\mathrm{cc}}$	V <sub>cc</sub>	0.05	_		V/ms		*2



Notes: Connect the TEST pin to  $V_{ss}$ .

- 1. Applies to the mask-ROM version.
- 2. Applies to the flash memory version.
- 3. Pin states when supply current is measured.

Mode	RES Pin	Internal State	Other Pins	LCD Power Supply	Oscillator Pins
Active (high-speed) mode (I <sub>OPE1</sub> )	V <sub>cc</sub>	Only CPU operates	V <sub>cc</sub>	Stops	System clock: crystal resonator
Active (medium- speed) mode (I <sub>OPE2</sub> )					Subclock: Pin X, = GND
Sleep mode	V <sub>cc</sub>	Only all on-chip timers operate	V <sub>cc</sub>	Stops	
Subactive mode	V <sub>cc</sub>	Only CPU operates	V <sub>cc</sub>	Stops	System clock:
Subsleep mode	V <sub>cc</sub>	Only all on-chip timers operate	V <sub>cc</sub>	Stops	crystal resonator Subclock: crystal resonator
		CPU stops			
Watch mode	V <sub>cc</sub>	Only clock time base operates	V <sub>cc</sub>	Stops	_
		CPU stops			
Standby mode	V <sub>cc</sub>	CPU and timers both stop	V <sub>cc</sub>	Stops	System clock: crystal resonator
					Subclock: Pin X <sub>1</sub> = GND

4. Except current which flows to the pull-up MOS or output buffer.

5. Used when user mode or boot mode is determined after canceling a reset in the flash memory version.

6. Voltage maintained in standby mode.
# 17.2.3 AC Characteristics

Table 17.3 lists the control signal timing and table 17.4 lists the serial interface timing.

# Table 17.3 Control Signal Timing

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

		Applicable	Values				Reference	
Item	Symbol	Pins	Min	Тур	Max	Unit	Test Condition	Figure
System clock	f <sub>osc</sub>	$OSC_1, OSC_2$	2.0	_	20.0	MHz		
oscillation frequency			0.7	—	2.0		On-chip oscillator selected	*2
OSC clock ( $\phi_{osc}$ ) cycle time	t <sub>osc</sub>	OSC <sub>1</sub> , OSC <sub>2</sub>	50.0	—	500	ns		Figure 17.1
			500	—	1429		On-chip oscillator selected	
System clock (ø)	t <sub>cyc</sub>		2		128	t <sub>osc</sub>		
cycle time			_		182	μs	_	
Subclock oscillation frequency	f <sub>w</sub>	X <sub>1</sub> , X <sub>2</sub>	—	32.768	_	kHz		
Watch clock $(\phi_w)$ cycle time	t <sub>w</sub>	X <sub>1</sub> , X <sub>2</sub>	_	30.5	_	μs		Figure 17.1
Subclock ( $\phi_{SUB}$ ) cycle time	$t_{_{subcyc}}$		2	_	8	t <sub>w</sub>		*1
Instruction cycle time			2	—	_	t <sub>cyc</sub> t <sub>subcyc</sub>		
Oscillation stabilization time	t <sub>rc</sub>	OSC <sub>1</sub> , OSC <sub>2</sub>		_	20	ms		
	t <sub>rc</sub>	X <sub>1</sub> , X <sub>2</sub>	—	_	2.0	S		
External clock high width	t <sub>CPH</sub>	OSC,	20	_	_	ns		Figure 17.1
External clock low width	t <sub>CPL</sub>	OSC <sub>1</sub>	20	_	_	ns		Figure 17.1
External clock rise time	t <sub>CPr</sub>	OSC <sub>1</sub>	—	—	5	ns		Figure 17.1
External clock fall time	$t_{\rm CPf}$	OSC <sub>1</sub>	-	—	5	ns		Figure 17.1
RES pin low width	t <sub>rel</sub>	RES	10	_	_	t <sub>cyc</sub>		Figure 17.2

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		Applicable		Value	es			Reference
Item	Symbol	Pins	Min	Тур	Max	Unit	Test Condition	Figure
Input pin high width	t <sub>⊪</sub>	IRQ0, IRQ1, IRQ3, IRQ4, IRQAEC, WKP0 to WKP7, TMIC, TMIF, TMIG, ADTRG	2	_	_	t <sub>cyc</sub> t <sub>subcyc</sub>		Figure 17.3
		AEVL, AEVH	0.5	_	_	t <sub>osc</sub>	_	
Input pin low width	t <sub>iL</sub>	IRQ0, IRQ1, IRQ3, IRQ4, IRQAEC, WKP0 to WKP7, TMIC, TMIF, TMIG, ADTRG	2	_	_	t <sub>cyc</sub> t <sub>subcyc</sub>		Figure 17.3
		AEVL, AEVH	0.5	_	—	t <sub>osc</sub>		
UD pin minimum transition width	t <sub>udh</sub> t <sub>udl</sub>	UD	4	—	—	t <sub>cyc</sub> t <sub>subcyc</sub>		Figure 17.6

Notes: 1. Determined by the SA1 and SA0 bits in the system control register 2 (SYSCR2).

2. These characteristics are given as ranges between minimum and maximum values in order to account for factors such as temperature, power supply voltage, and variation among production lots. When designing systems, make sure to give due consideration to the SPEC range. Please contact a Renesas sales or support representative for actual performance data on the product.

### Table 17.4 Serial Interface (SCI3) Timing

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

			١	/alues	5		Test	Reference
ltem		Symbol	Min	Тур	Max	Unit	Condition	Figure
•	Asynchronous	t <sub>scyc</sub>	4	—	—	$t_{_{\rm cyc}}  \text{or}  t_{_{\rm subcyc}}$		Figure 17.4
cycle	Clocked synchronous	_	6	_	_	_		
Input clock	pulse width	t <sub>scкw</sub>	0.4	_	0.6	t <sub>scyc</sub>		Figure 17.4
	ata delay time (nchronous)	t <sub>TXD</sub>	_	—	1	$\rm t_{_{cyc}}~or~t_{_{subcyc}}$		Figure 17.5
	ta setup time (nchronous)	t <sub>RXS</sub>	150.0			ns		Figure 17.5
	ita hold time vnchronous)	t <sub>RXH</sub>	150.0	_	—	ns		Figure 17.5



# 17.2.4 A/D Converter Characteristics

Table 17.5 shows the A/D converter characteristics.

#### Table 17.5 A/D Converter Characteristics

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

		Applicable	Values			Test	Reference	
Item	Symbol	Pins	Min	Тур	Max	Unit	Condition	Figure
Analog power supply voltage	$AV_{cc}$	$AV_{cc}$	2.7	_	5.5	V		*1
Analog input voltage	$AV_{IN}$	AN <sub>o</sub> to AN <sub>7</sub>	- 0.3	_	$AV_{cc} + 0.3$	V		
Analog power supply	Al	AV <sub>cc</sub>	_	_	1.5	mA	$AV_{cc} = 5.0 V$	
current	AI <sub>STOP1</sub>	AV <sub>cc</sub>	_	600	_	μA		*2 Reference value
	AI <sub>STOP2</sub>	AV <sub>cc</sub>	_	_	5.0	μA		*3
Analog input capacitance	C <sub>AIN</sub>	AN <sub>o</sub> to AN <sub>7</sub>	_	_	15.0	pF		
Allowable signal source impedance	R <sub>AIN</sub>		_	_	10.0	kΩ		
Resolution (data length)			_	_	10	bit		
Nonlinearity error			_	_	±3.5	LSB	AV <sub>cc</sub> = 4.0 V to 5.5 V	
			_	_	±7.5	_	AV <sub>cc</sub> = 2.7 V to 5.5 V	-
Quantization error			—	—	±0.5	LSB		
Absolute accuracy			_	±2.0	±4.0	LSB	AV <sub>cc</sub> = 4.0 V to 5.5 V	
			_	±2.0	±8.0	_	AV <sub>cc</sub> = 2.7 V to 5.5 V	-
Conversion time			6.2	_	124	μs		

Notes: 1. Set  $AV_{cc} = V_{cc}$  when the A/D converter is not used.

2. Al<sub>STOP1</sub> is the current in active and sleep modes while the A/D converter is idle.

3. Al<sub>STOP2</sub> is the current at reset and in standby, watch, subactive, and subsleep modes while the A/D converter is idle.

#### 17.2.5 LCD Characteristics

Table 17.6 shows the LCD characteristics.

#### Table 17.6 LCD Characteristics

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

		Applicable		Value	s			Reference
ltem	Symbol	Pins	Min	Тур	Max	Unit	Test Condition	Figure
Segment driver step-down voltage	$V_{\rm ds}$	SEG <sub>1</sub> to SEG <sub>32</sub>	_	_	0.6	V	$I_{D} = 2 \ \mu A$ V1 = 2.7 V to 5.5 V	*1
Common driver step-down voltage	V <sub>DC</sub>	$COM_1$ to $COM_4$		_	0.3	V	$I_{D} = 2 \mu A$ V1 = 2.7 V to 5.5 V	*1
LCD power supply split-resistance	$R_{\text{LCD}}$		1.5	3.0	7.0	MΩ	Between V1 and $V_{ss}$	
Liquid crystal display voltage	$V_{\text{LCD}}$	V <sub>1</sub>	2.7	—	5.5	V		*2

Notes: 1. The voltage step-down from power supply pins V1, V2, V3, and V<sub>ss</sub> to each segment pin or common pin.

2. When the liquid crystal display voltage is supplied from an external power supply, ensure that the following relationship is maintained:  $V_{cc} \ge V1 \ge V2 \ge V3 \ge V_{ss}$ .



# 17.2.6 Flash Memory Characteristics

# Table 17.7 Flash Memory Characteristics

Condition:

ition:  $AV_{cc} = 2.7 V$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 V$ ,  $V_{cc} = 2.7 V$  to 5.5 V (range of operating voltage when reading),  $V_{cc} = 3.0 V$  to 5.5 V (range of operating voltage when programming/erasing),  $T_a = -20^{\circ}$ C to  $+75^{\circ}$ C (range of operating temperature when programming/erasing: product with regular specifications, product with wide-range temperature specifications)

			Values				Test
Item		Symbol	Min	Тур	Max	Unit	Conditions
Programming t	Programming time <sup>*1*2*4</sup>		_	7	200	ms/128 bytes	
Erase time*1*3*	\$5	t <sub>e</sub>	—	100	1200	ms/block	
Reprogrammin	g count	$N_{_{WEC}}$	1000**	10000 <sup>*9</sup>	_	times	
Data retain per	iod	t <sub>DRP</sub>	10 <sup>*10</sup>	_	_	year	
Programming	Wait time after SWE-bit setting <sup>*1</sup>	x	1	_	_	μs	
	Wait time after PSU-bit setting <sup>*1</sup>	У	50	—	_	μs	
	Wait time after	z1	28	30	32	μs	$1 \le n \le 6$
	P-bit setting <sup>*1*4</sup>	z2	198	200	202	μs	$7 \le n \le 1000$
		z3	8	10	12	μs	Additional programming
	Wait time after P-bit clear <sup>*1</sup>	α	5	—	_	μs	
	Wait time after PSU-bit clear <sup>*1</sup>	β	5	_	_	μs	
	Wait time after PV-bit setting <sup>*1</sup>	γ	4	_	—	μs	
	Wait time after dummy write <sup>*1</sup>	8	2	_	_	μs	
	Wait time after PV-bit clear <sup>*1</sup>	η	2	—	—	μs	
	Wait time after SWE-bit clear <sup>*1</sup>	θ	100	—	—	μs	
	Maximum programming count <sup>*1*4*5</sup>	N	_	_	1000	times	

		Values				Test	
Item		Symbol	Min	Тур	Max	Unit	Conditions
Erase	Wait time after SWE-bit setting <sup>*1</sup>	х	1	—	—	μs	
	Wait time after ESU-bit setting <sup>*1</sup>	у	100	—	—	μs	
	Wait time after E-bit setting <sup>*1*6</sup>	Z	10	—	100	ms	
	Wait time after E-bit clear <sup>*1</sup>	α	10	—	—	μs	
	Wait time after ESU-bit clear <sup>*1</sup>	β	10	_	_	μs	
	Wait time after EV-bit setting <sup>*1</sup>	γ	20	_	_	μs	
	Wait time after dummy write <sup>*1</sup>	ε	2	_	_	μs	
	Wait time after EV-bit clear*1	η	4	_	_	μs	
	Wait time after SWE-bit clear <sup>*1</sup>	θ	100	_	_	μs	
	Maximum erase count <sup>*1*6*7</sup>	Ν	—	_	120	times	

Notes: 1. Set the times according to the program/erase algorithms.

