

GENERAL DESCRIPTION

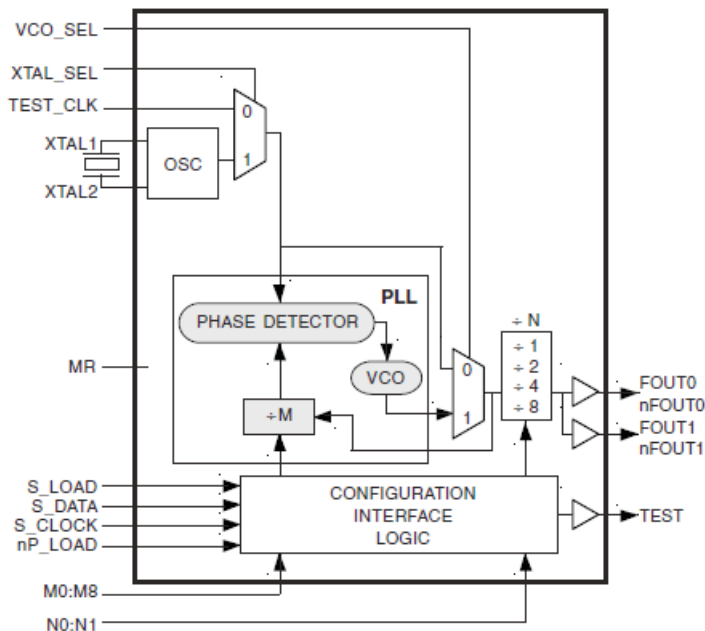
The 84320I-01 is a general purpose, dual output Crystal-to-3.3V Differential LVPECL High Frequency Synthesizer. The 84320I-01 has a selectable TEST_CLK or crystal inputs. The VCO operates at a frequency range of 620MHz to 780MHz. The VCO frequency is programmed in steps equal to the value of the input reference or crystal frequency. The VCO and output frequency can be programmed using the serial or parallel interfaces to the configuration logic. The low phase noise characteristics of the 84320I-01 make it an ideal clock source for 10 Gigabit Ethernet, SONET, and Serial Attached SCSI applications.

FEATURES

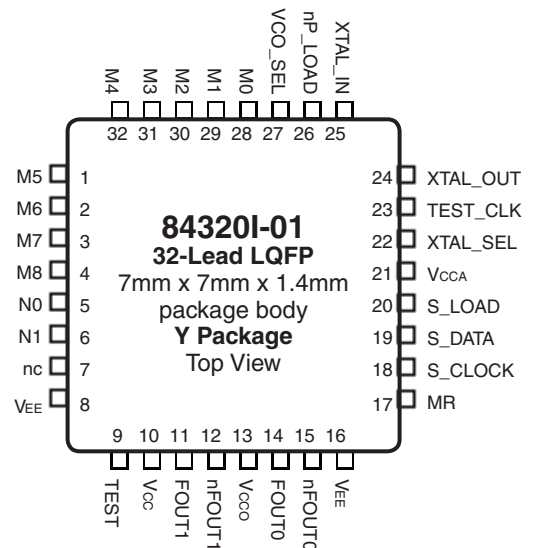
- Dual differential 3.3V LVPECL outputs
- Selectable crystal oscillator interface or LVCMOS/LVTTL TEST_CLK
- Output frequency range: 77.5MHz to 780MHz
- Crystal input frequency range: 14MHz to 40MHz
- VCO range: 620MHz to 780MHz
- Parallel or serial interface for programming counter and output dividers
- Duty cycle: 44% - 56% (N > 1)
- RMS period jitter: 2.0ps (typical)
- RMS phase jitter at 155.52MHz, using a 38.88MHz crystal (12kHz to 20MHz): 2.38ps (typical)
- RMS phase noise at 155.52MHz (typical)

| Offset | Noise Power |
|--------|---------------|
| 100Hz | -90.5 dBc/Hz |
| 1KHz | -114.2 dBc/Hz |
| 10KHz | -123.6 dBc/Hz |
| 100KHz | -128.1 dBc/Hz |
- 3.3V supply voltage
- -40°C to 85°C ambient operating temperature
- Available in lead-free (RoHS 6) package

BLOCK DIAGRAM



PIN ASSIGNMENT



FUNCTIONAL DESCRIPTION

NOTE: The functional description that follows describes operation using a 25MHz crystal. Valid PLL loop divider values for different crystal or input frequencies are defined in the Input Frequency Characteristics, Table 5, NOTE 1.

The 84320I-01 features a fully integrated PLL and therefore requires no external components for setting the loop bandwidth. A fundamental crystal is used as the input to the on-chip oscillator. The output of the oscillator is fed into the phase detector. A 25MHz crystal provides a 25MHz phase detector reference frequency. The VCO of the PLL operates over a range of 620MHz to 780MHz. The output of the M divider is also applied to the phase detector.

The phase detector and the M divider force the VCO output frequency to be M times the reference frequency by adjusting the VCO control voltage. Note that for some values of M (either too high or too low), the PLL will not achieve lock. The output of the VCO is scaled by a divider prior to being sent to each of the LVPECL output buffers. The divider provides a 50% output duty cycle.

The programmable features of the 84320I-01 support two input modes to program the M divider and N output divider. The two input operational modes are parallel and serial. Figure 1 shows the timing diagram for each mode. In parallel mode, the nP_LOAD input is initially LOW. The data on inputs M0 through M8 and N0 and N1 is passed directly to the M divider and N output divider. On the LOW-to-HIGH transition of the nP_LOAD input, the data is latched and the M divider remains loaded until the next LOW transition on nP_LOAD or until a serial event occurs. As a result, the M and N bits can be hardwired to set the M divider

and N output divider to a specific default state that will automatically occur during power-up. The TEST output is LOW when operating in the parallel input mode. The relationship between the VCO frequency, the crystal frequency and the M divider is defined as follows:

$$f_{VCO} = f_{xtal} \times M$$

The M value and the required values of M0 through M8 are shown in Table 3B to program the VCO Frequency Function Table. Valid M values for which the PLL will achieve lock for a 25MHz reference are defined as $25 \leq M \leq 31$. The frequency output is defined as follows:

$$F_{OUT} = \frac{f_{VCO}}{N} = \frac{f_{xtal} \times M}{N}$$

Serial operation occurs when nP_LOAD is HIGH and S_LOAD is LOW. The shift register is loaded by sampling the S_DATA bits with the rising edge of S_CLOCK. The contents of the shift register are loaded into the M divider and N output divider when S_LOAD transitions from LOW-to-HIGH. The M divide and N output divide values are latched on the HIGH-to-LOW transition of S_LOAD. If S_LOAD is held HIGH, data at the S_DATA input is passed directly to the M divider and N output divider on each rising edge of S_CLOCK. The serial mode can be used to program the M and N bits and test bits T1 and T0. The internal registers T0 and T1 determine the state of the TEST output as follows:

| T1 | T0 | TEST Output |
|----|----|------------------------------|
| 0 | 0 | LOW |
| 0 | 1 | S_Data, Shift Register Input |
| 1 | 0 | Output of M divider |
| 1 | 1 | CMOS Fout |

