

# μPD48011318 μPD48011336

# 1.1G-BIT Low Latency DRAM-III Common I/O Burst Length of 2

R10DS0012EJ0200 Rev.2.00 Feb 01, 2013

#### **Description**

The  $\mu$ PD48011318 is a 67,108,864-word by 18-bit and the  $\mu$ PD48011336 is a 33,554,432-word by 36-bit synchronous double data rate Low Latency RAM fabricated with advanced CMOS technology using one-transistor eDRAM memory cell.

The Low Latency DRAM-III chip is a 1.1Gbit DRAM capable of a sustained throughput of approximately 43.2 Gbps for burst length of 2 (approximately 51.2 Gbps for applications implementing error correction), excluding refresh overhead and data bus turn-around

With a bus speed of 600 MHz, a burst length of 2, and a tRC of 13.3 ns, the Low Latency DRAM-III chip is capable of achieving this rate when accesses to at least 6 banks of memory are overlapped.

These products are packaged in 180-pin TAPE FBGA.

# **Specification**

- Density: 1,1Gbit
- Organization: 8M words x 18 bits x 8 banks
   4M words x 36 bits x 8 banks
- Operating frequency
   600MHz (MAX.) @ trc=13.3 ns
- trc

13.3 ns tRC ( and 13.3 ns tRFC )

- Burst length: 2
- Address bus

1 cycle DDR address

Package

180-pin FBGA (Ball Array: 1 mm x 1 mm Pitch)

Package size: 18.5 mm x 14 mm

ROHS 6/6 compliance

- Power supply
  - 2.5 V VEXT
  - 1.5 V V<sub>DD</sub>
  - 1.0 V VDDQ
- Refresh command
  - Auto Refresh : 16384 cycles / 2 ms for each bank
  - Overlapped Refresh with DPR# pin
- Operating case temperature: 0 to 95 °C

# **Features**

- 1 cycle 600MHz DDR Muxed Address
- Optional data bus inversion to reduce SSO, SSN, maximum I/O current, and average I/O power
- Training sequence for per-bit deskew
- Selectable Refresh Mode: Auto or Overlapped Refresh
- Programmable PVT-compensated output impedance
- Programmable PVT-compensated on-die input termination
- PLL for improved input jitter tolerance and wide output data valid window

# **Ordering Information**

Part number	Cycle	Clock	Random	<b>Output Supply</b>	Burst	Address	Organization	Package
	Time	Frequency	Cycle	Voltage	Length	Туре	(word x bit)	
				(V <sub>DD</sub> Q)				
	ns	MHz	ns	V				
μPD48011318FF-FH16-FF1-A	1.66	600	13.3	1.0	2	DDR	64 M x 18	180-pin
μPD48011318FF-FH19-FF1-A	1.90	525						TAPE FBGA
μPD48011336FF-FH16-FF1-A	1.66	600					32 M x 36	(18.5 x14)
μPD48011336FF-FH19-FF1-A	1.90	525						Lead-free

#### **Pin Configurations**

# 180-pin TAPE FBGA (18.5 x 14) (Top View) [μPD48011318] ( x18 )

	1	2	3	4	5	6	7	8	9	10	11	12	13
Α	TMS	$V_{DD}$	TCK	VddQ	Vss				Vss	V <sub>DD</sub> Q	VEXT	V <sub>DD</sub>	QK0
В	Vss	DNU, Vss	Vss	DQ6	V <sub>DD</sub>		x18		V <sub>DD</sub>	DNU, Vss	Vss	DQ8	Vss
С	QVLD	V <sub>DD</sub> Q	DNU, Vss	V <sub>DD</sub> Q	DQ4		Burst of 2		DNU, Vss	V <sub>DD</sub> Q	DQ7	V <sub>DD</sub> Q	QK0#
D	Vss	DNU, Vss	Vss	DQ2	Vss		259 sq. mm		Vss	DNU, Vss	Vss	DQ5	Vss
Е	DNU, Vss	VddQ	DNU, Vss	V <sub>DD</sub> Q	DQ0				DNU, Vss	V <sub>DD</sub> Q	DQ1	V <sub>DD</sub> Q	DQ3
F	Vss	DINV0	Vss	A12	Vss				Vss	A13	Vss	DNU, Vss	Vss
G	V <sub>DD</sub>	V <sub>DD</sub> Q	A10	V <sub>DD</sub> Q	A8				A7	V <sub>DD</sub> Q	A11	V <sub>DD</sub> Q	V <sub>DD</sub>
Н	Vss	RST#	Vss	A6	Vss		Top View		Vss	A5	Vss	DM	Vss
J	MF	V <sub>DD</sub> Q	DK0	Vss	V <sub>DD</sub>				СК	Vss	DNU, Vss	V <sub>DD</sub> Q	LBK#
К	VREF	Vss	DK0#	V <sub>DD</sub> Q	V <sub>DD</sub>				CK#	V <sub>DD</sub> Q	DNU, Vss	Vss	VREF
L	TRST#	CS#	Vss	A4	Vss				Vss	А3	Vss	WE#	Vss
М	V <sub>DD</sub>	V <sub>DD</sub> Q	DPR#	$V_{DD}Q$	A2				REF#	V <sub>DD</sub> Q	A9	V <sub>DD</sub> Q	V <sub>DD</sub>
N	Vss	DINV1	Vss	A0	Vss	l			Vss	A1	Vss	DNU, Vss	Vss
Р	DNU, Vss	V <sub>DD</sub> Q	DNU, Vss	V <sub>DD</sub> Q	DQ9				DNU, Vss	V <sub>DD</sub> Q	DQ10	V <sub>DD</sub> Q	DQ12
R	Vss	DNU, Vss	Vss	DQ11	Vss				Vss	DNU, Vss	Vss	DQ14	Vss
Т	QK1#	V <sub>DD</sub> Q	DNU, Vss	V <sub>DD</sub> Q	DQ13				DNU, Vss	V <sub>DD</sub> Q	DQ16	V <sub>DD</sub> Q	TDO
U	Vss	DNU, Vss	Vss	DQ15	V <sub>DD</sub>				V <sub>DD</sub>	DNU, Vss	Vss	DQ17	Vss
٧	QK1	V <sub>DD</sub>	ZQ	V <sub>DD</sub> Q	Vss				Vss	V <sub>DD</sub> Q	VEXT	V <sub>DD</sub>	TDI

DQ0-DQ17: Data inputs / outputREF#: Refresh EnableDINV0-DINV1: Data inversionLBK#: Loopback ModeDM: Write data maskMF: Mirror Function

A0-A13 : Address **TMS** : IEEE 1149.1 Test input CK, CK# : Input clock TDI : IEEE 1149.1 Test input DK0, DK0# : Input data clock **TCK** : IEEE 1149.1 Clock input QVLD : Read Data Valid TDO : IEEE 1149.1 Test output TRST# QK0-QK1, : Output data clock : IEEE 1149.1 Test reset input QK0#-QK1# VREF : HSTL input reference input

ZQ : Output impedance & Input  $V_{DD}$  : Power Supply, 1.5 V nominal termination control  $V_{DD}Q$  : DQ Power Supply, 1.0 V nominal

RST# : Master reset VEXT : Power Supply, 2.5 V nominal

CS# : Chip select Vss : Ground

WE# : Write Enable DNU,  $V_{SS}$  : Must not be used, or must be connected to  $V_{SS}$  DQ0-DQ17 : Data inputs / output DNU,  $V_{DD}Q$  : Must not be used, or must be connected to  $V_{DD}Q$ 

DPR# : Overlapped Refresh

# 180-pin TAPE FBGA (18.5 x 14) (Top View) [μPD48011336] ( x36 )

	1	2	3	4	5	6	7	8	9	10	11	12	13
Α	TMS	$V_{DD}$	TCK	VddQ	Vss				Vss	VddQ	VEXT	$V_{\text{DD}}$	QK0
В	Vss	DQ9	Vss	DQ6	V <sub>DD</sub>		x36		V <sub>DD</sub>	DQ11	Vss	DQ8	Vss
С	QVLD	V <sub>DD</sub> Q	DQ10	V <sub>DD</sub> Q	DQ4		Burst of 2		DQ13	V <sub>DD</sub> Q	DQ7	V <sub>DD</sub> Q	QK0#
D	Vss	DQ12	Vss	DQ2	Vss		259 sq. mm		Vss	DQ15	Vss	DQ5	Vss
Е	DQ16	VddQ	DQ14	V <sub>DD</sub> Q	DQ0				DQ17	VDDQ	DQ1	VDDQ	DQ3
F	Vss	DINV0	Vss	A12	Vss				Vss	A13	Vss	DINV1	Vss
G	V <sub>DD</sub>	V <sub>DD</sub> Q	A10	V <sub>DD</sub> Q	A8				A7	V <sub>DD</sub> Q	A11	V <sub>DD</sub> Q	V <sub>DD</sub>
Н	Vss	RST#	Vss	A6	Vss		Top View		Vss	A5	Vss	DM	Vss
J	MF	V <sub>DD</sub> Q	DK0	Vss	V <sub>DD</sub>				СК	Vss	DK1	V <sub>DD</sub> Q	LBK#
К	VREF	Vss	DK0#	V <sub>DD</sub> Q	V <sub>DD</sub>				CK#	V <sub>DD</sub> Q	DK1#	Vss	VREF
L	TRST#	CS#	Vss	A4	Vss				Vss	A3	Vss	WE#	Vss
М	V <sub>DD</sub>	V <sub>DD</sub> Q	DPR#	V <sub>DD</sub> Q	A2				REF#	V <sub>DD</sub> Q	A9	V <sub>DD</sub> Q	V <sub>DD</sub>
N	Vss	DINV2	Vss	A0	Vss				Vss	A1	Vss	DINV3	Vss
Р	DQ35	V <sub>DD</sub> Q	DQ34	V <sub>DD</sub> Q	DQ18				DQ33	V <sub>DD</sub> Q	DQ19	V <sub>DD</sub> Q	DQ21
R	Vss	DQ32	Vss	DQ20	Vss				Vss	DQ31	Vss	DQ23	Vss
Т	QK1#	V <sub>DD</sub> Q	DQ30	V <sub>DD</sub> Q	DQ22				DQ29	V <sub>DD</sub> Q	DQ25	V <sub>DD</sub> Q	TDO
U	Vss	DQ28	Vss	DQ24	V <sub>DD</sub>				V <sub>DD</sub>	DQ27	Vss	DQ26	Vss
٧	QK1	V <sub>DD</sub>	ZQ	V <sub>DD</sub> Q	Vss				Vss	V <sub>DD</sub> Q	VEXT	V <sub>DD</sub>	TDI

DQ0-DQ35: Data inputs / outputREF#: Refresh EnableDINV0-DINV3: Data inversionLBK#: Loopback ModeDM: Write data maskMF: Mirror Function

A0-A13 : Address TMS : IEEE 1149.1 Test input
CK, CK# : Input clock TDI : IEEE 1149.1 Test input
DK0-DK1, : Input data clock TCK : IEEE 1149.1 Clock input
DK0#-DK1# TDO : IEEE 1149.1 Test output

OVID: : Read Data Valid

QVLD : Read Data Valid TRST# : IEEE 1149.1 Test reset input QK0-QK1, : Output data clock  $V_{\mathsf{REF}}$ : HSTL input reference input QK0#-QK1#  $V_{DD}$ : Power Supply, 1.5 V nominal ZQ : Output impedance & Input  $V_{DD}Q$ : DQ Power Supply, 1.0 V nominal termination control  $V_{EXT}$ : Power Supply, 2.5 V nominal

RST# : Master reset Vss : Ground

CS# : Chip select DNU, Vss : Must not be used, or must be connected to Vss
WE# : Write Enable DNU, VDDQ : Must not be used, or must be connected to VDDQ

DPR# : Overlapped Refresh

Pin Identification (1/2)