- 2. Programming time per 128 bytes (Shows the total period for which the P bit in FLMCR1 is set. It does not include the programming verification time.)
- 3. Block erase time (Shows the total period for which the E bit in FLMCR1 is set. It does not include the erase verification time.)
- Maximum programming time (t<sub>p</sub> (max)) t<sub>p</sub> (max) = Wait time after P-bit setting (z) × maximum number of writes (N)
- 5. The maximum number of writes (N) should be set according to the actual set value of z1, z2, and z3 to allow programming within the maximum programming time (t<sub>p</sub> (max)). The wait time after P-bit setting (z1 and z2) should be alternated according to the number of writes (n) as follows:

$$1 \le n \le 6$$
  $z1 = 30 \ \mu s$ 

$$7 \le n \le 1000$$
  $z2 = 200 \ \mu s$ 

6. Maximum erase time  $(t_{E} (max))$ 

 $t_{E}$  (max) = Wait time after E-bit setting (z) × maximum erase count (N)

- 7. The maximum number of erases (N) should be set according to the actual set value of z to allow erasing within the maximum erase time ( $t_{E}$  (max)).
- 8. This minimum value guarantees all characteristics after reprogramming (the guaranteed range is from 1 to the minimum value).
- Reference value when the temperature is 25°C (normally reprogramming will be performed by this count).
- 10. This is a data retain characteristic when reprogramming is performed within the specification range including this minimum value.

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#### 17.2.7 Power Supply Voltage Detection Circuit Characteristics

#### Table 17.8 Power Supply Voltage Detection Circuit Characteristics (1)

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

			Rated Va	alues		
Item	Symbol	Min	Тур	Max	Unit	Test Conditions
LVDR operation drop voltage*	$V_{\scriptscriptstyle LVDRmin}$	1.0	—	—	V	
LVD stabilization time	T	150		—	μs	
Standby mode	I <sub>stby</sub>	—		100	μA	LVDE = 1
supply current						$V_{cc} = 5.0 V$
						32 oscillator not used

Note: \* In some cases no reset may occur if the power supply voltage,  $V_{cc}$ , drops below VLVDRmin = 1.0 V and then rises, so thorough evaluation is called for.

#### Table 17.9 Power Supply Voltage Detection Circuit Characteristics (2)

Using on-chip reference voltage and ladder resistor (VREFSEL = VINTDSEL = VINTUSEL = 0)

			Rated Va	alues		
Item	Symbol	Min	Тур	Max	Unit	Test Conditions
Power supply drop detection voltage	Vint(D) <sup>*3</sup>	3.3	3.7	4.2	V	LVDSEL = 0
Power supply rise detection voltage	Vint(U)*3	3.6	4.0	4.5	V	LVDSEL = 0
Reset detection voltage 1*1	Vreset1*3	2.0	2.3	2.7	V	LVDSEL = 0
Reset detection voltage 2 <sup>*2</sup>	Vreset2*3	2.7	3.3	3.9	V	LVDSEL = 1

Notes: 1. The above function should be used in conjunction with the voltage drop/rise detection function.

2. Low-voltage detection reset should be selected for low-voltage detection reset only.

3. The values of Vint(D), Vint(U), Vreset1, and Vreset2 change relative to each other.

Example: If Vint(D) is the minimum value, Vint(U), Vreset1, and Vreset2 are also the minimum values.

#### Table 17.10 Power Supply Voltage Detection Circuit Characteristics (3)

Using on-chip reference voltage and detect voltage external input (VREFSEL = 0, VINTDSEL and VINTUSEL = 1)

		Rated Values				
ltem	Symbol	Min	Тур	Мах	Unit	Test Condition
extD/extU interrupt detection level	Vexd	0.80	1.20	1.60	V	
extD/extU pin input voltage <sup>*2</sup>	VextD <sup>*1</sup> VextU <sup>*1</sup>	-0.3	_	$V_{\rm cc}$ + 0.3 or AV <sub>cc</sub> + 0.3, whichever is lower	V	$V_{cc}$ = 2.7 to 3.3 V
		-0.3	_	3.6 or $AV_{cc}$ + 0.3, whichever is lower	V	$V_{cc} = 3.3 \text{ to } 5.5 \text{ V}$

Notes: 1. The VextD voltage must always be greater than the VextU voltage.

2. The maximum input voltage of the extD and extU pins is 3.6 V.



#### Table 17.11 Power Supply Voltage Detection Circuit Characteristics (4)

Using external reference voltage and ladder resistor (VREFSEL = 1, VINTDSEL = VINTUSEL = 0)

			Rated Values	S		Test
Item	Symbol	Min	Тур	Мах	Unit	Condition
Power supply drop detection voltage	Vint(D) <sup>*1</sup>	3.08 * (Vref1 - 0.1)	3.08 * Vref1	3.08 * (Vref1 + 0.1)	V	LVDSEL = 0
Vref input voltage (Vint(D))	Vref1 <sup>*2</sup>	0.98	_	1.68	V	Vint(D)
Power supply rise detection voltage	Vint(U)*1	3.33 * (Vref2 - 0.1)	3.33 * Vref2	3.33 * (Vref2 + 0.1)	V	LVDSEL = 0
Vref input voltage (Vint(U))	Vref2 <sup>*2</sup>	0.91	_	1.55	V	Vint(U)
Reset detection voltage 1	Vreset1*1	1.91 * (Vref3 – 0.1)	1.91 * Vref3	1.91 * (Vref3 + 0.1)	V	LVDSEL = 0
Vref input voltage (Vreset1)	Vref3 <sup>*2</sup>	0.89	_	2.77	V	Vreset1
Reset detection voltage 2	Vreset2*1	2.76 * (Vref4 - 0.1)	2.76 * Vref4	2.76 * (Vref4 + 0.1)	V	LVDSEL = 1
Vref input voltage (Vreset2)	Vref4 <sup>*2</sup>	1.08	_	1.89	V	Vreset2

Notes: 1. The values of Vint(D), Vint(U), Vreset1, and Vreset2 change relative to each other.

Example: If Vint(D) is the minimum value, Vint(U), Vreset1, and Vreset2 are also the minimum values.

2. The Vref input voltage is calculated using the following formula.

2.7 V (= V <sub>cc</sub> min)	< Vint(D), Vint(U), Vreset2	$< 5.5 V (= V_{cc} max)$
1.5 V (= RAM retention voltage)	< Vreset1	$< 5.5 V (= V_{cc} max)$
Vref1: 2.7 < 3.08 * (Vref1 - 0.1), 3.0	08 * (Vref1 + 0.1) < 5.5 $\rightarrow$ 0.98 <	< Vref1 < 1.68
Vref2: 2.7 < 3.33 * (Vref2 - 0.1), 3.3	33 * (Vref2 + 0.1) < 5.5 $\rightarrow$ 0.91 <	< Vref2 < 1.55
Vref3: 1.5 < 1.91 * (Vref3 - 0.1), 1.9	91 * (Vref3 + 0.1) < 5.5 $ ightarrow$ 0.89 <	< Vref3 < 2.77
Vref4: 2.7 < 2.76 * (Vref4 - 0.1), 2.7	76 * (Vref4 + 0.1) < 5.5 $ ightarrow$ 1.08 <	< Vref4 < 1.89

### Table 17.12 Power Supply Voltage Detection Circuit Characteristics (5)

Using external reference voltage and detect voltage external input (VREFSEL = VINTDSEL = VINTUSEL = 1)

			Rat	ed Values		
ltem	Symbol	Min	Тур	Max	Unit	Test Condition
Comparator detection accuracy	Vcdl	0.1	—	_	V	VextU – Vref     VextD – Vref
extD/extU pin input voltage	VextD* VextU*	-0.3	_	$V_{cc}$ + 0.3 or AV <sub>cc</sub> + 0.3, whichever is lower	V	$V_{cc}$ = 2.7 to 3.3 V
		-0.3	_	3.6 or AV <sub>cc</sub> + 0.3, whichever is lower	V	$V_{cc} = 3.3 \text{ to } 5.5 \text{ V}$
Vref pin input voltage	Vref5	0.8		2.8	V	$V_{\rm cc}$ = 2.7 to 5.5 V

Note: \* The VextD voltage must always be greater than the VextU voltage.

#### 17.2.8 Power-On Reset Circuit Characteristics

#### Table 17.13 Power-On Reset Circuit Characteristics

 $V_{cc} = 2.7 \text{ V}$  to 5.5 V,  $AV_{cc} = 2.7 \text{ V}$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 \text{ V}$ , unless otherwise specified

			Rated Va	alues		
Item	Symbol	Min	Тур	Max	Unit	<b>Test Condition</b>
RES pin pull-up resistance	$R_{res}$	65	100	—	kΩ	
Power-on reset start voltage	$V_{\rm por}$			100	mV	

Note: Make sure to drop the power supply voltage,  $V_{cc}$ , to below Vpor = 100 mV and then raise it after the RES pin load had thoroughly dissipated. To drain the load of the RES pin, attaching a diode to the  $V_{cc}$  side is recommended. The power-on reset function may not work properly if the power supply voltage,  $V_{cc}$ , is raised from a level exceeding 100 mV.



# 17.2.9 Watchdog Timer Characteristics

# Table 17.14 Watchdog Timer Characteristics

 $AV_{cc} = 2.7 V$  to 5.5 V,  $V_{ss} = AV_{ss} = 0.0 V$ , unless otherwise specified

		Applicable	R	ated Va	alues			Test
Item	Symbol	Pins	Min	Тур	Max	Unit	Note	Condition
On-chip oscillator overflow time	t <sub>ovf</sub>		0.2	0.4	_	S	*	$V_{cc} = 5 V$
Note: * When the	on-chip os	cillator is selec	ted, the	e timer o	counts fi	rom 0 to	255, in	dicating the

time remaining until an internal reset is generated.

# **17.3** Operation Timing

Figures 17.1 to 17.6 show timing diagrams.



Figure 17.1 Clock Input Timing







Figure 17.3 Input Timing



Figure 17.4 SCK3 Input Clock Timing





Figure 17.5 SCI3 Synchronous Mode Input/Output Timing



Figure 17.6 UD Pin Minimum Transition Width Timing

# 17.4 Output Load Circuit





# 17.5 Resonator Equivalent Circuit



# Figure 17.8 Resonator Equivalent Circuit (1)

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Figure 17.9 Resonator Equivalent Circuit (2)

# 17.6 Usage Note

The flash memory and mask ROM versions satisfy the electrical characteristics shown in this manual, but actual electrical characteristic values, operating margins, noise margins, and other properties may vary due to differences in manufacturing process, on-chip ROM, layout patterns, and so on.

When system evaluation testing is carried out using the flash memory version, the same evaluation testing should also be conducted for the mask ROM version when changing over to that version.



