RENESAS

μΡD48011318 μΡD48011336

1.1G-BIT Low Latency DRAM-III Common I/O Burst Length of 2

R10DS0012EJ0200 Rev.2.00 Feb 01, 2013

Description

The μ PD48011318 is a 67,108,864-word by 18-bit and the μ PD48011336 is a 33,554,432-word by 36-bit synchronous double data rate Low Latency RAM fabricated with advanced CMOS technology using one-transistor eDRAM memory cell.

The Low Latency DRAM-III chip is a 1.1Gbit DRAM capable of a sustained throughput of approximately 43.2 Gbps for burst length of 2 (approximately 51.2 Gbps for applications implementing error correction), excluding refresh overhead and data bus turn-around

With a bus speed of 600 MHz, a burst length of 2, and a trc of 13.3 ns, the Low Latency DRAM-III chip is capable of achieving this rate when accesses to at least 6 banks of memory are overlapped.

These products are packaged in 180-pin TAPE FBGA.

Specification

- Density: 1,1Gbit
- Organization: 8M words x 18 bits x 8 banks 4M words x 36 bits x 8 banks
- Operating frequency 600MHz (MAX.) @ tRc=13.3 ns
- t_{RC}
 - 13.3 ns tRC (and 13.3 ns tRFC)
- Burst length: 2
- Address bus
 - 1 cycle DDR address

Package

180-pin FBGA (Ball Array: 1 mm x 1 mm Pitch) Package size: 18.5 mm x 14 mm ROHS 6/6 compliance

- Power supply
 - 2.5 V VEXT
 - 1.5 V VDD
 - 1.0 V VDDQ
- Refresh command
 - Auto Refresh : 16384 cycles / 2 ms for each bank
 - Overlapped Refresh with DPR# pin
- Operating case temperature: 0 to 95 °C



Features

- 1 cycle 600MHz DDR Muxed Address
- Optional data bus inversion to reduce SSO, SSN, maximum I/O current, and average I/O power
- Training sequence for per-bit deskew
- Selectable Refresh Mode: Auto or Overlapped Refresh
- Programmable PVT-compensated output impedance
- Programmable PVT-compensated on-die input termination
- PLL for improved input jitter tolerance and wide output data valid window

Ordering Information

Part number	Cycle Time	Clock Frequency	Random Cycle	Output Supply Voltage	Burst Length	Address Type	Organization (word x bit)	Package
				(VDDQ)				
	ns	MHz	ns	v				
μPD48011318FF-FH16-FF1-A	1.66	600	13.3	1.0	2	DDR	64 M x 18	180-pin
μPD48011318FF-FH19-FF1-A	1.90	525						TAPE FBGA
μPD48011336FF-FH16-FF1-A	1.66	600					32 M x 36	(18.5 x14)
μPD48011336FF-FH19-FF1-A	1.90	525						Lead-free



Pin Configurations

	1	2	3	4	5	6	7	8	9	10	11	12	13
А	TMS	VDD	тск	VddQ	Vss				Vss	VddQ	VEXT	VDD	QK0
В	Vss	DNU, Vss	Vss	DQ6	Vdd		x18		VDD	DNU, Vss	Vss	DQ8	Vss
С	QVLD	VDDQ	DNU, Vss	VddQ	DQ4		Burst of 2		DNU, Vss	VDDQ	DQ7	VddQ	QK0#
D	Vss	DNU, Vss	Vss	DQ2	Vss		259 sq. mm		Vss	DNU, Vss	Vss	DQ5	Vss
Е	DNU, Vss	VddQ	DNU, Vss	VddQ	DQ0				DNU, Vss	VddQ	DQ1	VddQ	DQ3
F	Vss	DINV0	Vss	A12	Vss				Vss	A13	Vss	DNU, Vss	Vss
G	Vdd	VddQ	A10	VddQ	A8				A7	VddQ	A11	VDDQ	Vdd
Н	Vss	RST#	Vss	A6	Vss		Top View		Vss	A5	Vss	DM	Vss
J	MF	VddQ	DK0	Vss	Vdd				СК	Vss	DNU, Vss	VDDQ	LBK#
К	VREF	Vss	DK0#	VDDQ	Vdd				CK#	VDDQ	DNU, Vss	Vss	VREF
L	TRST#	CS#	Vss	A4	Vss				Vss	A3	Vss	WE#	Vss
М	Vdd	VDDQ	DPR#	VddQ	A2				REF#	VddQ	A9	VDDQ	Vdd
Ν	Vss	DINV1	Vss	A0	Vss				Vss	A1	Vss	DNU, Vss	Vss
Ρ	DNU, Vss	VDDQ	DNU, Vss	VddQ	DQ9				DNU, Vss	VddQ	DQ10	VDDQ	DQ12
R	Vss	DNU, Vss	Vss	DQ11	Vss				Vss	DNU, Vss	Vss	DQ14	Vss
Т	QK1#	VDDQ	DNU, Vss	VddQ	DQ13				DNU, Vss	VDDQ	DQ16	VDDQ	TDO
U	Vss	DNU, Vss	Vss	DQ15	Vdd				Vdd	DNU, Vss	Vss	DQ17	Vss
V	QK1	VDD	ZQ	VddQ	Vss				Vss	VddQ	VEXT	VDD	TDI
										l	l	1	1
)Q0-DQ1)INV0-DI			nputs / o nversion	-		REF#		Refresh E				
	ли v 0–Di)M						LBK# MF		.oopback /lirror Fui				
	0–A13		: Addres	data mas	21		TMS		EEE 114		innut		
	CK, CK#		: Input o				TDI		EEE 114		•		
	0K0, DK0	#		lata cloc	k		TCK		EEE 114		•		
	QVLD		-	Data Val			TDO		EEE 114		•		
	VLD 0K0–QK1			t data clo			TRST#		EEE 114		•	nut	
	2K0#-QK		. Guipu				Vref		ISTL inp				
	200#–QN 20		· Outou	timpeda	nce & In	out	VREF		Power Su		•		
2	. 					Jui							
			termin	ation cor	100		VddQ	: L	Q Powe	Supply	, i.u v r	ominal	

180-pin TAPE FBGA (18.5 x 14) (Top View) [μPD48011318] (x18)

: Master reset

: Chip select

: Write Enable

: Data inputs / output

RST#

CS#

WE#

DQ0-DQ17



Vext

Vss

DNU, Vss

DPR#

: Power Supply, 2.5 V nominal

: Overlapped Refresh

DNU, $V_{DD}Q$: Must not be used, or must be connected to $V_{DD}Q$

: Must not be used, or must be connected to Vss

: Ground

180-pin TAPE FBGA (18.5 x 14) (Top View) [μPD48011336] (x36)

	1	2	3	4	5	6	7	8	9	10	11	12	13
А	TMS	Vdd	тск	VddQ	Vss				Vss	VddQ	VEXT	Vdd	QK0
В	Vss	DQ9	Vss	DQ6	Vdd		x36		Vdd	DQ11	Vss	DQ8	Vss
С	QVLD	VddQ	DQ10	VddQ	DQ4		Burst of 2		DQ13	VddQ	DQ7	VddQ	QK0#
D	Vss	DQ12	Vss	DQ2	Vss		259 sq. mm		Vss	DQ15	Vss	DQ5	Vss
Е	DQ16	VDDQ	DQ14	VddQ	DQ0				DQ17	VddQ	DQ1	VddQ	DQ3
F	Vss	DINV0	Vss	A12	Vss				Vss	A13	Vss	DINV1	Vss
G	Vdd	VDDQ	A10	VddQ	A8				A7	VddQ	A11	VddQ	Vdd
н	Vss	RST#	Vss	A6	Vss		Top View		Vss	A5	Vss	DM	Vss
J	MF	VddQ	DK0	Vss	Vdd				СК	Vss	DK1	VddQ	LBK#
к	VREF	Vss	DK0#	VddQ	VDD				CK#	VddQ	DK1#	Vss	VREF
L	TRST#	CS#	Vss	A4	Vss				Vss	A3	Vss	WE#	Vss
М	Vdd	VDDQ	DPR#	VddQ	A2				REF#	VDDQ	A9	VddQ	Vdd
Ν	Vss	DINV2	Vss	A0	Vss				Vss	A1	Vss	DINV3	Vss
Ρ	DQ35	VDDQ	DQ34	VddQ	DQ18				DQ33	VddQ	DQ19	VddQ	DQ21
R	Vss	DQ32	Vss	DQ20	Vss				Vss	DQ31	Vss	DQ23	Vss
т	QK1#	VDDQ	DQ30	VddQ	DQ22				DQ29	VddQ	DQ25	VddQ	TDO
U	Vss	DQ28	Vss	DQ24	VDD				VDD	DQ27	Vss	DQ26	Vss
v	QK1	VDD	ZQ	VddQ	Vss				Vss	VddQ	VEXT	VDD	TDI

DQ0–DQ35 DINV0–DINV3 DM A0–A13 CK, CK# DK0–DK1,	: Data inputs / output : Data inversion : Write data mask : Address : Input clock : Input data clock	REF# LBK# MF TMS TDI TCK	: Refresh Enable : Loopback Mode : Mirror Function : IEEE 1149.1 Test input : IEEE 1149.1 Test input
DK0–DK1, DK0#–DK1# QVLD	: Input data clock : Read Data Valid	TDO TRST#	: IEEE 1149.1 Clock input : IEEE 1149.1 Test output : IEEE 1149.1 Test reset input
QK0–QK1, QK0#–QK1# ZQ	: Output data clock : Output impedance & Input	Vref Vdd VddQ	: HSTL input reference input : Power Supply, 1.5 V nominal : DQ Power Supply, 1.0 V nominal
RST#	termination control : Master reset	VEXT Vss	: Power Supply, 2.5 V nominal : Ground
CS# WE#	: Chip select : Write Enable	DNU, Vss DNU, VddQ DPR#	: Must not be used, or must be connected to Vss : Must not be used, or must be connected to VddQ : Overlapped Refresh



