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April 1st, 2010
Renesas Electronics Corporation

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300/309-OUTPUT TFT-LCD SOURCE DRIVER (64-GRAY SCALE)

The μ PD16640B is a source driver for TFT-LCD 64 gray scale displays. Its logic circuit operates at 3.3 V and the driver circuit operates at 3.3 or 5.0 V (selectable). The input data is digital data at 6 bits \times 3 dots, and 260,000 colors can be displayed in 64-value outputs γ -corrected by the internal D/A converter and 11 external power supplies.

The clock frequency is 55 MHz MIN. By switching over the number of outputs between 300 and 309, the μ PD16640B can be used in TFT-LCD panels conforming to the SVGA/XGA standards.

★ FEATURES

- CMOS level input
- 6 bits (gray scale data) \times 3 dots input
- 64-value output by 11 external power supplies and internal D/A converter
- Output voltage range: $V_{SS2} + 0.1$ V to $V_{DD2} - 0.1$ V
- High-speed data transfer: $f_{MAX.} = 55$ MHz MIN. (internal data transfer speed when operating at 3.0 V)
- Precharge-less output buffer
- Level of γ -corrected power supply can be inverted
- Number of outputs selectable ($O_{sel} = H$: 300 outputs, $O_{sel} = L$: 309 outputs)
- Supply voltage of driver circuit selectable ($V_{sel} = H$: 3.3 V, $V_{sel} = L$: 5.0 V)
- Slim TCP
- Input data inversion function (INV)
- Logic power supply (V_{DD1}): 3.3 V \pm 0.3 V
- Driver power supply (V_{DD2}): 3.3 V \pm 0.3 V ($V_{sel} = H$)
5.0 V \pm 0.5 V ($V_{sel} = L$)

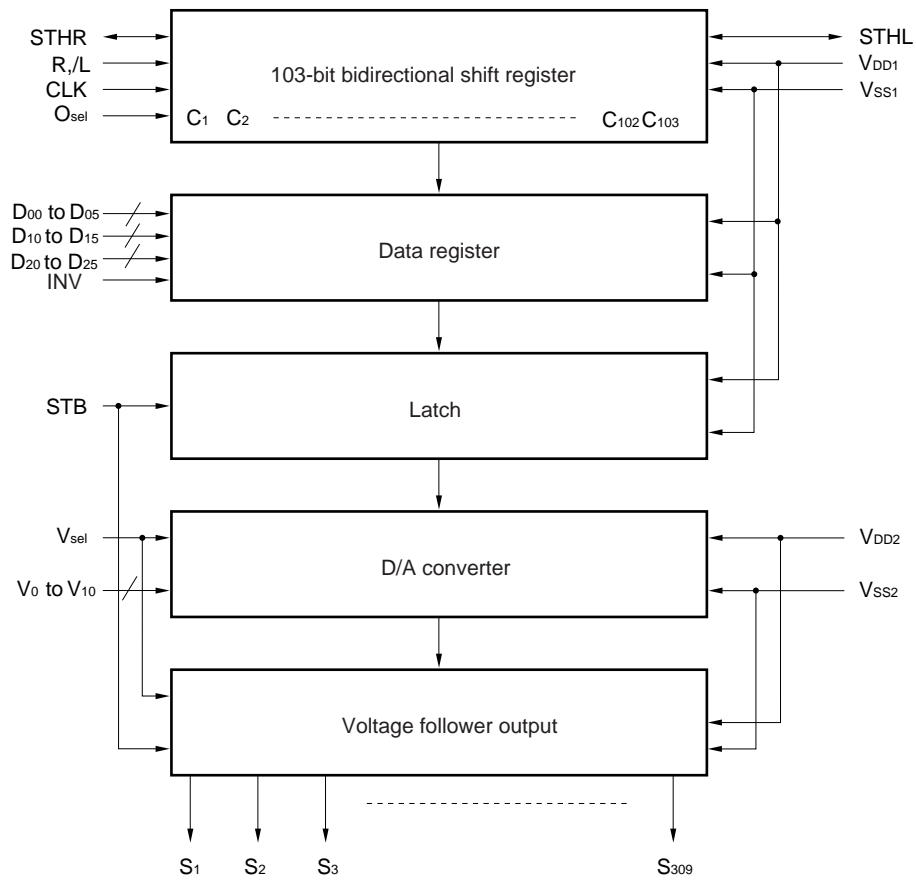
ORDERING INFORMATION

Part Number	Package
μ PD16640BN- xxx	TCP (TAB package)

Remark The TCP's external shape is custom-order item. Users are requested to consult with a NEC sales representative.