# Appendix

# A. Instruction Set

# A.1 Instruction List

### **Condition Code**

Symbol	Description
Rd	General destination register
Rs	General source register
Rn	General register
ERd	General destination register (address register or 32-bit register)
ERs	General source register (address register or 32-bit register)
ERn	General register (32-bit register)
(EAd)	Destination operand
(EAs)	Source operand
PC	Program counter
SP	Stack pointer
CCR	Condition-code register
Ν	N (negative) flag in CCR
Z	Z (zero) flag in CCR
V	V (overflow) flag in CCR
С	C (carry) flag in CCR
disp	Displacement
$\rightarrow$	Transfer from the operand on the left to the operand on the right, or transition from the state on the left to the state on the right
+	Addition of the operands on both sides
-	Subtraction of the operand on the right from the operand on the left
×	Multiplication of the operands on both sides
÷	Division of the operand on the left by the operand on the right
^	Logical AND of the operands on both sides
$\vee$	Logical OR of the operands on both sides
$\oplus$	Logical exclusive OR of the operands on both sides

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Symbol	Description
7	NOT (logical complement)
(), < >	Contents of operand

Note: General registers include 8-bit registers (R0H to R7H and R0L to R7L) and 16-bit registers (R0 to R7 and E0 to E7).

# **Condition Code Notation (cont)**

Symbol	Description
$\updownarrow$	Changed according to execution result
*	Undetermined (no guaranteed value)
0	Cleared to 0
1	Set to 1
_	Not affected by execution of the instruction
Δ	Varies depending on conditions, described in notes



# Table A.1Instruction Set

#### 1. Data Transfer Instructions

									le a 1 (by		)								No. Stat	
	Mnemonic	<b>Operand Size</b>	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa		Operation	1	Con H	ditio	n Co	v	с	Normal	Advanced
MOV	MOV.B #xx:8, Rd	В	2									#xx:8 → Rd8	—	-	\$	\$	0	-	2	2
	MOV.B Rs, Rd	В		2								$Rs8 \rightarrow Rd8$	—	—	\$	\$	0	—	2	2
	MOV.B @ERs, Rd	В			2							$@ERs \rightarrow Rd8$	-	-	\$	\$	0	-	2	1
	MOV.B @(d:16, ERs), Rd	В				4						@(d:16, ERs) → Rd8	—	-	\$	\$	0	—	6	6
Í	MOV.B @(d:24, ERs), Rd	В				8						@(d:24, ERs) → Rd8	—	_	\$	\$	0	—	1	0
	MOV.B @ERs+, Rd	в					2					@ERs → Rd8 ERs32+1 → ERs32	-	-	\$	\$	0	-	6	
	MOV.B @aa:8, Rd	В						2				@aa:8 $\rightarrow$ Rd8	-	-	\$	\$	0	-	2	1
	MOV.B @aa:16, Rd	В						4				@aa:16 $\rightarrow$ Rd8	—	—	\$	\$	0	—	6	6
	MOV.B @aa:24, Rd	В						6				@aa:24 $\rightarrow$ Rd8	—	-	\$	\$	0	—	8	3
	MOV.B Rs, @ERd	В			2							$Rs8 \rightarrow @ERd$	—	-	\$	$\updownarrow$	0	—	4	1
	MOV.B Rs, @(d:16, ERd)	В				4						$Rs8 \rightarrow @(d:16, ERd)$	—	—	\$	\$	0	—	6	6
	MOV.B Rs, @(d:24, ERd)	В				8						$Rs8 \rightarrow @(d:24, ERd)$	—	-	\$	$\uparrow$	0	—	1	0
	MOV.B Rs, @-ERd	В					2					$\begin{array}{l} ERd32-1 \rightarrow ERd32 \\ Rs8 \rightarrow @ERd \end{array}$	-	-	\$	\$	0	-	6	6
	MOV.B Rs, @aa:8	В						2				$Rs8 \rightarrow @aa:8$	-	-	\$	$\uparrow$	0	-	4	1
	MOV.B Rs, @aa:16	В						4				Rs8 → @aa:16	—	—	\$	\$	0	—	6	3
	MOV.B Rs, @aa:24	В						6				$Rs8 \rightarrow @aa:24$	—	-	\$	\$	0	-	8	3
	MOV.W #xx:16, Rd	W	4									#xx:16 → Rd16	—	—	\$	\$	0	—	4	1
	MOV.W Rs, Rd	W		2								$Rs16 \rightarrow Rd16$	—	—	\$	\$	0	—	2	2
	MOV.W @ERs, Rd	W			2							@ERs $\rightarrow$ Rd16	—	-	\$	$\updownarrow$	0	—	2	1
	MOV.W @(d:16, ERs), Rd	W				4						@(d:16, ERs) → Rd16	—	-	\$	\$	0	—	6	6
	MOV.W @(d:24, ERs), Rd	W				8						@(d:24, ERs) → Rd16	—	-	\$	\$	0	-	1	0
	MOV.W @ERs+, Rd	w					2					@ERs → Rd16 ERs32+2 → @ERd32	_	-	\$	\$	0	-	6	6
	MOV.W @aa:16, Rd	W						4				@aa:16 $\rightarrow$ Rd16	-	-	\$	\$	0	-	6	6
	MOV.W @aa:24, Rd	w						6				@aa:24 $\rightarrow$ Rd16	-	-	\$	\$	0	-	8	3
	MOV.W Rs, @ERd	W			2							$Rs16 \rightarrow @ERd$	-	-	\$	\$	0	-	4	1
	MOV.W Rs, @(d:16, ERd)	w				4						$Rs16 \rightarrow @(d:16, ERd)$	-	-	\$	\$	0	-	6	6
	MOV.W Rs, @(d:24, ERd)	W				8						$Rs16 \rightarrow @(d:24, ERd)$	-	-	\$	\$	0	—	1	0





									le ai 1 (by		)								No Stat	. of tes <sup>*1</sup>
	Mnemonic	<b>Operand Size</b>	***	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa	I	Operation	1	Con H	ditio	n Co	v	с	Normal	Advanced
MOV	MOV.W Rs, @-ERd	w					2					$ERd32-2 \rightarrow ERd32$ Rs16 $\rightarrow @ERd$	-	-	\$	\$	0	-	(	6
	MOV.W Rs, @aa:16	w						4				Rs16 → @aa:16	_	-	\$	\$	0	-	(	6
	MOV.W Rs, @aa:24	w						6				Rs16 → @aa:24	-	-	\$	1	0	-	8	8
	MOV.L #xx:32, ERd	L	6									#xx:32 → ERd32	-	-	\$	\$	0	-	(	6
	MOV.L ERs, ERd	L		2								$ERs32 \rightarrow ERd32$	-	-	\$	\$	0	-	2	2
	MOV.L @ERs, ERd	L			4							@ERs $\rightarrow$ ERd32	-	-	\$	\$	0	-	8	8
	MOV.L @(d:16, ERs), ERd	L				6						@(d:16, ERs) → ERd32	-	-	\$	\$	0	-	1	0
	MOV.L @(d:24, ERs), ERd	L				10						@(d:24, ERs) → ERd32	-	-	\$	\$	0	-	1	4
	MOV.L @ERs+, ERd	L					4					@ERs → ERd32 ERs32+4 → ERs32	-	-	\$	\$	0	-	1	0
	MOV.L @aa:16, ERd	L						6				@aa:16 $\rightarrow$ ERd32	-	-	\$	\$	0	-	1	0
	MOV.L @aa:24, ERd	L						8				@aa:24 $\rightarrow$ ERd32	-	-	\$	\$	0	-	1	2
	MOV.L ERs, @ERd	L			4							$ERs32 \rightarrow @ERd$	-	-	\$	\$	0	-	8	8
	MOV.L ERs, @(d:16, ERd)	L				6						$ERs32 \rightarrow @(d:16, ERd)$	-	-	\$	\$	0	-	1	0
	MOV.L ERs, @(d:24, ERd)	L				10						$ERs32 \rightarrow @(d:24, ERd)$	-	—	\$	\$	0	1-	1	4
	MOV.L ERs, @-ERd	L					4					$\begin{array}{l} ERd324 \rightarrow ERd32\\ ERs32 \rightarrow @ERd \end{array}$	-	-	\$	\$	0	-	1	0
	MOV.L ERs, @aa:16	L						6				ERs32 → @aa:16	-	-	\$	\$	0	-	1	0
	MOV.L ERs, @aa:24	L						8				$ERs32 \rightarrow @aa:24$	-	-	\$	\$	0	-	1	2
POP	POP.W Rn	W									2	$\begin{array}{l} @SP \to Rn16 \\ SP+2 \to SP \end{array}$	-	-	\$	\$	0	-	(	6
	POP.L ERn	L									4	$\begin{array}{l} @SP \to ERn32 \\ SP+4 \to SP \end{array}$	-	-	\$	\$	0	-	1	0
PUSH	PUSH.W Rn	W									2	$\begin{array}{l} SP-2 \rightarrow SP \\ Rn16 \rightarrow @SP \end{array}$	-	-	\$	\$	0	-	(	6
	PUSH.L ERn	L									4	$SP-4 \rightarrow SP$ ERn32 $\rightarrow @SP$	-	-	\$	\$	0	-	1	0
MOVFPE	MOVFPE @aa:16, Rd	В						4				Cannot be used in this LSI		annc s LS		use	ed ir	ו		
MOVTPE	MOVTPE Rs, @aa:16	В						4				Cannot be used in this LSI		anno s LS		use	ed ir	۱		

# 2. Arithmetic Instructions

						essi tion				nd /tes	)								No Stat	. of tes <sup>*1</sup>
	Mnemonic	Operand Size	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	0 @ aa		Operation		Con	ditio	n Co	v	с	Normal	Advanced
ADD	ADD.B #xx:8, Rd	В	++= 2	-							-	$Rd8+#xx:8 \rightarrow Rd8$	_	п \$	1	∠ ↓	▼ ↓	€ ‡		2
	ADD.B Rs, Rd	В		2								$Rd8+Rs8 \rightarrow Rd8$	-	1	\$	\$	1	1	:	2
	ADD.W #xx:16, Rd	w	4									Rd16+#xx:16 $\rightarrow$ Rd16	-	(1)	\$	\$	1	1	-	4
	ADD.W Rs, Rd	w		2					-			Rd16+Rs16 $\rightarrow$ Rd16	1_	(1)	\$	\$	1	1	1	2
	ADD.L #xx:32, ERd	L	6									ERd32+#xx:32 → ERd32	-	(2)	\$	\$	\$	\$	6	6
	ADD.L ERs, ERd	L		2								ERd32+ERs32 → ERd32	-	(2)	\$	\$	\$	\$	2	2
ADDX	ADDX.B #xx:8, Rd	В	2									$Rd8+#xx:8 + C \rightarrow Rd8$	-	↕	\$	(3)	\$	\$	2	2
	ADDX.B Rs, Rd	В		2								$Rd8+Rs8 + C \rightarrow Rd8$	-	\$	\$	(3)	\$	\$	2	2
ADDS	ADDS.L #1, ERd	L		2								$ERd32+1 \rightarrow ERd32$	-	—	—	-	-	—	2	2
	ADDS.L #2, ERd	L		2								$ERd32+2 \rightarrow ERd32$	-	—	—	-	—	—	2	2
	ADDS.L #4, ERd	L		2								$ERd32+4 \rightarrow ERd32$	—	—	—	—	—	—	2	2
INC	INC.B Rd	В		2								$Rd8+1 \rightarrow Rd8$	-	—	\$	\$	\$	—	2	2
	INC.W #1, Rd	W		2								$Rd16+1 \rightarrow Rd16$	-	—	$\updownarrow$	\$	€	—	2	2
	INC.W #2, Rd	W		2								$Rd16+2 \rightarrow Rd16$	-	—	\$	\$	\$	—	2	2
	INC.L #1, ERd	L		2								$ERd32+1 \rightarrow ERd32$	-	—	\$	\$	\$	—	2	2
	INC.L #2, ERd	L		2								$ERd32+2 \rightarrow ERd32$	-	—	\$	\$	\$	—	2	2
DAA	DAA Rd	В		2								Rd8 decimal adjust $\rightarrow$ Rd8	-	*	\$	\$	*	-	2	2
SUB	SUB.B Rs, Rd	В		2								$Rd8-Rs8 \rightarrow Rd8$	-	\$	\$	\$	\$	\$	2	2
	SUB.W #xx:16, Rd	W	4									Rd16–#xx:16 $\rightarrow$ Rd16	-	(1)	\$	\$	\$	\$	4	4
	SUB.W Rs, Rd	W		2								$Rd16-Rs16 \rightarrow Rd16$	-	(1)	\$	\$	\$	\$	2	2
	SUB.L #xx:32, ERd	L	6									$ERd32-\#xx:32 \rightarrow ERd32$	-	(2)	\$	\$	\$	\$	f	6
	SUB.L ERs, ERd	L		2								$ERd32-ERs32 \rightarrow ERd32$	—	(2)	\$	\$	\$	\$	2	2
SUBX	SUBX.B #xx:8, Rd	В	2									$Rd8-#xx:8-C \rightarrow Rd8$	-	\$	\$	(3)	\$	\$	2	2
	SUBX.B Rs, Rd	В		2								$Rd8-Rs8-C \rightarrow Rd8$	-	\$	\$	(3)	\$	\$	2	2
SUBS	SUBS.L #1, ERd	L		2								$ERd32-1 \rightarrow ERd32$	-	—	—	—	—	—	2	2
	SUBS.L #2, ERd	L		2								$ERd32-2 \rightarrow ERd32$	-	-	—	-	—	—	2	2
	SUBS.L #4, ERd	L		2								$ERd32-4 \rightarrow ERd32$	-	—	—	-	—	—	2	2
DEC	DEC.B Rd	В		2								$Rd8-1 \rightarrow Rd8$	-	-	\$	\$	\$	—	2	2
	DEC.W #1, Rd	W		2								$Rd16-1 \rightarrow Rd16$	-	—	$\updownarrow$	€	\$	—	2	2
	DEC.W #2, Rd	W		2								$Rd16-2 \rightarrow Rd16$	-	-	\$	\$	\$	—	1	2