Pin Identification

in Identificatio				(1/2
Symbol	Direction	I/О Туре	Freq MHz	Description
CK, CK#	Input	1.0 V HSIO	600 / 525	Clock inputs:
				CK and CK# are differential clock inputs. This input clock pair
				registers address and control inputs on the rising edge of CK. CK# is
				ideally 180 degrees out of phase with CK.
DK0-DK1,	Input	1.0 V HSIO	600 / 525	Differential clocks for DQ inputs:
DK0#-DK1#				DK0, DK0# => DQ0-DQ17, DM
				DK1, DK1# => DQ18-DQ35
				Note that DK1,DK1# are only active on x36 parts.
CS#	Input	1.0 V HSIO	600 / 525	Chip select
				CS# enables the commands when CS# is LOW and disables them
				when CS# is HIGH. When the command is disabled, new commands
				are ignored, but internal operations continue.
WE#, REF#	Input	1.0 V HSIO	600 / 525	Write enable, Refresh enable :
				WE#, REF# are sampled at the positive edge of CK, WE#, and REF#
				define (together with CS#) the command to be executed.
RST#	Input	1.0 V HSIO	-	Master reset
		(no ODT)		Note that this pin has no on-die termination.
DPR#	Input	1.0 V HSIO	600 / 525	Overlapped Refresh
LBK#	Input	1.0 V HSIO	-	Loopback mode for control pin de-skew training
		(no ODT)		Note that this pin has no on-die termination.
MF	Input	1.0 V HSIO	DC	Causes "mirroring" of certain pins as described in
		(no ODT)		2.14 Clam-shell support.
				Note that this pin has no on-die termination.
A0-A13	Input	1.0 V HSIO	600 / 525	Address inputs for DDR Address
			DDR	Address bus, including bank select bits.
				A13 is veserved for future use.
DQ0-DQ35	Input	1.0 V HSIO	600 / 525	Data input/output:
	/Output			The DQ signals form the 36 bit data bus. During READ commands,
				the data is referenced to both edges of QKx. During WRITE
				commands, the data is sampled at both edges of DKx.
DINV0-DINV3	Input	1.0 V HSIO	600 / 525	Data inversion state for DQ inputs:
	/Output			DINV0 => DQ0-DQ8
				DINV1 => DQ9-DQ17
				DINV2 => DQ18-DQ26
				DINV3 => DQ27-DQ35
				Note that DINV2-DINV3 are only active on x36 parts
DM	Input	1.0 V HSIO	600 / 525	Write data mask; disables writing of the corresponding data value.
			5001 020	Clocked by DK0.



				(2/2)
Symbol	Direction	I/O Type	Freq MHz	Description
QK0-QK1,	Output	1.0 V HSIO	600 / 525	Output data clocks:
QK0#-QK1#				Differential clocks for DQ, QVLD outputs:
				QK0/QK0# => x36: DQ0-DQ17, x18: DQ0-DQ8, QVLD
				QK1/QK1# => x36: DQ18-DQ35, x18: DQ9-DQ17
QVLD	Output	1.0 V HSIO	600 / 525	Data valid;
				The QVLD indicates valid output data. QVLD is edge-aligned with QKx and
				QKx#.
TMS, TDI	Input	1.0 V HSIO		JTAG function pins:
				IEEE 1149.1 test inputs: These balls may be left as no connects if the
				JTAG function is not used in the circuit
тск	Input	1.0 V HSIO		JTAG function pin;
				IEEE 1149.1 clock input: This ball must be tied to V_{SS} if the JTAG function
				is not used in the circuit.
TDO	Output	1.0 V HSIO		JTAG function pin;
				IEEE 1149.1 test output: JTAG output.
				This ball may be left as no connect if JTAG function is not used.
TRST#	Input	1.0 V HSIO		JTAG reset;
				IEEE 1149.1 test rest: This ball must be tied to V_{SS} if the JTAG function is
				not used in the circuit.
ZQ	Analog			Output impedance and input termination control
Vref	Ref			VDDQx0.7 V I/O reference voltage.
VEXT	Supply			Power supply; 2.5 V nominal. See Recommended DC Operating
				Conditions for range.
Vdd	Supply			Power supply; 1.5 V nominal. See Recommended DC Operating
				Conditions for range.
VddQ	Supply			DQ power supply;
				Nominally, 1.0 V. Isolated on the device for improved noise immunity.
				See Recommended DC Operating Conditions for range.
Vss	Supply			Ground
DNU, Vss				Do Not Use, Or must not be used, or must be connected to Vss.
DNU, VDDQ				Do Not Use, Or must not be used, or must be connected to $V_{\text{DD}}Q$

HSIO is a single-ended 1.0 V high-side terminated I/O described in 1. Electrical Specifications.



Block Diagram



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1. Electrical Specifications

Absolute Maximum Ratings

Parameter	Symbol	Conditions	Rating	Unit
Supply voltage	VEXT	2.5 V nominal	-0.3 to +2.8	V
Supply voltage	VDD	1.5 V nominal	-0.3 to +1.95.	V
Output supply voltage,	VDDQ	1.0 V nominal	-0.3 to +1.35	V
Input voltage, Input / Output voltage				
Input / Output voltage	VIH / VIL	1.0 V nominal	–0.3 to +1.35	V
Junction temperature	Tj MAX.		105	°C
Storage temperature	Tstg		-55 to +125	°C

Caution Exposing the device to stress above those listed in Absolute Maximum Ratings could cause permanent damage. The device is not meant to be operated under conditions outside the limits described in the operational section of this specification. Exposure to Absolute Maximum Rating conditions for extended periods may affect device reliability.

Recommended DC Operating Conditions

$0^\circ C \leq T \texttt{c} \leq 95^\circ C$

Parameter	Symbol	MIN.	TYP.	MAX.	Unit	Comments	Note
Supply voltage	VEXT	2.3	2.5	2.7	V		1
Supply voltage	Vdd	1.395	1.5	1.605	V		1
Output supply voltage	VDDQ	0.95	1.0	1.05	V		1
Reference voltage	VREF	VDDQ*0.69	VDDQ*0.7	VDDQ*0.71	V		1,2,4
High level output voltage	Voh (DC)	VddQ-0.025			V		1
Low level output voltage	Vol (DC)		VDDQ*0.4		V	Zol=40 Ω,	1
						R∟=60 Ω	
High level input voltage	VIH (DC)	VREF+0.07		VDDQ+0.3	V		1,4
Low level input voltage	VIL (DC)	-0.3		VREF-0.07	V		1,4
Clock input voltage	VIn	-0.3		VDDQ+0.3	V		
Clock differential voltage	VID	0.2		V _{DD} Q+0.6	V		
Output Impedance HIGH	Zoh		60		Ω		3
Output Impedance LOW	Zol		40		Ω		3
Input Impedance LOW	Zi		60		Ω		4

Notes 1. All voltage referenced to Vss (GND)

- 2. Peak to Peak AC noise on VREF must not exceed +/- 2 % VDDQ (DC)
- 3. Programmable via ZQ and Reset Impedance Control
- 4. High-side termination (programmable via ZQ and Reset/MRS)

DC Characteristics

 $0^{\circ}C \leq T_{C} \leq 95^{\circ}C; \ 1.395 \ V \leq V_{DD} \leq 1.605 \ V,$ unless otherwise noted

Parameter	Symbol	Test condition	MIN.	MAX.	Unit	Note
Input leakage current	lu		0	50	μA	1
Output leakage current	Ilo		0	50	μA	1
3 state leakage current	loz		0	50	μA	1

Notes 1. Outputs Driver High-Z and ODT Disabled.

Capacitance (TA = 25 °C, f = 1MHz)

Parameter	Symbol	Test conditions	MIN.	MAX.	Unit
Input capacitance	CIN	V _{IN} = 0 V		1.5	pF
(CK,DK, CS#, WE#, REF#, A, DM)					
I/O, Output, Other capacitance	Cı/o	V1/0 = 0 V		1.8	pF
(DQ, DINV, QVLD)					
Input capacitance delta between	Срск	V _{IN} = 0 V		0.15	pF
differential clock pins					
Input capacitance delta between DQ pins	CDDQ	V1/0 = 0 V		0.2	pF
Input capacitance delta between CS#,	CDCTL	V _{IN} = 0 V		0.2	pF
WE#, REF#, A, DM pins					

Remark These parameters are periodically sampled and not 100% tested. Capacitance is not tested on ZQ pin.

Recommended AC Operating Conditions

 $0^{\circ}C \leq T_{C} \leq 95^{\circ}C; \ 1.395 \ V \leq V_{\text{DD}} \leq 1.605 \ \text{V},$ unless otherwise noted

Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Input HIGH voltage	VIH (AC)		V _{REF} +0.13	VDDQ+0.3	V	1
Input LOW voltage	VIL (AC)		-0.3	VREF-0.13	V	1

Note 1. Overshoot: VIH (AC) \leq VDDQ+0.3 V for t \leq tck/5

Undershoot: VIL (AC) ≥ -0.3 V for $t \leq t c \kappa/5$

Control input signals may not have pulse widths less than t_{CKH} (MIN.) or operate at cycle rates less than t_{CKH} (MIN.).



DC Characteristics

Parameter	Symbol	Test condition		Max	Unit	Note
Total power consumption	PD	Nominal supply voltage Sequential bank access Overlapped refresh mode Data inversion enabled Half address and data transitions 30% Write and 70% Read operation ODT=60Ω, Zoh/Zol = 60Ω/40Ω Tc=95°C	x18 x36	- 2.4	w	1

Note 1. Including all the power supply (VDD, VDDQ and VEXT)



AC Characteristics

Normal bus timing

Standard bus timing waveforms and values for command and data are shown in **Figure 1-1** through **Figure 1-6** and **Interface AC Parameters**. All timing is measured to/from the crossing point on differential clocks and to/from the VREF crossing point on single-ended signals.





Figure 1-2 Data Input Timing Waveforms





Figure 1-3 Data Output Timing Waveforms



Figure 1-4 Output Driver Enable/Disable Timing



Figure 1-5 Reset Timing Waveforms



Note The clock must be with specification and all other chip inputs must be driven to legal values throughout the tRSS period.