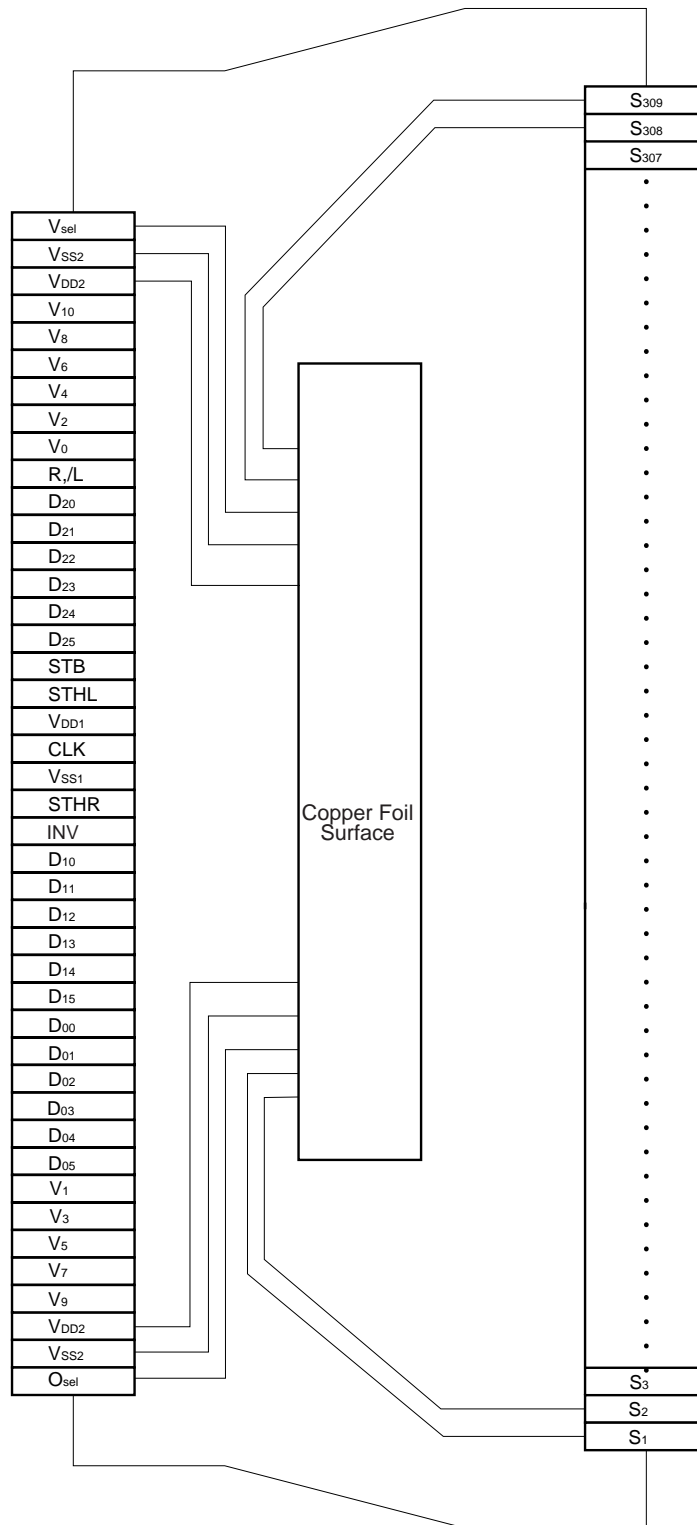
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Not all devices/types available in every country. Please check with local NEC representative for availability and additional information.

★ 1. BLOCK DIAGRAM



Remark /xxx indicates active low signal.

2. PIN CONFIGURATION (μ PD16640BN-xxx)



Remark O_{sel} and V_{sel} pins are internally pulled up.
 Therefore, the number of input pins can be reduced by opening or short-circuiting these pins to V_{SS2} by means of TCP wiring.