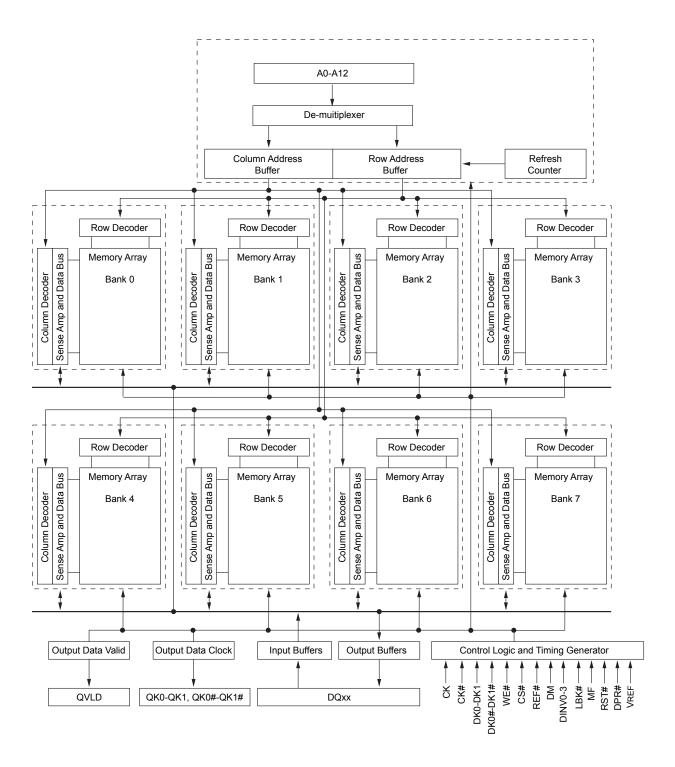
Symbol	Direction	I/O Type	Freq MHz	Description
CK, CK#	Input	1.0 V HSIO	600 / 525	Clock inputs:
				CK and CK# are differential clock inputs. This input clock pair
				registers address and control inputs on the rising edge of CK. CK# is
				ideally 180 degrees out of phase with CK.
DK0-DK1,	Input	1.0 V HSIO	600 / 525	Differential clocks for DQ inputs:
DK0#-DK1#				DK0, DK0# => DQ0-DQ17, DM
				DK1, DK1# => DQ18-DQ35
				Note that DK1,DK1# are only active on x36 parts.
CS#	Input	1.0 V HSIO	600 / 525	Chip select
				CS# enables the commands when CS# is LOW and disables them
				when CS# is HIGH. When the command is disabled, new commands
				are ignored, but internal operations continue.
WE#, REF#	Input	1.0 V HSIO	600 / 525	Write enable, Refresh enable :
				WE#, REF# are sampled at the positive edge of CK, WE#, and REF#
				define (together with CS#) the command to be executed.
RST#	Input	1.0 V HSIO	-	Master reset
		(no ODT)		Note that this pin has no on-die termination.
DPR#	Input	1.0 V HSIO	600 / 525	Overlapped Refresh
LBK#	Input	1.0 V HSIO	-	Loopback mode for control pin de-skew training
		(no ODT)		Note that this pin has no on-die termination.
MF	Input	1.0 V HSIO	DC	Causes "mirroring" of certain pins as described in
		(no ODT)		2.14 Clam-shell support.
				Note that this pin has no on-die termination.
A0-A13	Input	1.0 V HSIO	600 / 525	Address inputs for DDR Address
			DDR	Address bus, including bank select bits.
				A13 is veserved for future use.
DQ0-DQ35	Input	1.0 V HSIO	600 / 525	Data input/output:
	/Output			The DQ signals form the 36 bit data bus. During READ commands,
				the data is referenced to both edges of QKx. During WRITE
				commands, the data is sampled at both edges of DKx.
DINV0-DINV3	Input	1.0 V HSIO	600 / 525	Data inversion state for DQ inputs:
	/Output			DINV0 => DQ0-DQ8
				DINV1 => DQ9-DQ17
				DINV2 => DQ18-DQ26
				DINV3 => DQ27-DQ35
				Note that DINV2-DINV3 are only active on x36 parts
DM	Input	1.0 V HSIO	600 / 525	Write data mask;disables writing of the corresponding data value.
	1			Clocked by DK0.

(2/2)

Symbol	Direction	I/O Type	Freq MHz	Description
QK0-QK1,	Output	1.0 V HSIO	600 / 525	Output data clocks:
QK0#-QK1#				Differential clocks for DQ, QVLD outputs:
				QK0/QK0# => x36: DQ0-DQ17, x18: DQ0-DQ8, QVLD
				QK1/QK1# => x36: DQ18-DQ35, x18: DQ9-DQ17
QVLD	Output	1.0 V HSIO	600 / 525	Data valid;
				The QVLD indicates valid output data. QVLD is edge-aligned with QKx and
				QKx#.
TMS, TDI	Input	1.0 V HSIO		JTAG function pins:
				IEEE 1149.1 test inputs: These balls may be left as no connects if the
				JTAG function is not used in the circuit
TCK	Input	1.0 V HSIO		JTAG function pin;
				IEEE 1149.1 clock input: This ball must be tied to Vss if the JTAG function
				is not used in the circuit.
TDO	Output	1.0 V HSIO		JTAG function pin;
				IEEE 1149.1 test output: JTAG output.
				This ball may be left as no connect if JTAG function is not used.
TRST#	Input	1.0 V HSIO		JTAG reset;
				IEEE 1149.1 test rest: This ball must be tied to Vss if the JTAG function is
				not used in the circuit.
ZQ	Analog			Output impedance and input termination control
V <sub>REF</sub>	Ref			VddQx0.7 V I/O reference voltage.
VEXT	Supply			Power supply; 2.5 V nominal. See Recommended DC Operating
				Conditions for range.
V <sub>DD</sub>	Supply			Power supply; 1.5 V nominal. See Recommended DC Operating
				Conditions for range.
V <sub>DD</sub> Q	Supply			DQ power supply;
				Nominally, 1.0 V. Isolated on the device for improved noise immunity.
				See Recommended DC Operating Conditions for range.
Vss	Supply			Ground
DNU, Vss				Do Not Use, Or must not be used, or must be connected to Vss.
DNU, VDDQ				Do Not Use, Or must not be used, or must be connected to V <sub>DD</sub> Q

 $\label{eq:hsions} \mbox{HSIO} \mbox{ is a single-ended 1.0 V high-side terminated I/O described in \textbf{1. Electrical Specifications}.$ 

# **Block Diagram**



# **Contents**

1.	Elec	trical Specifications	9
2.	Ope	ration	17
	2.1	Interface Overview	17
	2.2	Clocking	17
	2.3	Address bus	18
	2.4	Command encoding	18
	2.5	Data mask	21
	2.6	Data inversion	21
	2.7	Data bus turn-around	21
	2.8	Command cycles	21
	2.9	Write data cycles	21
	2.10	Read data cycles	. 22
	2.11	READ and WRITE command protocol	22
	2.12	Automatic Refresh	.24
	2.13	Overlapped Refresh pin (DPR#)	. 25
	2.14	Clam-shell support	.26
3.		alization	
		Power-on	
		Reset	
		nitial impedance settings	
		Per-bit de-skew training sequence	
	3.5 (	Configuration	.33
4.	JTA	G Specification	. 38
		Fest Pins	
		JTAG AC Test Conditions	
		Boundary Scan	
		ITAG Instructions	
		FAP Controller State Diagram	
	7.0 1	- Controllor Cutto Diagram	. 70
5.	Pacl	kage Drawing	49

#### 1. Electrical Specifications

**Absolute Maximum Ratings** 

Parameter	Symbol	Conditions	Rating	Unit
Supply voltage	VEXT	2.5 V nominal	-0.3 to +2.8	V
Supply voltage	V <sub>DD</sub>	1.5 V nominal	-0.3 to +1.95.	V
Output supply voltage,	V <sub>DD</sub> Q	1.0 V nominal	–0.3 to +1.35	V
Input voltage, Input / Output voltage				
Input / Output voltage	VIH / VIL	1.0 V nominal	–0.3 to +1.35	V
Junction temperature	T <sub>j</sub> MAX.		105	°C
Storage temperature	Tstg		-55 to +125	°C

Caution Exposing the device to stress above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational section of this specification. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

# **Recommended DC Operating Conditions**

 $0^{\circ}C \le T_{C} \le 95^{\circ}C$ 

Parameter	Symbol	MIN.	TYP.	MAX.	Unit	Comments	Note
Supply voltage	VEXT	2.3	2.5	2.7	V		1
Supply voltage	V <sub>DD</sub>	1.395	1.5	1.605	V		1
Output supply voltage	V <sub>DD</sub> Q	0.95	1.0	1.05	٧		1
Reference voltage	VREF	VDDQ*0.69	V <sub>DD</sub> Q*0.7	V <sub>DD</sub> Q*0.71	V		1,2,4
High level output voltage	Voh (DC)	V <sub>DD</sub> Q-0.025			V		1
Low level output voltage	Vol (DC)		V <sub>DD</sub> Q*0.4		V	Z <sub>ol</sub> =40 Ω,	1
						RL=60 Ω	
High level input voltage	VIH (DC)	V <sub>REF</sub> +0.07		V <sub>DD</sub> Q+0.3	V		1,4
Low level input voltage	VIL (DC)	-0.3		VREF-0.07	V		1,4
Clock input voltage	VIn	-0.3		V <sub>DD</sub> Q+0.3	V		
Clock differential voltage	VID	0.2		V <sub>DD</sub> Q+0.6	V		
Output Impedance HIGH	Zoh		60		Ω		3
Output Impedance LOW	Zol		40		Ω		3
Input Impedance LOW	Zi		60		Ω		4

**Notes 1.** All voltage referenced to Vss (GND)

- 2. Peak to Peak AC noise on VREF must not exceed +/- 2 % VDDQ (DC)
- 3. Programmable via ZQ and Reset Impedance Control
- 4. High-side termination (programmable via ZQ and Reset/MRS)

#### **DC Characteristics**

 $0^{\circ}C \leq T_{\text{C}} \leq 95^{\circ}C; \ \ 1.395 \ V \leq V_{\text{DD}} \leq 1.605 \ V, \ unless \ otherwise \ noted$ 

Parameter	Symbol	Test condition	MIN.	MAX.	Unit	Note
Input leakage current	lu		0	50	μΑ	1
Output leakage current	Іьо		0	50	μΑ	1
3 state leakage current	loz		0	50	μΑ	1

Notes 1. Outputs Driver High-Z and ODT Disabled.

# Capacitance (TA = 25 °C, f = 1MHz)

Parameter	Symbol	Test conditions	MIN.	MAX.	Unit
Input capacitance	Cin	V <sub>IN</sub> = 0 V		1.5	pF
(CK,DK, CS#, WE#, REF#, A, DM)					
I/O, Output, Other capacitance	C <sub>I/O</sub>	V <sub>I/O</sub> = 0 V		1.8	pF
(DQ, DINV, QVLD)					
Input capacitance delta between	Срск	V <sub>IN</sub> = 0 V		0.15	pF
differential clock pins					
Input capacitance delta between DQ pins	CDDQ	V <sub>I/O</sub> = 0 V		0.2	pF
Input capacitance delta between CS#,	CDCTL	V <sub>IN</sub> = 0 V		0.2	pF
WE#, REF#, A, DM pins					

**Remark** These parameters are periodically sampled and not 100% tested.

Capacitance is not tested on ZQ pin.

# **Recommended AC Operating Conditions**

 $0^{\circ}C \leq T_{\text{C}} \leq 95^{\circ}C; \ 1.395 \ V \leq V_{\text{DD}} \leq 1.605 \ V,$  unless otherwise noted

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Input HIGH voltage	VIH (AC)		V <sub>REF</sub> +0.13	VDDQ+0.3	V	1
Input LOW voltage	VIL (AC)		-0.3	VREF-0.13	V	1

**Note 1.** Overshoot:  $V_{IH (AC)} \le V_{DD}Q + 0.3 V$  for  $t \le t_{CK}/5$ 

Undershoot: VIL (AC)  $\geq -0.3$  V for  $t \leq t \text{ck/}5$ 

Control input signals may not have pulse widths less than tckH (MIN.) or operate at cycle rates less than tck (MIN.).

# **DC Characteristics**

Parameter	Symbol	Test condition		Max	Unit	Note
Total power consumption	PD	Nominal supply voltage Sequential bank access Overlapped refresh mode Data inversion enabled Half address and data transitions 30% Write and 70% Read operation ODT= $60\Omega$ , Zoh/Zol = $60\Omega/40\Omega$ Tc= $95$ °C	x18 x36	- 2.4	w	1

Note 1. Including all the power supply ( $V_{DD}$ ,  $V_{DD}Q$  and  $V_{EXT}$ )

#### **AC Characteristics**

#### Normal bus timing

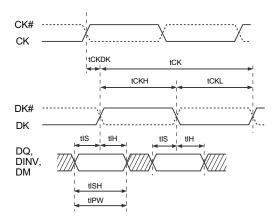
Standard bus timing waveforms and values for command and data are shown in Figure 1-1 through Figure 1-6 and Interface AC Parameters. All timing is measured to/from the crossing point on differential clocks and to/from the VREF crossing point on single-ended signals.

tCK tCKH tCKL CK#

Figure 1-1 Command and DDR Address Input Timing Waveforms

tAS tASH tASH tCS CS#, WE#, REF#, DPR# tCSH

**Figure 1-2 Data Input Timing Waveforms** 



**Figure 1-3 Data Output Timing Waveforms** 

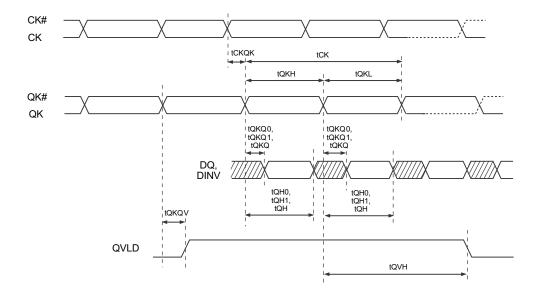


Figure 1-4 Output Driver Enable/Disable Timing

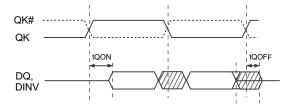
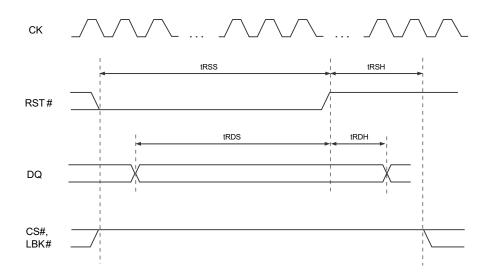


Figure 1-5 Reset Timing Waveforms



**Note** The clock must be with specification and all other chip inputs must be driven to legal values throughout the tass period.