									le a 1 (by		)								No Stat	. of es <sup>*1</sup>
	Mnemonic	Operand Size	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa		Operation			1	n Co			Normal	Advanced
DEC	DEC.L #1, ERd	L	#	2							-	ERd32–1 $\rightarrow$ ERd32	۱ 	н	N ↓	z ≎	v ≎	c		<b>4</b> 2
DLC	DEC.L #2, ERd	L		2								ERd32–2 $\rightarrow$ ERd32	_	_	↓	↓	↓	_		2
DAS	DAS.Rd	В		2								Rd8 decimal adjust $\rightarrow$ Rd8	-	*	\$	\$	*	-		2
MULXU	MULXU. B Rs, Rd	В		2								$Rd8 \times Rs8 \rightarrow Rd16$ (unsigned multiplication)	—	—	-	—		—	1	4
	MULXU. W Rs, ERd	W		2								$Rd16 \times Rs16 \rightarrow ERd32$ (unsigned multiplication)	—	-	-	-	—	-	2	2
MULXS	MULXS. B Rs, Rd	В		4								$Rd8 \times Rs8 \rightarrow Rd16$ (signed multiplication)	-	-	\$	\$	—	-	1	6
	MULXS. W Rs, ERd	W		4								$Rd16 \times Rs16 \rightarrow ERd32$ (signed multiplication)	_	-	\$	\$	_	-	2	4
DIVXU	DIVXU. B Rs, Rd	В		2								Rd16 $\div$ Rs8 $\rightarrow$ Rd16 (RdH: remainder, RdL: quotient) (unsigned division)			(6)	(7)		_	1	4
	DIVXU. W Rs, ERd	W		2								ERd32 ÷ Rs16 → ERd32 (Ed: remainder, Rd: quotient) (unsigned division)	-	-	(6)	(7)	_	_	2	2
DIVXS	DIVXS. B Rs, Rd	В		4								$Rd16 \div Rs8 \rightarrow Rd16$ (RdH: remainder, RdL: quotient) (signed division)	-	-	(8)	(7)	_	-	1	6
	DIVXS. W Rs, ERd	W		4								$ERd32 \div Rs16 \rightarrow ERd32$ (Ed: remainder, Rd: quotient) (signed division)	-	-	(8)	(7)	_	-	2	4
CMP	CMP.B #xx:8, Rd	В	2									Rd8–#xx:8	—	\$	\$	\$	\$	\$	2	2
	CMP.B Rs, Rd	В		2								Rd8–Rs8	—	\$	\$	€	$\updownarrow$	\$	2	2
	CMP.W #xx:16, Rd	W	4									Rd16-#xx:16	_	(1)	\$	\$	\$	\$	4	1
	CMP.W Rs, Rd	W		2								Rd16-Rs16	_	(1)	\$	\$	$\updownarrow$	\$	2	2
	CMP.L #xx:32, ERd	L	6									ERd32-#xx:32	_	(2)	\$	\$	$\updownarrow$	\$	4	4
	CMP.L ERs, ERd	L		2								ERd32–ERs32		(2)	\$	\$	$\updownarrow$	\$	2	2

							ing Lei			nd ytes	)								No Stat	. of es <sup>*1</sup>
	Mnemonic	Operand Size	~		@ERn	@(d, ERn)	-ERn/@ERn+	aa	@(d, PC)	@ aa		Operation		Con	ditio	n Co	ode		Normal	Advanced
		do	xx#	Rn	0	0	0	@ aa	0	0	Ι		Т	н	N	z	v	с	°	Ad
NEG	NEG.B Rd	В		2								$0-Rd8 \rightarrow Rd8$	—	\$	\$	\$	\$	\$	2	2
	NEG.W Rd	W		2								$0-Rd16 \rightarrow Rd16$	—	\$	\$			2	2	
	NEG.L ERd	L		2								$0-ERd32 \rightarrow ERd32$	—	\$	\$	\$	\$	\$	2	2
EXTU	EXTU.W Rd	W		2								$0 \rightarrow (\text{} \text{of Rd16})$	—	—	0	\$	0	_	2	2
	EXTU.L ERd	L		2								$0 \rightarrow$ ( <bits 16="" 31="" to=""> of ERd32)</bits>	-	-	0	\$	0	-	2	2
EXTS	EXTS.W Rd	W		2								( <bit 7=""> of Rd16) <math>\rightarrow</math> (<bits 15="" 8="" to=""> of Rd16)</bits></bit>	-	-	\$	\$	0	-	1	2
	EXTS.L ERd	L		2								$( of ERd32) \rightarrow$ ( <bits 16="" 31="" to=""> of ERd32)</bits>	-	-	\$	\$	0	-	2	2



# 3. Logic Instructions

							ng I Ler				)								No. Stat	
	Mnemonic	Operand Size	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	0 0 aa	1	Operation	1	Con	ditio	n Co	v	с	Normal	Advanced
AND	AND.B #xx:8, Rd	В	++= 2					•			1	Rd8∧#xx:8 → Rd8	-	- -	1	2	0	с —		2
AND	AND.B Rs, Rd	В	2	2								$Rd8 \land Rs8 \rightarrow Rd8$	_	_	↓ ↓	↓	0	_		2
	AND.W #xx:16, Rd	w	4	-								$Rd16 \neq xx:16 \rightarrow Rd16$	_	_	↓ ↓	↓	0	_		-
	AND.W Rs, Rd	w		2								$Rd16 \land Rs16 \rightarrow Rd16$	_	_	↓	↓	0	_	2	
	AND.L #xx:32, ERd	L	6	-								ERd32 $\wedge$ #xx:32 $\rightarrow$ ERd32	_	_	↓	↓	0	_	6	-
	AND.L ERs, ERd	L		4								$ERd32 \land ERs32 \rightarrow ERd32$	_	_	\$	\$	0	_	4	1
OR	OR.B #xx:8, Rd	В	2									Rd8∨#xx:8 → Rd8	_	_	\$	\$	0	_	2	2
	OR.B Rs, Rd	в		2								$Rd8 \lor Rs8 \rightarrow Rd8$	_	-	\$	\$	0	-	2	2
	OR.W #xx:16, Rd	w	4									Rd16∨#xx:16 → Rd16	_	—	\$	\$	0	—	4	1
	OR.W Rs, Rd	W		2								$Rd16 \lor Rs16 \rightarrow Rd16$	_	—	\$	€	0	—	2	2
	OR.L #xx:32, ERd	L	6									$ERd32 \lor \#xx:32 \rightarrow ERd32$	_	—	\$	€	0	—	6	3
	OR.L ERs, ERd	L		4								$ERd32{\lor}ERs32 \rightarrow ERd32$	—	—	\$	\$	0	—	4	1
XOR	XOR.B #xx:8, Rd	В	2									Rd8⊕#xx:8 → Rd8	_	—	\$	\$	0	—	2	2
	XOR.B Rs, Rd	В		2								Rd8⊕Rs8 → Rd8	—	—	\$	$\uparrow$	0	—	2	2
	XOR.W #xx:16, Rd	W	4									$Rd16 \oplus #xx:16 \rightarrow Rd16$	-	—	\$	€	0	—	4	1
	XOR.W Rs, Rd	W		2								Rd16⊕Rs16 → Rd16	—	—	\$	€	0	—	2	2
	XOR.L #xx:32, ERd	L	6									$ERd32{\oplus} \texttt{\#xx:32} \rightarrow ERd32$	—	—	\$	$\updownarrow$	0	—	6	3
	XOR.L ERs, ERd	L		4								$ERd32{\oplus}ERs32 \rightarrow ERd32$	_	—	↕	$\updownarrow$	0	—	4	1
NOT	NOT.B Rd	В		2								$\neg \text{ Rd8} \rightarrow \text{ Rd8}$	—	-	↕	↕	0	-	2	2
	NOT.W Rd	W		2								$\neg$ Rd16 $\rightarrow$ Rd16	_	_	↕	$\updownarrow$	0	_	2	2
	NOT.L ERd	L		2								$\neg \text{Rd32} \rightarrow \text{Rd32}$	-	—	\$	$\updownarrow$	0	—	2	2

# 4. Shift Instructions

				A Inst			ng Lei				)								No. State	
	Mnemonic	Operand Size	×		@ERn	@(d, ERn)	@-ERn/@ERn+	@ aa	@(d, PC)	@ aa		Operation		Con	ditio	n Co	ode		Normal	Advanced
		ŏ	XX#	R	0	0	0	0	0	0			I	н	N	z	v	с	Ň	Ad
SHAL	SHAL.B Rd	В		2									_	-	↕	$\uparrow$	↕	$\updownarrow$	2	2
	SHAL.W Rd	W		2									—	-	$\uparrow$	\$	\$	$\updownarrow$	2	2
	SHAL.L ERd	L		2								MSB LSB	_	-	$\uparrow$	€	\$	$\updownarrow$	2	2
SHAR	SHAR.B Rd	В		2								r∍[	_	_	$\uparrow$	\$	0	$\updownarrow$	2	2
	SHAR.W Rd	W		2									_	-	$\uparrow$	$\uparrow$	0	$\updownarrow$	2	2
	SHAR.L ERd	L		2								MSB LSB	—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
SHLL	SHLL.B Rd	В		2									—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
	SHLL.W Rd	W		2									—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
	SHLL.L ERd	L		2								MSB LSB	—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
SHLR	SHLR.B Rd	В		2									_	-	$\uparrow$	€	0	$\updownarrow$	2	2
	SHLR.W Rd	W		2								0 → + C	—	—	$\updownarrow$	$\updownarrow$	0	$\updownarrow$	2	2
	SHLR.L ERd	L		2								MSB LSB	—	-	$\updownarrow$	\$	0	$\updownarrow$	2	2
ROTXL	ROTXL.B Rd	В		2									—	—	$\updownarrow$	$\updownarrow$	0	$\updownarrow$	2	2
	ROTXL.W Rd	W		2									—	-	$\updownarrow$	\$	0	$\updownarrow$	2	2
	ROTXL.L ERd	L		2								MSB <del>&lt;</del> LSB	—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
ROTXR	ROTXR.B Rd	В		2									—	—	$\updownarrow$	$\updownarrow$	0	$\updownarrow$	2	2
	ROTXR.W Rd	W		2									—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
	ROTXR.L ERd	L		2								MSB → LSB	—	-	$\updownarrow$	$\uparrow$	0	$\updownarrow$	2	2
ROTL	ROTL.B Rd	В		2									_	-	$\uparrow$	$\updownarrow$	0	$\updownarrow$	2	2
	ROTL.W Rd	W		2									_	-	\$	\$	0	\$	2	2
	ROTL.L ERd	L		2								MSB 🗲 — LSB	_	-	\$	\$	0	$\updownarrow$	2	2
ROTR	ROTR.B Rd	В		2									—	-	\$	\$	0	$\uparrow$	2	2
	ROTR.W Rd	W		2									_	-	\$	\$	0	$\updownarrow$	2	2
	ROTR.L ERd	L		2								MSB ──► LSB	_	—	\$	\$	0	\$	2	2



