

## GENERAL DESCRIPTION

The 843031-01 is an 10Gb Ethernet Clock Generator and a member of the family of high performance devices from IDT. The 843031-01 uses an 18pF parallel resonant crystal. The 843031-01 has excellent <1ps phase jitter performance, over the 1.875MHz - 20MHz integration range. The 843031-01 is packaged in a small 8-pin TSSOP, making it ideal for use in systems with limited board space.

## FEATURES

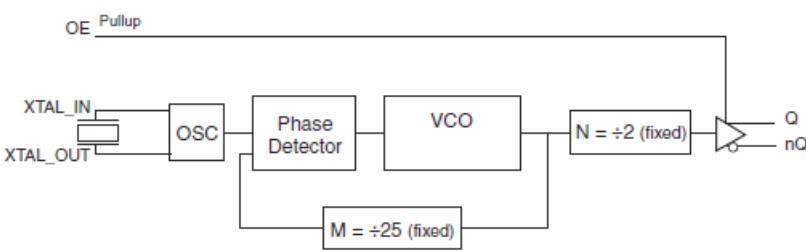
- One differential 3.3V or 2.5V LVPECL output

- Crystal oscillator interface designed for 25MHz, 18pF parallel resonant crystal
- Output frequencies: 280MHz – 340MHz
- VCO range: 560MHz - 680MHz
- RMS phase jitter @ 312.5MHz, using a 25MHz crystal (1.875MHz - 20MHz): 0.46ps (typical)
- Full 3.3V or 2.5V operating supply
- 0°C to 70°C ambient operating temperature
- Available in lead-free (RoHS 6) package

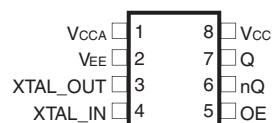
## COMMON CONFIGURATION TABLE

Inputs				Output Frequency (MHz)
Crystal Frequency (MHz)	M	N	Multiplication Value M/N	
25	25	2	12.5	312.5

## BLOCK DIAGRAM



## PIN ASSIGNMENT



**843031-01**  
8-Lead TSSOP  
4.40mm x 3.0mm x 0.925mm  
package body  
**G Package**  
Top View

**TABLE 1. PIN DESCRIPTIONS**

Number	Name	Type		Description
1	$V_{CCA}$	Power		Analog supply pin.
2	$V_{EE}$	Power		Negative supply pin.
3, 4	XTAL_OUT, XTAL_IN	Input		Crystal oscillator interface. XTAL_IN is the input, XTAL_OUT is the output.
5	OE	Input	Pullup	Output Enable pin. LVCMOS/LVTTL interface levels.
6, 7	nQ, Q	Output		Differential clock outputs. LVPECL interface levels.
8	$V_{CC}$	Power		Power supply pin.

NOTE: *Pullup* refers to internal input resistors. See Table 2, Pin Characteristics, for typical values.

**TABLE 2. PIN CHARACTERISTICS**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$C_{IN}$	Input Capacitance			4		pF
$R_{PULLDOWN}$	Input Pullup Resistor			51		k $\Omega$

**TABLE 3. OE FUNCTION TABLE**

Input	Outputs
OE	Q/nQ
0	Hi-Z
1	Enabled

## ABSOLUTE MAXIMUM RATINGS

Supply Voltage, $V_{cc}$	4.6V
Inputs, $V_i$	-0.5V to $V_{cc}$ + 0.5V
Outputs, $I_o$	
Continuous Current	50mA
Surge Current	100mA
Package Thermal Impedance, $\theta_{JA}$	101.7°C/W (0 mps)
Storage Temperature, $T_{STG}$	-65°C to 150°C

NOTE: Stresses beyond those listed under Absolute Maximum Ratings may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

TABLE 4A. POWER SUPPLY DC CHARACTERISTICS,  $V_{cc} = 3.3V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{cc}$	Power Supply Voltage		3.135	3.3	3.465	V
$V_{CCA}$	Analog Supply Voltage		$V_{cc} - 0.12$	3.3	3.465	V
$I_{CCA}$	Analog Supply Current				12	mA
$I_{EE}$	Power Supply Current				105	mA