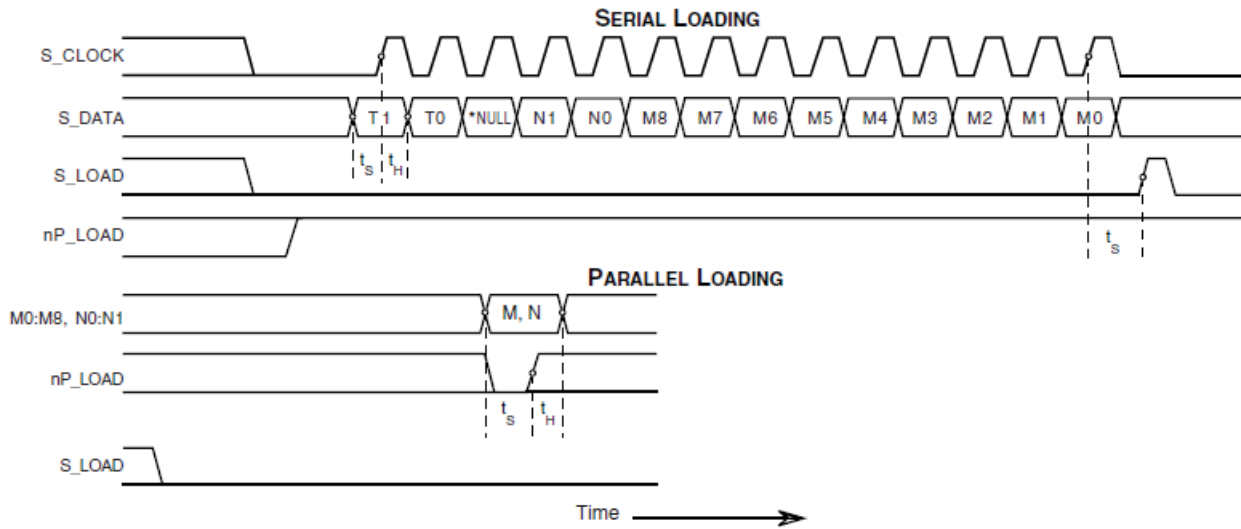


FIGURE 1. PARALLEL & SERIAL LOAD OPERATIONS

*NOTE: The NULL timing slot must be observed.

TABLE 1. PIN DESCRIPTIONS

| Number | Name | Type | | Description |
|-----------------------------|--------------------------------|--------|----------|--|
| 1 | M5 | Input | Pullup | M divider inputs. Data latched on LOW-to-HIGH transition of nP_LOAD input. LVCMOS / LVTTTL interface levels. |
| 2, 3, 4, 28, 29, 30, 31, 32 | M6, M7, M8, M0, M1, M2, M3, M4 | Input | Pulldown | |
| 5, 6 | N0, N1 | Input | Pulldown | Determines output divider value as defined in Table 3C, Function Table. LVCMOS / LVTTTL interface levels. |
| 7 | nc | Unused | | No connect. |
| 8, 16 | V _{EE} | Power | | Negative supply pins. |
| 9 | TEST | Output | | Test output which is ACTIVE in the serial mode of operation. Output driven LOW in parallel mode. LVCMOS/LVTTTL interface levels. |
| 10 | V _{CC} | Power | | Core supply pin. |
| 11, 12 | FOUT1, nFOUT1 | Output | | Differential output for the synthesizer. LVPECL interface levels. |
| 13 | V _{CCO} | Power | | Output supply pin. |
| 14, 15 | FOUT0, nFOUT0 | Output | | Differential output for the synthesizer. LVPECL interface levels. |
| 17 | MR | Input | Pulldown | Active High Master Reset. When logic HIGH, forces the internal dividers are reset causing the true outputs FOUTx to go low and the inverted outputs nFOUTx to go high. When logic LOW, the internal dividers and the outputs are enabled. Assertion of MR does not affect loaded M, N, and T values. LVCMOS / LVTTTL interface levels. |
| 18 | S_CLOCK | Input | Pulldown | Clocks in serial data present at S_DATA input into the shift register on the rising edge of S_CLOCK. LVCMOS/LVTTTL interface levels. |
| 19 | S_DATA | Input | Pulldown | Shift register serial input. Data sampled on the rising edge of S_CLOCK. LVCMOS/LVTTTL interface levels. |
| 20 | S_LOAD | Input | Pulldown | Controls transition of data from shift register into the dividers. LVCMOS / LVTTTL interface levels. |
| 21 | V _{CCA} | Power | | Analog supply pin. |
| 22 | XTAL_SEL | Input | Pullup | Selects between crystal or test inputs as the PLL reference source. Selects XTAL inputs when HIGH. Selects TEST_CLK when LOW. LVCMOS / LVTTTL interface levels. |
| 23 | TEST_CLK | Input | Pulldown | Test clock input. LVCMOS / LVTTTL interface levels. |
| 24, 25 | XTAL_OUT, XTAL_IN | Input | | Crystal oscillator interface. XTAL_IN is the input. XTAL_OUT is the output. |
| 26 | nP_LOAD | Input | Pulldown | Parallel load input. Determines when data present at M8:M0 is loaded into M divider, and when data present at N1:N0 sets the N output divider value. LVCMOS / LVTTTL interface levels. |
| 27 | VCO_SEL | Input | Pullup | Determines whether synthesizer is in PLL or bypass mode. LVCMOS / LVTTTL interface levels. |

NOTE: Pullup and Pulldown refer to internal input resistors. See Table 2, Pin Characteristics, for typical values.

TABLE 2. PIN CHARACTERISTICS

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-----------------------|-------------------------|-----------------|---------|---------|---------|-------|
| C _{IN} | Input Capacitance | | | 4 | | pF |
| R _{PULLUP} | Input Pullup Resistor | | | 51 | | kΩ |
| R _{PULLDOWN} | Input Pulldown Resistor | | | 51 | | kΩ |

TABLE 3A. PARALLEL AND SERIAL MODE FUNCTION TABLE

| Inputs | | | | | | | Conditions |
|--------|---------|------|------|--------|---------|--------|---|
| MR | nP_LOAD | M | N | S_LOAD | S_CLOCK | S_DATA | |
| H | X | X | X | X | X | X | Reset. Forces outputs LOW. |
| L | L | Data | Data | X | X | X | Data on M and N inputs passed directly to the M divider and N output divider. TEST output forced LOW. |
| L | ↑ | Data | Data | L | X | X | Data is latched into input registers and remains loaded until next LOW transition or until a serial event occurs. |
| L | H | X | X | L | ↑ | Data | Serial input mode. Shift register is loaded with data on S_DATA on each rising edge of S_CLOCK. |
| L | H | X | X | ↑ | L | Data | Contents of the shift register are passed to the M divider and N output divider. |
| L | H | X | X | ↓ | L | Data | M divider and N output divider values are latched. |
| L | H | X | X | L | X | X | Parallel or serial input do not affect shift registers. |
| L | H | X | X | H | ↑ | Data | S_DATA passed directly to M divider as it is clocked. |