3. PIN DESCRIPTION

Pin Symbol	Pin Name	Description
S ₁ to S ₃₀₉	Driver output	Output 64 gray scale analog voltages converted from digital signals. O _{sel} = H: 300 outputs (S ₁ → S ₁₅₀ , S ₁₆₀ → S ₃₀₉) O _{sel} = L: 309 outputs (S ₁ → S ₃₀₉) Output pins S ₁₅₁ to S ₁₅₉ are invalid in 300-output mode.
D ₀₀ to D ₀₅	Display data input	Inputs 18-bit-wide display gray scale data (6 bits) × 3 dots (RGB). D _{X0} : LSB, D _{X5} : MSB
D ₁₀ to D ₁₅		
D ₂₀ to D ₂₅		
R,/L	Shift direction select input	This pin inputs/outputs start pulses in cascade mode. Shift direction of shift register is as follows: R/L = H: STHR input, S ₁ → S ₃₀₉ , STHL output R/L = L: STHL input, S ₃₀₉ → S ₁ , STHR output
STHR	Right shift start pulse I/O	R/L = H: Inputs start pulse. R/L = L: Outputs start pulse.
STHL	Left shift start pulse I/O	R/L = H: Outputs start pulse. R/L = L: Inputs start pulse.
O _{sel}	Number of output selection	Selects number of outputs. This pin is internally pulled up by V _{DD1} power supply. O _{sel} = H: 300 outputs O _{sel} = L: 309 outputs
V _{sel}	Driver voltage selection	Selects driver voltage. This pin is internally pulled up by V _{DD2} power supply. V _{sel} = H: V _{DD2} = 3.3 V V _{sel} = L: V _{DD2} = 5.0 V
CLK	Shift clock input	Inputs shift clock to shift register. Display data is loaded to data register at rising edge of this pin. When O _{sel} = H, start pulse output goes high at rising edge of 100th clock after start pulse has been input, and serves as start pulse to driver in next stage. 100th clock of driver in first stage serves as start pulse of driver in next stage. When O _{sel} = L, start pulse output goes high at rising edge of 103rd clock after start pulse has been input, and serves as start pulse to driver in next stage. 103rd clock of driver in first stage serves as start pulse of driver in next stage.
STB	Latch input	Contents of data register are latched at rising edge, transferred to D/A converter, and output as analog voltage corresponding to display data. Contents of internal shift register are cleared after STB has been input. One pulse of this signal is input when μ PD16640B is started, and then device operates normally. For STB input timing, refer to 8. Switching Characteristic Waveform .
★ V ₀ to V ₁₀	γ-corrected power supply	Inputs γ-corrected power from external source. V _{SS2} + 0.1 V ≤ V ₁₀ ≤ V ₉ ≤ V ₈ ≤ V ₇ ≤ V ₆ ≤ V ₅ ≤ V ₄ ≤ V ₃ ≤ V ₂ ≤ V ₁ ≤ V ₀ ≤ V _{DD2} - 0.1 V or V _{SS2} + 0.1 V ≤ V ₀ ≤ V ₁ ≤ V ₂ ≤ V ₃ ≤ V ₄ ≤ V ₅ ≤ V ₆ ≤ V ₇ ≤ V ₈ ≤ V ₉ ≤ V ₁₀ ≤ V _{DD2} - 0.1 V Maintain gray scale power supply during gray scale voltage output.
INV	Data inversion input	Input data can be inverted when display data is loaded. INV = H: Inverts and loads input data. INV = L: Does not invert input data.
V _{DD1}	Logic circuit power supply	3.3 V ± 0.3 V
V _{DD2}	Driver circuit power supply	V _{sel} = H: V _{DD2} = 3.3 V ± 0.3 V V _{sel} = L: V _{DD2} = 5.0 V ± 0.5 V
V _{SS1}	Logic ground	Ground
V _{SS2}	Driver ground	Ground

Caution Be sure to turn on power in the order V_{DD1}, logic input, V_{DD2}, and gray scale power (V₀ to V₁₀), and turn off power in the reverse order, to prevent the μ PD16640B from being damaged by latchup. Be sure to observe this power sequence even during a transition period.

★ 4. RELATION BETWEEN INPUT DATA AND OUTPUT VOLTAGE VALUE

The 11 major points on the γ characteristic curve of the LCD panel are arbitrarily set by external power supplies V_0 through V_{10} . If the display data is 00H or 3FH, gray scale voltage V_0 or V_{10} is output. If the display data is in the range 01H to 3EH, the high-order 3 bits select an external powers pair V_{n+1} , V_n . The low-order 3 bits evenly divide the range of V_{n+1} to V_n into eight segments by means of D/A conversion (however, the ranges from V_9 to V_8 and from V_2 to V_1 are divided into seven segments) to output a 64-grayscale voltage.