Interface AC Parameters

Parameter	Symbol		H16 MHz)		H19 MHz)	Unit	(1/2 Note
		MIN.	MAX.	MIN.	MAX.	_	
CK, DK, QK clock period (maximum 600 MHz	tск	1.587	2.500	1.813	2.500	ns	
clock frequency, minimum 400 MHz)							
CK, DK LOW time (assumes 5% duty cycle	t cĸ∟	0.45*		0.45*		tск(avg)	5
distortion at 600 MHz)							
CK, DK HIGH time (assumed 5% duty cycle	tскн	0.45*		0.45*		tск(avg)	5
distortion at 600 MHz)							
Clock period jitter	tJIT(per)	-0.070	0.070	-0.070	0.070	ns	6,7
Cycle-to-cycle clock jitter	tJIT(cc)		0.140		0.140	ns	
A to CK setup	tas	0.213		0.244		ns	2
CK to A hold	tан	0.213		0.244		ns	1
CK to A setup/hold window	tash	0.213		0.244		ns	4
A input pulse width	tapw	0.266		0.305		ns	
CS#, WE#, REF#, DPR# to CK setup	tcs	0.240		0.275		ns	2
CK to CS#, WE#, REF#, DPR# hold	tсн	0.240		0.275			1
CK to CS#, WE#, REF#, DPR# setup/hold	tcsн	0.400		0.458		ns	4
window							
CS#, WE#, REF#, DPR# input pulse width	t _{CPW}	0.533		0.610		ns	
CK to DK skew	t ckdk	-0.266	0.266	-0.304	0.304	ns	
DQ, DINV, DM to DK setup	tıs	0.213		0.244		ns	2
DK to DQ, DINV, DM hold	tıн	0.213		0.244		ns	1
DK to DQ, DINV, DM setup/hold window	tısн	0.213		0.244		ns	4
DQ, DINV, DM input pulse width	tipw	0.266		0.305		ns	
Output signal rise time [See note]	t _{Rise}	2	5	2	5	V/ns	3
Output signal rise time [See note]	tFall	2	5	2	5	V/ns	3
QK LOW time	tqĸ∟	0.45*		0.45*		tск(avg)	5
(assumes 5% duty cycle distortion at 600 MHz)							
QK HIGH time	tакн	0.45*		0.45*		tск(avg)	5
(assumed 5% duty cycle distortion at 600 MHz)							
CK to QK skew	tскак	-0.400	0.400	-0.458	0.458	ns	
Additional tRL delay in loopback	tcp		5		5	ns	
QK0 to DQ[17:0], DINV[1:0] (x36) or DQ[8:0],	tακαυ		0.133		0.152	ns	8
DINV[0] (x18)							
QK0 to DQ[17:0], DINV[1:0] (x36) or DQ[8:0],	tqнo	0.4*		0.4*		tск(avg)	5
DINV[0] (x18)							
QK1 to DQ[35:18], DINV[3:2] (x36) or	t QKQ1		0.133		0.152	ns	8
DQ[17:9],DINV[1] (x18)							
QK1 to DQ[35:18], DINV[3:2] (x36) or	t _{QH1}	0.4*		0.4*		tск(avg)	5
DQ[17:9],DINV[1] (x18)							



							(2/2)
Parameter	Symbol		H16 MHz)		H19 MHz)	Unit	Note
		MIN.	MAX.	MIN.	MAX.		
Any QK to any DQ, DINV	tακα		0.200		0.288	ns	8
Any QK to any DQ, DINV	tqн	0.35*		0.35*		tск(avg)	5
Any QK to QVLD	tακαν		0.200		0.288	ns	
Any QK to QVLD	tqvн	0.85*		0.85*		tск(avg)	5
QK to DQ output driver turn-on time	tqon		0.133		0.152	ns	
and ODT turn-on time							
QK to DQ output driver turn-off time	t QOFF	-0.1*	0.133	-0.1*	0.152	tск(avg)	5
and ODT turn-on time							
DQ to RST# setup	trds	1000 *		1000 *		tск	2
DQ to RST# hold	t RDH	5 *		5 *		tск	1
RST# pulse length	trss	200		200		μs	
RST# deasserted to CS#, DPR#,	trsн	400000 *		400000 *		tск	
or LBK# asserted							
Time for PLL to stabilize after being	tpll		20		20	μs	
enabled							
MRS command start to next CS#, DPR#,	tmrd	24 *		24 *		tск	
or LBK# assertion							

Notes 1. All input hold timing assumes rising edge slew rate of 2 V/ns measured from VIL/VIH (DC) to VREF.

- 2. All input setup timing assumes falling edge slew rate of 2 V/ns measured from V_{REF} to $V_{IL}/V_{IH}(AC)$
- 3. All output timing assumes the load shown in Figure 1-6.
- **4.** Setup/hold windows, tASH, tCSH, tISH are used for de-skew timing budgeting and are based on electrical simulations. These cannot be directly measured without performing de-skew training.
- 5. tck (avg) is the value of tck averaged over 200 clock cycles.
- 6. Clock phase jitter is the variance from clock rising edge to the next expected clock rising edge.
- 7. Frequency drift is not allowed.
- 8. taka, takax are guaranteed by design.

Figure 1-6 Output Load





Temperature and Thermal Impedance

Temperature Limits

Parameter	Symbol	MIN.	MAX.	Unit	Note
Reliability junction temperature	Тı	0	+105	°C	1
Operating junction temperature	TJ	0	+100	°C	2
Operating case temperature	Tc	0	+95	°C	3

Notes 1. Temperatures greater than 105°C may cause permanent damage to the device. This is a stress rating only and functional operation of the device at or above this is not implied. Exposure to absolute maximum rating conditions for extended periods may affect reliability of the part.

- 2. Junction temperature depends upon cycle time, loading, ambient temperature, and airflow.
- **3.** MAX operating case temperature; Tc is measured in the center of the package. Device functionality is not guaranteed if the device exceeds maximum Tc during operation.

Thermal Impedance

Substrate	bstrate θja (°C/W)				θjc
	Air Flow = 0 m/s	Air Flow = 1 m/s	Air Flow = 2 m/s	(°C/W)	(°C/W)
4 - Layer	17.5	14.0	13.1	8.2	0.8

Impedance Controls

The Low Latency DRAM-III includes programmable impedance control that affects the output impedance of all output and bidirectional pins, as well as the termination of all input and bidirectional pins. The output impedance control affects all output pins and the input termination control affects those pins grouped together as shown in **Table 3-5** and **Table 3-6**. The impedance control affects the drive strengths of both high and low values, as well as the high-side termination impedance. The impedance is controlled via a precision resistor connected between the ZQ and Vss pins. The nominal value of the precision (1%) resistor is 240 Ω , with a supported range of 200 to 240 Ω .

The output impedance accuracy is within 15% of the programmed nominal value, with the linearity measured at the points described in **Impedance Test Parameters**. The actual impedance must remain within the range specified by the linear interpolation of the values at those points.

Normally, the input termination and output impedance are continuously adjusted, such that only minor instantaneous variations in the impedance occur. However, when the FZ bit in the configuration register is set, then the input termination and output impedance will be "frozen" at their current values. If the IM bit in the configuration register is set, then the output impedance will have a nominal value that is not dependent on the value of the resistor connected to the ZQ pin. In this mode, the ZQ pin is ignored by the Low Latency DRAM-III, and the output impedances will not be PVT compensated

Impedance Test Parameters

Parameter	Symbol	MIN.	NORM.	MAX.	Unit
Output HIGH voltage with forced I_oh = -(V_DDQ -0.85 V)/ Z_{oh}	Voh	0.723	0.85	0.978	V
Output HIGH voltage with forced I _{oh} = -(V _{DD} Q -0.7 V) / Z_{oh}	Voh	0.595	0.7	0.805	V
Output LOW voltage with forced I_{ol} = 0.7 V / Z_{ol}	Vol	0.595	0.7	0.805	V
Output LOW voltage with forced I_{ol} = 0.55 V / Z_{ol}	Vol	0.468	0.55	0.633	V



2. Operation

2.1 Interface Overview

The primary Low Latency DRAM-III interface consists of a unidirectional command and address bus and a bidirectional data bus. This type of data bus is often referred to as common I/O or CIO. **Pin Identification** contains the list of pins on the Low Latency DRAM-III.

The command bus is a single data rate (SDR) bus consisting of CS#, WE#, and REF#. The µPD480113xx have a double data rate (DDR) address bus consisting of multiplexed. Both the command bus and address bus are clocked by the differential clock pair CK and CK#. READ and WRITE commands can be issued at a rate of one every cycle in burst length 2 mode.

The data interface is a double-rate (DDR) interface that transfers 36 bits of data ^{Note} on each clock edge for 36-bit parts and 18 bits of data on each clock edge for 18-bit parts. The data interface also includes 4 data inversion pins. In addition to 36 data bits, the inputs to the Low Latency DRAM -III include a DDR data mask, DM, and two differential clock pairs, DK0-DK1 and DK0#-DK1#. The Low Latency DRAM -III outputs two differential clock pairs, QK0-QK1 and QK0#-QK1# that are associated with read data, and a data valid signal, QVLD. Because the data bus is bidirectional, idle cycles are required to implement data bus turn-around as described in **2.7 Data bus turn-around**.

Note For parts with 18-bit data bus, all references in this section should refer to DQ0-DQ17 and DINV0- DINV1 unless otherwise noted. DQ0-DQn is used to represent DQ0-DQ35 or DQ0-DQ17 for 36-bit and 18-bit parts, respectively, and DINV0-DINVm is used to represent either DINV0-DINV3 or DINV0-DINV1 for 36-bit and 18-bit parts.

2.2 Clocking

There are three groups of clock signals: CK/CK#, DK0-DK1/DK0#-DK1#, and QK0-QK1/QK0#-QK1#. The CK/CK# clock is associated with the address and command pins: A0-A13, CS#, WE#, REF#, and DPR#. At the Low Latency DRAM-III pins, the CK/CK# transitions are nominally centered with respect to address and command signal transitions.

The DK0-DK1/DK0#-DK1# clocks are associated with write data. DK0/DK0# is used as a source-centered clock for the double data rate DQ0-DQ17, DINV0-DINV1, and DM pins. DK1/DK1# is used as a source-centered clock for the double data rate DQ18-DQ35 and DINV2-DINV3 pins.

The DK0-DK1/DK0#-DK1# clocks must meet the specified tokdek skew with respect to the CK/CK# clock in order to ensure proper timing relationship between command and data cycles and to enable proper data bus turn-around. The QK0-QK1/QK0#-QK1# clocks are associated with read data. At the Low Latency DRAM-III pins, and for x36 devices, QK0/QK0# must be source-synchronous with the read data DQ0-DQ17, DINV0-DINV1, and QVLD pins. Similarly, for x36 devices, QK1/QK1# is used as a source-synchronous clock for the read data DQ18-DQ35 and DINV2-DINV3 pins. For x18 devices, QK0/QK0# must be source-synchronous with DQ0-DQ8, DINV0, and QVLD. Similarly, for x18 devices, QK1/QK1# must be source-synchronous with DQ9-DQ17 and DINV1.

The QK0-QK1/QK0#-QK1# clocks must meet the specified tckok skew with respect to the CK/CK# clock in order to ensure proper timing relationship between command and data cycles and to enable proper data bus turn-around.



2.3 Address bus

In Burst of 2 mode one clock cycle is required to load an address. In DDR Address mode address inputs are captured by the RAM in two beats per cycle. Three of the bits in the address field select which of the 8 banks in the RAM will be accessed. Note that a 4th bank address bit has been identified anticipating that a 16 bank version of the RAM might someday reach the market. The bit functions as an ordinary address bit in the devices described in this data sheet.