NOTE: L = LOW
H = HIGH
X = Don't care
↑ = Rising edge transition
↓ = Falling edge transition

TABLE 3B. PROGRAMMABLE VCO FREQUENCY FUNCTION TABLE

| VCO Frequency (MHz) | M Divide | 256 | 128 | 64 | 32 | 16 | 8 | 4 | 2 | 1 |
|---------------------|----------|-----|-----|----|----|----|----|----|----|----|
| | | M8 | M7 | M6 | M5 | M4 | M3 | M2 | M1 | M0 |
| 625 | 25 | 0 | 0 | 0 | 0 | 1 | 1 | 0 | 0 | 1 |
| • | • | • | • | • | • | • | • | • | • | • |
| 700 | 28 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 0 | 0 |
| • | • | • | • | • | • | • | • | • | • | • |
| 775 | 31 | 0 | 0 | 0 | 0 | 1 | 1 | 1 | 1 | 1 |

NOTE 1: These M divide values and the resulting frequencies correspond to crystal or TEST_CLK input frequency of 25MHz.

TABLE 3C. PROGRAMMABLE OUTPUT DIVIDER FUNCTION TABLE

| Inputs | | N Divider Value | Output Frequency (MHz) | |
|--------|----|-----------------|------------------------|---------|
| N1 | N0 | | Minimum | Maximum |
| 0 | 0 | 1 | 620 | 780 |
| 0 | 1 | 2 | 310 | 390 |
| 1 | 0 | 4 | 155 | 195 |
| 1 | 1 | 8 | 77.5 | 97.5 |

ABSOLUTE MAXIMUM RATINGS

| | |
|--|---------------------------|
| Supply Voltage, V_{CC} | 4.6V |
| Inputs, V_I | -0.5V to $V_{CC} + 0.5V$ |
| Outputs, V_O (LVCMOS) | -0.5V to $V_{CCO} + 0.5V$ |
| Outputs, I_O (LVPECL) | |
| Continuous Current | 50mA |
| Surge Current | 100mA |
| Package Thermal Impedance, θ_{JA} | 47.9°C/W (0 lfpm) |
| Storage Temperature, T_{STG} | -65°C to 150°C |

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 4A. POWER SUPPLY DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = -40^\circ\text{C}$ TO 85°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-----------|-----------------------|-----------------|---------|---------|---------|-------|
| V_{CC} | Core Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| V_{CCA} | Analog Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| V_{CCO} | Output Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| I_{EE} | Power Supply Current | | | | 155 | mA |
| I_{CCA} | Analog Supply Current | | | | 22 | mA |

TABLE 4B. LVCMOS / LVTTTL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = -40^\circ\text{C}$ TO 85°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|----------|---------------------|---|--------------------------------------|---------|----------------|---------------|
| V_{IH} | Input High Voltage | VCO_SEL, XTAL_SEL, MR, S_LOAD, nP_LOAD, N0:N1, S_DATA, S_CLOCK, M0:M8 | 2 | | $V_{CC} + 0.3$ | V |
| | | TEST_CLK | 2 | | $V_{CC} + 0.3$ | V |
| V_{IL} | Input Low Voltage | VCO_SEL, XTAL_SEL, MR, S_LOAD, nP_LOAD, N0:N1, S_DATA, S_CLOCK, M0:M8 | -0.3 | | 0.8 | V |
| | | TEST_CLK | -0.3 | | 1.3 | V |
| I_{IH} | Input High Current | M0-M4, M6-M8, N0, N1, MR, S_CLOCK, TEST_CLK, S_DATA, S_LOAD, nP_LOAD | $V_{CC} = V_{IN} = 3.465V$ | | 150 | μA |
| | | M5, XTAL_SEL, VCO_SEL | $V_{CC} = V_{IN} = 3.465V$ | | 5 | μA |
| I_{IL} | Input Low Current | M0-M4, M6-M8, N0, N1, MR, S_CLOCK, TEST_CLK, S_DATA, S_LOAD, nP_LOAD | $V_{CC} = 3.465V$, $V_{IN} = 0V$ | -5 | | μA |
| | | M5, XTAL_SEL, VCO_SEL | $V_{CC} = 3.465V$, $V_{IN} = 0V$ | -150 | | μA |
| V_{OH} | Output High Voltage | TEST; NOTE 1 | 2.6 | | | V |
| V_{OL} | Output Low Voltage | TEST; NOTE 1 | | | 0.5 | V |

NOTE 1: Outputs terminated with 50Ω to $V_{CCO}/2$.

TABLE 4C. LVPECL DC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = -40^\circ\text{C}$ TO 85°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-------------|-----------------------------------|-----------------|-----------------|---------|-----------------|-------|
| V_{OH} | Output High Voltage; NOTE 1 | | $V_{CCO} - 1.4$ | | $V_{CCO} - 0.9$ | V |
| V_{OL} | Output Low Voltage; NOTE 1 | | $V_{CCO} - 2.0$ | | $V_{CCO} - 1.7$ | V |
| V_{SWING} | Peak-to-Peak Output Voltage Swing | | 0.6 | | 1.0 | V |

NOTE 1: Outputs terminated with $50\ \Omega$ to $V_{CCO} - 2V$. See “Parameter Measurement Information” section, “3.3V Output Load Test Circuit”.

TABLE 5. INPUT FREQUENCY CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = -40^\circ\text{C}$ TO 85°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|----------|-----------------|---------------------------|---------|---------|---------|-------|
| f_{IN} | Input Frequency | TEST_CLK; NOTE 1 | 14 | | 40 | MHz |
| | | XTAL_IN, XTAL_OUT; NOTE 1 | 14 | | 40 | MHz |
| | | S_CLOCK | | | 50 | MHz |

NOTE 1: For the input crystal and TEST_CLK frequency range, the M value must be set for the VCO to operate within the 620MHz to 780MHz range. Using the minimum input frequency of 14MHz, valid values of M are $45 \leq M \leq 55$. Using the maximum frequency of 40MHz, valid values of M are $16 \leq M \leq 19$.