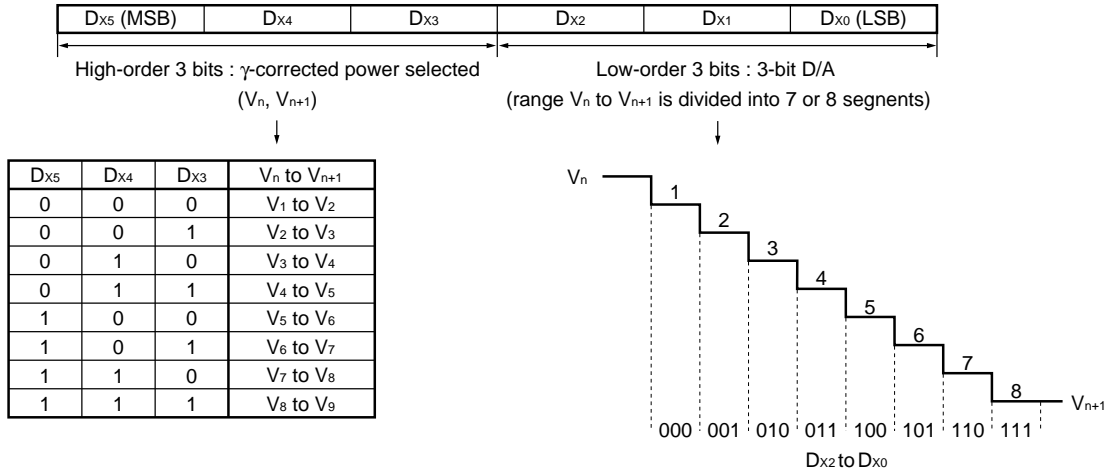


Figure4-1. Relationship between Input Data and γ -corrected Voltage