TABLE 4B. POWER SUPPLY DC CHARACTERISTICS,  $V_{cc} = 2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{cc}$	Power Supply Voltage		2.375	2.5	2.625	V
$V_{CCA}$	Analog Supply Voltage		$V_{cc} - 0.12$	2.5	2.625	V
$I_{CCA}$	Analog Supply Current				12	mA
$I_{EE}$	Power Supply Current				90	mA

TABLE 4C. LVCMOS/LVTTL DC CHARACTERISTICS,  $V_{cc} = 3.3V \pm 5\%$  OR  $2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$ 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage	$V_{cc} = 3.3V$	2		$V_{cc} + 0.3$	V
		$V_{cc} = 2.5V$	1.7		$V_{cc} + 0.3$	V
$V_{IL}$	Input Low Voltage	$V_{cc} = 3.3V$	-0.3		0.8	V
		$V_{cc} = 2.5V$	-0.3		0.7	V
$I_{IH}$	Input High Current	$V_{cc} = V_{IN} = 3.465V$ or $2.625V$			5	$\mu A$
$I_{IL}$	Input Low Current	$V_{cc} = 3.465V$ or $2.625V$ , $V_{IN} = 0V$	-150			$\mu A$

**TABLE 4D. LVPECL DC CHARACTERISTICS,  $V_{cc} = 3.3V \pm 5\%$  OR  $2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{OH}$	Output High Voltage; NOTE 1		$V_{cc} - 1.4$		$V_{cc} - 0.9$	V
$V_{OL}$	Output Low Voltage; NOTE 1		$V_{cc} - 2.0$		$V_{cc} - 1.7$	V
$V_{SWING}$	Peak-to-Peak Output Voltage Swing		0.6		1.0	V

NOTE 1: Outputs terminated with  $50\Omega$  to  $V_{cc} - 2V$ .

**TABLE 5. CRYSTAL CHARACTERISTICS**

Parameter	Test Conditions	Minimum	Typical	Maximum	Units
Mode of Oscillation			Fundamental		
Frequency		22.4	25	27.2	MHz
Equivalent Series Resistance (ESR)				40	$\Omega$
Shunt Capacitance				7	pF
Drive Level				300	mW

**TABLE 6A. AC CHARACTERISTICS,  $V_{cc} = 3.3V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{out}$	Output Frequency		280	312.5	340	MHz
$t_{jit}(\emptyset)$	RMS Phase Jitter ( Random); NOTE 1	312.5MHz @ Integration Range: 1.875MHz - 20MHz		0.46		ps
$t_r / t_f$	Output Rise/Fall Time	20% to 80%	150		500	ps
odc	Output Duty Cycle		48		52	%

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 Ifpm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: Please refer to the Phase Noise Plots following this section.