# **Interface AC Parameters**

(1/2)

							(1/2)
Parameter	Symbol		H16		H19	Unit	Note
		( 600	MHz )	( 525	MHz)		
		MIN.	MAX.	MIN.	MAX.		
CK, DK, QK clock period (maximum 600 MHz	tск	1.587	2.500	1.813	2.500	ns	
clock frequency, minimum 400 MHz)							
CK, DK LOW time (assumes 5% duty cycle	<b>t</b> ckL	0.45*		0.45*		tcк(avg)	5
distortion at 600 MHz)							
CK, DK HIGH time (assumed 5% duty cycle	<b>t</b> скн	0.45*		0.45*		tcк(avg)	5
distortion at 600 MHz)							
Clock period jitter	tлт(per)	-0.070	0.070	-0.070	0.070	ns	6,7
Cycle-to-cycle clock jitter	tлт(cc)		0.140		0.140	ns	
A to CK setup	tas	0.213		0.244		ns	2
CK to A hold	tан	0.213		0.244		ns	1
CK to A setup/hold window	<b>t</b> ash	0.213		0.244		ns	4
A input pulse width	<b>t</b> apw	0.266		0.305		ns	
CS#, WE#, REF#, DPR# to CK setup	tcs	0.240		0.275		ns	2
CK to CS#, WE#, REF#, DPR# hold	tсн	0.240		0.275			1
CK to CS#, WE#, REF#, DPR# setup/hold	<b>t</b> csH	0.400		0.458		ns	4
window							
CS#, WE#, REF#, DPR# input pulse width	<b>t</b> cpw	0.533		0.610		ns	
CK to DK skew	<b>t</b> CKDK	-0.266	0.266	-0.304	0.304	ns	
DQ, DINV, DM to DK setup	tıs	0.213		0.244		ns	2
DK to DQ, DINV, DM hold	tıн	0.213		0.244		ns	1
DK to DQ, DINV, DM setup/hold window	<b>t</b> ısh	0.213		0.244		ns	4
DQ, DINV, DM input pulse width	tıpw	0.266		0.305		ns	
Output signal rise time [See note]	tRise	2	5	2	5	V/ns	3
Output signal rise time [See note]	<b>t</b> Fall	2	5	2	5	V/ns	3
QK LOW time	<b>t</b> QKL	0.45*		0.45*		tcк(avg)	5
(assumes 5% duty cycle distortion at 600 MHz)							
QK HIGH time	tqкн	0.45*		0.45*		tcк(avg)	5
(assumed 5% duty cycle distortion at 600 MHz)							
CK to QK skew	<b>t</b> ckqk	-0.400	0.400	-0.458	0.458	ns	
Additional t <sub>RL</sub> delay in loopback	<b>t</b> cD		5		5	ns	
QK0 to DQ[17:0], DINV[1:0] (x36) or DQ[8:0],	takao		0.133		0.152	ns	8
DINV[0] (x18)							
QK0 to DQ[17:0], DINV[1:0] (x36) or DQ[8:0],	tано	0.4*		0.4*		tcк(avg)	5
DINV[0] (x18)							
QK1 to DQ[35:18], DINV[3:2] (x36) or	<b>t</b> QKQ1		0.133		0.152	ns	8
DQ[17:9],DINV[1] (x18)							
QK1 to DQ[35:18], DINV[3:2] (x36) or	<b>t</b> QH1	0.4*		0.4*		tск(avg)	5
DQ[17:9],DINV[1] (x18)							

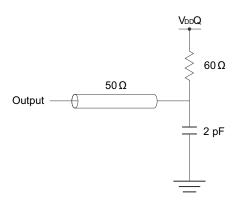
1	1	10	١
l	_	ız	J

Parameter	Symbol	–FI ( 600)	116 MHz )	–FI ( 525l	119 MHz )	Unit	Note
		MIN.	MAX.	MIN.	MAX.		
Any QK to any DQ, DINV	<b>t</b> ακα		0.200		0.288	ns	8
Any QK to any DQ, DINV	<b>t</b> QH	0.35*		0.35*		tск(avg)	5
Any QK to QVLD	<b>t</b> ακαν		0.200		0.288	ns	
Any QK to QVLD	tqvн	0.85*		0.85*		tск(avg)	5
QK to DQ output driver turn-on time	<b>t</b> QON		0.133		0.152	ns	
and ODT turn-on time							
QK to DQ output driver turn-off time	<b>t</b> qoff	-0.1*	0.133	-0.1*	0.152	tск(avg)	5
and ODT turn-on time							
DQ to RST# setup	<b>t</b> RDS	1000 *		1000 *		<b>t</b> cĸ	2
DQ to RST# hold	<b>t</b> RDH	5 *		5 *		<b>t</b> cĸ	1
RST# pulse length	trss	200		200		μs	
RST# deasserted to CS#, DPR#,	<b>t</b> rsh	400000 *		400000 *		<b>t</b> cĸ	
or LBK# asserted							
Time for PLL to stabilize after being	<b>t</b> PLL		20		20	μs	
enabled							
MRS command start to next CS#, DPR#,	<b>t</b> MRD	24 *		24 *		tск	
or LBK# assertion							

Notes 1. All input hold timing assumes rising edge slew rate of 2 V/ns measured from ViL/ViH (DC) to VREF.

- 2. All input setup timing assumes falling edge slew rate of 2 V/ns measured from VREF to VIL/VIH(AC)
- 3. All output timing assumes the load shown in Figure 1-6.
- **4.** Setup/hold windows, tash, tcsh, tish are used for de-skew timing budgeting and are based on electrical simulations. These cannot be directly measured without performing de-skew training.
- 5. tck (avg) is the value of tck averaged over 200 clock cycles.
- **6.** Clock phase jitter is the variance from clock rising edge to the next expected clock rising edge.
- 7. Frequency drift is not allowed.
- 8. taka, takax are guaranteed by design.

Figure 1-6 Output Load



#### **Temperature and Thermal Impedance**

#### **Temperature Limits**

Parameter	Symbol	MIN.	MAX.	Unit	Note
Reliability junction temperature	TJ	0	+105	°C	1
Operating junction temperature	TJ	0	+100	°C	2
Operating case temperature	Tc	0	+95	°C	3

- **Notes 1.** Temperatures greater than 105°C may cause permanent damage to the device. This is a stress rating only and functional operation of the device at or above this is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability of the part.
  - 2. Junction temperature depends upon cycle time, loading, ambient temperature, and airflow.
  - **3.** MAX operating case temperature; Tc is measured in the center of the package. Device functionality is not guaranteed if the device exceeds maximum Tc during operation.

#### Thermal Impedance

Substrate		θja (°C/W)	θjb	θjc		
	Air Flow = 0 m/s	Air Flow = 1 m/s	Air Flow = 2 m/s	(°C/W)	(°C/W)	
4 - Layer	17.5	14.0	13.1	8.2	0.8	

#### **Impedance Controls**

The Low Latency DRAM-III includes programmable impedance control that affects the output impedance of all output and bidirectional pins, as well as the termination of all input and bidirectional pins. The output impedance control affects all output pins and the input termination control affects those pins grouped together as shown in **Table 3-5** and **Table 3-6**. The impedance control affects the drive strengths of both high and low values, as well as the high-side termination impedance. The impedance is controlled via a precision resistor connected between the ZQ and Vss pins. The nominal value of the precision (1%) resistor is 240  $\Omega$ , with a supported range of 200 to 240  $\Omega$ .

The output impedance accuracy is within 15% of the programmed nominal value, with the linearity measured at the points described in **Impedance Test Parameters**. The actual impedance must remain within the range specified by the linear interpolation of the values at those points.

Normally, the input termination and output impedance are continuously adjusted, such that only minor instantaneous variations in the impedance occur. However, when the FZ bit in the configuration register is set, then the input termination and output impedance will be "frozen" at their current values. If the IM bit in the configuration register is set, then the output impedance will have a nominal value that is not dependent on the value of the resistor connected to the ZQ pin. In this mode, the ZQ pin is ignored by the Low Latency DRAM-III, and the output impedances will not be PVT compensated

#### **Impedance Test Parameters**

Parameter	Symbol	MIN.	NORM.	MAX.	Unit
Output HIGH voltage with forced Ioh = -(VDDQ -0.85 V)/ Zoh	Voh	0.723	0.85	0.978	V
Output HIGH voltage with forced Ioh = -(VDDQ -0.7 V) / Zoh	Voh	0.595	0.7	0.805	V
Output LOW voltage with forced Iol = 0.7 V / Zol	Vol	0.595	0.7	0.805	V
Output LOW voltage with forced Iol = 0.55 V / Zol	Vol	0.468	0.55	0.633	V

# 2. Operation

#### 2.1 Interface Overview

The primary Low Latency DRAM-III interface consists of a unidirectional command and address bus and a bidirectional data bus. This type of data bus is often referred to as common I/O or CIO. **Pin Identification** contains the list of pins on the Low Latency DRAM-III.

The command bus is a single data rate (SDR) bus consisting of CS#, WE#, and REF#. The  $\mu$ PD480113xx have a double data rate (DDR) address bus consisting of multiplexed. Both the command bus and address bus are clocked by the differential clock pair CK and CK#. READ and WRITE commands can be issued at a rate of one every cycle in burst length 2 mode.

The data interface is a double-rate (DDR) interface that transfers 36 bits of data Note on each clock edge for 36-bit parts and 18 bits of data on each clock edge for 18-bit parts. The data interface also includes 4 data inversion pins. In addition to 36 data bits, the inputs to the Low Latency DRAM -III include a DDR data mask, DM, and two differential clock pairs, DK0-DK1 and DK0#-DK1#. The Low Latency DRAM -III outputs two differential clock pairs, QK0-QK1 and QK0#-QK1# that are associated with read data, and a data valid signal, QVLD. Because the data bus is bidirectional, idle cycles are required to implement data bus turn-around as described in **2.7 Data bus turn-around**.

**Note** For parts with 18-bit data bus, all references in this section should refer to DQ0-DQ17 and DINV0- DINV1 unless otherwise noted. DQ0-DQn is used to represent DQ0-DQ35 or DQ0-DQ17 for 36-bit and 18-bit parts, respectively, and DINV0-DINVm is used to represent either DINV0-DINV3 or DINV0-DINV1 for 36-bit and 18-bit parts.

# 2.2 Clocking

There are three groups of clock signals: CK/CK#, DK0-DK1/DK0#-DK1#, and QK0-QK1/QK0#-QK1#.

The CK/CK# clock is associated with the address and command pins: A0-A13, CS#, WE#, REF#, and DPR#. At the Low Latency DRAM-III pins, the CK/CK# transitions are nominally centered with respect to address and command signal transitions.

The DK0-DK1/DK0#-DK1# clocks are associated with write data. DK0/DK0# is used as a source-centered clock for the double data rate DQ0-DQ17, DINV0-DINV1, and DM pins. DK1/DK1# is used as a source-centered clock for the double data rate DQ18-DQ35 and DINV2-DINV3 pins.

The DK0-DK1/DK0#-DK1# clocks must meet the specified toxdx skew with respect to the CK/CK# clock in order to ensure proper timing relationship between command and data cycles and to enable proper data bus turn-around. The QK0-QK1/QK0#-QK1# clocks are associated with read data. At the Low Latency DRAM-III pins, and for x36 devices, QK0/QK0# must be source-synchronous with the read data DQ0-DQ17, DINV0-DINV1, and QVLD pins. Similarly, for x36 devices, QK1/QK1# is used as a source-synchronous clock for the read data DQ18-DQ35 and DINV2-DINV3 pins. For x18 devices, QK0/QK0# must be source-synchronous with DQ0-DQ8, DINV0, and QVLD. Similarly, for x18 devices, QK1/QK1# must be source-synchronous with DQ9-DQ17 and DINV1.

The QK0-QK1/QK0#-QK1# clocks must meet the specified tokok skew with respect to the CK/CK# clock in order to ensure proper timing relationship between command and data cycles and to enable proper data bus turn-around.

#### 2.3 Address bus

In Burst of 2 mode one clock cycle is required to load an address. In DDR Address mode address inputs are captured by the RAM in two beats per cycle. Three of the bits in the address field select which of the 8 banks in the RAM will be accessed. Note that a 4th bank address bit has been identified anticipating that a 16 bank version of the RAM might someday reach the market. The bit functions as an ordinary address bit in the devices described in this data sheet.

Device A13 Note1 A12 Note1 Command Beat A11 A10 Α9 **A2** Α1 A0 **A8** Α7 **A6** Α5 Α4 А3 Width **Addres READ or WRITE** x36 0 9.7Gb 2.4Gb Bank s X 4.8Gb 1 **Address Addres** A Note2 READ or WRITE x18 0 4.8Gb Bank 9.7Gb 2.4Gb Address AUTO REFRESH ANY Χ Χ Χ Χ Х Χ Χ X X Bank 0 X X

Table 2-1 Address Bit Encoding

- **Note 1.** A13 are reserved for future expansion. For smaller capacity devices, the value is a don't care, but for devices of the designated capacity or larger, the value will be used.
  - **2.** Address bit A3 in beat zero is used as an address bit in the current devices, but may be used as a bank address bit in future generations of the part.

#### 2.4 Command encoding

The Low Latency DRAM-III supports five types of command cycles as show in Table 2-3.

The NOP command must be used on any cycle where no other commands are requested.

The READ command is used to initiate a burst read from the Low Latency DRAM -III.

The WRITE command is used to initiate a burst write from the Low Latency DRAM -III.

The AUTO REFRESH command is used to initiate a refresh operation on a particular bank of the memory.

The OVERLAPPED REFRESH command is used to initiate a refresh operation, but may be overlapped with other commands.

The MRS command is used to configure the Low Latency DRAM -III following a reset.

Because the overlapped refresh feature can be enabled or disabled, the command decoding and it's associated state diagram have two different forms. **Table 2-2** and **Figure 2-1** show the command encoding for the case when the overlapped refresh feature is disabled. **Figure 2-1** only shows the state diagram for the command decoder; there may be many other states internal to the device.

Because READ and WRITE commands only require a single cycle. When overlapped refresh is enabled, the modified command encoding is described in **Table 2-3** and the state diagram in **Figure 2-2**. In burst length 2 mode, overlapped refresh is accomplished using the DPR# pin, and is independent of other command encodings.