Command	Device Width	Beat	A13 Note1	A12 Note1	A11	A10	A9	A8	A7	A6	A5	A 4	A3	A2	A1	A0
READ or WRITE	x36	0	9.7Gb	2.4Gb	Addres A No.								A Note2	Note2 Bank		
		1	X	4.8Gb	8Gb Address											
READ or WRITE	x18	0	4.8Gb					Addro s	es				A Note2		Bank	
		1	9.7Gb	2.4Gb	Gb Address											
AUTO REFRESH	ANY	0	Х	х	X X X X X X X X X Bar						Bank					

Table 2-1 Address Bit Encoding

Note 1. A13 are reserved for future expansion. For smaller capacity devices, the value is a don't care, but for devices of the designated capacity or larger, the value will be used.

2. Address bit A3 in beat zero is used as an address bit in the current devices, but may be used as a bank address bit in future generations of the part.

2.4 Command encoding

The Low Latency DRAM-III supports five types of command cycles as show in Table 2-3.

The NOP command must be used on any cycle where no other commands are requested.

The READ command is used to initiate a burst read from the Low Latency DRAM -III.

The WRITE command is used to initiate a burst write from the Low Latency DRAM -III.

The AUTO REFRESH command is used to initiate a refresh operation on a particular bank of the memory.

The OVERLAPPED REFRESH command is used to initiate a refresh operation, but may be overlapped with other commands.

The MRS command is used to configure the Low Latency DRAM -III following a reset.

Because the overlapped refresh feature can be enabled or disabled, the command decoding and it's associated state diagram have two different forms. **Table 2-2** and **Figure 2-1** show the command encoding for the case when the overlapped refresh feature is disabled. **Figure 2-1** only shows the state diagram for the command decoder; there may be many other states internal to the device.

Because READ and WRITE commands only require a single cycle. When overlapped refresh is enabled, the modified command encoding is described in **Table 2-3** and the state diagram in **Figure 2-2**. In burst length 2 mode, overlapped refresh is accomplished using the DPR# pin, and is independent of other command encodings.



State	Command	CS#	WE#	REF#	LBK#	RST#	Length (cycles)	Description
IDLE	NOP	1	-	-	1	1	1	No operation
	READ	0	1	1	1	1	1	Read
	WRITE	0	0	1	1	1	1	Write
	AUTO REFRESH	0	1	0	1	1	1	Auto refresh
	MRS	0	0	0	1	1	2	Mode register set
MRS	NOP	1	-	-	1	1	-	No operation
	LOOPBACK	-	-	-	0	1	-	Address/control loopback
	RESET	-	-	-	-	0	-	Reset

Table 2-2 Command Encoding (Overlapped Refresh Disabled)

Figure 2-1 Command Decode State Diagram (Overlapped Refresh Disabled)





State	Command	CS#	WE#	REF#	DPR#	LBK#	RST#	Length (cycles)	Description
IDLE	NOP	1	-	-	1	1	1	1	No operation
	READ	0	1	1	1	1	1	1	Read
	WRITE	0	0	1	1	1	1	1	Write
	OVERLAPPED REFRESH	1	-	-	0	1	1	1	Overlapped refresh
	READ + REFRESH	0	1	1	0	1	1	1	Read and overlapped refresh
	WRITE + REFRESH	0	0	1	0	1	1	1	Write and overlapped refresh
	MRS	0	0	0	1	1	1	2	Mode register set
OR1,	NOP Note	1	-	-	- Note	1	1	-	No operation
OR2,	READ Note	0	1	1	- Note	1	1	1	Read (overlapped)
OR3	WRITE Note	0	0	1	- Note	1	1	1	Write (overlapped)
MRS	NOP	1	-	-	1	1	1	-	No operation
	LOOPBACK	-	-	-	-	0	1	-	Address/control loopback
	RESET	-	-	-	-	-	0	-	Reset

Table 2-3 Command Encoding (Overlapped Refresh Enabled)

Note These indicate that the value of the DPR# pin is ignored for the purpose of command decoding (the pin is used for conveying bank number during this cycle).

Figure 2-2 Command Decode State Diagram (Overlapped Refresh Enabled)





2.5 Data mask

The DM pin is used during write cycles to indicate that the corresponding data are to be masked off from writing to the memory. The value of the DM pin is sampled on both edges of DK0/DK0#, and corresponds to the data sampled on that same edge of either DK0/DK0# or DK1/DK1#.

When the DM pin has a logic 0 value, the data will be written to the memory. When the DM pin has a logic 1 value, the data will be ignored and no update to the corresponding memory data will be made.

The value of the DM pin is ignored in all other circumstances. Thus, driving a constant logic 0 to the DM pin can be used in systems where all data cycles in a burst result in writes to the memory.

When the Auto-DM Function (ADM) bit is set via an MRS command, the value of the DM pin is ignored and the device will behave as if it were a logic 0 on only the clock edge specified by the value of the DMP field.

2.6 Data inversion

In order to reduce simultaneous switch noise, I/O current and average I/O power, the Low Latency DRAM-III provides the ability to invert all data pins. Because the nominal design for Low Latency DRAM-III I/O signals is high-side termination to V_{DD}Q, signals driven to a logic high state will consume less power than those in a logic low state.

The DINVm-DINV0 pins indicate whether corresponding DQn-DQ0 pins, in groups of nine, represent the true logic value or an inverted logic value. With the ability to invert DQn-DQ0 pins in groups of nine, each group is guaranteed to be driving no more than five pins low on any given cycle. As a result, no more than five pins in each group can switch in the same direction during each bit time, reducing simultaneous switching noise effects.

When data inversion is enabled (DI=1) and a DINV pin is a logic 1, the corresponding DQ pins have been inverted. When data inversion is disabled (DI=0) then the Atris device will ignore the value of the DINV pins for write data and will drive DINV pins to logic 1 for read data; however, in such a case, the DQ pins are not inverted, regardless of the values of the DINV pins.

2.7 Data bus turn-around

Because the DQn-DQ0 and DINVm-DINV0 pins are bidirectional, care must be taken to ensure that the Low Latency DRAM-III and the system chip connected to it do not drive these pins simultaneously. In order to guarantee this, a rapid DQ and DINV turn-off time is required. The actual bus turn-around time is dependent on many system parameters, including BL, RL, WL, pre-amble, post-amble, controller I/O timing, DRAM I/O timing, PCB delays; and the system must ensure that neither the Low Latency DRAM-III, nor the system chip is driving the bus during that time. This is accomplished by inserting a sufficient number of NOP commands between each READ-to-WRITE command transition and between each WRITE-to-READ transition. Additional NOP commands may be required to allow settling of the bus in order to insure no adverse effect on I/O timing.

Note that a REFRESH command can be used in place of a NOP command to effect the insertion of an idle cycle on the data bus.

2.8 Command cycles

A command to the Low Latency DRAM-III is initiated by driving the CS# pin low at the rising edge of CK and simultaneously driving the WE#, REF#, and A0-A13 to the appropriate state corresponding to the command. READ and WRITE commands require only a single cycle. The NOP and AUTO REFRESH commands are each a single cycle in length. The OVERLAPPED REFRESH command is four cycles in length.

2.9 Write data cycles

DQ0-DQn, DINV0-DINVm, and DM associated with a WRITE command are received by the Low Latency DRAM-III in a DDR burst, starting on a rising edge of DK that is offset WL clock cycles from rising edge of the CK signal corresponding

to the first cycle that the WRITE command was initiated. This delay of WL cycles is intended to ensure that the write data are not driven at the same time that data from a previous READ command are being driven from the Low Latency DRAM-III.

2.10 Read data cycles

DQ0-DQn and DINV0-DINVm associated with a READ command are driven by the Low Latency DRAM-III in a DDR burst by the Low Latency DRAM-III RL clock cycles from the rising edge of the CK signal corresponding to the first cycle that the READ command was initiated. This delay of RL cycles is equal to the delay required for the internal logic and memory within the Low Latency DRAM-III to read data and make it available on the bus.

2.11 READ and WRITE command protocol

Figure 2-3 through Figure 2-4 show the READ and WRITE command sequences.



Figure 2-3 READ Command Sequence





Figure 2-4 WRITE Command Sequence

Figure 2-5 shows an example of a sequence of 2 READ, 2 WRITE, and 2 READ commands, with the following configuration:

Burst length 2

Speed Config 3: tRC=8, tRL=12, tWL=13

Data inversion is enabled

Some observations of this sequence are:

A NOP is inserted between READ2 and WRITE1 to allow for read-to-write bus turn-around

3 NOPs are inserted between WRITE2 and READ3 to allow for write-to-read bus turn-around

Because tRc is 8, READ3 may access the same bank as READ1; however, none of READ2, WRITE1, or WRITE2 may access the same bank as READ1





Figure 2-5 READ/WRITE/READ Command Sequence (Speed Config 3, trc=8, trL=12, twL=13, DI=1)

2.12 Automatic Refresh

Refresh cycles of the internal memory are initiated via the AUTO REFRESH command. Each refresh command causes a single refresh cycle to occur on the bank specified in the command.

The minimum data retention time is 2 ms over the temperature range specified in **1**. Electrical Specifications, and there are 16384 words in each bank. Therefore, each bank must receive 16384 refresh commands every 2 ms in order to ensure proper data retention.

This means that a refresh command must be received at an average rate of one every 13.3 ns. At a frequency of 600 MHz, this represents 8.2% of the usable bandwidth to the Low Latency DRAM-III.

The refresh command waveforms are shown in Figure 2-6.







2.13 Overlapped Refresh pin (DPR#)

In burst length 2 mode, overlapped refresh is accomplished using the DPR# pin, instead of the REF# pin. This extra pin is required, in order to allow READ and WRITE commands to completely consume the standard command bus (pins CS#, WE#, and REF#).

The DPR# pin uses the same protocol as the REF# pin does for burst length 4 mode.

The DPR# pin can only be used in burst length 2 mode and when the MRS register specifies overlapped refresh mode. In all other cases, the DPR# pin must be connected to $V_{DD}Q$.

An example of Overlapped Refresh overlapped with five READ commands can be seen in Figure 2-7.



Figure 2-7 Overlapped Refresh with READ Commands

Note tRc for overlapped refresh begins 3 cycles after the start of the command.



2.14 Clam-shell support

In order to support clam-shell mounting on the board, the pin-out of the Low Latency DRAM-III is designed so that most pins that will be interchanged via mirroring are functionally interchangeable. A few pins require their function to be swapped, however. The MF pin causes the pin assignment to be modified to effect those pin swaps. In order to support proper signal integrity, package trace lengths for mirrored pins are matched to within +/- 0.1mm. **Table2-4** shows the pins that are mirrored by the MF pin.