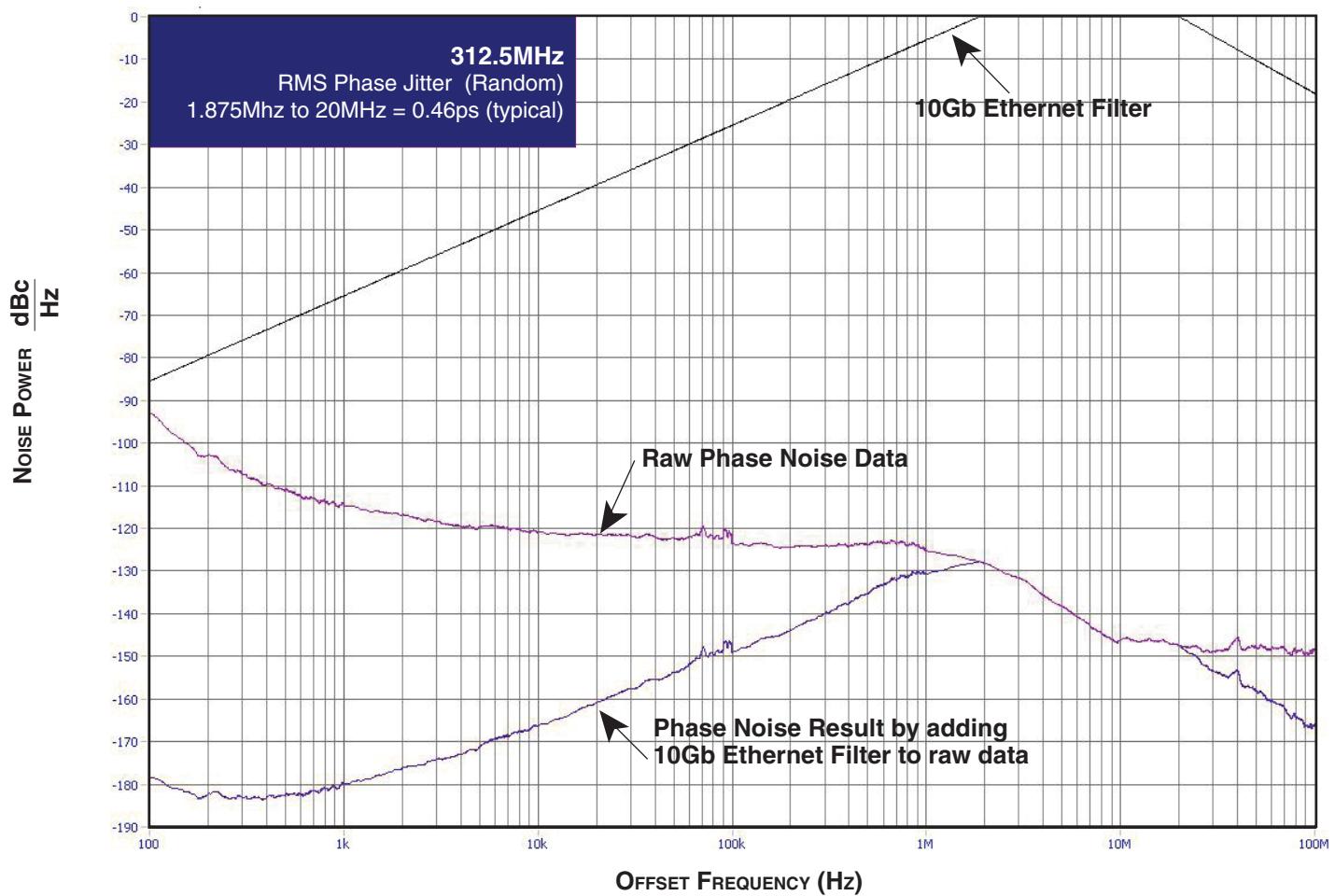
**TABLE 6B. AC CHARACTERISTICS,  $V_{cc} = 2.5V \pm 5\%$ ,  $T_A = 0^\circ C$  TO  $70^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{out}$	Output Frequency		280	312.5	340	MHz
$t_{jit}(\emptyset)$	RMS Phase Jitter ( Random); NOTE 1	312.5MHz @ Integration Range: 1.875MHz - 20MHz		0.48		ps
$t_r / t_f$	Output Rise/Fall Time	20% to 80%	150		500	ps
odc	Output Duty Cycle		48		52	%

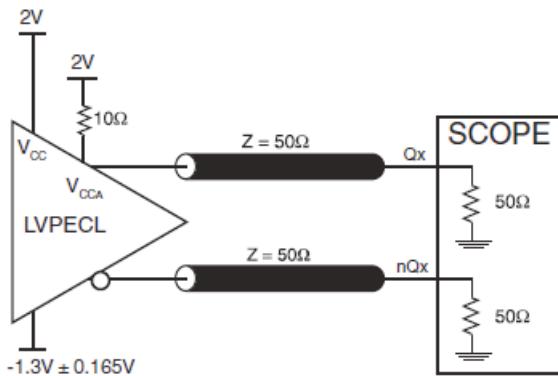
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NOTE 1: Please refer to the Phase Noise Plots following this section.