TABLE 6. CRYSTAL CHARACTERISTICS

| Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|------------------------------------|-----------------|-------------|---------|---------|----------|
| Mode of Oscillation | | Fundamental | | | |
| Frequency | | 14 | | 40 | MHz |
| Equivalent Series Resistance (ESR) | | | | 50 | Ω |
| Shunt Capacitance | | | | 7 | pF |
| Drive Level | | | | 1 | mW |

TABLE 7. AC CHARACTERISTICS, $V_{CC} = V_{CCA} = V_{CCO} = 3.3V \pm 5\%$, $T_A = -40^\circ\text{C}$ TO 85°C

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|----------------|----------------------------|---------------------------|----------------------|---------|----------------------|-------|
| F_{OUT} | Output Frequency | | 77.5 | | 780 | MHz |
| $f_{jit(per)}$ | Period Jitter, RMS; NOTE 1 | $f_{OUT} > 100\text{MHz}$ | | 2.0 | 2.6 | ps |
| $t_{sk(o)}$ | Output Skew; NOTE 2, 3 | | | | 15 | ps |
| t_R / t_F | Output Rise/Fall Time | 20% to 80% | 100 | | 700 | ps |
| t_S | Setup Time | M, N to nP_LOAD | 5 | | | ns |
| | | S_DATA to S_CLOCK | 5 | | | ns |
| | | S_CLOCK to S_LOAD | 5 | | | ns |
| t_H | Hold Time | M, N to nP_LOAD | 5 | | | ns |
| | | S_DATA to S_CLOCK | 5 | | | ns |
| | | S_CLOCK to S_LOAD | 5 | | | ns |
| odc | Output Duty Cycle | $N > 1$ | 49 | | 51 | % |
| | | $f_{OUT} \leq 625$ | 44 | | 56 | % |
| t_{PW} | Output Pulse Width | $f > 625$ | $t_{PERIOD}/2 - 150$ | | $t_{PERIOD}/2 + 150$ | ps |
| t_{LOCK} | PLL Lock Time | | | | 1 | ms |

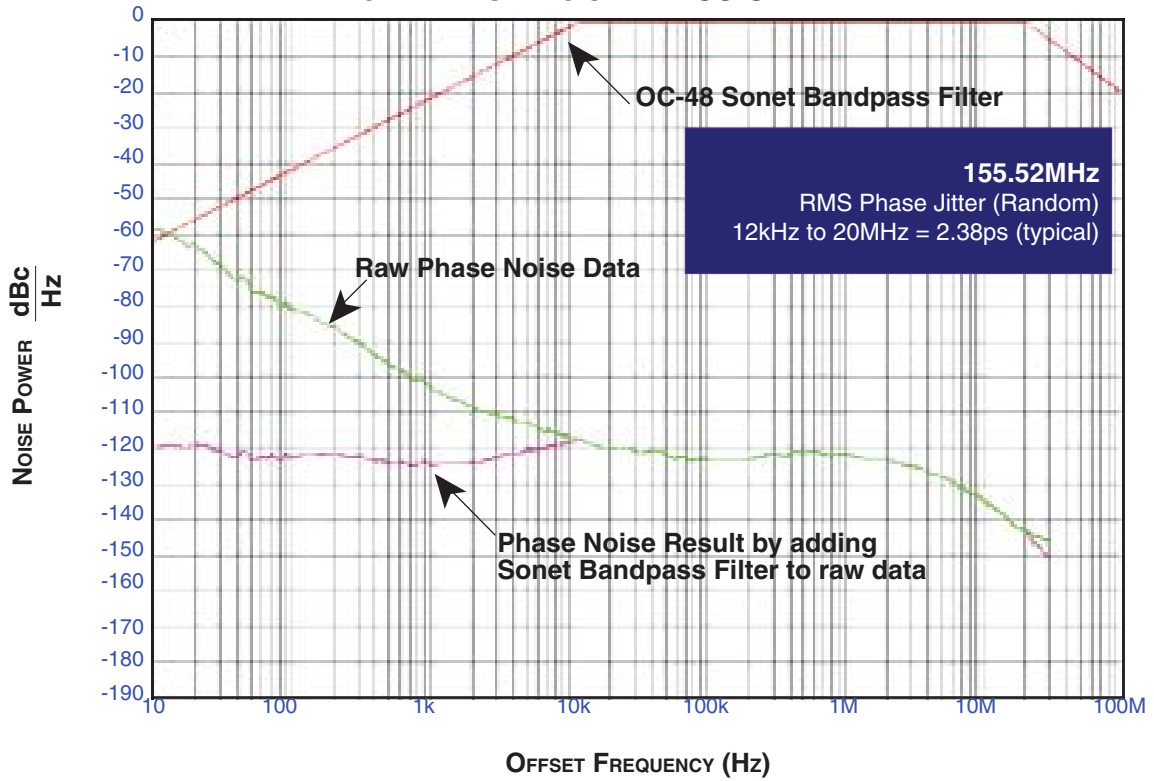
See Parameter Measurement Information section.

NOTE 1: Jitter performance using XTAL inputs.

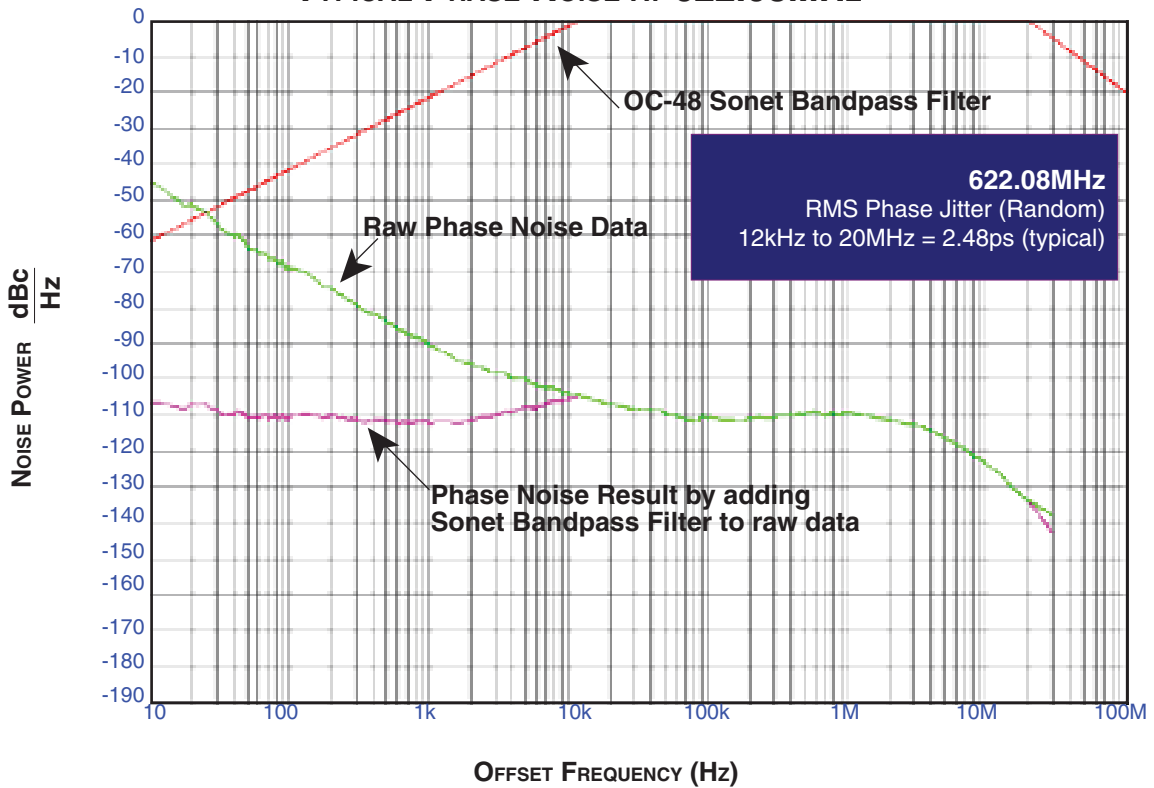
NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

NOTE 3: This parameter is defined in accordance with JEDEC Standard 65.