Due to the mirroring of A5 and A6, all devices are required to include an I/O cell on A6 in order to support the mirror function. For lower capacity devices, the A6 pin would remain DNU, VDDQ from a memory controller standpoint, so the system need not actively drive this functionally unused pin.

Similarly, for devices that require A12, pin A13 is also required to include an I/O cell.

A3 and A4 are mirrored to support future devices that may use A3 as an additional bank select bit.

All of A0 through A5 must be mirrored in order for MRS commands to be interpreted correctly.

Pin	MF=0	MF=1
N4	A0	A1
N10	A1	A0
L4	A4	A3
L10	A3	A4
M5	A2	REF#
M9	REF#	A2
L2	CS#	WE#
L12	WE#	CS#
H10	A5	A6
H4	A6	A5
F4	A12	A13
F10	A13	A12
M11	A9	DPR#
M3	DPR#	A9
H2	RST#	DM
H12	DM	RST#

Table 2-4 Pins Affected by MF Pin



3. Initialization

Prior to functional use, the Low Latency DRAM-III must be initialized and configured. The steps described in this chapter will ensure that the internal logic of the Low Latency DRAM-III has been properly reset and that the functional timing parameters of the chip have been configured.

Figure 3-1 shows the overall initialization sequence required by the Low Latency DRAM-III. An example of an initialization sequence, showing each of the different phases, is shown in **Figure 3-7**.







3.1 Power-on

- 1 There are no restrictions on the sequence of applying the V_{DD}, V_{EXT}, and V_{DD}Q power supplies as long as maintaining RST# = Low and fix MF input state before and during any of the power supplies ramp up.
- 2 In case when RST# can not be asserted Low and fixed MF input state before and during any of the power supplies ramp up, make sure to ramp up following manners.
 - (1) Ramp up VDD and VEXT
 - (no ramp up sequence restriction on VDD and VEXT)
 - (2) Ramp up VDDQ and VREF following to the VDD and VEXT
 - (no ramp up sequence restriction on VDDQ and VREF)
 - (3) Make sure to assert RST# for tRSS specified period before initiating any operations



Figure 3-2 Power-on 1





3.2 Reset

Before the Low Latency DRAM-III can be configured, it must be properly reset, to ensure that the Low Latency DRAM-III is in a known and functional state, regardless of any power on anomalies and power supply ramp rates. Reset requires the assertion of the RST# pin for at least 200 μ s, during which time the device logic is reset to a known state, and the I/O impedances are programmed as described in **3.3 Initial impedance settings**.

The contents of memory are not guaranteed to be retained when the chip is reset.

3.3 Initial impedance settings

When the RST# pin is deasserted, the value of specific DQ pins is used to configure the input termination and output impedance as shown in **Table 3-1**. The meaning of the bits is the same as the corresponding bits described in **3.5 Configuration**, although their positions in the DQ pins are unrelated to their positions in the corresponding MRS registers.

While RST# is asserted, the impedance values are continuously updated based on the values of the DQ pins and on the sampled value of the ZQ impedance programming pin. On the rising edge of RST#, the values will be latched into the device and stored in the appropriate positions in the MRS registers. **Figure 3-4** shows the waveforms associated with this function.

Note that the DINV pins are ignored during reset time and the values of the DQ pins are considered to never be inverted during this time.

						-		-	-		-		-				-	
Pins	U12	T11	U4	R12	T5	P13	R4	P11	P5	B12	C11	B4	D12	C5	E13	D4	E11	E5
DQ (x18)	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
DQ (x36)	26	25	24	23	22	21	20	19	18	8	7	6	5	4	3	2	1	0
Reset Reg	17	16	15	14	13	12	11	10	9	8	7	6	5	4	3	2	1	0
$MSB\toLSB$																		
Cell ID			K	D	К	U	С	D	С	U	D	D	D	U	ODT	QD	QU	IM





Figure 3-4 Reset Sequence

3.4 Per-bit de-skew training sequence

The Low Latency DRAM-III device provides support that allows a memory controller to de-skew signals for high-speed operation. If per-bit de-skew is desired, the memory controller must provide the actual de-skew functionality. Per-pin de-skew is normally performed after deassertion of the RST# pin, but may be performed at any time, without affecting the contents of the memory within the device. Furthermore, resetting the device does not necessarily affect the skew characteristics of the device pins; therefore, a reset of the device does not necessarily require repeating the de-skew training sequence that was previously performed.

Per-pin de-skew training is anticipated to be implemented in three steps:

De-skew of the control, address, and clock pins: A0-A13, CS#, WE#, REF#, DPR#, and DK0-DK1 with respect to CK De-skew of the read path: DQ0-DQn, DINV0- DINVm and QVLD with respect to QK0-QK1

De-skew of the write path: DQ0-DQn, DINV0- DINVm and DM with respect to DK0-DK1

The first phase, de-skew of the control and related pins, requires support from the device by driving the LBK# pin to a logic 0. When the LBK# pin is a logic 0, the DK0-DK1, CS#, WE#, REF#, DPR#, and A0-A13 pins are looped back to the various DQ and DINV pins. All pins are sampled by the CK clock, including DK0-DK1. The specific mapping of input pins to output pins during loopback mode is shown in **Table 3-2**.

			x36 part				
		Input		Output	Inj	Input	
Input Pin	Output Pin	MF=0	MF=1		MF=0	MF=1	
G5	C5	A8	A8	DQ4	A8	A8	DQ4
H4	E5	A6	A5	DQ0	A6	A5	DQ0
J1, K1	E11	DK0, DK0#	DK0, DK0#	DQ1	DK0, DK0#	DK0, DK0#	DQ1
L4	P5	A4	A3	DQ18	A4	A3	DQ9
L2	R4	CS#	WE#	DQ20	CS#	WE#	DQ11
M5	T5	A2	REF#	DQ22	A2	REF#	DQ13
N4	U4	A0	A1	DQ24	A0	A1	DQ15
H10	E13	A5	A6	DQ3	A5	A6	DQ3
J11, K11	E3	DK1, DK1#	DK1, DK1#	DQ14	DNU, Vss	DNU, Vss	DNU, Vss
L10	P11	A3	A4	DQ19	A3	A4	DQ10
L12	P13	WE#	CS#	DQ21	WE#	CS#	DQ12
M9	T11	REF#	A2	DQ25	REF#	A2	DQ16



			x36 part			x18 part			
		In	Input		In	put	Output		
Input Pin	Output Pin	MF=0	MF=1		MF=0	MF=1			
N10	U12	A1	A0	DQ26	A1	A0	DQ17		
F4	B4	A12	A13	DQ6	A12	A13	DQ6		
G3	D4	A10	A10	DQ2	A10	A10	DQ2		
M3	N2	DPR#	A9	DINV2	DPR#	A9	DINV1		
F10	B12	A13	A12	DQ8	A13	A12	DQ8		
G9	C11	A7	A7	DQ7	A7	A7	DQ7		
G11	D12	A11	A11	DQ5	A11	A11	DQ5		
M11	R12	A9	DPR#	DQ23	A9	DPR#	DQ14		

Notes:

DK0, DK0# and DK1, DK1# have differential receivers, so the output of the receiver is used as signal being looped back.

For each pin that is looped back, the input pin is sampled on both the rising edge and falling edges of the corresponding input clock. The value output on the rising edge of the corresponding output clock will be the value that was sampled on the rising edge of the input clock. The value output on the falling edge of the corresponding output clock will be the *inverted* value that was sampled on the falling edge of the input clock. Data inversion is *not used* during loopback mode and the value of the corresponding MRS bits are ignored.

The delay from input pins to DQ/DINV output pins, when loopback mode is enabled, is 5 cycles of CK.

In burst length 4 mode, while LBK# pin is a logic 0, the DQ/DINV output pins corresponding to unused input pins (see **Table 3-2**) will still meet the AC timing specification (See **Interface AC Parameters**), but they may have arbitrary logic state and may or may not have a constant logic value.

While the LBK# pin is a logic 0, the device will ignore any apparent commands being presented on the CS#, WE#, REF#, and DPR# pins and will perform self refresh operations to prevent loss of data during extended periods of de-skew training. In order to ensure that commands are not inadvertently received during entry and exit from de-skew training, no commands may be sent to the device for a period of 16 cycles of CK prior to the LBK# pin being driven to 0, nor for 16 cycles after the LBK# pin is returned to a logic 1. Also, no valid input signals for looped back must be inserted for 16 cycles after LBK# pin being driven to 0 when device just enters into loopback mode.

While LBK# is asserted, the on-chip PLL is ignored; therefore the phase of the QK output clock will be undefined with respect to the CK input clock. However, the QK to DQ timing will still meet its specifications. Note that the PLL will remain enabled or disabled, per its state prior to LBK# being asserted, and after LBK# is deasserted. An example of a loopback sequence is show in **Figure 3-5**.

Figure 3-5 Loopback De-skew Training Sequence





Following de-skew of the control, address, and DK pins, the memory controller drives the LBK# pin to a logic 1, then will likely perform read data and write data de-skew training.

Prior to beginning read path de-skew, the memory controller should issue an MRS command to the Low Latency DRAM-III device to enable the PLL. After the PLL is enabled, the timing of the QK, DQ outputs, and QVLD will be unspecified for a period of tPLL. The MRS command is described in **3.5 Configuration**. While the PLL is stabilizing, refresh commands must continue to be issued to the device in order to ensure proper retention of data.

Read data de-skew requires that a training pattern be written to Low Latency DRAM-III memory. Complex data patterns can be written to the Low Latency DRAM-III memory using the non-de-skewed DQ signals and the auto-DM functions. Auto-DM functions are enabled through MRS, as described in **3.5 Configuration**.

Read data de-skew can be performed by using the de-skewed control and related pins, including the DK pin, and nonswitching DQ and DINV pins to write 36-bit values to memory locations in the device. Because the data are held at constant values, the auto-DM modes of the device are used to specify which 36-bit beat (or 18-bit beat for x18 devices) of data on the bus is to be written to the memory. Once written to the memory, these data can then be read out using standard BL2, READ commands to effect any desired pattern on the DQ bus. This permits the system to de-skew the DQ, DINV, and QVLD signals with respect to the QK clocks.

Write data de-skew can be performed by issuing WRITE commands to the device, then using the de-skewed read bus to determine whether or not data were correctly received by the device. This permits the system to de-skew the DQ, DINV, and DM signals with respect to the DK clocks.