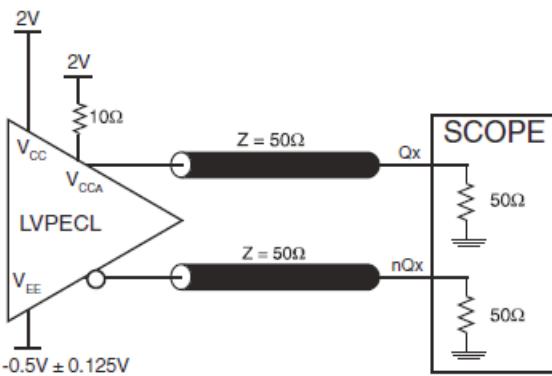
## TYPICAL PHASE NOISE AT 312.5MHz AT 3.3V



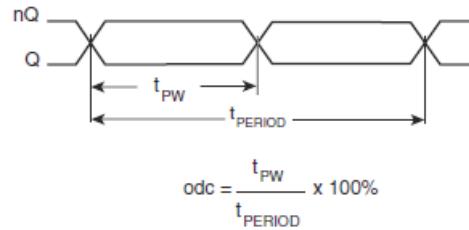
# PARAMETER MEASUREMENT INFORMATION



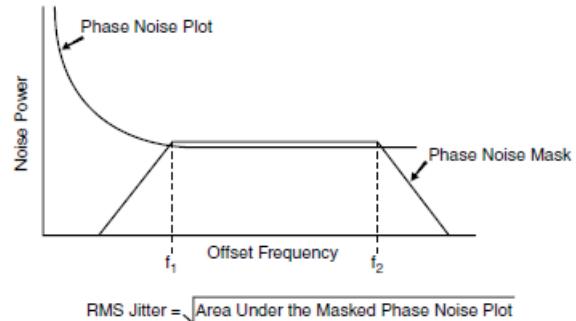
**3.3V OUTPUT LOAD AC TEST CIRCUIT**



**2.5V OUTPUT LOAD AC TEST CIRCUIT**



**OUTPUT DUTY CYCLE/PULSE WIDTH/PERIOD**



**RMS PHASE JITTER**



**OUTPUT RISE/FALL TIME**

## APPLICATION INFORMATION

### POWER SUPPLY FILTERING TECHNIQUES

As in any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The 843031-01 provides separate power supplies to isolate any high switching noise from the outputs to the internal PLL.  $V_{cc}$  and  $V_{cca}$  should be individually connected to the power supply plane through vias, and  $0.01\mu F$  bypass capacitors should be used for each pin. Figure 1 illustrates this for a generic  $V_{cc}$  pin and also shows that  $V_{cca}$  requires that an additional  $10\Omega$  resistor along with a  $10\mu F$  bypass capacitor be connected to the  $V_{cca}$  pin.

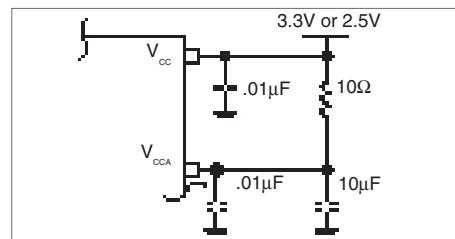


FIGURE 1. POWER SUPPLY FILTERING

### CRYSTAL INPUT INTERFACE

The 843031-01 has been characterized with  $18pF$  parallel resonant crystals. The capacitor values,  $C1$  and  $C2$ , shown in Figure 2 below were determined using a  $25MHz$ ,  $18pF$  parallel

resonant crystal and were chosen to minimize the ppm error. The optimum  $C1$  and  $C2$  values can be slightly adjusted for different board layouts.

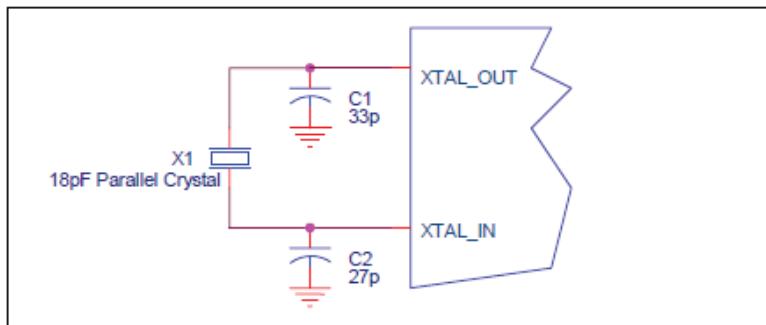


FIGURE 2. CRYSTAL INPUT INTERFACE

## LVC MOS TO XTAL INTERFACE

The XTAL\_IN input can accept a single-ended LVC MOS signal through an AC coupling capacitor. A general interface diagram is shown in *Figure 3*. The XTAL\_OUT pin can be left floating. The input edge rate can be as slow as 10ns. For LVC MOS inputs, it is recommended that the amplitude be reduced from full swing to half swing in order to prevent signal interference with the power rail and to reduce noise. This configuration requires that the output