Table 2-2 Command Encoding (Overlapped Refresh Disabled)

State	Command	CS#	WE#	REF#	LBK#	RST#	Length (cycles)	Description
IDLE	NOP	1	-	-	1	1	1	No operation
	READ	0	1	1	1	1	1	Read
	WRITE	0	0	1	1	1	1	Write
	AUTO REFRESH	0	1	0	1	1	1	Auto refresh
	MRS	0	0	0	1	1	2	Mode register set
MRS	NOP	1	-	-	1	1	-	No operation
	LOOPBACK	-	-	-	0	1	-	Address/control loopback
	RESET	-	-	-	-	0	-	Reset

Figure 2-1 Command Decode State Diagram ( Overlapped Refresh Disabled )

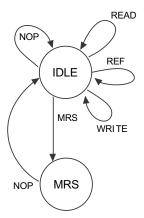
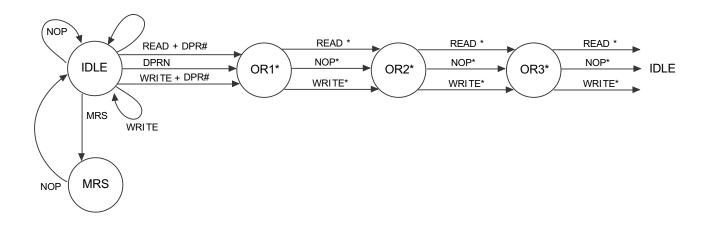


Table 2-3 Command Encoding (Overlapped Refresh Enabled)

State	Command	CS#	WE#	REF#	DPR#	LBK#	RST#	Length (cycles)	Description
IDLE	NOP	1	1	-	1	1	1	1	No operation
	READ	0	1	1	1	1	1	1	Read
	WRITE	0	0	1	1	1	1	1	Write
	OVERLAPPED REFRESH	1	-	-	0	1	1	1	Overlapped refresh
	READ + 0 1 REFRESH		1	1	0	1 1		1	Read and overlapped refresh
	WRITE + REFRESH	0	0	1	0	1	1	1	Write and overlapped refresh
	MRS	0	0	0	1	1	1	2	Mode register set
OR1,	NOP Note	1	-	-	_ Note	1	1	-	No operation
OR2,	READ Note	0	1	1	_ Note	1	1	1	Read (overlapped)
OR3	WRITE Note	0	0	1	_ Note	1	1	1	Write (overlapped)
MRS	NOP	1	-	-	1	1	1	-	No operation
	LOOPBACK	-	1	-	-	0	1	-	Address/control loopback
	RESET	-	-	-	-	-	0	-	Reset

**Note** These indicate that the value of the DPR# pin is ignored for the purpose of command decoding (the pin is used for conveying bank number during this cycle).

Figure 2-2 Command Decode State Diagram (Overlapped Refresh Enabled)



#### 2.5 Data mask

The DM pin is used during write cycles to indicate that the corresponding data are to be masked off from writing to the memory. The value of the DM pin is sampled on both edges of DK0/DK0#, and corresponds to the data sampled on that same edge of either DK0/DK0# or DK1/DK1#.

When the DM pin has a logic 0 value, the data will be written to the memory. When the DM pin has a logic 1 value, the data will be ignored and no update to the corresponding memory data will be made.

The value of the DM pin is ignored in all other circumstances. Thus, driving a constant logic 0 to the DM pin can be used in systems where all data cycles in a burst result in writes to the memory.

When the Auto-DM Function (ADM) bit is set via an MRS command, the value of the DM pin is ignored and the device will behave as if it were a logic 0 on only the clock edge specified by the value of the DMP field.

#### 2.6 Data inversion

In order to reduce simultaneous switch noise, I/O current and average I/O power, the Low Latency DRAM-III provides the ability to invert all data pins. Because the nominal design for Low Latency DRAM-III I/O signals is high-side termination to V<sub>DD</sub>Q, signals driven to a logic high state will consume less power than those in a logic low state.

The DINVm-DINV0 pins indicate whether corresponding DQn-DQ0 pins, in groups of nine, represent the true logic value or an inverted logic value. With the ability to invert DQn-DQ0 pins in groups of nine, each group is guaranteed to be driving no more than five pins low on any given cycle. As a result, no more than five pins in each group can switch in the same direction during each bit time, reducing simultaneous switching noise effects.

When data inversion is enabled (DI=1) and a DINV pin is a logic 1, the corresponding DQ pins have been inverted. When data inversion is disabled (DI=0) then the Atris device will ignore the value of the DINV pins for write data and will drive DINV pins to logic 1 for read data; however, in such a case, the DQ pins are not inverted, regardless of the values of the DINV pins.

#### 2.7 Data bus turn-around

Because the DQn-DQ0 and DINVm-DINV0 pins are bidirectional, care must be taken to ensure that the Low Latency DRAM-III and the system chip connected to it do not drive these pins simultaneously. In order to guarantee this, a rapid DQ and DINV turn-off time is required. The actual bus turn-around time is dependent on many system parameters, including BL, RL, WL, pre-amble, post-amble, controller I/O timing, DRAM I/O timing, PCB delays; and the system must ensure that neither the Low Latency DRAM-III, nor the system chip is driving the bus during that time. This is accomplished by inserting a sufficient number of NOP commands between each READ-to-WRITE command transition and between each WRITE-to-READ transition. Additional NOP commands may be required to allow settling of the bus in order to insure no adverse effect on I/O timing.

Note that a REFRESH command can be used in place of a NOP command to effect the insertion of an idle cycle on the data bus.

# 2.8 Command cycles

A command to the Low Latency DRAM-III is initiated by driving the CS# pin low at the rising edge of CK and simultaneously driving the WE#, REF#, and A0-A13 to the appropriate state corresponding to the command.

READ and WRITE commands require only a single cycle. The NOP and AUTO REFRESH commands are each a single cycle in length. The OVERLAPPED REFRESH command is four cycles in length.

#### 2.9 Write data cycles

DQ0-DQn, DINV0-DINVm, and DM associated with a WRITE command are received by the Low Latency DRAM-III in a DDR burst, starting on a rising edge of DK that is offset WL clock cycles from rising edge of the CK signal corresponding



to the first cycle that the WRITE command was initiated. This delay of WL cycles is intended to ensure that the write data are not driven at the same time that data from a previous READ command are being driven from the Low Latency DRAM-III.

#### 2.10 Read data cycles

DQ0-DQn and DINV0-DINVm associated with a READ command are driven by the Low Latency DRAM-III in a DDR burst by the Low Latency DRAM-III RL clock cycles from the rising edge of the CK signal corresponding to the first cycle that the READ command was initiated. This delay of RL cycles is equal to the delay required for the internal logic and memory within the Low Latency DRAM-III to read data and make it available on the bus.

#### 2.11 READ and WRITE command protocol

Figure 2-3 through Figure 2-4 show the READ and WRITE command sequences.

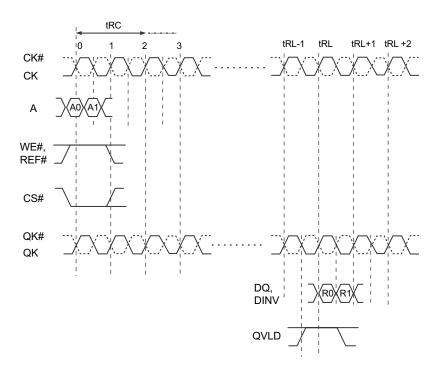


Figure 2-3 READ Command Sequence

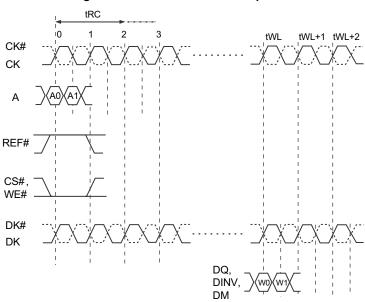


Figure 2-4 WRITE Command Sequence

**Figure 2-5** shows an example of a sequence of 2 READ, 2 WRITE, and 2 READ commands, with the following configuration:

Burst length 2

Speed Config 3: trc=8, trL=12, twL=13

Data inversion is enabled

Some observations of this sequence are:

A NOP is inserted between READ2 and WRITE1 to allow for read-to-write bus turn-around

3 NOPs are inserted between WRITE2 and READ3 to allow for write-to-read bus turn-around Because tRc is 8, READ3 may access the same bank as READ1; however, none of READ2, WRITE1, or WRITE2 may access the same bank as READ1

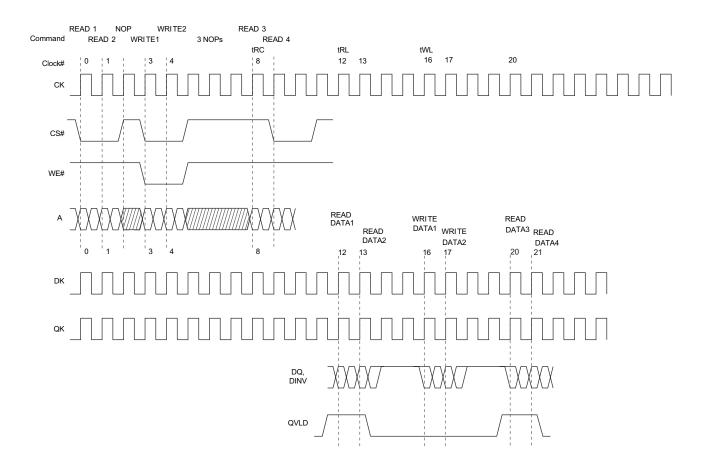


Figure 2-5 READ/WRITE/READ Command Sequence (Speed Config 3, trc=8, trl=12, twl=13, DI=1)

#### 2.12 Automatic Refresh

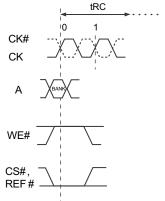
Refresh cycles of the internal memory are initiated via the AUTO REFRESH command. Each refresh command causes a single refresh cycle to occur on the bank specified in the command.

The minimum data retention time is 2 ms over the temperature range specified in **1. Electrical Specifications**, and there are 16384 words in each bank. Therefore, each bank must receive 16384 refresh commands every 2 ms in order to ensure proper data retention.

This means that a refresh command must be received at an average rate of one every 13.3 ns. At a frequency of 600 MHz, this represents 8.2% of the usable bandwidth to the Low Latency DRAM-III.

The refresh command waveforms are shown in Figure 2-6.

Figure 2-6 Automatic Refresh Command Sequence



#### 2.13 Overlapped Refresh pin (DPR#)

In burst length 2 mode, overlapped refresh is accomplished using the DPR# pin, instead of the REF# pin. This extra pin is required, in order to allow READ and WRITE commands to completely consume the standard command bus (pins CS#, WE#, and REF#).

The DPR# pin uses the same protocol as the REF# pin does for burst length 4 mode.

The DPR# pin can only be used in burst length 2 mode and when the MRS register specifies overlapped refresh mode. In all other cases, the DPR# pin must be connected to VDDQ.

An example of Overlapped Refresh overlapped with five READ commands can be seen in Figure 2-7.

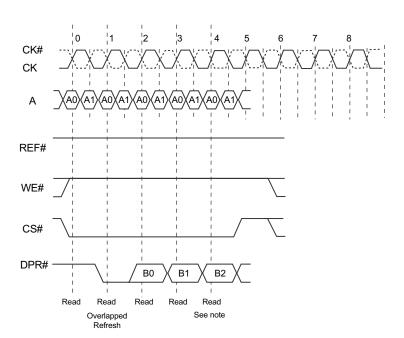


Figure 2-7 Overlapped Refresh with READ Commands

 $\textbf{Note} \ \ \text{tRc for overlapped refresh begins 3 cycles after the start of the command}.$ 

# 2.14 Clam-shell support

In order to support clam-shell mounting on the board, the pin-out of the Low Latency DRAM-III is designed so that most pins that will be interchanged via mirroring are functionally interchangeable. A few pins require their function to be swapped, however. The MF pin causes the pin assignment to be modified to effect those pin swaps. In order to support proper signal integrity, package trace lengths for mirrored pins are matched to within +/- 0.1mm. **Table2-4** shows the pins that are mirrored by the MF pin.

Due to the mirroring of A5 and A6, all devices are required to include an I/O cell on A6 in order to support the mirror function. For lower capacity devices, the A6 pin would remain DNU, VDDQ from a memory controller standpoint, so the system need not actively drive this functionally unused pin.

Similarly, for devices that require A12, pin A13 is also required to include an I/O cell.

A3 and A4 are mirrored to support future devices that may use A3 as an additional bank select bit.

All of A0 through A5 must be mirrored in order for MRS commands to be interpreted correctly.

Table 2-4 Pins Affected by MF Pin

Pin	MF=0	MF=1
N4	A0	A1
N10	A1	A0
L4	A4	A3
L10	A3	A4
M5	A2	REF#
M9	REF#	A2
L2	CS#	WE#
L12	WE#	CS#
H10	A5	A6
H4	A6	A5
F4	A12	A13
F10	A13	A12
M11	A9	DPR#
M3	DPR#	A9
H2	RST#	DM
H12	DM	RST#

#### 3. Initialization

Prior to functional use, the Low Latency DRAM-III must be initialized and configured. The steps described in this chapter will ensure that the internal logic of the Low Latency DRAM-III has been properly reset and that the functional timing parameters of the chip have been configured.

**Figure 3-1** shows the overall initialization sequence required by the Low Latency DRAM-III. An example of an initialization sequence, showing each of the different phases, is shown in **Figure 3-7**.