# 5. Bit-Manipulation Instructions

				A Inst		essi tion	-				)								No Stat	. of tes <sup>*1</sup>
	Mnemonic	<b>Operand Size</b>	#xx	Rn	@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @aa	I	Operation	1	Con H	ditio	n Co	v	с	Normal	Advanced
BSET	BSET #xx:3, Rd	В		2								(#xx:3 of Rd8) ← 1	-	—	—	-	—	-	2	2
	BSET #xx:3, @ERd	В			4							(#xx:3 of @ERd) ← 1	-	—	—	-	-	-	8	8
	BSET #xx:3, @aa:8	В						4				(#xx:3 of @aa:8) ← 1	-	-	_	-	—	-	8	8
	BSET Rn, Rd	В		2								(Rn8 of Rd8) ← 1	—	-	—	-	—	-	2	2
	BSET Rn, @ERd	В			4							(Rn8 of @ERd) ← 1	—	—	—	-	—	-	8	В
	BSET Rn, @aa:8	В						4				(Rn8 of @aa:8) ← 1	-	-	_	-	—	-	8	8
BCLR	BCLR #xx:3, Rd	В		2								(#xx:3 of Rd8) ← 0	_	-	_	-	—	-	2	2
	BCLR #xx:3, @ERd	В			4							(#xx:3 of @ERd) ← 0	-	-	_	-	—	-	8	8
	BCLR #xx:3, @aa:8	В						4				(#xx:3 of @aa:8) ← 0	-	-	_	-	_	-	8	8
	BCLR Rn, Rd	В		2								(Rn8 of Rd8) ← 0	_	-	_	-	_	-	2	2
	BCLR Rn, @ERd	В			4							(Rn8 of @ERd) ← 0	-	-	—	-	_	1_	8	8
	BCLR Rn, @aa:8	В						4				(Rn8 of @aa:8) ← 0	-	-	_	-	_	-	8	8
BNOT	BNOT #xx:3, Rd	В		2								(#xx:3 of Rd8) ← ¬ (#xx:3 of Rd8)	-	-	—	-	-	-	2	2
	BNOT #xx:3, @ERd	В			4							(#xx:3 of @ERd) ← ¬ (#xx:3 of @ERd)	-	-	—	-	—	-	8	8
	BNOT #xx:3, @aa:8	В						4				(#xx:3 of @aa:8) ← ¬ (#xx:3 of @aa:8)	-	-	—	-	—	-	8	8
	BNOT Rn, Rd	В		2								(Rn8 of Rd8) ← ¬ (Rn8 of Rd8)	-	-	-	-	—	-	2	2
	BNOT Rn, @ERd	В			4							(Rn8 of @ERd) ← ¬ (Rn8 of @ERd)	-	-	—	-	—	-	8	8
	BNOT Rn, @aa:8	В						4				(Rn8 of @aa:8) ← ¬ (Rn8 of @aa:8)	-	-	—	-	—	-	8	8
BTST	BTST #xx:3, Rd	в		2								¬ (#xx:3 of Rd8) → Z	_	_	_	\$	_	_	2	2
	BTST #xx:3, @ERd	в			4							¬ (#xx:3 of @ERd) → Z	-	-	—	\$	—	-	6	6
	BTST #xx:3, @aa:8	в						4				¬ (#xx:3 of @aa:8) → Z	-	-	-	\$	—	-	6	6
	BTST Rn, Rd	В		2								¬ (Rn8 of @Rd8) → Z	-	—	_	\$	—	_	2	2
	BTST Rn, @ERd	В			4							¬ (Rn8 of @ERd) → Z	-	-	-	\$	-	-	6	6
	BTST Rn, @aa:8	в						4				¬ (Rn8 of @aa:8) → Z	-	-	-	\$	-	-	6	6
BLD	BLD #xx:3, Rd	В		2								(#xx:3 of Rd8) $\rightarrow$ C	_	_	_	_	-	\$	2	2

						essi tion				nd /tes	)								No Stat	
	Mnemonic	Operand Size	×		@ERn	@(d, ERn)	@-ERn/@ERn+	@ aa	@(d, PC)	@aa		Operation		Con	ditio	n Co	ode		Normal	Advanced
		ŏ	xx#	Rn	0	0	0	0	0	0	Ι		Т	н	Ν	z	v	С	ž	Ac
BLD	BLD #xx:3, @ERd	В			4							(#xx:3 of @ERd) $\rightarrow$ C	-	_	—	-	-	€	6	6
	BLD #xx:3, @aa:8	В						4				(#xx:3 of @aa:8) $\rightarrow$ C	—	-	—	_	-	\$	6	6
BILD	BILD #xx:3, Rd	В		2								$\neg$ (#xx:3 of Rd8) $\rightarrow$ C	—	-	—	-	—	$\uparrow$	2	2
	BILD #xx:3, @ERd	В			4							$\neg$ (#xx:3 of @ERd) $\rightarrow$ C	—	-	—	—	—	$\uparrow$	6	6
	BILD #xx:3, @aa:8	В						4				¬ (#xx:3 of @aa:8) → C	—	-	—	-	-	$\uparrow$	6	6
BST	BST #xx:3, Rd	В		2								$C \rightarrow$ (#xx:3 of Rd8)	-	-	—	-	-	-	2	2
	BST #xx:3, @ERd	В			4							$C \rightarrow$ (#xx:3 of @ERd24)	—	-	—	-	-	—	8	3
	BST #xx:3, @aa:8	В						4				$C \rightarrow$ (#xx:3 of @aa:8)	—	—	—	—	—	—	8	3
BIST	BIST #xx:3, Rd	В		2								$\neg C \rightarrow (\#xx:3 \text{ of } Rd8)$	—	-	—	-	—	—	2	2
	BIST #xx:3, @ERd	В			4							$\neg C \rightarrow (\#xx:3 \text{ of } @ERd24)$	—	—	—	—	—	—	8	3
	BIST #xx:3, @aa:8	В						4				$\neg C \rightarrow (\#xx:3 \text{ of } @aa:8)$	—	-	—	-	-	-	8	3
BAND	BAND #xx:3, Rd	В		2								$C_{\wedge}(\#xx:3 \text{ of } Rd8) \rightarrow C$	—	-	—	—	-	\$	2	2
	BAND #xx:3, @ERd	В			4							$C_{\wedge}(\#xx:3 \text{ of } @ERd24) \rightarrow C$	—	-	—	—	-	\$	6	6
	BAND #xx:3, @aa:8	В						4				$C_{\wedge}(\#xx:3 \text{ of } @aa:8) \rightarrow C$	_	-		_	-	\$	6	6
BIAND	BIAND #xx:3, Rd	В		2								$C_{\wedge} \neg$ (#xx:3 of Rd8) $\rightarrow C$	—	-	—	—	-	\$	2	2
	BIAND #xx:3, @ERd	В			4							$C_{\wedge} \neg$ (#xx:3 of @ERd24) $\rightarrow$ C	_	-	—	-	-	\$	6	6
	BIAND #xx:3, @aa:8	в						4				$C \land \neg$ (#xx:3 of @aa:8) $\rightarrow C$	—	-	—	_	—	\$	6	6
BOR	BOR #xx:3, Rd	В		2								$C_{\vee}(\#xx:3 \text{ of } Rd8) \rightarrow C$	—	-		—	-	\$	2	2
	BOR #xx:3, @ERd	В			4							$C_{\vee}(\#xx:3 \text{ of } @ERd24) \rightarrow C$	_	-		-	-	\$	6	6
	BOR #xx:3, @aa:8	В						4				$C_{\vee}(\#xx:3 \text{ of } @aa:8) \rightarrow C$	—	-	—	—	-	\$	6	6
BIOR	BIOR #xx:3, Rd	В		2								$C \lor \neg$ (#xx:3 of Rd8) $\rightarrow C$	—	-	—	-	-	\$	2	2
	BIOR #xx:3, @ERd	В			4							$C \lor \neg$ (#xx:3 of @ERd24) $\rightarrow C$	_	-	—	—	-	\$	6	6
	BIOR #xx:3, @aa:8	В						4				C∨ ¬ (#xx:3 of @aa:8) → C	_	-	—	—	—	\$	6	3
BXOR	BXOR #xx:3, Rd	В		2								C⊕(#xx:3 of Rd8) → C	-	-	-	-	-	\$	2	2
	BXOR #xx:3, @ERd	В			4							C⊕(#xx:3 of @ERd24) → C	_	-	-	_	-	\$	6	6
	BXOR #xx:3, @aa:8	В						4				C⊕(#xx:3 of @aa:8) $\rightarrow$ C	—	-	-	-	-	\$	6	6
BIXOR	BIXOR #xx:3, Rd	В		2								C⊕ ¬ (#xx:3 of Rd8) → C	_	-	-	-	-	\$	2	2
	BIXOR #xx:3, @ERd	В			4							C⊕ ¬ (#xx:3 of @ERd24) → C	—	-	-	-	-	\$	6	6
	BIXOR #xx:3, @aa:8	в						4				C⊕ ¬ (#xx:3 of @aa:8) → C	_	-	_	_	-	\$	6	6



# 6. Branching Instructions

									le ai (by		)									No Stat	. of es <sup>*1</sup>
	Mnemonic	Operand Size	×		@ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@aa		Oper	ation		Con	ditio	n Co	ode		Normal	Advanced
	1	ŏ	xx#	Rn	0	0	0	0		0			Condition	Т	н	Ν	z	v	С	ž	Ac
Bcc	BRA d:8 (BT d:8)	—							2			If condition	Always	_	-	-	_	-	-	4	4
	BRA d:16 (BT d:16)	—							4			is true then		_	-	—	-	—	-	6	5
	BRN d:8 (BF d:8)	—							2			$PC \leftarrow PC+d$ else next:	Never	_	-	—	-	—	-	4	4
	BRN d:16 (BF d:16)	—							4			eise next,		_	—	-	-	—	-	6	6
	BHI d:8	—							2				C∨ Z = 0	_	—	—	-	—		4	4
	BHI d:16	—							4					_	—	—	-	—		6	6
	BLS d:8	_							2				C∨ Z = 1	_	_	_	_	_	-	4	4
	BLS d:16	—							4					—	-	—	-	-	-	6	6
	BCC d:8 (BHS d:8)	—							2				C = 0	-	-	—	-	—	-	4	4
	BCC d:16 (BHS d:16)	_							4					—	—	—	-	—	-	6	6
	BCS d:8 (BLO d:8)	—							2				C = 1	-	—	—	-	—	-	4	4
	BCS d:16 (BLO d:16)	_							4					—	—	—	-	—	-	6	6
	BNE d:8	—							2				Z = 0	-	—	—	-	—	-	4	4
	BNE d:16	—							4					—	-	—	-	—	-	6	6
	BEQ d:8	_							2				Z = 1	—	—	—	—	—	—	4	4
	BEQ d:16	_							4					—	—	—	-	—	-	6	6
	BVC d:8	_							2				V = 0	-	—	—	-	—	-	4	4
	BVC d:16	_							4					—	—	—	-	—	—	6	6
	BVS d:8	_							2				V = 1	-	-	—	-	—	-	4	4
	BVS d:16	_							4					—	—	—	—	—	—	6	6
	BPL d:8	_							2			1	N = 0	-	-	-	-	-	-	4	4
	BPL d:16	_							4			1		_	-	—	-	—	-	6	6
	BMI d:8	_							2			1	N = 1	-	—	—	-	—	-	4	4
	BMI d:16	—							4			1		-	-	—	-	—	-	6	6
	BGE d:8	_							2			1	N⊕V = 0	-	_	-	-	—	-	4	4
	BGE d:16	—							4			1		-	-	-	-	—	-	6	6
	BLT d:8	_							2			1	N⊕V = 1	-	-	-	-	—	-	4	4
	BLT d:16	—							4			1		-	-	-	-	-	-	6	6
	BGT d:8	_							2			1	Z∨ (N⊕V) = 0	-	-	-	-	-	-	4	4
	BGT d:16	_							4			1		_	_	_	-	_	_	6	6
	BLE d:8	_							2			1	Z∨ (N⊕V) = 1	-	-	-	-	-	-	4	4
	BLE d:16	_							4			1		_	_	_	-	_	-	6	6