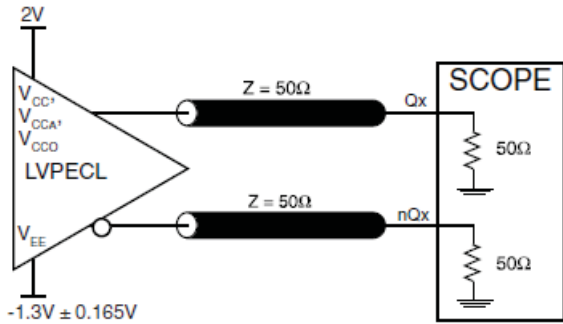
TYPICAL PHASE NOISE AT 155.52MHz



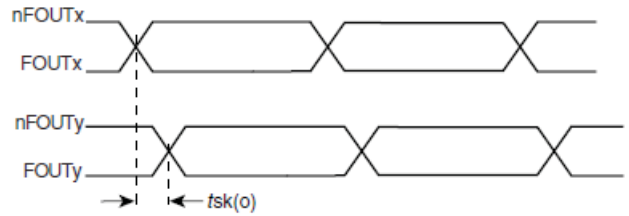
TYPICAL PHASE NOISE AT 622.08MHz



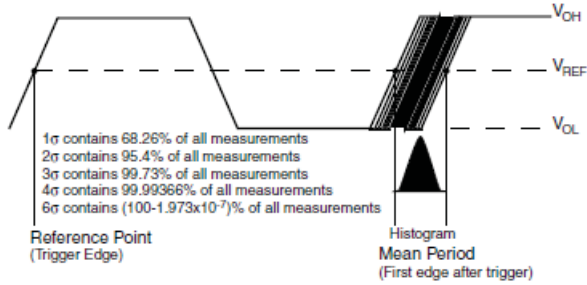
PARAMETER MEASUREMENT INFORMATION



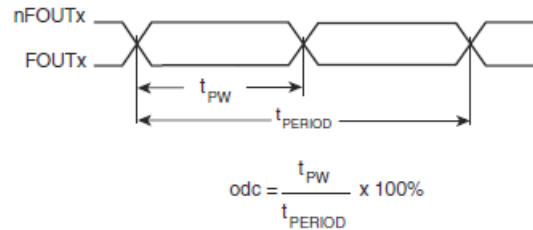
3.3V OUTPUT LOAD AC TEST CIRCUIT



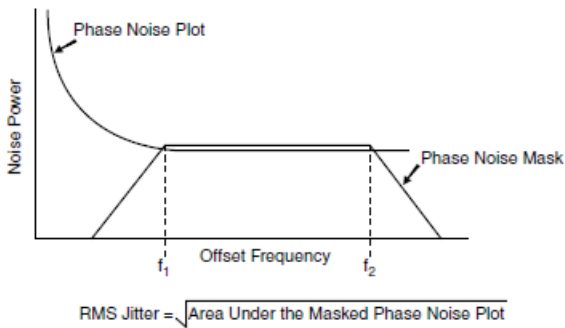
OUTPUT SKEW



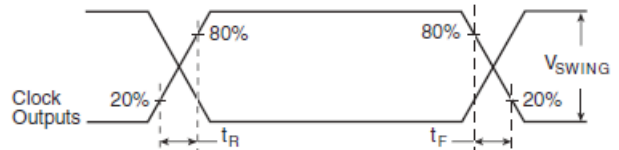
PERIOD JITTER



OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD



RMS PHASE JITTER



OUTPUT RISE/FALL TIME

APPLICATION INFORMATION

POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. The 843201-01 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL. V_{CC} , V_{CCA} , and V_{CCO} should be individually connected to the power supply plane through vias, and bypass capacitors should be used for each pin. To achieve optimum jitter performance, power supply isolation is required. *Figure 2* illustrates how a 24Ω resistor along with a $10\mu\text{F}$ and a $.01\mu\text{F}$ bypass capacitor should be connected to each V_{CCA} pin.

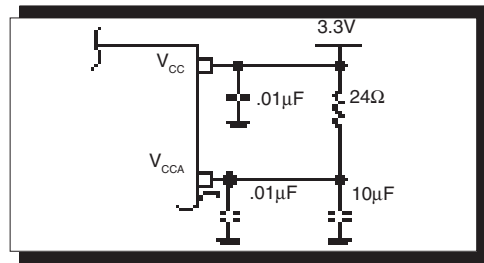


FIGURE 2. POWER SUPPLY FILTERING

RECOMMENDATIONS FOR UNUSED INPUT AND OUTPUT PINS

INPUTS:

CRYSTAL INPUT

For applications not requiring the use of the crystal oscillator input, both XTAL_IN and XTAL_OUT can be left floating. Though not required, but for additional protection, a $1\text{k}\Omega$ resistor can be tied from XTAL_IN to ground.

TEST_CLK INPUT

For applications not requiring the use of the test clock, it can be left floating. Though not required, but for additional protection, a $1\text{k}\Omega$ resistor can be tied from the TEST_CLK to ground.

SELECT PINS

All select pins have internal pull-ups and pull-downs; additional resistance is not required but can be added for additional protection. A $1\text{k}\Omega$ resistor can be used.

OUTPUTS:

LVC MOS OUTPUT

All unused LVC MOS output can be left floating. We recommend that there is no trace attached.

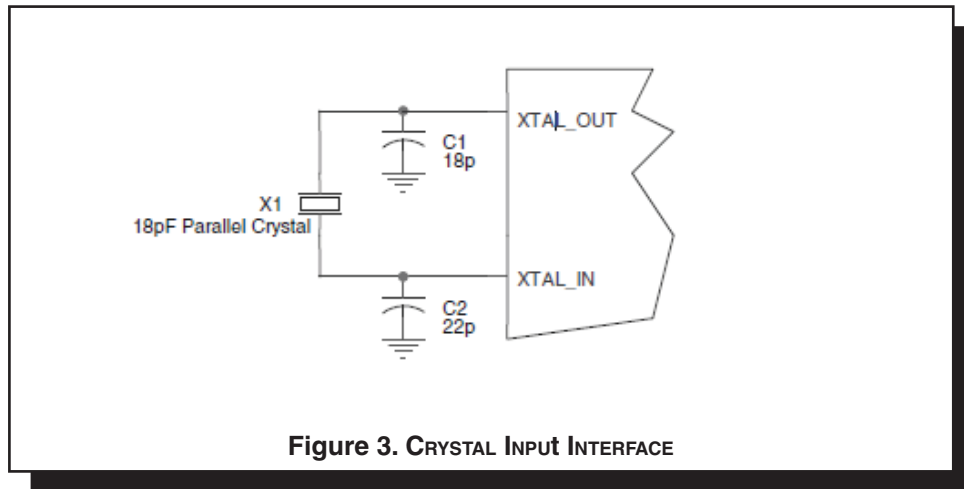
LVPECL OUTPUTS

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

CRYSTAL INPUT INTERFACE

A crystal can be characterized for either series or parallel mode operation. The 84320I-01 has a built-in crystal oscillator circuit. This interface can accept either a series or parallel crystal without additional components and generate frequencies with

accuracy suitable for most applications. Additional accuracy can be achieved by adding two small capacitors C1 and C2 as shown in *Figure 3*.