3.5 Configuration

The MRS command is used to configure the Low Latency DRAM-III. Configuration is used to specify the following parameters:

Address Mode

The Address Mode of Low Latency DRAM-III devices from some vendors may be electrically configurable. The default Address Mode may vary from vendor-to-vendor as well. Therefore, systems must configure the desired Address Mode prior to issuing READ or WRITE commands

Mode select for READ latency, WRITE latency, and tRc The version of the Low Latency DRAM-III covered by this specification only supports modes 0 through 5

Select driver pull-up impedance of 1/4 or 1/6 of reference resistor (default 1/4) Select driver pull-down impedance of 1/4 or 1/6 reference resistor (default 1/6) Select DQ, DINV, DM, DK, DKN, CK, CK#, RST# terminator pull-up impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 1/4) Select DQ, DINV, DM, DK, DKN, CK, CK#, RST# terminator pull-down impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 0FF) Select CS#, WE#, REF#, A, LBK#, terminator pull-up impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 1/4) Select CS#, WE#, REF#, A, LBK#, terminator pull-down impedance of OFF, 1x, 1/2, or 1/4 of reference resistor (default 0FF)

Enable/freeze dynamic PVT compensation (default enabled) Enable/disable PLL (default disabled) Enable/disable read/write data inversion (default disabled)

Refresh mode (default AUTO REFRESH)

Select auto-DM modes (default disabled)



The MRS command allows the configuration of one of eight different registers to be configured using SDR timing on the address pins. The MRS command waveforms are shown in **Figure 3-5** and the values of the MRS registers are described in **Table 3-3** through **Table 3-6**. Because an MRS command may change values that significantly affect the behavior of the device, no other commands should be issued to the device for a period of t_{MRD} clock cycles following the MRS command.

MRS commands use the address pins specified in **Table 3-3**. The actual pins used depend on the value of the mirror function (MF) pin.

				Beat 0							Beat 2			
Pins (MF=0)	H4	H10	L4	L10	M5	N10	N4	H4	H10	L4	L10	M5	N10	N4
Pins (MF=1)	H10	H4	L10	L4	M9	N4	N10	H10	H4	L10	L4	M9	N4	N10
Address	A6	A5	A4	A3	A2	A1	A0	A6	A5	A4	A3	A2	A1	A0
Mode Register LSB→MSB					М	ode Reg	g #							
Active	Х	Co	onfigura	tion	0	0	0	Х	0	DI	FZ	RM	BL	
Active	Х	DMF	2	ADM	0	0	1	Х	0	0	0	0	0	PLL
Active	Х	KU	_	ODT	0	1	0	Х	CD		CU		KD	
Active	Х	QD	QU	IM	0	1	1	Х	DD	_	DU		0	0
Reserved	Х	0	0	0	1	0	0	Х	0	0	0	0	0	0
	Х	0	0	0	1	0	1	Х	0	0	0	0	0	0
	Х	0	0	0	1	1	0	Х	0	0	0	0	0	0
	Х	0	0	0	1	1	1	Х	0	0	0	0	0	0

Table 3-3 MRS Configuration Register Bit Assignments



Speed Configuration					Clock Cycle	S	
Config #	Configuration			DI	tRC	tRL	tWL
0	0	0	0	0		Not supported	
1	0	0	1	0		Not supported	
2	0	1	0	0	7	9	10
3	0	1	1	0	8	10	11
4	1	0	0	0		Not supported	
5	1	0	1	0	Not supported		
6	1	1	0	0	Not supported		
7	1	1	1	0	Not supported		
0	0	0	0	1	Not supported		
1	0	0	1	1		Not supported	1
2	0	1	0	1	7	11	12
3	0	1	1	1	8	12	13
4	1	0	0	1	Not supported		
5	1	0	1	1	Not supported		
6	1	1	0	1	Not supported		
7	1	1	1	1		Not supported	

Table 3-4 MRS Configuration Bit Definitions

Refresh Mode

RM	Operation					
0	Auto Refresh Mode (Default)					
1	Overlapped Refresh/DPR# Mode					

Auto-DM Function

ADM Operation				
0	Off / Normal Operation (Default)			
1	On			

Note: When ADM=0, DMP=Don't Care

Auto-DM Pattern

Burst of 2	DMP						
Pattern #	00	10	01	11			
Write Beat 0	1	0	0	0			
Write Beat 1	0	1	0	0			

Note: Data to be written =1, Date to be masked =0

PLL Control

PLL	Operation					
0	PLL Off / Reset (Default)					
1	PLL On					

Burst Length

E	3L	Operation
0	0	Not supported
0	1	2
1	0	Reserved
1	1	Reserved

Note: Setting 00,10 & 11 not supported

Impedance Freeze

FZ	Operation					
0	Update active (Default)					
1	Update frozen					

Data Inversion

DI	Operation						
0	Disable data inversion (Default)						
1	Enable data inversion						



Pins	Data I	nput O	n-Die	Termir	nation	- Up		Data Ir	nput Oi	n-Die T	ermina	tion - I	Down	
DQ, DIV,	ODT	DU		Div	200	240	RQ Ohms	ODT	DD		Div	200	240	RQ Ohms
DM,	1	Х	Х	Off	~	~		1	Х	Х	Off	8	~	
DK1, DK1#	Х	1	1	Off	8	8		Х	1	1	Off	8	8	
DK0, DK0#	0	1	0	1	200	240		0	1	0	1	8	∞	(Not Supported)
	0	0	1	1/2	100	120		0	0	1	1/2	∞	~	(Not Supported)
	0	0	0	1/4	50	60		0	0	0	1/4	∞	∞	(Not Supported)
														-
Pins	Control Input On-Die Termination - Up Control Input On-Die Termination - Down									'n				
Address,	ODT	CU		Div	200	240	RQ Ohms	ODT	CD		Div	200	240	RQ Ohms
CS#, WE#,	1	Х	Х	Off	∞	∞		1	Х	Х	Off	~	∞	
REF#	Х	1	1	Off	∞	∞		Х	1	1	Off	∞	∞	
	0	1	0	1	200	240		0	1	0	1	8	~	(Not Supported)
	0	0	1	1/2	100	120		0	0	1	1/2	8	∞	(Not Supported)
	0	0	0	1/4	50	60		0	0	0	1/4	8	~	(Not Supported)
Pins	Clock Input On-Die Termination - Up							Clock Input On-Die Termination - Down						
	ODT	KU		Div	200	240	RQ Ohms	ODT	KD		Div	200	240	RQ Ohms
	1	Х	Х	Off	∞	∞		1	Х	Х	Off	∞	∞	
CK, CK#	Х	1	1	Off	∞	∞		Х	1	1	Off	∞	∞	
	0	1	0	1	200	240		0	1	0	1	∞	∞	(Not Supported)
	0	0	1	1/2	100	120		0	0	1	1/2	∞	∞	(Not Supported)
	0	0	0	1/4	50	60		0	0	0	1/4	~	~	(Not Supported)
	Note	. Inte	rnal R	Q is s	et to 2	40 Ω w	hen IM=1.							

Table 3-5 MRS Input Termination Bit Definitions

Table 3-6 MRS Output Impedance Bit Definitions

Pins					
DQ, DINV,					
QK0, QK0#					
QK1, QK1#					
QVLD					

0	0	1/6	NA	40.0				
Note: $60^* = 60 \Omega + - 20\%$ no PVT compensation								

200

50

60*

240

60

60*

Output Driver Impedance - Up

Div

1/4

Х

QU

1

Х

IM

0

1

	_					
RQ Ohms	IM	QD	Div	200	240	RQ Ohms
	0	1	1/4	50	60	
	1	Х	Х	40*	40*	
	0	0	1/6	33.3	40.0	

Note: $40^* = 40 \Omega + - 20\%$ no PVT compensation

Figure 3-6 MRS Sequence








4.JTAG Specification

These products support a limited set of JTAG functions as in IEEE standard 1149.1.

4.1 Test Pins

Pin Name	Test Use	Functional Mode Value	Pull Up/Down in Chip	Pull Up/Down on Board
TDI	JTAG serial chain input	1	1	1
TDO	JTAG serial chain output	1		
TCK	JTAG clock	0	0	0
TMS	JTAG mode	1	1	1
TRST#	JTAG reset	0	0	0

Table 4-1 Low Latency DRAM-III Test Pins

Table 4-2. Test Access Port (TAP) Pins

Pin Name	Pin Assignment	Description
ТСК	3A	Test Clock Input. All input are captured on the rising edge of TCK and all
		outputs propagate from the falling edge of TCK.
TMS	1A	Test Mode Select. This is the command input for the TAP controller state machine.
TDI	13V	Test Data Input. This is the input side of the serial registers placed between TDI
		and TDO. The register placed between TDI and TDO is determined by the state
		of the TAP controller state machine and the instruction that is currently loaded in
		the TAP instruction.
TDO	13T	Test Data Output. This is the output side of the serial registers placed between
		TDI and TDO. Output changes in response to the falling edge of TCK.
TRST#	1L	Test Reset Input. This is active-low input asynchronously reset the TAP controller.
		This pin should be driven low or tied to GND when JTAG function is not in use.

Table 4-3. JTAG DC Characteristics (0°C ≤ Tc ≤ 95°C, 1.395 V ≤ VDD ≤ 1.605 V, unless otherwise noted)

Parameter	Symbol	Conditions	MIN.	MAX.	Unit
JTAG Input leakage current	lu	$0~V \leq V_{\text{IN}} \leq V_{\text{DD}}$	-100	+100	μA
JTAG I/O leakage current	Ilo	$0 \ V \leq V_{\text{IN}} \leq V_{\text{DD}} Q \ ,$	-100	+100	μA
		Outputs disabled			
JTAG input HIGH voltage	VIH (DC)		V _{REF} + 0.15	V _{DD} + 0.3	V
JTAG input LOW voltage	VIL (DC)		VssQ - 0.3	V _{REF} - 0.15	V
JTAG output HIGH voltage	VOH (DC)	Іонс = 100 μА	VDDQ-0.25		V
JTAG output LOW voltage	Vol (DC)	Ιοις = 100 μΑ		0.25	V

Note 1. All voltages referenced to Vss (GND).

Table 4-4. JTAG Capacitance

Parameter	Symbol	Teat Conditions	MIN.	MAX.	Unit
JTAG Input Capacitance	Cinj	V _{IN} = 0V		2.5	pF
(TCK, TMS and TRST#)					
JTAG I/O leakage current	Сој	V _{IO} = 0V		2.5	pF
(TDO)					

Note 1. Including package.