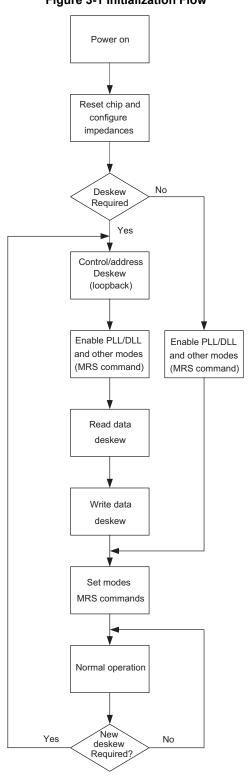


Figure 3-1 Initialization Flow

#### 3.1 Power-on

- 1 There are no restrictions on the sequence of applying the V<sub>DD</sub>, V<sub>EXT</sub>, and V<sub>DD</sub>Q power supplies as long as maintaining RST# = Low and fix MF input state before and during any of the power supplies ramp up.
- 2 In case when RST# can not be asserted Low and fixed MF input state before and during any of the power supplies ramp up, make sure to ramp up following manners.
  - (1) Ramp up VDD and VEXT(no ramp up sequence restriction on VDD and VEXT)
  - (2) Ramp up VDDQ and VREF following to the VDD and VEXT (no ramp up sequence restriction on VDDQ and VREF)
  - (3) Make sure to assert RST# for tRSS specified period before initiating any operations

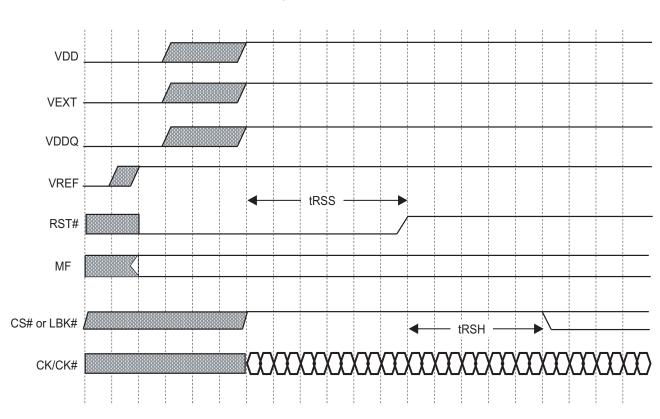
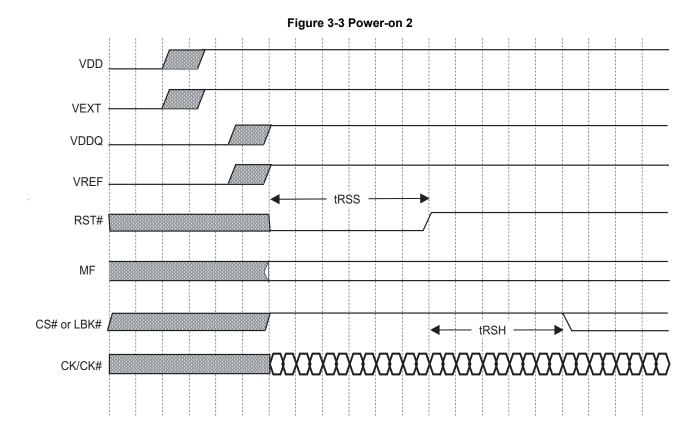


Figure 3-2 Power-on 1



#### 3.2 Reset

Before the Low Latency DRAM-III can be configured, it must be properly reset, to ensure that the Low Latency DRAM-III is in a known and functional state, regardless of any power on anomalies and power supply ramp rates. Reset requires the assertion of the RST# pin for at least 200  $\mu$ s, during which time the device logic is reset to a known state, and the I/O impedances are programmed as described in **3.3 Initial impedance settings**.

The contents of memory are not guaranteed to be retained when the chip is reset.

# 3.3 Initial impedance settings

When the RST# pin is deasserted, the value of specific DQ pins is used to configure the input termination and output impedance as shown in **Table 3-1**. The meaning of the bits is the same as the corresponding bits described in **3.5 Configuration**, although their positions in the DQ pins are unrelated to their positions in the corresponding MRS registers.

While RST# is asserted, the impedance values are continuously updated based on the values of the DQ pins and on the sampled value of the ZQ impedance programming pin. On the rising edge of RST#, the values will be latched into the device and stored in the appropriate positions in the MRS registers. **Figure 3-4** shows the waveforms associated with this function.

Note that the DINV pins are ignored during reset time and the values of the DQ pins are considered to never be inverted during this time.

D12 Pins U12 T11 U4 R12 T5 P13 R4 P11 P5 B12 C11 B4 C5 E13 D4 E11 E5 DQ (x18) 17 16 15 14 13 12 11 10 9 8 7 6 5 4 3 2 0 DQ (x36) 18 7 4 26 25 24 23 22 21 20 19 8 6 5 3 2 0 Reset Reg 17 16 15 14 13 12 11 10 9 8 7 6 5 4 3 2 0  $\mathsf{MSB} \to \mathsf{LSB}$ Cell ID KD KU CD CU DD DU ODT QD QU IM

**Table 3-1 Reset Impedance Control Assignments** 

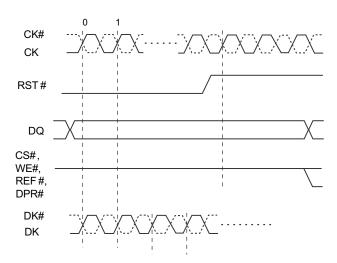


Figure 3-4 Reset Sequence

#### 3.4 Per-bit de-skew training sequence

The Low Latency DRAM-III device provides support that allows a memory controller to de-skew signals for high-speed operation. If per-bit de-skew is desired, the memory controller must provide the actual de-skew functionality. Per-pin de-skew is normally performed after deassertion of the RST# pin, but may be performed at any time, without affecting the contents of the memory within the device. Furthermore, resetting the device does not necessarily affect the skew characteristics of the device pins; therefore, a reset of the device does not necessarily require repeating the de-skew training sequence that was previously performed.

Per-pin de-skew training is anticipated to be implemented in three steps:

De-skew of the control, address, and clock pins: A0-A13, CS#, WE#, REF#, DPR#, and DK0-DK1 with respect to CK De-skew of the read path: DQ0-DQn, DINV0- DINVm and QVLD with respect to QK0-QK1

De-skew of the write path: DQ0-DQn, DINV0- DINVm and DM with respect to DK0-DK1

The first phase, de-skew of the control and related pins, requires support from the device by driving the LBK# pin to a logic 0. When the LBK# pin is a logic 0, the DK0-DK1, CS#, WE#, REF#, DPR#, and A0-A13 pins are looped back to the various DQ and DINV pins. All pins are sampled by the CK clock, including DK0-DK1. The specific mapping of input pins to output pins during loopback mode is shown in **Table 3-2**.

x36 part x18 part Input Output Input Output **Output Pin Input Pin** MF=0 MF=1 MF=0 MF=1 G5 C5 Α8 Α8 DQ4 DQ4 Α8 Α8 H4 E5 Α6 A5 DQ0 Α6 A5 DQ0 DK0, DK0# DK0, DK0# DK0, DK0# DK0, DK0# J1, K1 E11 DQ1 DQ1 L4 P5 A4 A3 DQ18 A4 A3 DQ9 L2 R4 CS# WE# DQ20 CS# WE# DQ11 REF# REF# М5 T5 A2 DQ22 A2 **DQ13** N4 U4 A0 Α1 DQ24 A0 DQ15 A1 H10 E13 Α5 A6 DQ3 Α5 A6 DQ3 DNU, Vss E3 DK1, DK1# DK1, DK1# DNU, Vss DNU, Vss J11, K11 **DQ14** L10 P11 A3 A4 **DQ19** A3 A4 DQ10 CS# CS# L12 P13 WE# DQ21 WE# DQ12 M9 T11 REF# A2 DQ25 REF# A2 DQ16

Table 3-2 Control/Address Pin Mapping When LBK#=0

			x36 part		x18 part				
		Inj	out	Output	Inj	Output			
Input Pin	Output Pin	MF=0	MF=1		MF=0	MF=1			
N10	U12	A1	A0	DQ26	A1	A0	DQ17		
F4	B4	A12	A13	DQ6	A12	A13	DQ6		
G3	D4	A10	A10	DQ2	A10	A10	DQ2		
M3	N2	DPR#	A9	DINV2	DPR#	A9	DINV1		
F10	B12	A13	A12	DQ8	A13	A12	DQ8		
G9	C11	A7	A7	DQ7	A7	A7	DQ7		
G11	D12	A11	A11	DQ5	A11	A11	DQ5		
M11	R12	A9	DPR#	DQ23	A9	DPR#	DQ14		

#### Notes:

DK0, DK0# and DK1, DK1# have differential receivers, so the output of the receiver is used as signal being looped back.

For each pin that is looped back, the input pin is sampled on both the rising edge and falling edges of the corresponding input clock. The value output on the rising edge of the corresponding output clock will be the value that was sampled on the rising edge of the input clock. The value output on the falling edge of the corresponding output clock will be the inverted value that was sampled on the falling edge of the input clock. Data inversion is not used during loopback mode and the value of the corresponding MRS bits are ignored.

The delay from input pins to DQ/DINV output pins, when loopback mode is enabled, is 5 cycles of CK.

In burst length 4 mode, while LBK# pin is a logic 0, the DQ/DINV output pins corresponding to unused input pins (see Table 3-2) will still meet the AC timing specification (See Interface AC Parameters), but they may have arbitrary logic state and may or may not have a constant logic value.

While the LBK# pin is a logic 0, the device will ignore any apparent commands being presented on the CS#, WE#, REF#, and DPR# pins and will perform self refresh operations to prevent loss of data during extended periods of de-skew training. In order to ensure that commands are not inadvertently received during entry and exit from de-skew training, no commands may be sent to the device for a period of 16 cycles of CK prior to the LBK# pin being driven to 0, nor for 16 cycles after the LBK# pin is returned to a logic 1. Also, no valid input signals for looped back must be inserted for 16 cycles after LBK# pin being driven to 0 when device just enters into loopback mode.

While LBK# is asserted, the on-chip PLL is ignored; therefore the phase of the QK output clock will be undefined with respect to the CK input clock. However, the QK to DQ timing will still meet its specifications. Note that the PLL will remain enabled or disabled, per its state prior to LBK# being asserted, and after LBK# is deasserted. An example of a loopback sequence is show in Figure 3-5.

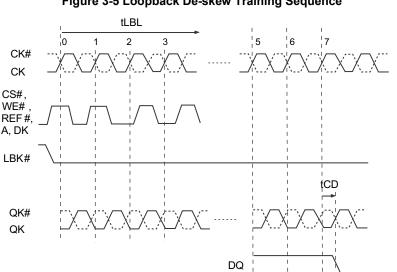


Figure 3-5 Loopback De-skew Training Sequence

Following de-skew of the control, address, and DK pins, the memory controller drives the LBK# pin to a logic 1, then will likely perform read data and write data de-skew training.

Prior to beginning read path de-skew, the memory controller should issue an MRS command to the Low Latency DRAM-III device to enable the PLL. After the PLL is enabled, the timing of the QK, DQ outputs, and QVLD will be unspecified for a period of tpll. The MRS command is described in **3.5 Configuration**. While the PLL is stabilizing, refresh commands must continue to be issued to the device in order to ensure proper retention of data.

Read data de-skew requires that a training pattern be written to Low Latency DRAM-III memory. Complex data patterns can be written to the Low Latency DRAM-III memory using the non-de-skewed DQ signals and the auto-DM functions. Auto-DM functions are enabled through MRS, as described in **3.5 Configuration**.

Read data de-skew can be performed by using the de-skewed control and related pins, including the DK pin, and non-switching DQ and DINV pins to write 36-bit values to memory locations in the device. Because the data are held at constant values, the auto-DM modes of the device are used to specify which 36-bit beat (or 18-bit beat for x18 devices) of data on the bus is to be written to the memory. Once written to the memory, these data can then be read out using standard BL2, READ commands to effect any desired pattern on the DQ bus. This permits the system to de-skew the DQ, DINV, and QVLD signals with respect to the QK clocks.

Write data de-skew can be performed by issuing WRITE commands to the device, then using the de-skewed read bus to determine whether or not data were correctly received by the device. This permits the system to de-skew the DQ, DINV, and DM signals with respect to the DK clocks.

#### 3.5 Configuration

The MRS command is used to configure the Low Latency DRAM-III. Configuration is used to specify the following parameters:

Address Mode

The Address Mode of Low Latency DRAM-III devices from some vendors may be electrically configurable. The default Address Mode may vary from vendor-to-vendor as well. Therefore, systems must configure the desired Address Mode prior to issuing READ or WRITE commands

Mode select for READ latency, WRITE latency, and tRC

The version of the Low Latency DRAM-III covered by this specification only supports modes 0 through 5

Select driver pull-up impedance of 1/4 or 1/6 of reference resistor (default 1/4)

Select driver pull-down impedance of 1/4 or 1/6 reference resistor (default 1/6)

Select DQ, DINV, DM, DK, DKN, CK, CK#, RST# terminator pull-up impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 1/4)

Select DQ, DINV, DM, DK, DKN, CK, CK#, RST# terminator pull-down impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default OFF)

Select CS#, WE#, REF#, A, LBK#, terminator pull-up impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 1/4) Select CS#, WE#, REF#, A, LBK#, terminator pull-down impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default OFF)

Enable/freeze dynamic PVT compensation (default enabled)

Enable/disable PLL (default disabled)

Enable/disable read/write data inversion (default disabled)

Refresh mode (default AUTO REFRESH)

Select auto-DM modes (default disabled)

The MRS command allows the configuration of one of eight different registers to be configured using SDR timing on the address pins. The MRS command waveforms are shown in **Figure 3-5** and the values of the MRS registers are described in **Table 3-3** through **Table 3-6**. Because an MRS command may change values that significantly affect the behavior of the device, no other commands should be issued to the device for a period of tmrD clock cycles following the MRS command.

MRS commands use the address pins specified in **Table 3-3**. The actual pins used depend on the value of the mirror function (MF) pin.