TERMINATION FOR LVPECL OUTPUTS

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

FOUT and nFOUT are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to

drive 50Ω transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

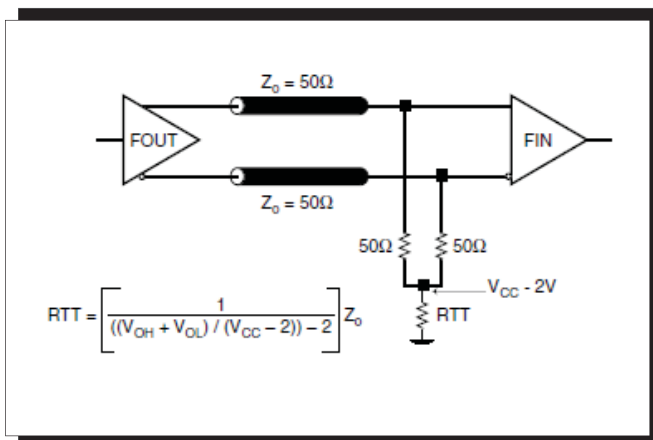


FIGURE 4A. LVPECL OUTPUT TERMINATION

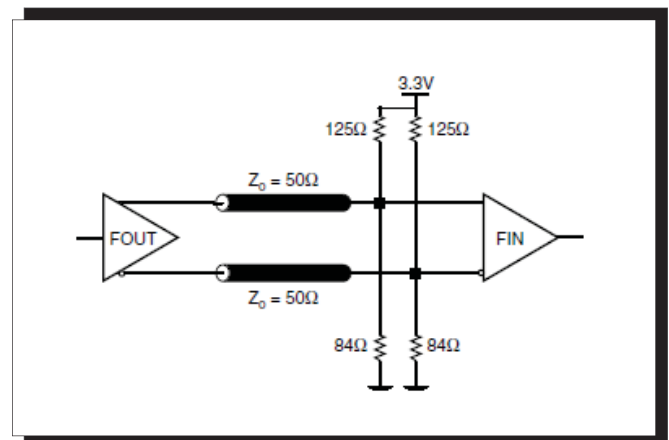


FIGURE 4B. LVPECL OUTPUT TERMINATION

LAYOUT GUIDELINE

The schematic of the 84320I-01 layout example used in this layout guideline is shown in Figure 5A. The 84320I-01 recommended PCB board layout for this example is shown in Figure 5B. This layout example is used as a general guideline. The

layout in the actual system will depend on the selected component types, the density of the components, the density of the traces, and the stack up of the P.C. board.

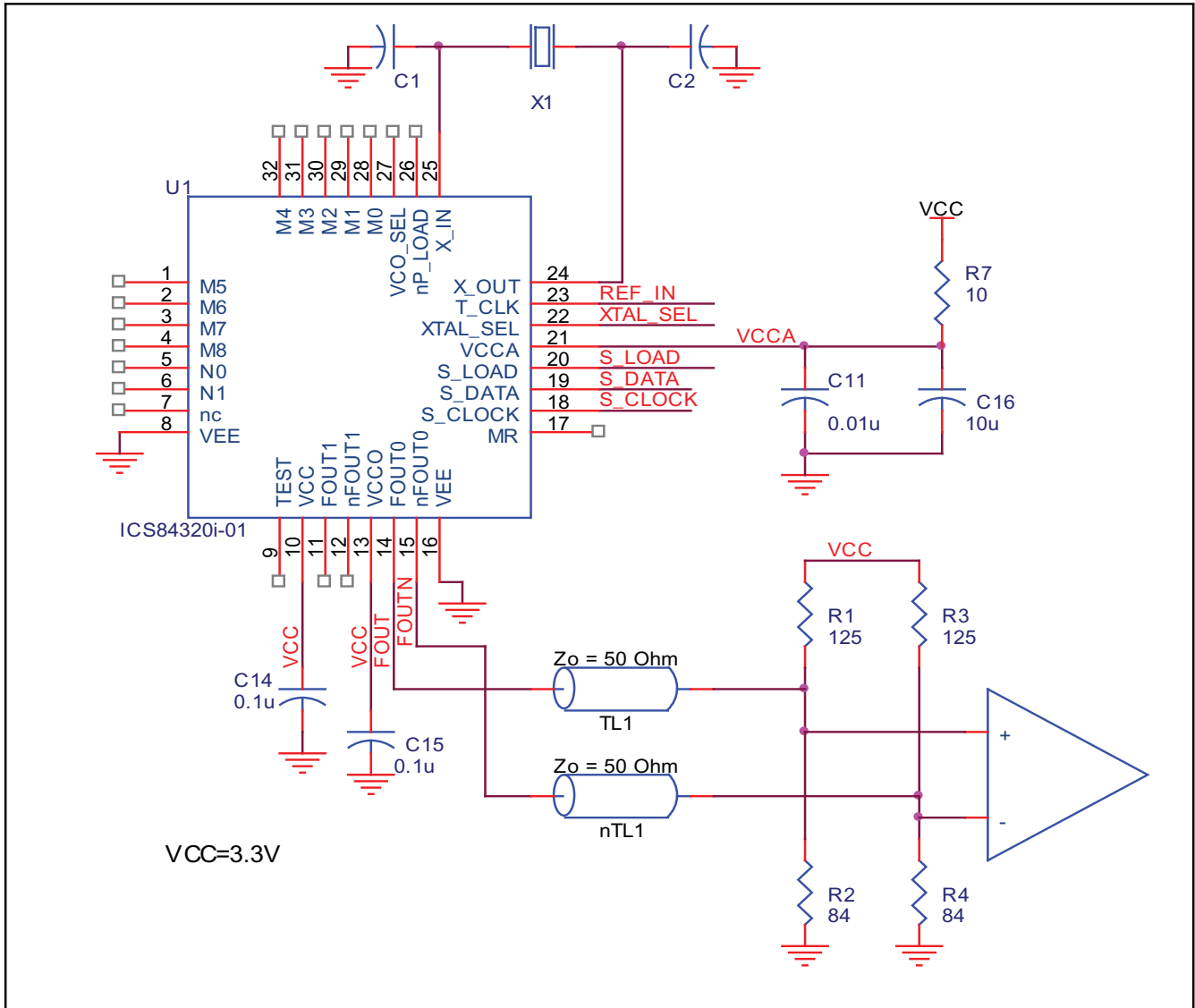


FIGURE 5A. SCHEMATIC OF RECOMMENDED LAYOUT

The following component footprints are used in this layout example:

All the resistors and capacitors are size 0603.