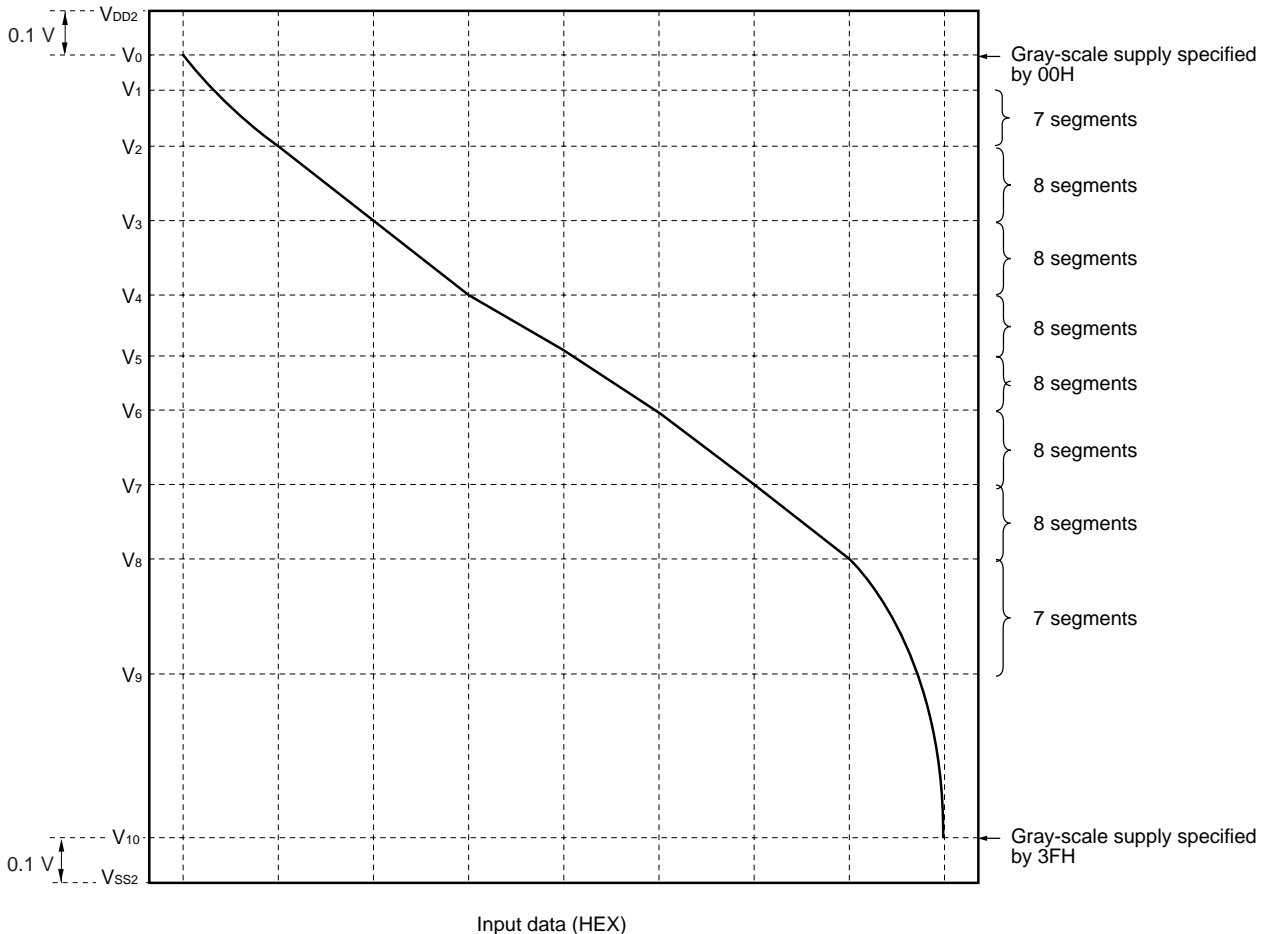


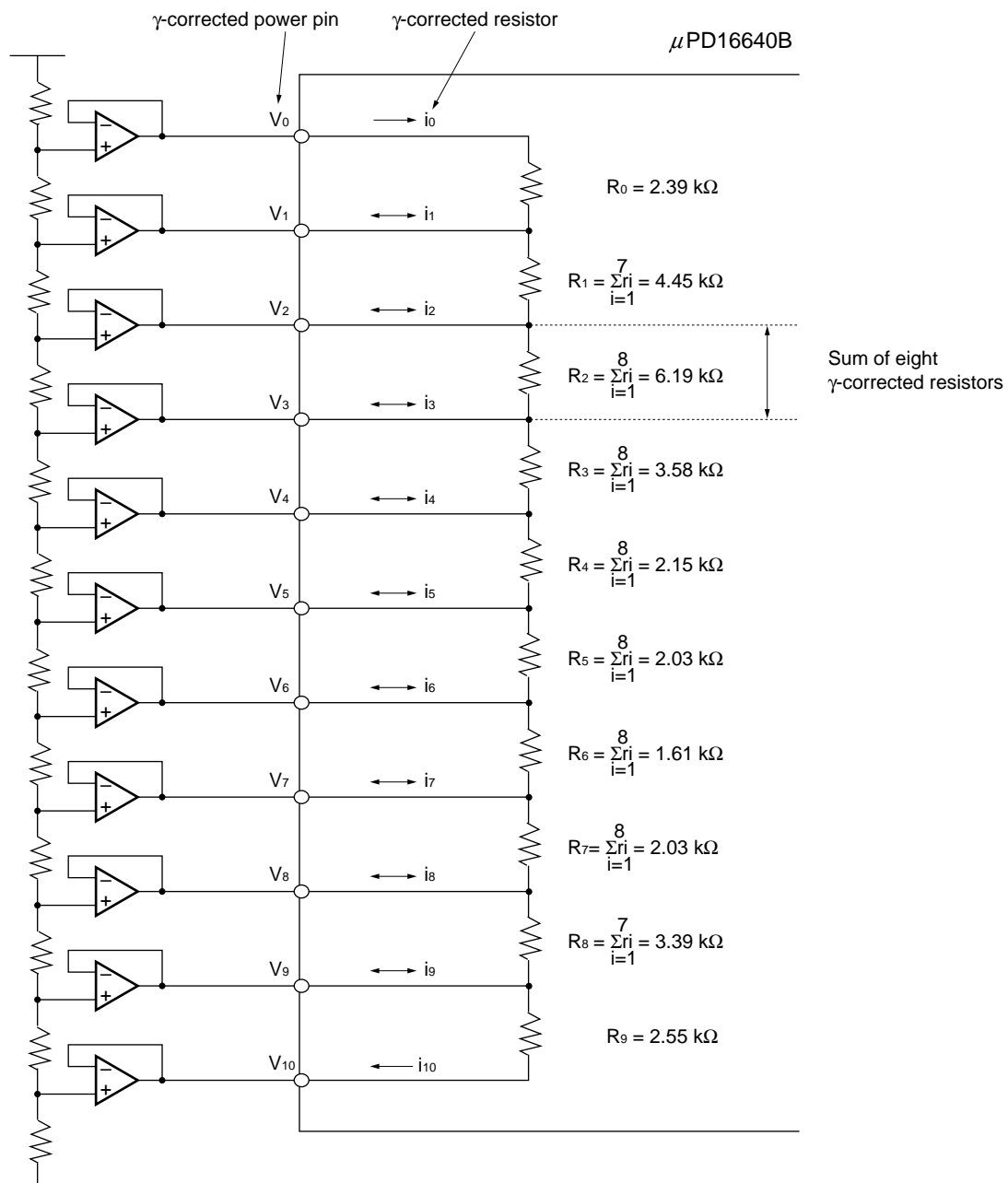
Table4-1. Relation between Input Data and Output Voltage