**Table 3-3 MRS Configuration Register Bit Assignments** 

				Beat 0							Beat 2			
Pins (MF=0)	H4	H10	L4	L10	M5	N10	N4	H4	H10	L4	L10	M5	N10	N4
Pins (MF=1)	H10	H4	L10	L4	М9	N4	N10	H10	H4	L10	L4	М9	N4	N10
Address	A6	A5	A4	A3	A2	A1	A0	A6	A5	A4	A3	A2	A1	A0
Mode Register LSB→MSB					M	lode Re	g #							
Active	Х	Configuration		0	0	0	Χ	0	DI	FZ	RM	BL	_	
Active	Х	DMF	)	ADM	0	0	1	Χ	0	0	0	0	0	PLL
Active	Х	KU	_	ODT	0	1	0	Χ	CD		CU	l	KD	_
Active	Х	QD	QU	IM	0	1	1	Χ	DD	_	DU	ļ	0	0
Reserved	Х	0	0	0	1	0	0	Χ	0	0	0	0	0	0
	Х	0	0	0	1	0	1	Х	0	0	0	0	0	0
	Х	0	0	0	1	1	0	Χ	0	0	0	0	0	0
	Х	0	0	0	1	1	1	Х	0	0	0	0	0	0

**Table 3-4 MRS Configuration Bit Definitions** 

Speed Conf	figuration				Clock Cycles				
Config #		Configuration	า	DI	tRC	tRL	tWL		
0	0	0	0	0		Not supported			
1	0	0	1	0		Not supported			
2	0	1	0	0	7	9	10		
3	0	1	1	0	8	10	11		
4	1	0	0	0		Not supported			
5	1	0	1	0		Not supported			
6	1	1	0	0	Not supported				
7	1	1	1	0		Not supported			
0	0	0	0	1		Not supported			
1	0	0	1	1		Not supported			
2	0	1	0	1	7	11	12		
3	0	1	1	1	8	12	13		
4	1	0	0	1	Not supported				
5	1	0	1	1	Not supported				
6	1	1	0	1	Not supported				
7	1	1	1	1		Not supported			

#### Refresh Mode

RM	Operation	
0	Auto Refresh Mode (Default)	
1	Overlapped Refresh/DPR# Mode	

#### **Auto-DM Function**

ADM	Operation
0	Off / Normal Operation (Default)
1	On

Note: When ADM=0, DMP=Don't Care

# Auto-DM Pattern

Burst of 2	DMP			
Pattern #	00	10	01	11
Write Beat 0	1	0	0	0
Write Beat 1	0	1	0	0

Note: Data to be written =1, Date to be masked =0

# **PLL Control**

PLL	Operation		
0	PLL Off / Reset (Default)		
1	PLL On		

# **Burst Length**

BL		Operation	
0	0	Not supported	
0	1	2	
1	0	Reserved	
1	1	Reserved	

Note: Setting 00,10 & 11 not supported

#### Impedance Freeze

FZ	Operation		
0	Update active (Default)		
1	Update frozen		

# **Data Inversion**

DI	Operation		
0	Disable data inversion (Default)		
1	Enable data inversion		

Table 3-5 MRS Input Termination Bit Definitions

Pins
DQ, DIV,
DM,
DK1, DK1#
DK0, DK0#

Data Input On-Die Termination - Up					
ODT	DU		Div	200	240
1	Χ	Χ	Off	8	8
Χ	1	1	Off	8	8
0	1	0	1	200	240
0	0	1	1/2	100	120
0	0	0	1/4	50	60

Data Input On-Die Termination						
RQ Ohms	ODT	DD	DD		20	
	1	Χ	Χ	Off	8	
	Χ	1	1	Off	8	
	0	1	0	1	~	
	0	0	1	1/2	~	
	0	0	0	1/4	∞	

	DOWII	_
00	240	RQ Ohms
0	∞	
0	∞	
0	8	(Not Supported)
0	∞	(Not Supported)

Pins
Address,
CS#, WE#,
REF#

Control Input On-Die Termination - Up						
ODT	CU		Div	200	240	
1	Χ	Χ	Off	∞	8	
Χ	1	1	Off	8	∞	
0	1	0	1	200	240	
0	0	1	1/2	100	120	
0	0	0	1/4	50	60	

Control Input On-Die Termination - Down							
RQ Ohms	ODT	CD	)	Div	200	240	
	1	Χ	Χ	Off	∞	8	
	Χ	1	1	Off	∞	8	
	0	1	0	1	∞	8	1)
	0	0	1	1/2	∞	8	1)
	0	0	0	1/4	∞	8	(1

RQ Ohms
(Not Supported)
(Not Supported)

(Not Supported)

**RQ Ohms** 

(Not Supported)

Pins			
CK, CK#			

Clock Input On-Die Termination - Up						
ODT	KU		Div	200	240	
1	Χ	Χ	Off	∞	8	
Χ	1	1	Off	8	8	
0	1	0	1	200	240	
0	0	1	1/2	100	120	
0	0	0	1/4	50	60	
Note: Interest DO in college						

**Clock Input On-Die Termination - Down RQ Ohms** ODT Div 200 KD 240 Χ Χ Off ∞ Χ 1 1 Off 1 0 1 ∞ 0 0 1/2 1 0 0 0 1/4

(Not Supported) (Not Supported) (Not Supported)

**RQ Ohms** 

Note . Internal RQ is set to  $240\Omega$  when IM=1.

**Table 3-6 MRS Output Impedance Bit Definitions** 

**RQ Ohms** 

Pins			
DQ, DINV,			
QK0, QK0#			
QK1, QK1#			
QVLD			

Output Driver Impedance - Up						
IM	QU	Div	200	240		
0	1	1/4	50	60		
1	X	Х	60*	60*		
0	0	1/6	NA	40.0		

Output Driver Impedance - Down						
IM	QD	Div	200	240		
0	1	1/4	50	60		
1	Х	X	40*	40*		
0	0	1/6	33.3	40.0		

Note:  $60^* = 60 \Omega$  +/- 20% no PVT compensation Note:  $40^* = 40 \Omega$  +/- 20% no PVT compensation

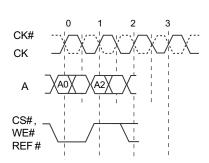
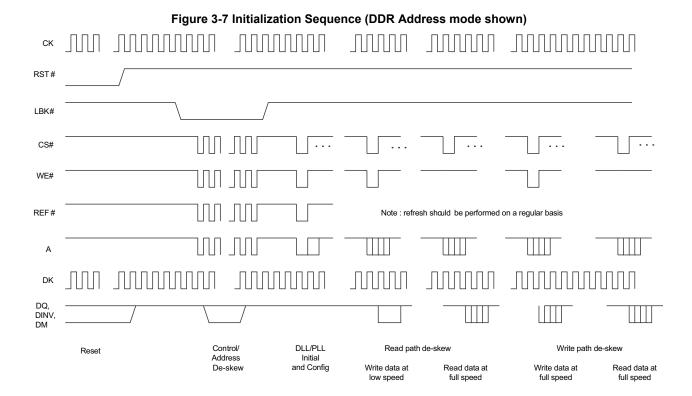


Figure 3-6 MRS Sequence



# 4.JTAG Specification

These products support a limited set of JTAG functions as in IEEE standard 1149.1.

#### 4.1 Test Pins

Table 4-1 Low Latency DRAM-III Test Pins

Pin Name	Test Use	Functional Mode Value	Pull Up/Down in Chip	Pull Up/Down on Board
TDI	JTAG serial chain input	1	1	1
TDO	JTAG serial chain output	1		
TCK	JTAG clock	0	0	0
TMS	JTAG mode	1	1	1
TRST#	JTAG reset	0	0	0

Table 4-2. Test Access Port (TAP) Pins

Pin Name	Pin Assignment	Description
TCK	3A	Test Clock Input. All input are captured on the rising edge of TCK and all
		outputs propagate from the falling edge of TCK.
TMS	1A	Test Mode Select. This is the command input for the TAP controller state machine.
TDI	13V	Test Data Input. This is the input side of the serial registers placed between TDI
		and TDO. The register placed between TDI and TDO is determined by the state
		of the TAP controller state machine and the instruction that is currently loaded in
		the TAP instruction.
TDO	13T	Test Data Output. This is the output side of the serial registers placed between
		TDI and TDO. Output changes in response to the falling edge of TCK.
TRST#	1L	Test Reset Input. This is active-low input asynchronously reset the TAP controller.
		This pin should be driven low or tied to GND when JTAG function is not in use.

Table 4-3. JTAG DC Characteristics (0°C ≤ Tc ≤ 95°C, 1.395 V ≤ VDD ≤ 1.605 V, unless otherwise noted)

Parameter	Symbol	Conditions	MIN.	MAX.	Unit
JTAG Input leakage current	lu	$0 \text{ V} \leq V_{IN} \leq V_{DD}$	-100	+100	μΑ
JTAG I/O leakage current	ILO	$0 \text{ V} \leq V_{IN} \leq V_{DD}Q$ ,	-100	+100	μΑ
		Outputs disabled			
JTAG input HIGH voltage	VIH (DC)		V <sub>REF</sub> + 0.15	V <sub>DD</sub> + 0.3	V
JTAG input LOW voltage	VIL (DC)		VssQ - 0.3	V <sub>REF</sub> - 0.15	V
JTAG output HIGH voltage	Voh (DC)	Ioнc   = 100 μA	V <sub>DD</sub> Q - 0.25		V
JTAG output LOW voltage	Vol (DC)	lo <sub>LC</sub> = 100 μA		0.25	V

Note 1. All voltages referenced to Vss (GND).

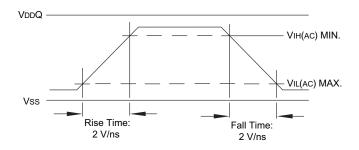
Table 4-4. JTAG Capacitance

Parameter	Symbol	Teat Conditions	MIN.	MAX.	Unit
JTAG Input Capacitance	CINJ	V <sub>IN</sub> = 0V		2.5	pF
(TCK, TMS and TRST#)					
JTAG I/O leakage current	Сол	V <sub>IO</sub> = 0V		2.5	pF
(TDO)					

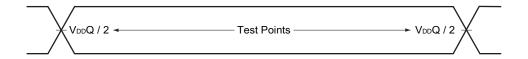
Note 1. Including package.

## 4.2 JTAG AC Test Conditions

# Input waveform (Rise / Fall time ≤ 0.3 ns)



# **Output waveform**



# **Output load condition**

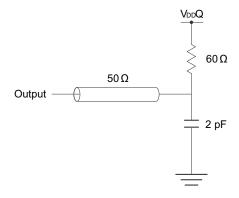


Table 4-5. JTAG AC Characteristics (0°C  $\leq$  Tc  $\leq$  95°C)

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Clock						
Clock cycle time	tтнтн		20		ns	
Clock frequency	f⊤⊧			50	MHz	
Clock HIGH time	tтнтL		10		ns	
Clock LOW time	tтьтн		10		ns	
Output time	1					
TCK LOW to TDO unknown	<b>t</b> TLOX		0		ns	
TCK LOW to TDO valid	<b>t</b> tlov			10	ns	
Setup time	1					
TMS setup time	tмvтн		5		ns	
TDI valid to TCK HIGH	tоvтн		5		ns	
TRST High Level to	tкvтн		50		ns	
valid riseing edge of TCK						
Capture setup time	tcsJ		5		ns	1
Hold time	1					
TMS hold time	tтнмх		5		ns	
TCK HIGH to TDI invalid	<b>t</b> THDX		5		ns	
Capture hold time	tснл		5		ns	1
Reset	1					
TRST LOW pulse width	<b>t</b> TRSTW		100		ns	

Note 1. tcsJ and tcHJ refer to the setup and hold time requirements of latching data from the boundary scan register.

# **JTAG Timing Diagram**

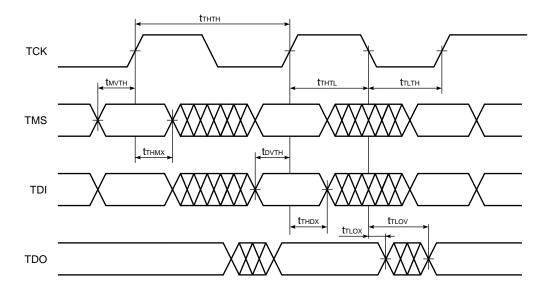


Table 4-6. Scan Register Definition (1)

Register name	Description
Instruction register	The 8 bit instruction registers hold the instructions that are executed by the TAP controller. The register can be loaded when it is placed between the TDI and TDO pins. The instruction register is automatically preloaded with the IDCODE instruction at power-up whenever the controller is placed in test-logic-reset state.
Bypass register	The bypass register is a single bit register that can be placed between TDI and TDO. It allows serial test data to be passed through the RAMs TAP to another device in the scan chain with as little delay as possible. The bypass register is set LOW (Vss) when the bypass instruction is executed.
ID register	The ID Register is a 32 bit register that is loaded with a device and vendor specific 32 bit code when the controller is put in capture-DR state with the IDCODE command loaded in the instruction register. The register is then placed between the TDI and TDO pins when the controller is moved into shift-DR state.
Boundary register	The boundary register, under the control of the TAP controller, is loaded with the contents of the RAMs I/O ring when the controller is in capture-DR state and then is placed between the TDI and TDO pins when the controller is moved to shift-DR state. Several TAP instructions can be used to activate the boundary register.  The Scan Exit Order tables describe which device bump connects to each boundary register location. The first column defines the bit's position in the boundary register. The second column is the name of the input or I/O at the bump and the third column is the bump number.

Table 4-7. Scan Register Definition (2)

Register name	Bit size	Unit
Instruction register	8	bit
Bypass register	1	bit
ID register	32	bit
Boundary register	113	bit

Table 4-8. ID Register Definition

Part number	Organization	ID [31:28] vendor revision no.	ID [27:12] part no.	ID [11:1] vendor ID no.	ID [0] fix bit
μPD48011318	64M x 18	0000	0000 0000 1101 1010	0000010000	1
μPD48011336	32M x 36	0000	0000 0000 1101 1011	00000010000	1

## 4.3 Boundary Scan

IEEE 1149.1 compliant boundary scan will be implemented on all component I/O pins. High speed I/Os that are differential or AC coupled will be fully compliant to the IEEE 1149.6 standard.

The actual boundary scan order is defined in an underscore (e.g. "CS#\_WE#").