							ing l Ler				)									. of tes <sup>*1</sup>
	Mnemonic	Operand Size	×		ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ @ aa		Operation		Con	ditic	on Co	ode		Normal	Advanced
		ŏ	XX#	Rn	0	0	0	0	0	0			Т	н	N	z	v	с	ž	Ad
JMP	JMP @ERn	-			2							$PC \gets ERn$	—	-	-	-	-	—	4	4
	JMP @aa:24	-						4				PC ← aa:24	—	-	-	—	-	—	(	6
	JMP @@aa:8	-								2		PC ← @aa:8	—	-	-	—	-	—	8	10
BSR	BSR d:8	-							2			$PC \rightarrow @-SP$ $PC \leftarrow PC+d:8$	—	-	-	-	-	—	6	8
	BSR d:16	-							4			$PC \rightarrow @-SP$ $PC \leftarrow PC+d:16$	—	-	-	-	-	—	8	10
JSR	JSR @ERn	-			2							$PC \rightarrow @-SP$ $PC \leftarrow ERn$	—	-	-	-	-	-	6	8
	JSR @aa:24	-						4				$PC \rightarrow @-SP$ $PC \leftarrow aa:24$	—	-	-	-	-	-	8	10
	JSR @@aa:8	-								2		$PC \rightarrow @-SP$ $PC \leftarrow @aa:8$	—	-	-	-	-	—	8	12
RTS	RTS	-									2	$PC \leftarrow @SP+$	_	-	-	_	-	—	8	10



# 7. System Control Instructions

					ddre					nd /tes	)								No. Stat	. of es <sup>*1</sup>
	Mnemonic	Operand Size	×	_	ERn	@(d, ERn)	@-ERn/@ERn+	@aa	@(d, PC)	@ aa		Operation		Con	ditio	n Co	ode		Normal	Advanced
	1	õ	XX#	Rn	0	0	0	0	0	0			Т	н	Ν	z	۷	С	ž	¥
RTE	RTE	-										$CCR \leftarrow @SP+$ PC $\leftarrow @SP+$	\$	\$	\$	\$	\$	\$	1	0
SLEEP	SLEEP	-										Transition to power- down state							2	2
LDC	LDC #xx:8, CCR	В	2									#xx:8 → CCR	\$	\$	\$	\$	\$	\$	2	2
	LDC Rs, CCR	В		2								$Rs8 \rightarrow CCR$	\$	\$	\$	\$	\$	\$	2	2
	LDC @ERs, CCR	W			4							$@ERs \rightarrow CCR$	\$	\$	\$	€	$\updownarrow$	\$	6	3
	LDC @(d:16, ERs), CCR	W				6						@(d:16, ERs) → CCR	\$	\$	\$	\$	\$	\$	8	}
	LDC @(d:24, ERs), CCR	W				10						@(d:24, ERs) → CCR	\$	\$	\$	\$	\$	\$	1	2
	LDC @ERs+, CCR	W					4					@ERs → CCR ERs32+2 → ERs32	\$	\$	\$	€	≎	\$	8	3
	LDC @aa:16, CCR	W						6				@aa:16 → CCR	\$	\$	\$	\$	\$	\$	8	}
	LDC @aa:24, CCR	W						8				@aa:24 $\rightarrow$ CCR	\$	\$	\$	\$	$\updownarrow$	\$	1	0
STC	STC CCR, Rd	В		2								$CCR \rightarrow Rd8$							2	2
	STC CCR, @ERd	W			4							$CCR \rightarrow @ERd$							6	3
	STC CCR, @(d:16, ERd)	W				6						$CCR \rightarrow @(d:16, ERd)$							8	}
	STC CCR, @(d:24, ERd)	W				10						$CCR \rightarrow @(d:24, ERd)$							1	2
	STC CCR, @-ERd	W					4					$ERd32-2 \rightarrow ERd32$ $CCR \rightarrow @ERd$							8	3
	STC CCR, @aa:16	W						6				CCR → @aa:16							8	}
	STC CCR, @aa:24	W						8				$CCR \rightarrow @aa:24$							1	0
ANDC	ANDC #xx:8, CCR	В	2									$CCR_{\wedge}\#xx:8 \rightarrow CCR$	\$	\$	\$	\$	\$	\$	2	2
ORC	ORC #xx:8, CCR	В	2									$CCR_{\vee}$ #xx:8 $\rightarrow$ CCR	\$	$\updownarrow$	$\updownarrow$	€	$\updownarrow$	\$	2	2
XORC	XORC #xx:8, CCR	В	2									CCR⊕#xx:8 → CCR	\$	\$	\$	\$	\$	\$	2	2
NOP	NOP	—									2	$PC \leftarrow PC+2$							2	2

#### 8. Block Transfer Instructions

								Moc ngth		nd ytes	)								No. Stat	
	Mnemonic	Operand Size	×		@ERn	@(d, ERn)	-ERn/@ERn+	@aa	@(d, PC)	@aa		Operation		Con	ditio	n Co	ode		Normal	Advanced
		ő	XX#	Rn	0	0	0	0	0	0			I	н	N	z	v	с	ŝ	Ad
EEPMOV	EEPMOV. B										4	$\begin{array}{l} \text{if } R4L \neq 0 \text{ then} \\ \text{repeat}  @R5 \rightarrow @R6 \\ &R5+1 \rightarrow R5 \\ &R6+1 \rightarrow R6 \\ &R4L-1 \rightarrow R4L \\ \text{until} \qquad R4L=0 \\ \\ \text{else next} \end{array}$							8+ 4n <sup>*2</sup>	
	EEPMOV. W										4	$\begin{array}{l} \text{if } R4 \neq 0 \text{ then} \\ \text{repeat}  @R5 \rightarrow @R6 \\ & R5+1 \rightarrow R5 \\ & R6+1 \rightarrow R6 \\ & R4-1 \rightarrow R4 \\ \text{until} \qquad R4=0 \\ \text{else next} \end{array}$							8+ 4n <sup>*2</sup>	

- Notes: 1. The number of states in cases where the instruction code and its operands are located in on-chip memory is shown here. For other cases, see appendix A.3, Number of Execution States.
  - 2. n is the value set in register R4L or R4.
  - (1) Set to 1 when a carry or borrow occurs at bit 11; otherwise cleared to 0.
  - (2) Set to 1 when a carry or borrow occurs at bit 27; otherwise cleared to 0.
  - (3) Retains its previous value when the result is zero; otherwise cleared to 0.
  - (4) Set to 1 when the adjustment produces a carry; otherwise retains its previous value.
  - (5) The number of states required for execution of an instruction that transfers data in synchronization with the E clock is variable.
  - (6) Set to 1 when the divisor is negative; otherwise cleared to 0.
  - (7) Set to 1 when the divisor is zero; otherwise cleared to 0.
  - (8) Set to 1 when the quotient is negative; otherwise cleared to 0.



# A.2 Operation Code Map

# Table A.2 Operation Code Map (1)

		ш	Table A.2 (2)	Table A.2 (2)			BLE											
		ш	ADDX Ta	SUBX Té			BGT	JSR		5								
			4	0,			BLT			Table A.2 (3)								
is 0.	is 1.		MOV	CMP			BGE	BSR										
ΒH	ΒH	U					B	BS	MOV									
t bit of	t bit of	ß	Table A.2 (2)	Table A.2 (2)			BMI		Σ	EEPMOV								
nifican	nifican	A	Table A.2 Table A.2 (2)	Table A.2 Table A.2 (2)			BPL	AML		Table A.2 (2)								
iost sig	tost sig	6					BVS			Table A.2 Table A.2 EEPMOV (2) (2)								
when m	when m	ω	ADD	SUB			BVC	Table A.2 (2)		NOM								
- Instruction when most significant bit of BH is 0.	uction	2	LDC	Table A.2 (2)		MOV.B	BEQ	F	BST BIST	BLD	ADD	ADDX	CMP	SUBX	OR	XOR	AND	MOV
— Instr	- Instruction when most significant bit of BH is 1.	9	ANDC	AND.B			BNE	RTE	AND	BAND BIAND								
	-	2	XORC	XOR.B			BCS	BSR	XOR	BXOR								
te	ļ	4	ORC	OR.B			BCC	RTS	OR	BOR								
2nd byte	BH BL	m	LDC	Table A.2 (2)			BLS	DIVXU		BIST								
byte	IAL	5	STC	Table A.2 (2)			BHI	MULXU		BCLR								
	AH	-	Table A.2 (2)	Table A.2         Table A.2 <thtable a.2<="" th=""> <thtable a.2<="" th=""> <tht< td=""><td></td><td></td><td>BRN</td><td>DIVXU</td><td></td><td>BNOT</td><td></td><td></td><td></td><td></td><td></td><td></td><td></td><td></td></tht<></thtable></thtable>			BRN	DIVXU		BNOT								
ion code		0	don	Table A.2 (2)			BRA	MULXU		BSET								
Instruction code:		AH	0	-	2	e	4	5	9	7	æ	6	A	В	U	۵	ш	ш





2nd byte BH 1st byte AL AH Instruction code:

BL

Ŀ	Table A-2 (3)		INC						EXTS		DEC		T BLE		
ш	N												BGT		
٥	Table A-2 Table A-2 (3) (3)		INC						EXTS		DEC		BLT		
ပ	Table A-2 (3)	ADD		MOV						SUB		CMP	BGE		
В		AL		W	SHAL	SHAR	ROTL	ROTR	NEG	S		ō	BMI		
A													BPL		
6			ADDS		SHAL	SHAR	ROTL	ROTR	NEG		SUB		BVS		
8	SLEEP		AD		Ŷ	SH	RC	RC	ÏZ		้ร		BVC		
7			INC						ЕХТИ		DEC		BEQ		
9													BNE	AND	
5			INC						ЕХТО		DEC		BCS	XOR	
4	LDC/STC												BCC	OR	
ĸ					SHLL	SHLR	ROTXL	ROTXR	NOT				BLS	SUB	
2													BH	CMP	
-					L L	R.	_XL	XR	Т				BRN	ADD	
0	MOV	INC	ADDS	DAA	SHLL	SHLR	ROTXL	ROTXR	NOT	DEC	SUBS	DAS	BRA	MOV	
AH AL	01	OA	OB	OF	10	11	12	13	17	1A	1B	1F	28	79	

#### Table A.2 **Operation Code Map (2)**





#### A.3 Number of Execution States

The status of execution for each instruction of the H8/300H CPU and the method of calculating the number of states required for instruction execution are shown below. Table A.4 shows the number of cycles of each type occurring in each instruction, such as instruction fetch and data read/write. Table A.3 shows the number of states required for each cycle. The total number of states required for execution of an instruction can be calculated by the following expression:

Execution states =  $I \times S_1 + J \times S_2 + K \times S_K + L \times S_L + M \times S_M + N \times S_N$ 

Examples: When instruction is fetched from on-chip ROM, and an on-chip RAM is accessed.