impedance of the driver ( $R_o$ ) plus the series resistance ( $R_s$ ) equals the transmission line impedance. In addition, matched termination at the crystal input will attenuate the signal in half. This can be done in one of two ways. First,  $R_1$  and  $R_2$  in parallel should equal the transmission line impedance. For most  $50\Omega$  applications,  $R_1$  and  $R_2$  can be  $100\Omega$ . This can also be accomplished by removing  $R_1$  and making  $R_2 50\Omega$ .

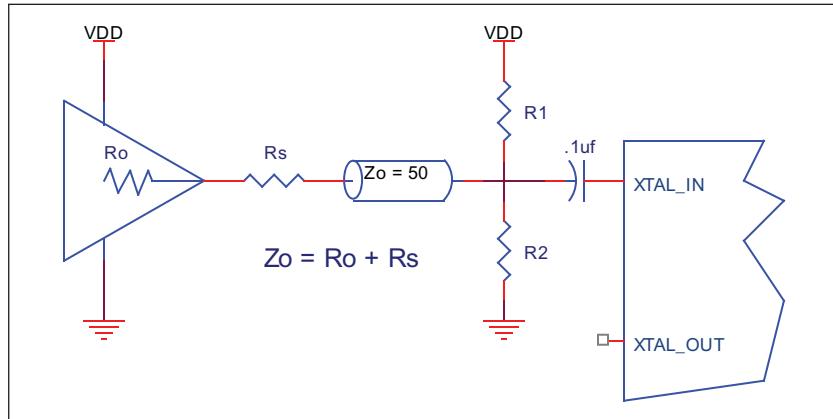


FIGURE 3. GENERAL DIAGRAM FOR LVC MOS DRIVER TO XTAL INPUT INTERFACE

## TERMINATION FOR 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

$F_{OUT}$  and  $nF_{OUT}$  are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive  $50\Omega$  transmission

lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A* and *4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

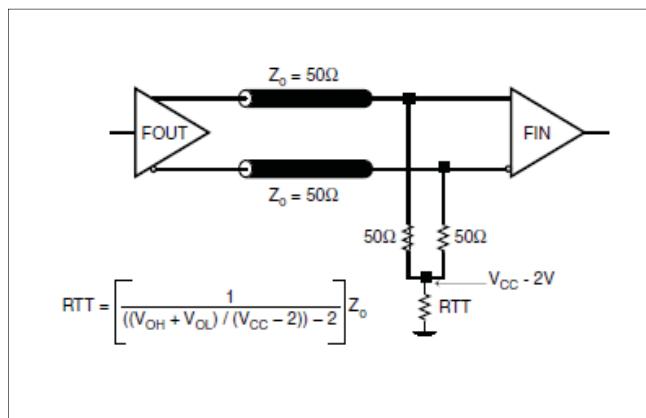


FIGURE 4A. LVPECL OUTPUT TERMINATION

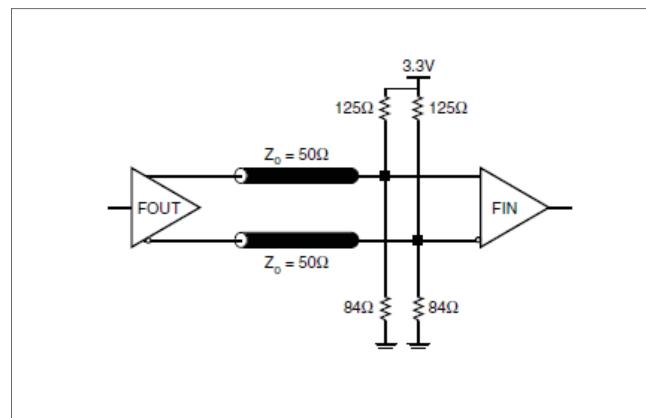


FIGURE 4B. LVPECL OUTPUT TERMINATION

## TERMINATION FOR 2.5V LVPECL OUTPUTS

Figure 5A and Figure 5B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating  $50\Omega$  to  $V_{cc} - 2V$ . For  $V_{cc} = 2.5V$ , the  $V_{cc} - 2V$  is very close to ground