#### Table 4-9 and Table 4-10.

Pins labeled DNU0 are DNU, Vss. Pins labeled DNU1 are DNU, VDDQ. Pins that are subject to the mirror function are named with the obverse name and reverse name, separated by an underscore (e.g. "CS#\_WE#").

Table 4-9 Boundary Scan Order, x18

Table 4-9 Boundary Scan Order, x18										
X18	Cell#	Cell	Pin	TYPE	Safe	Cntrl	Cntrl	Cntrl		
XIO	OCII II	OCII	Name		Bit	Cell	Value	State		
J3	0	BC_1	DK0	input	Х					
K3	1	BC_1	DK0#	input	Х					
L2	2	BC_1	CS#_WE#	input	Х					
L4	3	BC_1	A4_A3	input	Х					
M5	4	BC_1	A2_REF#	input	Х					
M3	5	BC_1	DPR#_A9	input	Χ					
	6	BC_2	*	control	0					
N2	7	BC_7	DINV2	bidir	Χ	6	0	Z		
N4	8	BC_1	A0_A1	input	Х					
	9	BC 2	*	control	0					
P5	10	BC 7	DQ9	bidir	Х	9	0	Z		
	11	BC 2	*	internal	0					
P3	12	BC_7	DNU0(1)	internal	Х					
	13	BC_2	*	internal	0					
P1	14	BC_7	DNU0(2)	internal	X					
	15	BC 2	*	internal	0					
R2	16	BC_2 BC 7	DNU0(3)	internal	X					
114	17	BC_7 BC 2	*	control	0					
DΛ						17	0	7		
R4	18	BC_7	DQ11 *	bidir	X	17	0	Z		
	19	BC_2		control	0	40		7		
T5	20	BC_7	DQ13 *	bidir	X	19	0	Z		
	21	BC_2		internal	0					
T3	22	BC_7	DNU0(4)	internal	X					
T1	23	BC_1	QK1#	output2	X					
	24	BC_2	*	internal	0					
U2	25	BC_7	DNU0(5)	internal	Х					
	26	BC_2	*	control	0					
U4	27	BC_7	DQ15	bidir	Х	26	0	Z		
V1	28	BC_1	QK1	output2	Χ					
	29	BC_2	*	control	0					
U12	30	BC_7	DQ17	bidir	Х	29	0	Z		
	31	BC_2	*	internal	0					
U10	32	BC_7	DNU0(6)	internal	Х					
	33	BC_2	*	internal	0					
Т9	34	BC_7	DNU0(7)	internal	Χ					
	35	BC_2	*	control	0					
T11	36	BC_7	DQ16	bidir	Х	35	0	Z		
	37	BC_2	*	control	0					
R12	38	BC_7	DQ14	bidir	Х	37	0	Z		
	39	BC_2	*	internal	0					
R10	40	BC_7	DNU0(8)	internal	Х					
	41	BC_2	*	internal	0					
P9	42	BC 7	DNU0(9)	internal	X					
	43	BC 2	*	control	0					
P11	44	BC 7	DQ10	bidir	X	43	0	Z		
<del>-                                    </del>	45	BC 2	*	control	0					
P13	46	BC_7	DQ12	bidir	X	45	0	Z		
	47	BC 2	*	internal	0			_		
N12	48	BC_2 BC 7	DNU0(10)	internal	X					
N10	49	BC_7	A1 A0	input	X					
M9	50	BC_1	REF#_A2	input	X					
M11	51	BC_1	A9 DPR#		X					
L12	52	BC_1	WE#_CS#	input	X					
	-		_	input						
L10	53	BC_1	A3_A4	input	X					
K9	54	BC_1	CK#	input	X					
K11	55	BC_1	DNU0(11)	internal	X					
J13	56	BC_1	LBK#	input	1					
J11	57	BC_1	DNU0(12)	internal	Х					
		1			1	1	1	1		

V40	Call #	Call	Pin	TVDE	Safe	Cntrl	Cntrl	Cntrl
X18	Cell #	Cell	Name	TYPE	Bit	Cell	Value	State
J9	58	BC_1	CK	input	X			
H10	59	BC_1	A5_A6	input	X			
H12	60	BC_1	DM_RST#	input	X			
G11	61	BC_1	A11	input	Х			
G9	62	BC_1	A7	input	Х			
F10	63	BC_1	A13-A12	input	Х			
	64	BC_2	*	internal	0			
F12	65	BC_7	DNU0(13)	internal	Х			
	66	BC_2	*	control	0			
E13	67	BC_7	DQ3	bidir	Х	66	0	Z
	68	BC_2	*	control	0			
E11	69	BC_7	DQ1	bidir	Х	68	0	Z
	70	BC_2	*	internal	0			
E9	71	BC_7	DNU0(14)	internal	Х			
	72	BC_2	*	internal	0			
D10	73	BC_7	DNU0(15)	internal	Х			
	74	BC_2	*	control	0			
D12	75	BC_7	DQ5	bidir	Х	74	0	Z
C13	76	BC_1	QK0#	output2	Х			
	77	BC_2	*	control	0			
C11	78	BC_7	DQ7	bidir	Χ	77	0	Z
	79	BC_2	*	internal	0			
C9	80	BC_7	DNU0(16)	internal	Χ			
	81	BC_2	*	internal	0			
B10	82	BC_7	DNU0(17)	internal	Χ			
	83	BC_2	*	control	0			
B12	84	BC_7	DQ8	bidir	Х	83	0	Z
A13	85	BC_1	QK0	output2	Х			
	86	BC_2	*	control	0			
B4	87	BC_7	DQ6	bidir	Х	86	0	Z
	88	BC_2	*	internal	0			
B2	89	BC_7	DNU0(18)	internal	Х			
C1	90	BC_1	QVLD	output2	Х			
	91	BC_2	*	internal	0			
C3	92	BC_7	DNU0(19)	internal	Х			
	93	BC_2	*	control	0			
C5	94	BC 7	DQ4	bidir	Х	93	0	Z
	95	BC_2	*	control	0			
D4	96	BC_7	DQ2	bidir	Х	95	0	Z
	97	BC_2	*	internal	0			
D2	98	BC_7	DNU0(20)	internal	Х			
	99	BC_2	*	internal	0			
E1	100	BC_7	DNU0(21)	internal	X			
	101	BC_2	*	internal	0			
E3	102	BC 7	DNU0(22)	internal	X			
	103	BC_2	*	control	0			
E5	104	BC_7	DQ0	bidir	X	103	0	Z
F4	105	BC 1	A12 A13	input	X		_	_
	106	BC_2	*	control	0			
F2	107	BC 7	DINV0	bidir	X	106	0	Z
G3	108	BC 1	A10	input	X	. 50		
G5	109	BC 1	A8	input	X			
H4	110	BC 1	A6 A5	input	X			
H2	111	BC_1	RST# DM	input	X			
J1	112	BC_1	MF	input	X			
JI	114	DO_1	IVII	πιραι	^			

Table 4-10 Boundary Scan Order, x36

Pin Safa Cottel Cottel Cottel											
x36	Cell #	Cell	Pin Name	TYPE	Safe Bit	Cntrl Cell	Cntrl Value	Cntrl State			
J3	0	BC_1	DK0	input	X	Cell	value	State			
K3	1	BC_1		input	X						
L2	2	BC_1	DK0# CS#_WE#	input	X						
L2 L4	3	_	_	input							
		BC_1	A4_A3	input	X						
M5	4	BC_1	A2_REF#	input	X						
M3	5	BC_1	DPR#_A9 *	input							
NO	6	BC_2		control	0		_	7			
N2	7	BC_7	DINV2	bidir	X	6	0	Z			
N4	8	BC_1	A0_A1 *	input	X						
	9	BC_2		control	0			-			
P5	10	BC_7	DQ18 *	bidir	X	9	0	Z			
	11	BC_2		control	0			_			
P3	12	BC_7	DQ34 *	bidir	X	11	0	Z			
	13	BC_2		control	0		_	_			
P1	14	BC_7	DQ35	bidir	Х	13	0	Z			
	15	BC_2	*	control	0		_	_			
R2	16	BC_7	DQ32	bidir	Х	15	0	Z			
	17	BC_2	*	control	0						
R4	18	BC_7	DQ20	bidir	Х	17	0	Z			
	19	BC_2	*	control	0						
T5	20	BC_7	DQ22	bidir	Х	19	0	Z			
	21	BC_2	*	control	0						
Т3	22	BC_7	DQ30	bidir	Х	21	0	Z			
T1	23	BC_1	QK1#	output2	Х						
	24	BC_2	*	control	0						
U2	25	BC_7	DQ28	bidir	Х	24	0	Z			
	26	BC_2	*	control	0						
U4	27	BC_7	DQ24	bidir	Х	26	0	Z			
V1	28	BC_1	QK1	output2	Х						
	29	BC_2	*	control	0						
U12	30	BC_7	DQ26	bidir	Х	29	0	Z			
	31	BC_2	*	control	0						
U10	32	BC 7	DQ27	bidir	Х	31	0	Z			
	33	BC 2	*	control	0						
Т9	34	BC_7	DQ29	bidir	X	33	0	Z			
	35	BC_2	*	control	0						
T11	36	BC_7	DQ25	bidir	X	35	0	Z			
	37	BC 2	*	control	0			_			
R12	38	BC 7	DQ23	bidir	X	37	0	Z			
	39	BC 2	*	control	0	<u> </u>		_			
R10	40	BC 7	DQ31	bidir	X	39	0	Z			
	41	BC_2	*	control	0			_			
P9	42	BC 7	DQ33	bidir	X	41	0	Z			
1.5	43	BC_7 BC_2	*	control	0	71	-				
P11	44	BC_2 BC 7	DQ19	bidir	X	43	0	Z			
' ' ' '	45	BC_7 BC 2	DQ19 *	control	0	70					
P13	46	BC_2 BC 7	DQ21	bidir	X	45	0	Z			
1 13	47	BC_7 BC 2	DQ21 *		0	70	J				
NIAO		BC_2 BC 7	DINV3	control bidir		47		7			
N12 N10	48 49	BC_7 BC 1	A1 A0		X	41	0	Z			
M9		BC_1	REF#_A2	input		1	1				
	50 51	_		input	X						
M11	51	BC_1	A9_DPR#	input	X						
L12	52	BC_1	WE#_CS#	input	X						
L10	53	BC_1	A3_A4	input	X						
K9	54	BC_1	CK#	input	X						
K11	55	BC_1	DK1#	input	X						
J13	56	BC_1	LBK#	input	1						
J11	57	BC_1	DK1	input	X						

			Pin		Safe	Cntrl	Cntrl	Cntrl
x36	Cell #	Cell	Name	TYPE	Bit	Cell	Value	State
J9	58	BC_1	CK	input	Х			
H10	59	BC_1	A5_A6	input	Х			
H12	60	BC_1	DM_RST#	input	Х			
G11	61	BC_1	A11	input	Х			
G9	62	BC_1	A7	input	Χ			
F10	63	BC_1	DNU1(1)	internal	Χ			
	64	BC_2	*	control	0			
F12	65	BC_7	DINV1	bidir	Х	64	0	Z
	66	BC_2	*	control	0			
E13	67	BC_7	DQ3	bidir	Х	66	0	Z
	68	BC_2	*	control	0			
E11	69	BC_7	DQ1	bidir	Χ	68	0	Z
	70	BC_2	*	control	0			
E9	71	BC_7	DQ17	bidir	Х	70	0	Z
	72	BC_2	*	control	0			
D10	73	BC_7	DQ15	bidir	Х	72	0	Z
	74	BC_2	*	control	0			
D12	75	BC_7	DQ5	bidir	Х	74	0	Z
C13	76	BC_1	QK0#	output2	Х			
	77	BC_2	*	control	0			
C11	78	BC_7	DQ7	bidir	Х	77	0	Z
	79	BC_2	*	control	0			
C9	80	BC_7	DQ13	bidir	Х	79	0	Z
	81	BC_2	*	control	0			
B10	82	BC_7	DQ11	bidir	Х	81	0	Z
	83	BC_2	*	control	0			
B12	84	BC_7	DQ8	bidir	Х	83	0	Z
A13	85	BC_1	QK0	output2	Х			
	86	BC_2	*	control	0			
B4	87	BC_7	DQ6	bidir	Х	86	0	Z
	88	BC_2	*	control	0			
B2	89	BC_7	DQ9	bidir	Х	88	0	Z
C1	90	BC_1	QVLD	output2	Х			
	91	BC_2	*	control	0		_	_
C3	92	BC_7	DQ10	bidir	Х	91	0	Z
	93	BC_2	*	control	0			
C5	94	BC_7	DQ4	bidir	X	93	0	Z
	95	BC_2	*	control	0		_	_
D4	96	BC_7	DQ2 *	bidir	X	95	0	Z
	97	BC_2		control	0	67		_
D2	98	BC_7	DQ12 *	bidir	X	97	0	Z
F4	99	BC_2		control	0	00		7
E1	100	BC_7	DQ16 *	bidir	X	99	0	Z
	101	BC_2		control	0	101		7
E3	102	BC_7	DQ14 *	bidir	X	101	0	Z
	103	BC_2		control	0	400		7
E5	104	BC_7	DQ0	bidir	X	103	0	Z
F4	105	BC_1	DNU1(2)	internal	X			
F0	106	BC_2		control	0	100	_	7
F2	107	BC_7	DINV0	bidir	X	106	0	Z
G3	108	BC_1	A10	input	X			
G5	109	BC_1	A8 A6_A5	input	X			1
H4 ⊔2	110	BC_1		input	X			
H2	111	BC_1	RST#_DM ME	input	X			
J1	112	BC_1	MF	input	Χ	j	<u> </u>	

#### 4.4 JTAG Instructions

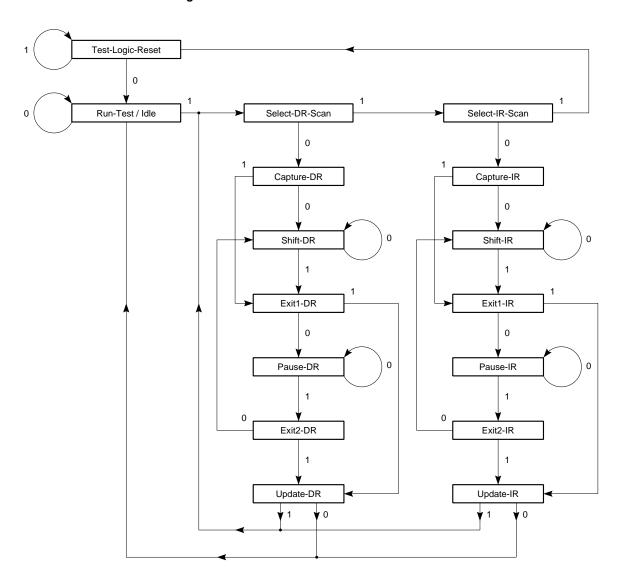
Many different instructions (2<sup>8</sup>) are possible with the 8-bit instruction register. All used combinations are listed in **Table 4-11**, Instruction Codes. These six instructions are described in detail below. The remaining instructions are reserved and should not be used.