4.2 JTAG AC Test Conditions

Input waveform (Rise / Fall time ≤ 0.3 ns)



Output waveform



Output load condition





Parameter	Symbol	Conditions	MIN.	MAX.	Unit	Note
Clock						
Clock cycle time	t тнтн		20		ns	
Clock frequency	ftf			50	MHz	
Clock HIGH time	tтнт∟		10		ns	
Clock LOW time	tтьтн		10		ns	
Output time	1					
TCK LOW to TDO unknown	t⊤∟ox		0		ns	
TCK LOW to TDO valid	t tlov			10	ns	
Setup time	1					
TMS setup time	tмvтн		5		ns	
TDI valid to TCK HIGH	t dvth		5		ns	
TRST High Level to	t RVTH		50		ns	
valid riseing edge of TCK						
Capture setup time	tcsJ		5		ns	1
Hold time	1					
TMS hold time	tтнмх		5		ns	
TCK HIGH to TDI invalid	t thdx		5		ns	
Capture hold time	tснл		5		ns	1
Reset	1					
TRST LOW pulse width	t TRSTW		100		ns	

Table 4-5. JTAG AC Characteristics ($0^{\circ}C \le T_{c} \le 95^{\circ}C$)

Note 1. tcsJ and tcHJ refer to the setup and hold time requirements of latching data from the boundary scan register.

JTAG Timing Diagram





Register name	Description
Instruction register	The 8 bit instruction registers hold the instructions that are executed by the TAP controller. The register can be loaded when it is placed between the TDI and TDO pins. The instruction register is automatically preloaded with the IDCODE instruction at power-up whenever the controller is placed in test-logic-reset state.
Bypass register	The bypass register is a single bit register that can be placed between TDI and TDO. It allows serial test data to be passed through the RAMs TAP to another device in the scan chain with as little delay as possible. The bypass register is set LOW (V_{SS}) when the bypass instruction is executed.
ID register	The ID Register is a 32 bit register that is loaded with a device and vendor specific 32 bit code when the controller is put in capture-DR state with the IDCODE command loaded in the instruction register. The register is then placed between the TDI and TDO pins when the controller is moved into shift-DR state.
Boundary register	The boundary register, under the control of the TAP controller, is loaded with the contents of the RAMs I/O ring when the controller is in capture-DR state and then is placed between the TDI and TDO pins when the controller is moved to shift-DR state. Several TAP instructions can be used to activate the boundary register. The Scan Exit Order tables describe which device bump connects to each boundary register location. The first column defines the bit's position in the boundary register. The second column is the name of the input or I/O at the bump and the third column is the bump number.

Table 4-6. Scan Register Definition (1)

Table 4-7. Scan Register Definition (2)

Register name	Bit size	Unit
Instruction register	8	bit
Bypass register	1	bit
ID register	32	bit
Boundary register	113	bit

Table 4-8. ID Register Definition

Part number	Organization	ID [31:28] vendor revision no.	ID [27:12] part no.	ID [11:1] vendor ID no.	ID [0] fix bit
μPD48011318	64M x 18	0000	0000 0000 1101 1010	00000010000	1
μPD48011336	32M x 36	0000	0000 0000 1101 1011	00000010000	1



4.3 Boundary Scan

IEEE 1149.1 compliant boundary scan will be implemented on all component I/O pins. High speed I/Os that are differential or AC coupled will be fully compliant to the IEEE 1149.6 standard.

The actual boundary scan order is defined in an underscore (e.g. "CS#_WE#").

Table 4-9 and Table 4-10.

Pins labeled DNU0 are DNU, Vss. Pins labeled DNU1 are DNU, VbDQ. Pins that are subject to the mirror function are named with the obverse name and reverse name, separated by an underscore (e.g. "CS#_WE#").



Table 4-9 Boundary Scan Order, X18									
X18	Cell #	Cell	Pin	TYPE	Safe	Cntrl	Cntrl	Cntrl	
Лю		001	Name		Bit	Cell	Value	State	
J3	0	BC_1	DK0	input	Х				
K3	1	BC_1	DK0#	input	Х				
L2	2	BC_1	CS#_WE#	input	Х				
L4	3	BC_1	A4_A3	input	Х				
M5	4	BC_1	A2_REF#	input	Х				
M3	5	BC_1	DPR#_A9	input	Х				
	6	BC_2	*	control	0				
N2	7	BC_7	DINV2	bidir	Х	6	0	Z	
N4	8	BC_1	A0_A1	input	Х				
	9	BC_2	*	control	0				
P5	10	BC_7	DQ9	bidir	Х	9	0	Z	
	11	BC_2	*	internal	0				
P3	12	BC_7	DNU0(1)	internal	Х				
	13	BC_2	*	internal	0				
P1	14	BC_7	DNU0(2)	internal	Х				
	15	BC_2	*	internal	0				
R2	16	BC_7	DNU0(3)	internal	Х				
	17	BC_2	*	control	0				
R4	18	BC_7	DQ11	bidir	Х	17	0	Z	
	19	BC_2	*	control	0				
T5	20	BC_7	DQ13	bidir	Х	19	0	Z	
	21	BC_2	*	internal	0				
Т3	22	BC 7	DNU0(4)	internal	Х				
T1	23	BC 1	QK1#	output2	Х				
	24	BC_2	*	internal	0				
U2	25	BC 7	DNU0(5)	internal	Х				
	26	BC 2	*	control	0				
U4	27	 BC_7	DQ15	bidir	Х	26	0	Z	
V1	28	BC_1	QK1	output2	Х				
	29	 BC_2	*	control	0				
U12	30	BC_7	DQ17	bidir	Х	29	0	Z	
	31	BC 2	*	internal	0				
U10	32	BC_7	DNU0(6)	internal	Х				
	33	BC_2	*	internal	0				
Т9	34	BC_7	DNU0(7)	internal	Х				
	35	 BC_2	*	control	0				
T11	36	BC 7	DQ16	bidir	Х	35	0	Z	
	37	BC_2	*	control	0				
R12	38	BC 7	DQ14	bidir	X	37	0	Z	
	39	BC_2	*	internal	0				
R10	40	BC_7	DNU0(8)	internal	X				
	41	BC_2	*	internal	0				
P9	42	BC_7	DNU0(9)	internal	Х				
	43	BC_2	*	control	0				
P11	44	BC_7	DQ10	bidir	Х	43	0	Z	
	45	BC_2	*	control	0				
P13	46	BC_7	DQ12	bidir	Х	45	0	Z	
	47	BC_2	*	internal	0				
N12	48	BC_7	DNU0(10)	internal	Х				
N10	49	BC_1	A1_A0	input	Х				
M9	50	 BC_1	 REF#_A2	input	Х				
M11	51	BC_1	 A9_DPR#	input	Х				
L12	52	BC_1	WE#_CS#	input	Х				
L10	53	BC_1	 A3_A4	input	Х				
K9	54	BC_1	CK#	input	Х				
K11	55	BC_1	DNU0(11)	internal	Х				
J13	56	BC_1	LBK#	input	1				
J11	57	BC_1	DNU0(12)	internal	X				
JII	57			internal	^				

Table 4-9 Boundary Scan Order, x18



X18	Cell #	Cell	Pin	TYPE	Safe	Cntrl	Cntrl	Cntrl
			Name		Bit	Cell	Value	State
J9	58	BC_1	CK	input	X			
H10	59	BC_1	A5_A6	input	X			
H12	60	BC_1	DM_RST#	input	X			
G11	61	BC_1	A11	input	X			
G9	62	BC_1	A7	input	X			
F10	63	BC_1	A13-A12 *	input	X			
E40	64	BC_2		internal	0			
F12	65	BC_7	DNU0(13) *	internal	X			
E 40	66	BC_2		control	0		0	7
E13	67	BC_7	DQ3 *	bidir	X	66	0	Z
F 11	68	BC_2		control	0	69	0	7
E11	69 70	BC_7	DQ1 *	bidir	X	68	0	Z
E9	70	BC_2		internal	0			
E9	71	BC_7	DNU0(14) *	internal	X			
D10	72	BC_2		internal	0			
D10	73	BC_7	DNU0(15)	internal	X			
D10	74	BC_2	DOF	control	0	74	0	7
D12	75	BC_7	DQ5	bidir	X	74	0	Z
C13	76	BC_1	QK0#	output2	X			
014	77	BC_2		control	0	77	0	7
C11	78	BC_7	DQ7	bidir	X	77	0	Z
<u></u>	79	BC_2		internal	0			
C9	80	BC_7	DNU0(16)	internal	X			
D 40	81	BC_2		internal	0			
B10	82	BC_7	DNU0(17)	internal	X			
D 40	83	BC_2		control	0			7
B12	84	BC_7	DQ8	bidir	X	83	0	Z
A13	85	BC_1	QK0 *	output2	X			
D 4	86	BC_2		control	0	00	0	7
B4	87	BC_7	DQ6 *	bidir	X	86	0	Z
DO	88	BC_2		internal	0			
B2	89	BC_7	DNU0(18)	internal	X			
C1	90	BC_1	QVLD *	output2	X			
<u></u>	91	BC_2		internal	0			
C3	92	BC_7	DNU0(19)	internal	X			
<u></u>	93	BC_2		control	0	00		7
C5	94	BC_7	DQ4 *	bidir	X	93	0	Z
D4	95	BC_2		control	0	05		7
D4	96	BC_7	DQ2	bidir	X	95	0	Z
50	97	BC_2		internal	0			
D2	98	BC_7	DNU0(20)	internal	X			
E 1	99 100	BC_2		internal	0			
E1	100	BC_7	DNU0(21)	internal	X			
E2	101	BC_2		internal	0			
E3	102	BC_7	DNU0(22)	internal	X			
F F	103	BC_2 BC 7		control	0	400		7
E5 F4	104	_	DQ0	bidir	X X	103	0	Z
Г4	105 106	BC_1	A12_A13 *	input	0			
ED		BC_2 BC_7		control		106	0	7
F2	107		DINV0	bidir	X	106	0	Z
G3 G5	108	BC_1	A10 A8	input	X			
G5 H4	109	BC_1		input	X			
	110	BC_1	A6_A5	input	X			
H2	111	BC_1	RST#_DM	input	X		<u> </u>	
J1	112	BC_1	MF	input	Х		1	