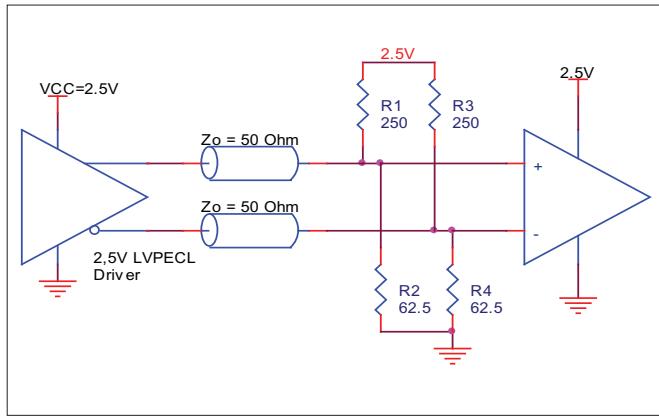


FIGURE 5A. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

level. The R3 in Figure 5B can be eliminated and the termination is shown in Figure 5C.

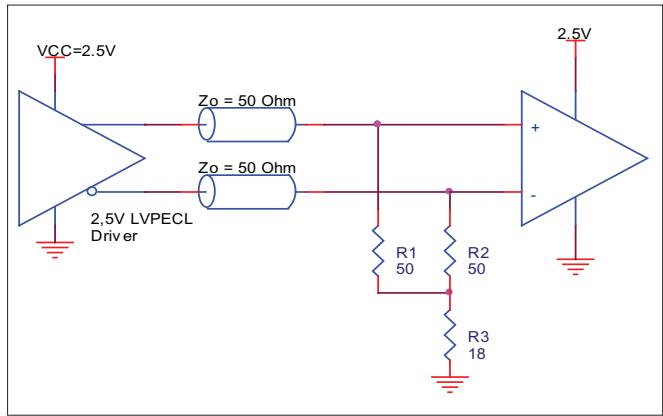


FIGURE 5B. 2.5V LVPECL DRIVER TERMINATION EXAMPLE

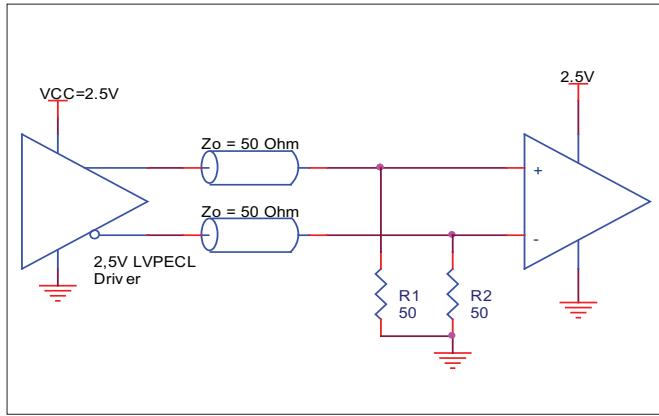


FIGURE 5C. 2.5V LVPECL TERMINATION EXAMPLE

## POWER CONSIDERATIONS

This section provides information on power dissipation and junction temperature for the 843031-01.

Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the 843031-01 is the sum of the core power plus the power dissipated in the load(s).

The following is the power dissipation for  $V_{CC} = 3.3V + 5\% = 3.465V$ , which gives worst case results.

**NOTE:** Please refer to Section 3 for details on calculating power dissipated in the load.

- Power (core)<sub>MAX</sub> =  $V_{CC\_MAX} * I_{EE\_MAX} = 3.465V * 105mA = 363.8\text{mW}$
- Power (outputs)<sub>MAX</sub> = **30mW/Loaded Output pair**

**Total Power<sub>MAX</sub>** (3.465V, with all outputs switching) =  $363.8\text{mW} + 30\text{mW} = 393.8\text{mW}$

### 2. Junction Temperature.

Junction temperature,  $T_J$ , is the temperature at the junction of the bond wire and bond pad and directly affects the reliability of the device. The maximum recommended junction temperature for HiPerClockS™ devices is 125°C.