POWER AND GROUNDING

Place the decoupling capacitors C14 and C15, as close as possible to the power pins. If space allows, placement of the decoupling capacitor on the component side is preferred. This can reduce unwanted inductance between the decoupling capacitor and the power pin caused by the via.

Maximize the power and ground pad sizes and number of vias capacitors. This can reduce the inductance between the power and ground planes and the component power and ground pins.

The RC filter consisting of R7, C11, and C16 should be placed as close to the V_{CCA} pin as possible.

CLOCK TRACES AND TERMINATION

Poor signal integrity can degrade the system performance or cause system failure. In synchronous high-speed digital systems, the clock signal is less tolerant to poor signal integrity than other signals. Any ringing on the rising or falling edge or excessive ring back can cause system failure. The shape of the trace and the trace delay might be restricted by the available space on the board and the component location. While routing the traces, the clock signal traces should be routed first and should be locked prior to routing other signal traces.

- The differential 50Ω output traces should have the same length.
- Avoid sharp angles on the clock trace. Sharp angle turns cause the characteristic impedance to change on the transmission lines.
- Keep the clock traces on the same layer. Whenever possible, avoid placing vias on the clock traces. Placement of vias on the traces can affect the trace characteristic impedance and hence degrade signal integrity.
- To prevent cross talk, avoid routing other signal traces in parallel with the clock traces. If running parallel traces is unavoidable, allow a separation of at least three trace widths between the differential clock trace and the other signal trace.
- Make sure no other signal traces are routed between the clock trace pair.
- The matching termination resistors should be located as close to the receiver input pins as possible.

CRYSTAL

The crystal X1 should be located as close as possible to the pins 25 (XTAL_IN) and 24 (XTAL_OUT). The trace length between the X1 and U1 should be kept to a minimum to avoid unwanted parasitic inductance and capacitance. Other signal traces should not be routed near the crystal traces.

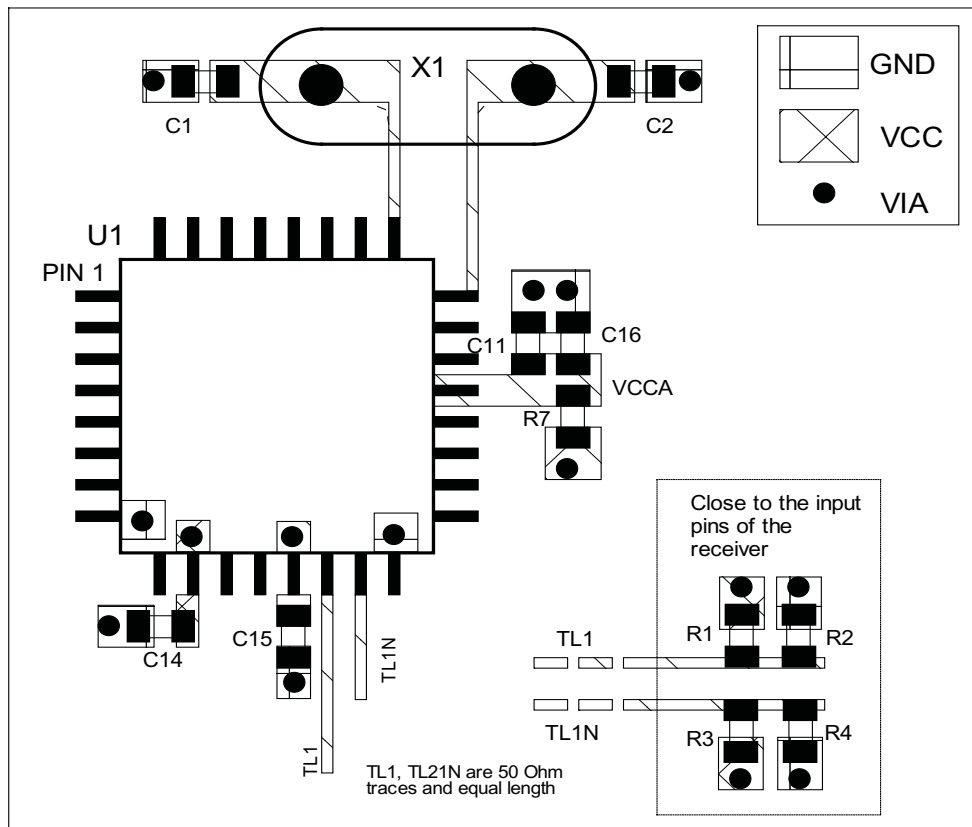


FIGURE 5B. PCB BOARD LAYOUT FOR 84320I-01

POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the 84320I-01. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the 84320I-01 is the sum of the core power plus the power dissipated in the load(s). The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)_{MAX} = $V_{CC_MAX} * I_{EE_MAX} = 3.465V * 155mA = 537.08mW$
- Power (outputs)_{MAX} = **30mW/Loaded Output pair**
If all outputs are loaded, the total power is $2 * 30mW = 60mW$

Total Power_{MAX} (3.465V, with all outputs switching) = $537.08mW + 60mW = 597.08mW$

2. Junction Temperature.

Junction temperature, T_j, is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for the devices is 125°C.

The equation for T_j is as follows: $T_j = \theta_{JA} * Pd_total + T_A$

T_j = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

Pd_{total} = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming a moderate air flow of 200 linear feet per minute and a multi-layer board, the appropriate value is 42.1°C/W per Table 8 below.

Therefore, T_j for an ambient temperature of 85°C with all outputs switching is:

$85^\circ C + 0.597W * 42.1^\circ C/W = 110.1^\circ C$. This is well below the limit of 125°C.

This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

TABLE 8. THERMAL RESISTANCE θ_{JA} FOR 32-PIN LQFP, FORCED CONVECTION

| θ_{JA} by Velocity (Linear Feet per Minute) | | | |
|--|----------|----------|----------|
| | 0 | 200 | 500 |
| Single-Layer PCB, JEDEC Standard Test Boards | 67.8°C/W | 55.9°C/W | 50.1°C/W |
| Multi-Layer PCB, JEDEC Standard Test Boards | 47.9°C/W | 42.1°C/W | 39.4°C/W |

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in Figure 6.

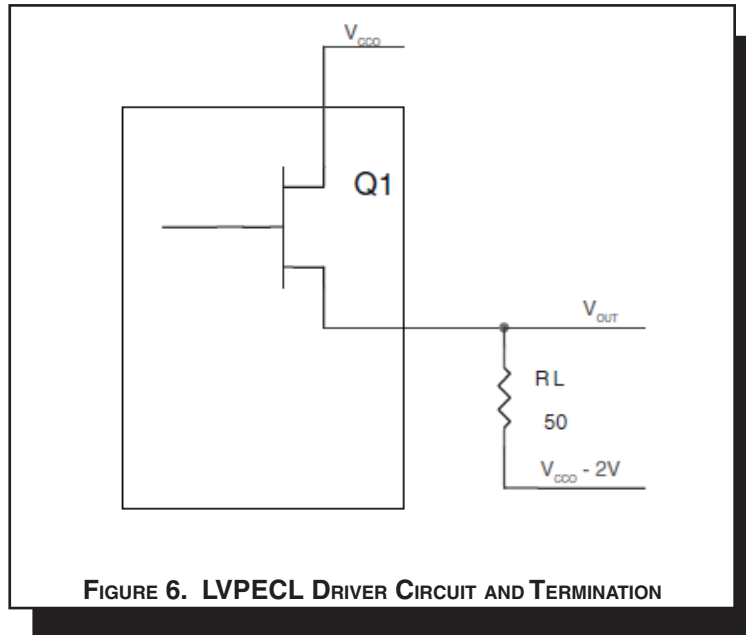


FIGURE 6. LVPECL DRIVER CIRCUIT AND TERMINATION

To calculate worst case power dissipation into the load, use the following equations which assume a 50Ω load, and a termination voltage of V_{CCO} - 2V.