BSET #0, @FF00

From table A.4:  $I=L=2, \quad J=K=M=N=0$ 

From table A.3:  $S_1 = 2$ ,  $S_1 = 2$ 

Number of states required for execution  $= 2 \times 2 + 2 \times 2 = 8$ 

When instruction is fetched from on-chip ROM, branch address is read from on-chip ROM, and on-chip RAM is used for stack area.

JSR @@ 30

From table A.4:  $I=2, \quad J=K=1, \quad L=M=N=0$ 

From table A.3:

$$\mathbf{S}_{\mathrm{I}} = \mathbf{S}_{\mathrm{J}} = \mathbf{S}_{\mathrm{K}} = 2$$

Number of states required for execution =  $2 \times 2 + 1 \times 2 + 1 \times 2 = 8$ 



Execution Status		Ad	ccess Location
(Instruction Cycle)		On-Chip Memory	On-Chip Peripheral Module
Instruction fetch	S,	2	_
Branch address read	S	_	
Stack operation	S <sub>K</sub>	_	
Byte data access	$S_{L}$	_	2 or 3*
Word data access	S <sub>M</sub>	_	_
Internal operation	S <sub>N</sub>		1

Note: \* Depends on which on-chip peripheral module is accessed. See section 16.1, Register Addresses (Address Order).


Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
ADD	ADD.B #xx:8, Rd	1					
	ADD.B Rs, Rd	1					
	ADD.W #xx:16, Rd	2					
	ADD.W Rs, Rd	1					
	ADD.L #xx:32, ERd	3					
	ADD.L ERs, ERd	1					
ADDS	ADDS #1/2/4, ERd	1					
ADDX	ADDX #xx:8, Rd	1					
	ADDX Rs, Rd	1					
AND	AND.B #xx:8, Rd	1					
	AND.B Rs, Rd	1					
	AND.W #xx:16, Rd	2					
	AND.W Rs, Rd	1					
	AND.L #xx:32, ERd	3					
	AND.L ERs, ERd	2					
ANDC	ANDC #xx:8, CCR	1					
BAND	BAND #xx:3, Rd	1					
	BAND #xx:3, @ERd	2			1		
	BAND #xx:3, @aa:8	2			1		
Bcc	BRA d:8 (BT d:8)	2					
	BRN d:8 (BF d:8)	2					
	BHI d:8	2					
	BLS d:8	2					
	BCC d:8 (BHS d:8)	2					
	BCS d:8 (BLO d:8)	2					
	BNE d:8	2					
	BEQ d:8	2					
	BVC d:8	2					
	BVS d:8	2					
	BPL d:8	2					
	BMI d:8	2					
	BGE d:8	2					

### Table A.4 Number of Cycles in Each Instruction



Appendix

Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
Bcc	BLT d:8	2					
	BGT d:8	2					
	BLE d:8	2					
	BRA d:16(BT d:16)	2					2
	BRN d:16(BF d:16)	2					2
	BHI d:16	2					2
	BLS d:16	2					2
	BCC d:16(BHS d:16)	2					2
	BCS d:16(BLO d:16)	2					2
	BNE d:16	2					2
	BEQ d:16	2					2
	BVC d:16	2					2
	BVS d:16	2					2
	BPL d:16	2					2
	BMI d:16	2					2
	BGE d:16	2					2
	BLT d:16	2					2
	BGT d:16	2					2
	BLE d:16	2					2
BCLR	BCLR #xx:3, Rd	1					
	BCLR #xx:3, @ERd	2			2		
	BCLR #xx:3, @aa:8	2			2		
	BCLR Rn, Rd	1					
	BCLR Rn, @ERd	2			2		
	BCLR Rn, @aa:8	2			2		
BIAND	BIAND #xx:3, Rd	1					
	BIAND #xx:3, @ERd	2			1		
	BIAND #xx:3, @aa:8	2			1		
BILD	BILD #xx:3, Rd	1					
	BILD #xx:3, @ERd	2			1		
	BILD #xx:3, @aa:8	2			1		

		Instruction Fetch	Addr. Read	•	Byte Data Access	Word Data Access	Operation
	Mnemonic	I	J	К	L	М	N
BIOR	BIOR #xx:3, Rd	1					
	BIOR #xx:3, @ERd	2			1		
	BIOR #xx:3, @aa:8	2			1		
BIST	BIST #xx:3, Rd	1					
	BIST #xx:3, @ERd	2			2		
	BIST #xx:3, @aa:8	2			2		
BIXOR	BIXOR #xx:3, Rd	1					
	BIXOR #xx:3, @ERd	2			1		
	BIXOR #xx:3, @aa:8	2			1		
BLD	BLD #xx:3, Rd	1					
	BLD #xx:3, @ERd	2			1		
	BLD #xx:3, @aa:8	2			1		
BNOT	BNOT #xx:3, Rd	1					
	BNOT #xx:3, @ERd	2			2		
	BNOT #xx:3, @aa:8	2			2		
	BNOT Rn, Rd	1					
	BNOT Rn, @ERd	2			2		
	BNOT Rn, @aa:8	2			2		
BOR	BOR #xx:3, Rd	1					
	BOR #xx:3, @ERd	2			1		
	BOR #xx:3, @aa:8	2			1		
BSET	BSET #xx:3, Rd	1					
	BSET #xx:3, @ERd	2			2		
	BSET #xx:3, @aa:8	2			2		
	BSET Rn, Rd	1					
	BSET Rn, @ERd	2			2		
	BSET Rn, @aa:8	2			2		
BSR	BSR d:8	2		1			
	BSR d:16	2		1			2
BST	BST #xx:3, Rd	1					
	BST #xx:3, @ERd	2			2		
	BST #xx:3, @aa:8	2			2		



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Appendix

Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
BTST	BTST #xx:3, Rd	1					
	BTST #xx:3, @ERd	2			1		
	BTST #xx:3, @aa:8	2			1		
	BTST Rn, Rd	1					
	BTST Rn, @ERd	2			1		
	BTST Rn, @aa:8	2			1		
BXOR	BXOR #xx:3, Rd	1					
	BXOR #xx:3, @ERd	2			1		
	BXOR #xx:3, @aa:8	2			1		
CMP	CMP.B #xx:8, Rd	1					
	CMP.B Rs, Rd	1					
	CMP.W #xx:16, Rd	2					
	CMP.W Rs, Rd	1					
	CMP.L #xx:32, ERd	3					
	CMP.L ERs, ERd	1					
DAA	DAA Rd	1					
DAS	DAS Rd	1					
DEC	DEC.B Rd	1					
	DEC.W #1/2, Rd	1					
	DEC.L #1/2, ERd	1					
DUVXS	DIVXS.B Rs, Rd	2					12
	DIVXS.W Rs, ERd	2					20
DIVXU	DIVXU.B Rs, Rd	1					12
	DIVXU.W Rs, ERd	1					20
EEPMOV	EEPMOV.B	2			2n+2*1		
	EEPMOV.W	2			2n+2*1		
EXTS	EXTS.W Rd	1					
	EXTS.L ERd	1					
EXTU	EXTU.W Rd	1					
	EXTU.L ERd	1					

		Instruction Fetch	Addr. Read	-	Byte Data Access	Word Data Access	Internal Operation
	Mnemonic	I	J	К	L	Μ	N
INC	INC.B Rd	1					
	INC.W #1/2, Rd	1					
	INC.L #1/2, ERd	1					
JMP	JMP @ERn	2					
	JMP @aa:24	2					2
	JMP @@aa:8	2	1				2
JSR	JSR @ERn	2		1			
	JSR @aa:24	2		1			2
	JSR @@aa:8	2	1	1			
LDC	LDC #xx:8, CCR	1					
	LDC Rs, CCR	1					
	LDC@ERs, CCR	2				1	
	LDC@(d:16, ERs), CCR	3				1	
	LDC@(d:24,ERs), CCR	5				1	
	LDC@ERs+, CCR	2				1	2
	LDC@aa:16, CCR	3				1	
	LDC@aa:24, CCR	4				1	
MOV	MOV.B #xx:8, Rd	1					
	MOV.B Rs, Rd	1					
	MOV.B @ERs, Rd	1			1		
	MOV.B @(d:16, ERs), Rd	2			1		
	MOV.B @(d:24, ERs), Rd	4			1		
	MOV.B @ERs+, Rd	1			1		2
	MOV.B @aa:8, Rd	1			1		
	MOV.B @aa:16, Rd	2			1		
	MOV.B @aa:24, Rd	3			1		
	MOV.B Rs, @Erd	1			1		
	MOV.B Rs, @(d:16, ERd)	2			1		
	MOV.B Rs, @(d:24, ERd)	4			1		
	MOV.B Rs, @-ERd	1			1		2
	MOV.B Rs, @aa:8	1			1		

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#### Appendix

Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
MOV	MOV.B Rs, @aa:16	2			1		
	MOV.B Rs, @aa:24	3			1		
	MOV.W #xx:16, Rd	2					
	MOV.W Rs, Rd	1					
	MOV.W @ERs, Rd	1				1	
	MOV.W @(d:16,ERs), Rd	2				1	
	MOV.W @(d:24,ERs), Rd	4				1	
	MOV.W @ERs+, Rd	1				1	2
	MOV.W @aa:16, Rd	2				1	
	MOV.W @aa:24, Rd	3				1	
	MOV.W Rs, @ERd	1				1	
	MOV.W Rs, @(d:16,ERd)	2				1	
	MOV.W Rs, @(d:24,ERd)	4				1	
MOV	MOV.W Rs, @-ERd	1				1	2
	MOV.W Rs, @aa:16	2				1	
	MOV.W Rs, @aa:24	3				1	
	MOV.L #xx:32, ERd	3					
	MOV.L ERs, ERd	1					
	MOV.L @ERs, ERd	2				2	
	MOV.L @(d:16,ERs), ERd	3				2	
	MOV.L @(d:24,ERs), ERd	5				2	
	MOV.L @ERs+, ERd	2				2	2
	MOV.L @aa:16, ERd	3				2	
	MOV.L @aa:24, ERd	4				2	
	MOV.L ERs,@ERd	2				2	
	MOV.L ERs, @(d:16,ERd)	3				2	
	MOV.L ERs, @(d:24,ERd)	5				2	
	MOV.L ERs, @-ERd	2				2	2
	MOV.L ERs, @aa:16	3				2	
	MOV.L ERs, @aa:24	4				2	
MOVFPE	MOVFPE @aa:16, Rd* <sup>2</sup>	2			1		
MOVTPE	MOVTPE Rs,@aa:16*2	2			1		

	<b>.</b> .	Instruction Fetch	Addr. Read	-	Byte Data Access	Word Data Access	Internal Operation
	Mnemonic	1	J	к	L	Μ	N
MULXS	MULXS.B Rs, Rd	2					12
	MULXS.W Rs, ERd	2					20
MULXU	MULXU.B Rs, Rd	1					12
	MULXU.W Rs, ERd	1					20
NEG	NEG.B Rd	1					
	NEG.W Rd	1					
	NEG.L ERd	1					
NOP	NOP	1					
NOT	NOT.B Rd	1					
	NOT.W Rd	1					
	NOT.L ERd	1					
OR	OR.B #xx:8, Rd	1					
	OR.B Rs, Rd	1					
	OR.W #xx:16, Rd	2					
	OR.W Rs, Rd	1					
	OR.L #xx:32, ERd	3					
	OR.L ERs, ERd	2					
ORC	ORC #xx:8, CCR	1					
POP	POP.W Rn	1				1	2
	POP.L ERn	2				2	2
PUSH	PUSH.W Rn	1				1	2
	PUSH.L ERn	2				2	2
ROTL	ROTL.B Rd	1					
	ROTL.W Rd	1					
	ROTL.L ERd	1					
ROTR	ROTR.B Rd	1					
	ROTR.W Rd	1					
	ROTR.L ERd	1					
ROTXL	ROTXL.B Rd	1					
	ROTXL.W Rd	1					
	ROTXL.L ERd	1					