The equation for  $T_J$  is as follows:  $T_J = \theta_{JA} * P_{d\_total} + T_A$

$T_J$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$P_{d\_total}$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming a moderate air flow of 1 meter per second and a multi-layer board, the appropriate value is 90.5°C/W per Table 7 below.

Therefore,  $T_J$  for an ambient temperature of 70°C with all outputs switching is:

$70^\circ\text{C} + 0.394\text{W} * 90.5^\circ\text{C/W} = 105.6^\circ\text{C}$ . This is well below the limit of 125°C.

This calculation is only an example.  $T_J$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow, and the type of board (single layer or multi-layer).

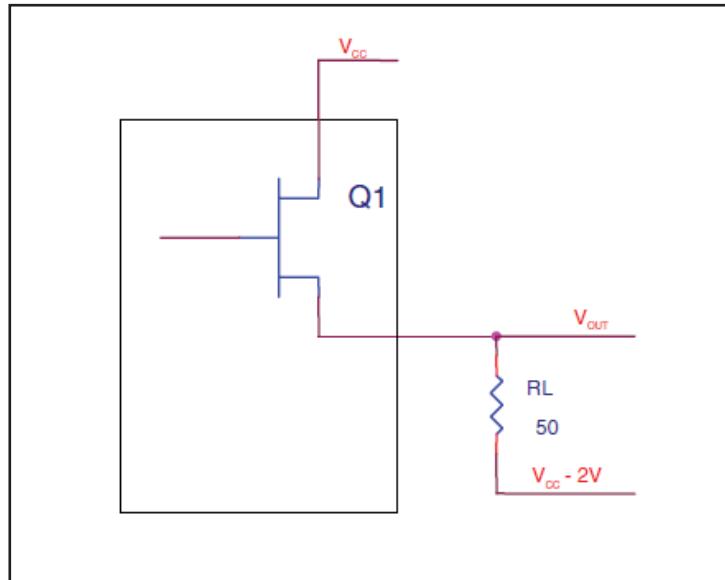
**TABLE 7. THERMAL RESISTANCE  $\theta_{JA}$  FOR 8-PIN TSSOP, FORCED CONVECTION**

<b><math>\theta_{JA}</math> by Velocity (Meters per Second)</b>			
	<b>0</b>	<b>1</b>	<b>2.5</b>
Multi-Layer PCB, JEDEC Standard Test Boards	101.7°C/W	90.5°C/W	89.8°C/W

### 3. Calculations and Equations.

The purpose of this section is to derive the power dissipated into the load.

LVPECL output driver circuit and termination are shown in *Figure 6*.



**FIGURE 6. LVPECL DRIVER CIRCUIT AND TERMINATION**

To calculate worst case power dissipation into the load, use the following equations which assume a  $50\Omega$  load, and a termination voltage of  $V_{CC} - 2V$ .

- For logic high,  $V_{OUT} = V_{OH_{MAX}} = V_{CC_{MAX}} - 0.9V$

$$(V_{CC_{MAX}} - V_{OH_{MAX}}) = 0.9V$$

- For logic low,  $V_{OUT} = V_{OL_{MAX}} = V_{CC_{MAX}} - 1.7V$

$$(V_{CC_{MAX}} - V_{OL_{MAX}}) = 1.7V$$

$Pd_H$  is power dissipation when the output drives high.

$Pd_L$  is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_{MAX}} - (V_{CC_{MAX}} - 2V))/R_L] * (V_{CC_{MAX}} - V_{OH_{MAX}}) = [(2V - (V_{CC_{MAX}} - V_{OH_{MAX}}))/R_L] * (V_{CC_{MAX}} - V_{OH_{MAX}}) = [(2V - 0.9V)/50\Omega] * 0.9V = 19.8mW$$

$$Pd_L = [(V_{OL_{MAX}} - (V_{CC_{MAX}} - 2V))/R_L] * (V_{CC_{MAX}} - V_{OL_{MAX}}) = [(2V - (V_{CC_{MAX}} - V_{OL_{MAX}}))/R_L] * (V_{CC_{MAX}} - V_{OL_{MAX}}) = [(2V - 1.7V)/50\Omega] * 1.7V = 10.2mW$$