- For logic high, V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 0.9V
 (V_{CCO_MAX} - V_{OH_MAX}) = 0.9V
- For logic low, V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.7V
 (V_{CCO_MAX} - V_{OL_MAX}) = 1.7V

Pd_H is power dissipation when the output drives high.
 Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

$$\text{Total Power Dissipation per output pair} = Pd_H + Pd_L = 30mW$$

RELIABILITY INFORMATION

TABLE 9. θ_{JA} VS. AIR FLOW TABLE FOR 32 LEAD LQFP

| θ_{JA} by Velocity (Linear Feet per Minute) | | | |
|--|----------|----------|----------|
| | 0 | 200 | 500 |
| Single-Layer PCB, JEDEC Standard Test Boards | 67.8°C/W | 55.9°C/W | 50.1°C/W |
| Multi-Layer PCB, JEDEC Standard Test Boards | 47.9°C/W | 42.1°C/W | 39.4°C/W |

NOTE: Most modern PCB designs use multi-layered boards. The data in the second row pertains to most designs.

TRANSISTOR COUNT

The transistor count for 84320I-01 is: 3776

PACKAGE OUTLINE - Y SUFFIX FOR 32 LEAD LQFP

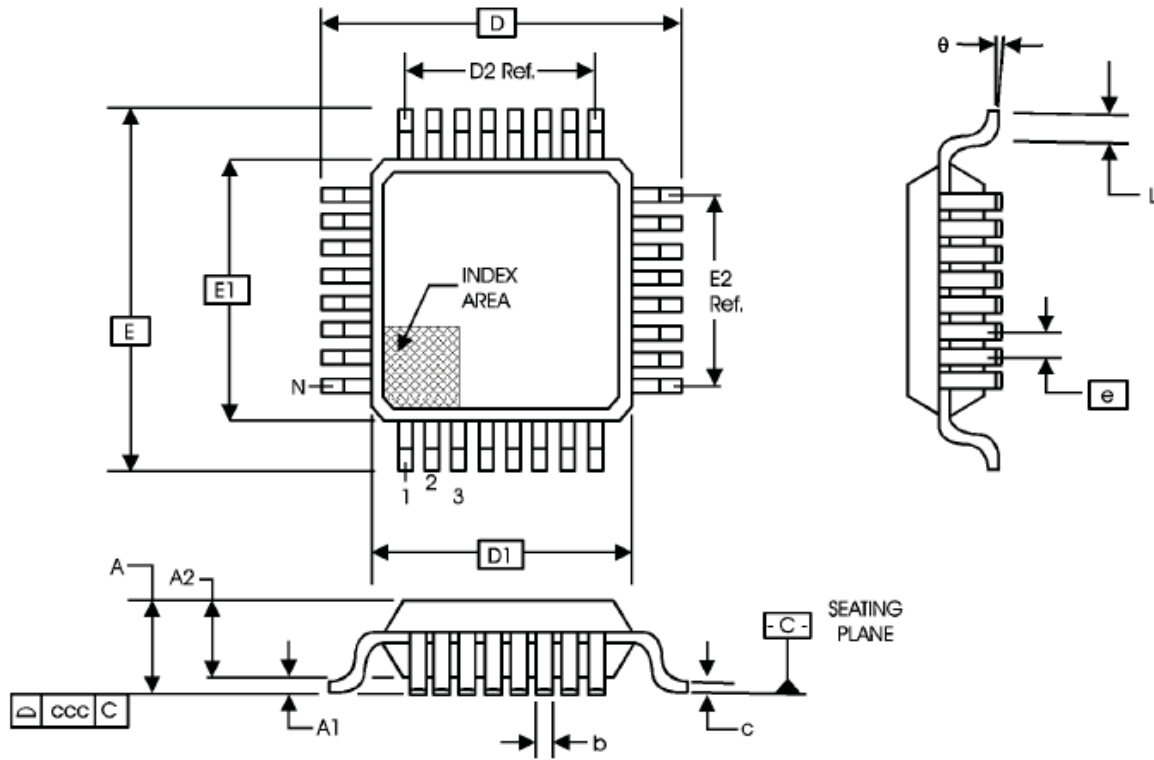


TABLE 10. PACKAGE DIMENSIONS

| JEDEC VARIATION ALL DIMENSIONS IN MILLIMETERS | | | |
|--|------------|---------|---------|
| SYMBOL | BBA | | |
| | MINIMUM | NOMINAL | MAXIMUM |
| N | 32 | | |
| A | -- | -- | 1.60 |
| A1 | 0.05 | -- | 0.15 |
| A2 | 1.35 | 1.40 | 1.45 |
| b | 0.30 | 0.37 | 0.45 |
| c | 0.09 | -- | 0.20 |
| D | 9.00 BASIC | | |
| D1 | 7.00 BASIC | | |
| D2 | 5.60 Ref. | | |
| E | 9.00 BASIC | | |
| E1 | 7.00 BASIC | | |
| E2 | 5.60 Ref. | | |
| e | 0.80 BASIC | | |
| L | 0.45 | 0.60 | 0.75 |
| theta | 0° | -- | 7° |
| ccc | -- | -- | 0.10 |

Reference Document: JEDEC Publication 95, MS-026

TABLE 11. ORDERING INFORMATION

| Part/Order Number | Marking | Package | Shipping Packaging | Temperature |
|-------------------|--------------|--------------------------|--------------------|---------------|
| 84320AYI-01LF | ICS4320AI01L | 32 Lead "Lead-Free" LQFP | tray | -40°C to 85°C |
| 84320AYI-01LFT | ICS4320AI01L | 32 Lead "Lead-Free" LQFP | 1000 tape & reel | -40°C to 85°C |

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

| REVISION HISTORY SHEET | | | | |
|-------------------------------|--------------|-------------|--|-------------|
| Rev | Table | Page | Description of Change | Date |
| A | T11 | 18 | Ordering Information Table - added LF marking | 9/5/08 |
| B | T11 | 18 20 | Updated datasheet's header/footer with IDT from ICS. Removed ICS prefix from Part/Order Number column. Added Contact Page. | 8/4/10 |
| C | T11 | 18 | Ordering Information Table - removed leaded devices, PDN CQ-13-02 expired | 12/17/14 |

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