Appendix

Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
ROTXR	ROTXR.B Rd	1					
	ROTXR.W Rd	1					
	ROTXR.L ERd	1					
RTE	RTE	2		2			2
RTS	RTS	2		1			2
SHAL	SHAL.B Rd	1					
	SHAL.W Rd	1					
	SHAL.L ERd	1					
SHAR	SHAR.B Rd	1					
	SHAR.W Rd	1					
	SHAR.L ERd	1					
SHLL	SHLL.B Rd	1					
	SHLL.W Rd	1					
	SHLL.L ERd	1					
SHLR	SHLR.B Rd	1					
	SHLR.W Rd	1					
	SHLR.L ERd	1					
SLEEP	SLEEP	1					
STC	STC CCR, Rd	1					
	STC CCR, @ERd	2				1	
	STC CCR, @(d:16,ERd)	3				1	
	STC CCR, @(d:24,ERd)	5				1	
	STC CCR,@-ERd	2				1	2
	STC CCR, @aa:16	3				1	
	STC CCR, @aa:24	4				1	
SUB	SUB.B Rs, Rd	1					
	SUB.W #xx:16, Rd	2					
	SUB.W Rs, Rd	1					
	SUB.L #xx:32, ERd	3					
	SUB.L ERs, ERd	1					
SUBS	SUBS #1/2/4, ERd	1					

Instruction	Mnemonic	Instruction Fetch I	Branch Addr. Read J	Stack Operation K	Byte Data Access L	Word Data Access M	Internal Operation N
SUBX	SUBX #xx:8, Rd	1					
	SUBX. Rs, Rd	1					
XOR	XOR.B #xx:8, Rd	1					
	XOR.B Rs, Rd	1					
	XOR.W #xx:16, Rd	2					
	XOR.W Rs, Rd	1					
	XOR.L #xx:32, ERd	3					
	XOR.L ERs, ERd	2					
XORC	XORC #xx:8, CCR	1					

Notes: 1. n: Specified value in R4L. The source and destination operands are accessed n+1 times respectively.

2. It cannot be used in this LSI.



### A.4 Combinations of Instructions and Addressing Modes

### Table A.5 Combinations of Instructions and Addressing Modes

			Addressing Mode											
Functions	Instructions	XX#	Rn	@ERn	@(d:16.ERn)	@(d:24.ERn)	@ERn+/@ERn	@aa:8	@aa:16	@aa:24	@(d:8.PC)	@(d:16.PC)	@ @ aa:8	I
Data	MOV	BWL	BWL	BWL	BWL	BWL	BWL	В	BWL	BWL	_		_	_
transfer	POP, PUSH	_	_	—	_	_	_	_	_	_	_	—	_	WL
instructions	MOVFPE,	_	_	—	_	_	_	_	_	_	_	—	_	_
	MOVTPE													
Arithmetic	ADD, CMP	BWL	BWL	—	_	—	—	_	—	—	—	—	—	_
operations	SUB	WL	BWL	-	—	—	—	_	—	—	_	-	_	—
	ADDX, SUBX	В	В	-	—	—	—	_	—	—	_	-	_	—
	ADDS, SUBS	—	L	-	—	—	—	—	—	—	_	—	_	—
	INC, DEC	—	BWL	_	—	—	—	—	—	—	_	—	_	—
	DAA, DAS	—	В	—	—	—	—	_	—	—	_	—	_	—
	MULXU,	—	BW	—	—	—	—	_	—	—	_	—	_	—
	MULXS,													
	DIVXU,													
	DIVXS													
	NEG	—	BWL	-	—	—	—	_	—	_	_	—	_	—
	EXTU, EXTS	—	WL	-	_	—	—	_	—	_	_	—	_	—
Logical	AND, OR, XOR	—	BWL	-	_	_	—	_	—	_	_		_	—
operations	NOT	—	BWL		_	_	—	_	—	—	_		_	—
Shift operation	ons	—	BWL	—	_	—	—	_	—	—	—	-	—	—
Bit manipulat	tions	—	В	В	—	—	—	В	—	—	_	-	_	—
Branching	BCC, BSR	—	—	-	—	—	—	_	—	—	_	—	_	—
instructions	JMP, JSR	—	—	0	—	—	—	—	—	—	0	0	_	—
	RTS	—	—	-	—	—	—	_	—	0	—	-	0	—
System	RTE	—	—	—	—	—	—	_	—	—	_	—	_	0
control	SLEEP	—	—	—	—	—	—	_	—	—	_	—	_	0
instructions	LDC	В	В	W	W	W	W		W	W	_	—	_	0
	STC	_	В	W	W	W	W		W	W		_	_	
	ANDC, ORC, XORC	В	—	-	—	-	—		—	—	_	-	—	—
	NOP					_								$\vdash$
Dia da da C				-		-	-					-	_	
BIOCK data tr	ansfer instructions	_	_	_	_	_	_	_	_	_	_	_	—	BW



## B. I/O Port Block Diagrams

#### B.1 Block Diagrams of Port 1



Figure B.1(a) Port 1 Block Diagram (Pins P1, and P1,)



Figure B.1(b) Port 1 Block Diagram (Pin P1<sub>3</sub>)



#### B.2 Block Diagrams of Port 3



Figure B.2(a) Port 3 Block Diagram (Pins P3, and P3,)





Figure B.2(b) Port 3 Block Diagram (Pin P3<sub>5</sub>)





Figure B.2(c) Port 3 Block Diagram (Pins P3<sub>4</sub> and P3<sub>3</sub>)











Figure B.2(e) Port 3 Block Diagram (Pin P3<sub>0</sub>)



### B.3 Block Diagrams of Port 4



Figure B.3(a) Port 4 Block Diagram (Pin P4<sub>3</sub>)





Figure B.3(b) Port 4 Block Diagram (Pin P4<sub>2</sub>)





Figure B.3(c) Port 4 Block Diagram (Pin P4<sub>1</sub>)



Figure B.3(d) Port 4 Block Diagram (Pin P4<sub>0</sub>)



#### B.4 Block Diagram of Port 5



Figure B.4 Port 5 Block Diagram



#### B.5 Block Diagram of Port 6



Figure B.5 Port 6 Block Diagram



### B.6 Block Diagram of Port 7



Figure B.6 Port 7 Block Diagram



### B.7 Block Diagram of Port 8



Figure B.7 Port 8 Block Diagram



#### B.8 Block Diagrams of Port 9



Figure B.8(a) Port 9 Block Diagram (Pins P9, and P9,)





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Figure B.8(c) Port 9 Block Diagram (Pins P9<sub>3</sub>)



### B.9 Block Diagram of Port A



Figure B.9 Port A Block Diagram



### B.10 Block Diagrams of Port B



Figure B.10(a) Port B Block Diagram







Figure B.10(b) Port B Block Diagram (Pin PB<sub>0</sub>)





Figure B.10(c) Port B Block Diagram (Pin PB<sub>1</sub>)



## C. Port States in the Different Processing States

#### Table C.1 Port States Overview

Port	Reset	Sleep	Subsleep	Standby	Watch	Subactive	Active
P1 <sub>7</sub> , P1₄, P1	High ₃ impedance	Retained	Retained	High impedance <sup>*</sup>	Retained	Functions	Functions
P3 <sub>7</sub> to P3 <sub>0</sub>	High impedance	Retained	Retained	High impedance <sup>*</sup>	Retained	Functions	Functions
$P4_{3}$ to $P4_{0}$	High impedance	Retained	Retained	High impedance	Retained	Functions	Functions
$P5_7$ to $P5_0$	High impedance	Retained	Retained	High impedance <sup>*</sup>	Retained	Functions	Functions
$P6_7$ to $P6_0$	High impedance	Retained	Retained	High impedance <sup>*</sup>	Retained	Functions	Functions
$P7_{7}$ to $P7_{0}$	High impedance	Retained	Retained	High impedance	Retained	Functions	Functions
P8 <sub>7</sub> to P8 <sub>0</sub>	High impedance	Retained	Retained	High impedance	Retained	Functions	Functions
P9₅ to P9₀	High impedance	Retained	Retained	High impedance <sup>*</sup>	Retained	Functions	Functions
$PA_{_3}$ to $PA_{_0}$	High impedance	Retained	Retained	High impedance	Retained	Functions	Functions
PB <sub>7</sub> to PB <sub>0</sub>	•	•	•	High impedance	High impedance	High impedance	High impedance

Note: \* High level output when MOS pull-up is in on state.

## D. List of Product Codes

### Table D.1 Product Code Lineup

Product Ty	pe			Product Code	Mark Code	Package (Package Code)
H8/38524	524 H8/38524 Flash memory versions		Regular	HD64F38524H	F38524H	80-pin QFP (FP-80A)
Group			specifications	HD64F38524W	F38524W	80-pin TQFP (TFP-80C)
			Wide-range	HD64F38524HW	F38524H	80-pin QFP (FP-80A)
			specifications	HD64F38524WW	F38524W	80-pin TQFP (TFP-80C)
		Mask ROM	Regular	HD64338524H	38524(***)H	80-pin QFP (FP-80A)
		versions	specifications	HD64338524W	38524(***)W	80-pin TQFP (TFP-80C
			Wide-range	HD64338524HW	38524(***)H	80-pin QFP (FP-80A)
			specifications	HD64338524WW	38524(***)W	80-pin TQFP (TFP-80C
	H8/38523	Mask ROM	Regular	HD64338523H	38523(***)H	80-pin QFP (FP-80A)
		versions	specifications	HD64338523W	38523(***)W	80-pin TQFP (TFP-80C
			Wide-range	HD64338523HW	38523(***)H	80-pin QFP (FP-80A)
		specifications	HD64338523WW	38523(***)W	80-pin TQFP (TFP-80C	
	H8/38522	Flash	Regular	HD64F38522H	F38522H	80-pin QFP (FP-80A)
		memory spe versions	specifications	HD64F38522W	F38522W	80-pin TQFP (TFP-80C
		Wide-range		HD64F38522HW	F38522H	80-pin QFP (FP-80A)
			specifications	HD64F38522WW	F38522W	80-pin TQFP (TFP-80C
		Mask ROM	Regular	HD64338522H	38522(***)H	80-pin QFP (FP-80A)
		versions	specifications	HD64338522W	38522(***)W	80-pin TQFP (TFP-80C
			Wide-range	HD64338522HW	38522(***)H	80-pin QFP (FP-80A)
			specifications	HD64338522WW	38522(***)W	80-pin TQFP (TFP-80C
	H8/38521	Mask ROM	Regular	HD64338521H	38521(***)H	80-pin QFP (FP-80A)
		versions	specifications	HD64338521W	38521(***)W	80-pin TQFP (TFP-80C
			Wide-range	HD64338521HW	38521(***)H	80-pin QFP (FP-80A)
			specifications	HD64338521WW	38521(***)W	80-pin TQFP (TFP-80C
	H8/38520	Mask ROM	Regular	HD64338520H	38520(***)H	80-pin QFP (FP-80A)
		versions	specifications	HD64338520W	38520(***)W	80-pin TQFP (TFP-80C
			Wide-range	HD64338520HW	38520(***)H	80-pin QFP (FP-80A)
			specifications	HD64338520WW	38520(***)W	80-pin TQFP (TFP-80C

Note: (\*\*\*) is the ROM code.



## E. Package Dimensions

Dimensional drawings of the packages FP-80A and TFP-80C are shown in figures E.1 and E.2, below.



Figure E.1 FP-80A Package Dimensions



Figure E.2 TFP-80C Package Dimensions



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TCFL	
TCG	
TCRF	
TCSRF	
TCSRW	
TCW	
TDR	
TLC	
ТМА	
TMC	
TMG	
TMW	
TSR	

## W

WEGR7	0
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## Renesas 16-Bit Single-Chip Microcomputer Hardware Manual H8/38524 Group

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