Input Data	D _{x5}	D _{x4}	D _{x3}	D _{x2}	D _{x1}	D _{x0}	Output Voltage
00H	0	0	0	0	0	0	V ₀
01H	0	0	0	0	0	1	$V_2 + (V_1 - V_2) \times 6/7$
02H	0	0	0	0	1	0	$V_2 + (V_1 - V_2) \times 5/7$
03H	0	0	0	0	1	1	$V_2 + (V_1 - V_2) \times 4/7$
04H	0	0	0	1	0	0	$V_2 + (V_1 - V_2) \times 3/7$
05H	0	0	0	1	0	1	$V_2 + (V_1 - V_2) \times 2/7$
06H	0	0	0	1	1	0	$V_2 + (V_1 - V_2) \times 1/7$
07H	0	0	0	1	1	1	V ₂
08H	0	0	1	0	0	0	$V_3 + (V_2 - V_3) \times 7/8$
09H	0	0	1	0	0	1	$V_3 + (V_2 - V_3) \times 6/8$
0AH	0	0	1	0	1	0	$V_3 + (V_2 - V_3) \times 5/8$
0BH	0	0	1	0	1	1	$V_3 + (V_2 - V_3) \times 4/8$
0CH	0	0	1	1	0	0	$V_3 + (V_2 - V_3) \times 3/8$
0DH	0	0	1	1	0	1	$V_3 + (V_2 - V_3) \times 2/8$
0EH	0	0	1	1	1	0	$V_3 + (V_2 - V_3) \times 1/8$
0FH	0	0	1	1	1	1	V ₃
10H	0	1	0	0	0	0	$V_4 + (V_3 - V_4) \times 7/8$
11H	0	1	0	0	0	1	$V_4 + (V_3 - V_4) \times 6/8$
12H	0	1	0	0	1	0	$V_4 + (V_3 - V_4) \times 5/8$
13H	0	1	0	0	1	1	$V_4 + (V_3 - V_4) \times 4/8$
14H	0	1	0	1	0	0	$V_4 + (V_3 - V_4) \times 3/8$
15H	0	1	0	1	0	1	$V_4 + (V_3 - V_4) \times 2/8$
16H	0	1	0	1	1	0	$V_4 + (V_3 - V_4) \times 1/8$
17H	0	1	0	1	1	1	V ₄
18H	0	1	1	0	0	0	$V_5 + (V_4 - V_5) \times 7/8$
19H	0	1	1	0	0	1	$V_5 + (V_4 - V_5) \times 6/8$
1AH	0	1	1	0	1	0	$V_5 + (V_4 - V_5) \times 5/8$
1BH	0	1	1	0	1	1	$V_5 + (V_4 - V_5) \times 4/8$
1CH	0	1	1	1	0	0	$V_5 + (V_4 - V_5) \times 3/8$
1DH	0	1	1	1	0	1	$V_5 + (V_4 - V_5) \times 2/8$
1EH	0	1	1	1	1	0	$V_5 + (V_4 - V_5) \times 1/8$
1FH	0	1	1	1	1	1	V ₅
20H	1	0	0	0	0	0	$V_6 + (V_5 - V_6) \times 7/8$
21H	1	0	0	0	0	1	$V_6 + (V_5 - V_6) \times 6/8$
22H	1	0	0	0	1	0	$V_6 + (V_5 - V_6) \times 5/8$
23H	1	0	0	0	1	1	$V_6 + (V_5 - V_6) \times 4/8$
24H	1	0	0	1	0	0	$V_6 + (V_5 - V_6) \times 3/8$
25H	1	0	0	1	0	1	$V_6 + (V_5 - V_6) \times 2/8$
26H	1	0	0	1	1	0	$V_6 + (V_5 - V_6) \times 1/8$
27H	1	0	0	1	1	1	V ₆
28H	1	0	1	0	0	0	$V_7 + (V_6 - V_7) \times 7/8$
29H	1	0	1	0	0	1	$V_7 + (V_6 - V_7) \times 6/8$
2AH	1	0	1	0	1	0	$V_7 + (V_6 - V_7) \times 5/8$
2BH	1	0	1	0	1	1	$V_7 + (V_6 - V_7) \times 4/8$
2CH	1	0	1	1	0	0	$V_7 + (V_6 - V_7) \times 3/8$
2DH	1	0	1	1	0	1	$V_7 + (V_6 - V_7) \times 2/8$
2EH	1	0	1	1	1	0	$V_7 + (V_6 - V_7) \times 1/8$
2FH	1	0	1	1	1	1	V ₇
30H	1	1	0	0	0	0	$V_8 + (V_7 - V_8) \times 7/8$
31H	1	1	0	0	0	1	$V_8 + (V_7 - V_8) \times 6/8$
32H	1	1	0	0	1	0	$V_8 + (V_7 - V_8) \times 5/8$
33H	1	1	0	0	1	1	$V_8 + (V_7 - V_8) \times 4/8$
34H	1	1	0	1	0	0	$V_8 + (V_7 - V_8) \times 3/8$
35H	1	1	0	1	0	1	$V_8 + (V_7 - V_8) \times 2/8$
36H	1	1	0	1	1	0	$V_8 + (V_7 - V_8) \times 1/8$
37H	1	1	0	1	1	1	V ₈
38H	1	1	1	0	0	0	$V_9 + (V_8 - V_9) \times 6/7$
39H	1	1	1	0	0	1	$V_9 + (V_8 - V_9) \times 5/7$
3AH	1	1	1	0	1	0	$V_9 + (V_8 - V_9) \times 4/7$
3BH	1	1	1	0	1	1	$V_9 + (V_8 - V_9) \times 3/7$
3CH	1	1	1	1	0	0	$V_9 + (V_8 - V_9) \times 2/7$
3DH	1	1	1	1	0	1	$V_9 + (V_8 - V_9) \times 1/7$
3EH	1	1	1	1	1	0	V ₉
3FH	1	1	1	1	1	1	V ₁₀