The TAP controller used in this RAM is fully compliant to the 1149.1 convention. Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO balls. To execute the instruction once it is shifted in, the TAP controller needs to be moved into the Update-IR state.

**Table 4-11** 

Instructions	Instruction Code [7:0]	Description	
EXTEST	0000 0000	The EXTEST instruction allows circuitry external to the component package to be tested. Boundary-scan register cells at output pins are used to apply test vectors, while those at input pins capture test results. Typically, the first test vector to be applied using the EXTEST instruction will be shifted into the boundary scan register using the PRELOAD instruction. Thus, during the update-IR state of EXTEST, the output drive is turned on and the PRELOAD data is driven onto the output pins.	
IDCODE	0010 0001	The IDCODE instruction causes the ID ROM to be loaded into the ID register when the controller is in capture-DR mode and places the ID register between the TDI and TDO pins in shift-DR mode. The IDCODE instruction is the default instruction loaded in at power up and any time the controller is placed in the test-logic-reset state.	
SAMPLE / PRELOAD	0000 0101	SAMPLE / PRELOAD is a Standard 1149.1 mandatory public instruction. When the SAMPLE / PRELOAD instruction is loaded in the instruction register, moving the TAP controller into the capture-DR state loads the data in the RAMs input and Q pins into the boundary scan register. Because the RAM clock(s) are independent from the TAP clock (TCK) it is possible for the TAP to attempt to capture the I/O ring contents while the input buffers are in transition (i.e., in a metastable state). Although allowing the TAP to sample metastable input will not harm the device, repeatable results cannot be expected. RAM input signals must be stabilized for long enough to meet the TAPs input data capture setup plus hold time (tcs plus tch). The RAMs clock inputs need not be paused for any other TAP operation except capturing the I/O ring contents into the boundary scan register. Moving the controller to shift-DR state then places the boundary scan register between the TDI and TDO pins.	
CLAMP	0000 0111	When the CLAMP instruction is loaded into the instruction register, the data driven by the output balls are determined from the values held in the boundary scan register. Selects the bypass register to be connected between TDI and TDO. Data driven by output balls are determined from values held in the boundary scan register.	
High-Z	0000 0011	The High-z instruction causes the boundary scan register to be connected between the TDI and TDO. This places all RAMs outputs into a High-Z state.  Selects the bypass register to be connected between TDI and TDO. All outputs are forced into high impedance state.	
BYPASS	1111 1111	When the BYPASS instruction is loaded in the instruction register, the bypass register is placed between TDI and TDO. This occurs when the TAP controller is moved to the shift-DR state. This allows the board level scan path to be shortened to facilitate testing of other devices in the scan path.	
Reserved for Future Use	-	The remaining instructions are not implemented but are reserved for future use. Do not use these instructions.	

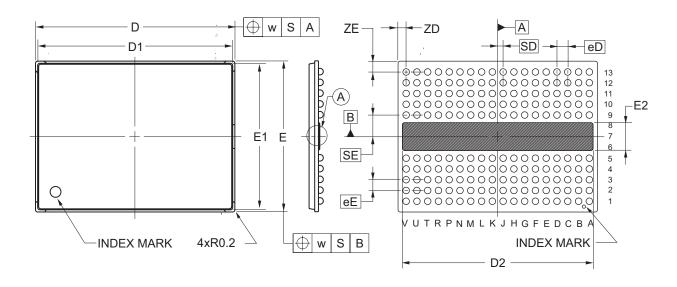
# 4.5 TAP Controller State Diagram

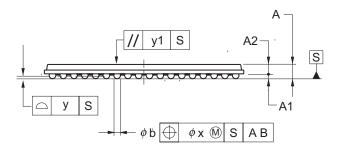


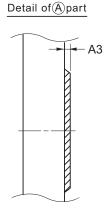
#### 5. Package Drawing

# 180-PIN TAPE FBGA (BGA) (18.5 x 14)

JEITA Package Code	RENESAS Code	Previous Code	MASS (TYP.) [g]
T-LBGA180-14x18.5-1.00	TLBG0180FA-A	T180FF-100-FF1	0.60







	(UNIT:mm)
ITEM	DIMENSIONS
D	18.50 ± 0.10
D1	18.10
D2	17.70
E	14.00 ± 0.10
E1	13.60
E2	2.59
W	0.20
eD	1.00
eЕ	1.00
SD	0.50
SE	2.00
Α	1.274 ± 0.10
A1	$0.47 \pm 0.05$
A2	0.804
А3	0.135
b	0.61 ± 0.05
х	0.15
У	0.10
y1	0.20
ZD	0.75
ZE	1.00
	T180FF-100-FF1

T180FF-100-FF1

**Revision History** 

# $\mu$ PD48011318, $\mu$ PD48011336 Series

		Description		
Rev.	Date	Page	Summary	
Rev.0.01	'10.08.18	-	New Preliminary Data Sheet	
Rev.1.00	'12.10.10	- P11 P16 P27 P48	New Data Sheet Update "DC Characteristics" Update "Thermal Impedance" Modified comments for "Power on" Update "Package Drawing"	
Rev.2.00	'13.02.01	P7	Update "Block Diagram"	

#### Notice

- 1. Descriptions of circuits, software and other related information in this document are provided only to illustrate the operation of semiconductor products and application examples. You are fully responsible for the incorporation of these circuits, software, and information in the design of your equipment. Renesas Electronics assumes no responsibility for any losses incurred by you or third parties arising from the use of these circuits, software, or information,
- 2. Renesas Electronics has used reasonable care in preparing the information included in this document, but Renesas Electronics does not warrant that such information is error free. Renesas Electronics assumes no liability whatsoever for any damages incurred by you resulting from errors in or omissions from the information included herein.
- 3. Renesas Electronics does not assume any liability for infringement of patents, copyrights, or other intellectual property rights of third parties by or arising from the use of Renesas Electronics products or technical information described in this document. No license, express, implied or otherwise, is granted hereby under any patents, copyrights or other intellectual property rights of Renesas Electronics or
- 4. You should not alter, modify, copy, or otherwise misappropriate any Renesas Electronics product, whether in whole or in part. Renesas Electronics assumes no responsibility for any losses incurred by you or third parties arising from such alteration, modification, copy or otherwise misappropriation of Renesas Electronics product.
- 5. Renesas Electronics products are classified according to the following two quality grades: "Standard" and "High Quality". The recommended applications for each Renesas Electronics product depends on the product's quality grade, as indicated below

"Standard": Computers; office equipment; communications equipment; test and measurement equipment; audio and visual equipment; home electronic appliances; machine tools; personal electronic equipment; and industrial robots etc.

"High Quality": Transportation equipment (automobiles, trains, ships, etc.); traffic control systems; anti-disaster systems; anti-crime systems; and safety equipment etc.

Renesas Electronics products are neither intended nor authorized for use in products or systems that may pose a direct threat to human life or bodily injury (artificial life support devices or systems, surgical implantations etc.), or may cause serious property damages (nuclear reactor control systems, military equipment etc.). You must check the quality grade of each Renesas Electronics product before using it in a particular application. You may not use any Renesas Electronics product for any application for which it is not intended. Renesas Electronics shall not be in any way liable for any damages or losses incurred by you or third parties arising from the use of any Renesas Electronics product for which the product is not intended by Renesas Electronics

- 6. You should use the Renesas Electronics products described in this document within the range specified by Renesas Electronics, especially with respect to the maximum rating, operating supply voltage range, movement power voltage range, heat radiation characteristics, installation and other product characteristics. Renesas Electronics shall have no liability for malfunctions or damages arising out of the use of Renesas Electronics products beyond such specified ranges.
- 7. Although Renesas Electronics endeavors to improve the quality and reliability of its products, semiconductor products have specific characteristics such as the occurrence of failure at a certain rate and malfunctions under certain use conditions. Further, Renesas Electronics products are not subject to radiation resistance design. Please be sure to implement safety measures to guard them against the possibility of physical injury, and injury or damage caused by fire in the event of the failure of a Renesas Electronics product, such as safety design for hardware and software including but not limited to redundancy, fire control and malfunction prevention, appropriate treatment for aging degradation or any other appropriate measures. Because the evaluation of microcomputer software alone is very difficult, please evaluate the safety of the final products or systems manufactured by you.
- 8. Please contact a Renesas Electronics sales office for details as to environmental matters such as the environmental compatibility of each Renesas Electronics product. Please use Renesas Electronics products in compliance with all applicable laws and regulations that regulate the inclusion or use of controlled substances, including without limitation, the EU RoHS Directive. Renesas Electronics assumes no liability for damages or losses occurring as a result of your noncompliance with applicable laws and regulations.
- 9. Renesas Electronics products and technology may not be used for or incorporated into any products or systems whose manufacture, use, or sale is prohibited under any applicable domestic or foreign laws or regulations. You should not use Renesas Electronics products or technology described in this document for any purpose relating to military applications or use by the military, including but not limited to the development of weapons of mass destruction. When exporting the Renesas Electronics products or technology described in this document, you should comply with the applicable export control laws and regulations and follow the procedures required by such laws and regulations.
- 10. It is the responsibility of the buyer or distributor of Renesas Electronics products, who distributes, disposes of, or otherwise places the product with a third party, to notify such third party in advance of the contents and conditions set forth in this document. Renesas Electronics assumes no responsibility for any losses incurred by you or third parties as a result of unauthorized use of Renesas Electronics products.
- 11. This document may not be reproduced or duplicated in any form, in whole or in part, without prior written consent of Renesas Electronics
- 12. Please contact a Renesas Electronics sales office if you have any questions regarding the information contained in this document or Renesas Electronics products, or if you have any other inquiries, (Note 1) "Renesas Electronics" as used in this document means Renesas Electronics Corporation and also includes its majority-owned subsidiaries.
- (Note 2) "Renesas Electronics product(s)" means any product developed or manufactured by or for Renesas Electronics



#### **SALES OFFICES**

Renesas Electronics Corporation

http://www.renesas.com

Refer to "http://www.renesas.com/" for the latest and detailed information

Renesas Electronics America Inc. 2880 Scott Boulevard Santa Clara, CA 95050-2554, U.S.A. Tel: +1-408-588-6000, Fax: +1-408-588-6130

Renesas Electronics Canada Limited 1101 Nicholson Road, Newmarket, Ontario L3Y 9C3, Canada Tel: +1-905-898-5441, Fax: +1-905-898-3220

Renesas Electronics Europe Limited
Dukes Meadow, Milliboard Road, Bourne End, Buckinghamshire, SL8 5FH, U.K
Tel: +44-1628-651-700, Fax: +44-1628-651-804

Renesas Electronics Europe GmbH

Arcadiastrasse 10, 40472 Düsseldorf, Germany Tel: +49-211-65030, Fax: +49-211-6503-1327

Renesas Electronics (China) Co., Ltd. 7th Floor, Quantum Plaza, No.27 ZhiChunLu Ha Tel: +86-10-8235-1155, Fax: +86-10-8235-7679 i. nunLu Haidian District. Beiiing 100083. P.R.China

Renesas Electronics (Shanghai) Co., Ltd.
Unit 204, 205, AZIA Center, No.1233 Lujiazui Ring Rd., Pudong District, Shanghai 200120, China Tel: +86-21-5877-1818, Fax: +86-21-6887-7858 / -7898

Renesas Electronics Hong Kong Limited
Unit 1601-1613, 16/F., Tower 2, Grand Century Place, 193 Prince Edward Road West, Mongkok, Kowloon, Hong Kong
Tel: +852-2868-9318, Fax: +852 2869-9022/9044

Renesas Electronics Taiwan Co., Ltd. 13F, No. 363, Fu Shing North Road, Taipei, Taiwan Tel: +886-2-8175-9600, Fax: +886 2-8175-9670

Renesas Electronics Singapore Pte. Ltd. 80 Bendemeer Road, Unit #06-02 Hyflux Innovation Centre Singapore 339949 Tel: +65-6213-0200, Fax: +65-6213-0300

Renesas Electronics Malaysia Sdn.Bhd. Unit 906, Block B, Menara Amcorp, Amcorp Trade Centre, No. 18, Jln Persiaran Barat, 46050 Petaling Jaya, Selangor Darul Ehsan, Malaysia Tel: +60-3-7955-9390, Fax: +60-3-7955-9510

Renesas Electronics Korea Co., Ltd. 11F., Samik Lavied' or Bldg., 720-2 Yeoksam-Dong, Kangnam-Ku, Seoul 135-080, Korea Tel: 482-2558-3737, Fax: 482-2558-5141

© 2013 Renesas Electronics Corporation. All rights reserved.