Total Power Dissipation per output pair =  $Pd_H + Pd_L = 30mW$

## RELIABILITY INFORMATION

**TABLE 8.  $\theta_{JA}$  VS. AIR FLOW TABLE FOR 8 LEAD TSSOP**

<b><math>\theta_{JA}</math> by Velocity (Meters per Second)</b>			
	<b>0</b>	<b>1</b>	<b>2.5</b>
Multi-Layer PCB, JEDEC Standard Test Boards	101.7°C/W	90.5°C/W	89.8°C/W

**TRANSISTOR COUNT**

The transistor count for 843031-01 is: 2377

## PACKAGE OUTLINE - G SUFFIX FOR 8 LEAD TSSOP

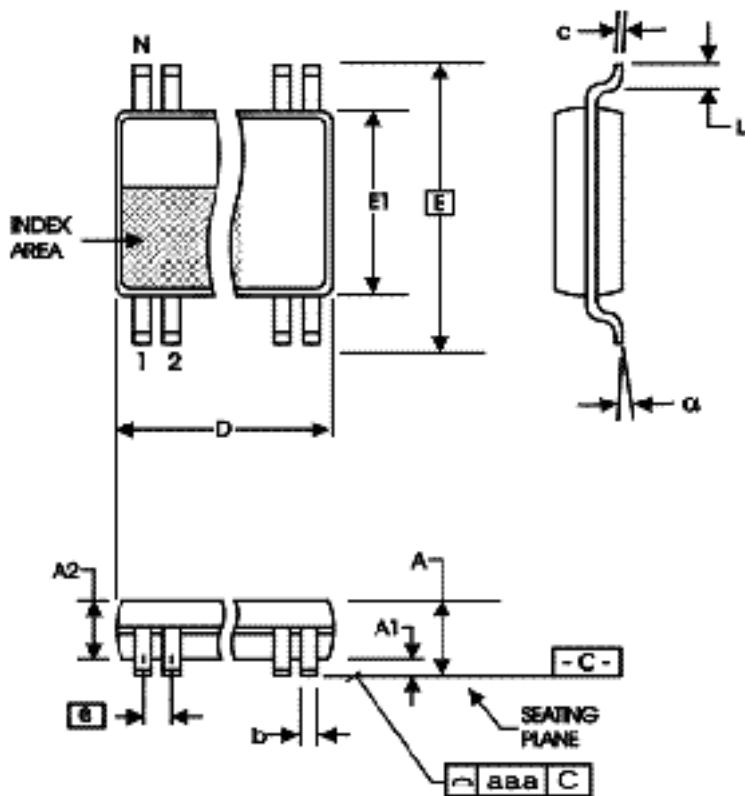


TABLE 9. PACKAGE DIMENSIONS

SYMBOL	Millimeters	
	Minimum	Maximum
N		8
A	--	1.20
A1	0.05	0.15
A2	0.80	1.05
b	0.19	0.30
c	0.09	0.20
D	2.90	3.10
E	6.40 BASIC	
E1	4.30	4.50
e	0.65 BASIC	
L	0.45	0.75
alpha	0°	8°
aaa	--	0.10

Reference Document: JEDEC Publication 95, MO-153

**TABLE 10. ORDERING INFORMATION**

Part/Order Number	Marking	Package	Shipping Packaging	Temperature
843031AG-01LF	1A01L	8 Lead "Lead-Free" TSSOP	tube	0°C to 70°C
843031AG-01LFT	1A01L	8 Lead "Lead-Free" TSSOP	tape & reel	0°C to 70°C

NOTE: Parts that are ordered with an "LF" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

## REVISION HISTORY SHEET

Rev	Table	Page	Description of Change	Date
A	T3	2	Added OE Function Table.	1/23/07
A	T10	14	Order Information Table - Added Lead Free marking	8/1/07
A		1	Common Configuration Table - corrected typo in Multiplication Value M/N column from 25 to 12.5.	11/11/08
A	T10	14	Ordering Information - removed leaded devices. Updated data sheet format.	10/15/15



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