x36	Cell #	Cell	Pin Name	TYPE	Safe	Cntrl	Cntrl	Cntrl	
J3	0	DC 1	DK0	innut	Bit	Cell	Value	State	
- 33 K3	1	BC_1 BC_1		input	X X				
L2	2	BC_1 BC_1	DK0#	input	X				
-			CS#_WE#	input					
L4	3	BC_1	A4_A3	input	X				
M5	4	BC_1	A2_REF#	input	X				
M3	5	BC_1	DPR#_A9 *	input	X				
NO	6	BC_2		control	0	0	0	7	
N2	7	BC_7	DINV2	bidir	X	6	0	Z	
N4	8	BC_1	A0_A1*	input	X				
	9	BC_2		control	0			_	
P5	10	BC_7	DQ18 *	bidir	X	9	0	Z	
5.0	11	BC_2		control	0			_	
P3	12	BC_7	DQ34 *	bidir	X	11	0	Z	
	13	BC_2		control	0				
P1	14	BC_7	DQ35	bidir	Х	13	0	Z	
	15	BC_2	*	control	0			<u> </u>	
R2	16	BC_7	DQ32	bidir	X	15	0	Z	
	17	BC_2	*	control	0				
R4	18	BC_7	DQ20	bidir	Х	17	0	Z	
	19	BC_2	*	control	0				
T5	20	BC_7	DQ22	bidir	Х	19	0	Z	
	21	BC_2	*	control	0				
Т3	22	BC_7	DQ30	bidir	Х	21	0	Z	
T1	23	BC_1	QK1#	output2	Х				
	24	BC_2	*	control	0				
U2	25	BC_7	DQ28	bidir	Х	24	0	Z	
	26	BC_2	*	control	0				
U4	27	BC_7	DQ24	bidir	Х	26	0	Z	
V1	28	BC_1	QK1	output2	Х				
	29	BC_2	*	control	0				
U12	30	BC_7	DQ26	bidir	Х	29	0	Z	
	31	BC_2	*	control	0				
U10	32	BC_7	DQ27	bidir	Х	31	0	Z	
	33	BC_2	*	control	0				
Т9	34	BC_7	DQ29	bidir	Х	33	0	Z	
	35	BC_2	*	control	0				
T11	36	BC_7	DQ25	bidir	Х	35	0	Z	
	37	BC_2	*	control	0				
R12	38	BC_7	DQ23	bidir	Х	37	0	Z	
	39	BC_2	*	control	0				
R10	40	BC_7	DQ31	bidir	Х	39	0	Z	
	41	BC_2	*	control	0				
P9	42	BC_7	DQ33	bidir	Х	41	0	Z	
	43	 BC_2	*	control	0				
P11	44	BC_7	DQ19	bidir	Х	43	0	Z	
	45	BC_2	*	control	0				
P13	46	BC_7	DQ21	bidir	Х	45	0	Z	
	47	 BC_2	*	control	0				
N12	48	BC_7	DINV3	bidir	Х	47	0	Z	
N10	49	BC_1	A1_A0	input	Х				
M9	50	 BC_1	 REF#_A2	input	Х				
M11	51	BC_1	 A9_DPR#	input	Х				
L12	52	BC 1	WE#_CS#	input	X				
L10	53	BC 1	A3 A4	input	X	1	1		
K9	54	BC 1	CK#	input	X				
K11	55	BC_1	DK1#	input	X				
J13	56	BC_1	LBK#	input	1				
J13	57	BC_1 BC_1	DK1	input	X		+		
JII	57		UNI	input	^	1	L	L	

Table 4-10 E	Boundary	Scan	Order,	x36
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x36	Cell #	Cell	Pin Name	TYPE	Safe Bit	Cntrl Cell	Cntrl Value	Cntrl State
J9	58	BC 1	CK	input	X	001	Value	Oldio
H10	59	BC 1	A5 A6	input	X			
H12	60	BC 1	DM_RST#	input	Х			
G11	61	BC 1	A11	input	Х			
G9	62	BC 1	A7	input	Х			
F10	63	BC 1	DNU1(1)	internal	Х			
	64	BC_2	*	control	0			
F12	65	BC_7	DINV1	bidir	Х	64	0	Z
	66	BC_2	*	control	0			
E13	67	BC_7	DQ3	bidir	Х	66	0	Z
	68	BC_2	*	control	0			
E11	69	BC 7	DQ1	bidir	Х	68	0	Z
	70	BC 2	*	control	0			
E9	71	BC 7	DQ17	bidir	Х	70	0	Z
	72	BC_2	*	control	0			
D10	73	BC 7	DQ15	bidir	Х	72	0	Z
	74	BC 2	*	control	0			
D12	75	BC 7	DQ5	bidir	Х	74	0	Z
C13	76	BC 1	QK0#	output2	Х			
	77	BC 2	*	control	0			
C11	78	BC 7	DQ7	bidir	Х	77	0	Z
	79	BC 2	*	control	0		-	
C9	80	BC 7	DQ13	bidir	X	79	0	Z
	81	BC 2	*	control	0			_
B10	82	BC 7	DQ11	bidir	X	81	0	Z
510	83	BC_2	*	control	0	01	Ű	_
B12	84	BC 7	DQ8	bidir	X	83	0	Z
A13	85	BC 1	QK0	output2	X	00	0	2
7110	86	BC 2	*	control	0			
B4	87	BC 7	DQ6	bidir	X	86	0	Z
	88	BC 2	*	control	0			_
B2	89	BC 7	DQ9	bidir	X	88	0	Z
C1	90	BC 1	QVLD	output2	X		Ű	-
	91	BC 2	*	control	0			
C3	92	BC 7	DQ10	bidir	X	91	0	Z
	93	BC_2	*	control	0	01	Ű	_
C5	94	BC_7	DQ4	bidir	X	93	0	Z
00	95	BC 2	*	control	0	00	Ű	_
D4	96	BC_7	DQ2	bidir	X	95	0	Z
	97	BC 2	*	control	0		, v	
D2	98	BC_7	DQ12	bidir	X	97	0	Z
	99	BC 2	*	control	0		, v	
E1	100	BC 7	DQ16	bidir	X	99	0	Z
<u> </u>	100	BC_2	*	control	0		Ť	-
E3	101	BC_7	DQ14	bidir	X	101	0	Z
	102	BC 2	*	control	0	101	Ť	-
E5	103	BC_7	DQ0	bidir	X	103	0	Z
5 F4	104	BC_1 BC_1	DQ0 DNU1(2)	internal	X	100	, v	<u> </u>
	105	BC_2	*	control	0		1	
F2	100	BC 7	DINV0	bidir	X	106	0	Z
G3	107	BC_1 BC_1	A10	input	X	100	0	<u> </u>
G5	100	BC 1	A8	input	X		+	
H4	109	BC 1	A6_A5	input	X		+	
H2	110	BC_1 BC_1	RST# DM	input	X	ļ		
J1	112	BC_1 BC_1	MF	input	X			
J 1	114	1_00	1011	input	~		1	

4.4 JTAG Instructions

Many different instructions (2^8) are possible with the 8-bit instruction register. All used combinations are listed in **Table 4-11**, Instruction Codes. These six instructions are described in detail below. The remaining instructions are reserved and should not be used.

The TAP controller used in this RAM is fully compliant to the 1149.1 convention. Instructions are loaded into the TAP controller during the Shift-IR state when the instruction register is placed between TDI and TDO. During this state, instructions are shifted through the instruction register through the TDI and TDO balls. To execute the instruction once it is shifted in, the TAP controller needs to be moved into the Update-IR state.

Instructions	Instruction Code [7:0]	Description
EXTEST	0000 0000	The EXTEST instruction allows circuitry external to the component package to be tested. Boundary-scan register cells at output pins are used to apply test vectors, while those at input pins capture test results. Typically, the first test vector to be applied using the EXTEST instruction will be shifted into the boundary scan register using the PRELOAD instruction. Thus, during the update-IR state of EXTEST, the output drive is turned on and the PRELOAD data is driven onto the output pins.
IDCODE	0010 0001	The IDCODE instruction causes the ID ROM to be loaded into the ID register when the controller is in capture-DR mode and places the ID register between the TDI and TDO pins in shift-DR mode. The IDCODE instruction is the default instruction loaded in at power up and any time the controller is placed in the test-logic-reset state.
SAMPLE / PRELOAD	0000 0101	SAMPLE / PRELOAD is a Standard 1149.1 mandatory public instruction. When the SAMPLE / PRELOAD instruction is loaded in the instruction register, moving the TAP controller into the capture-DR state loads the data in the RAMs input and Q pins into the boundary scan register. Because the RAM clock(s) are independent from the TAP clock (TCK) it is possible for the TAP to attempt to capture the I/O ring contents while the input buffers are in transition (i.e., in a metastable state). Although allowing the TAP to sample metastable input will not harm the device, repeatable results cannot be expected. RAM input signals must be stabilized for long enough to meet the TAPs input data capture setup plus hold time (tcs plus tcH). The RAMs clock inputs need not be paused for any other TAP operation except capturing the I/O ring contents into the boundary scan register. Moving the controller to shift-DR state then places the boundary scan register between the TDI and TDO pins.
CLAMP	0000 0111	When the CLAMP instruction is loaded into the instruction register, the data driven by the output balls are determined from the values held in the boundary scan register. Selects the bypass register to be connected between TDI and TDO. Data driven by output balls are determined from values held in the boundary scan register.
High-Z	0000 0011	The High-z instruction causes the boundary scan register to be connected between the TDI and TDO. This places all RAMs outputs into a High-Z state. Selects the bypass register to be connected between TDI and TDO. All outputs are forced into high impedance state.
BYPASS	1111 1111	When the BYPASS instruction is loaded in the instruction register, the bypass register is placed between TDI and TDO. This occurs when the TAP controller is moved to the shift-DR state. This allows the board level scan path to be shortened to facilitate testing of other devices in the scan path.
Reserved for Future Use	_	The remaining instructions are not implemented but are reserved for future use. Do not use these instructions.

Table 4-11

4.5 TAP Controller State Diagram





5. Package Drawing

180-PIN TAPE FBGA (BGA) (18.5 x 14)

JEITA Package Code	RENESAS Code	Previous Code	MASS (TYP.) [g]
T-LBGA180-14x18.5-1.00	TLBG0180FA-A	T180FF-100-FF1	0.60





Detail of (A) part

_	+	- A3
-		
_		

	(UNIT:mm)
ITEM	DIMENSIONS
D	18.50 ± 0.10
D1	18.10
D2	17.70
Е	14.00 ± 0.10
E1	13.60
E2	2.59
w	0.20
eD	1.00
еE	1.00
SD	0.50
SE	2.00
А	1.274 ±0.10
A1	0.47 ± 0.05
A2	0.804
A3	0.135
b	0.61±0.05
х	0.15
У	0.10
y1	0.20
ZD	0.75
ZE	1.00
	T180FF-100-FF1



Revision History

μPD48011318, μPD48011336 Series

		Description		
Rev.	Date	Page	Summary	
Rev.0.01	'10.08.18	-	New Preliminary Data Sheet	
Rev.1.00	'12.10.10	- P11 P16 P27 P48	New Data Sheet Update "DC Characteristics" Update "Thermal Impedance" Modified comments for "Power on" Update "Package Drawing"	
Rev.2.00	'13.02.01	P7	Update "Block Diagram"	

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