4.1 γ-Corrected Power Circuit

The reference power supply of the D/A converter consists of a ladder circuit with a total of 64 resistors, and resistance Σr_i between γ-corrected power pins differs depending on each pair of γ-corrected power pins. One pair of γ-corrected power pins consists of seven or eight series resistors, and resistance Σr_i in the figure below is indicated as the sum of the seven or eight resistors. The resistance ratio between the γ-corrected power pins (Σr_i ratio) is designed to be a value relatively close to the ratio of the γ-corrected voltages V_1 to V_9 (gray-scale voltages in 8 steps) used in an actual LCD panel. Under ideal conditions where there is no difference between the two, therefore, there is no voltage difference between the voltage of the γ-corrected power supplies and the gray-scale voltages in 8 steps of the resistor ladder circuits of the μ PD16640B, and no current flows into the γ-corrected power pins V_1 to V_9 . As a result, a voltage-follower circuit is not necessary.

Figure4-2. γ-Corrected Power Circuit



5. RELATIONSHIP BETWEEN OUTPUT CIRCUIT AND D/A CONVERTER

Data format : 1 pixel data (6 bits) × RGB (3 dots)

Input width : 18 bits

R,/L = H (right shift)

Output	S ₁	S ₂	S ₃	...	S ₃₀₈	S ₃₀₉
Data	D ₀₀ to D ₀₅	D ₁₀ to D ₁₅	D ₂₀ to D ₂₅	...	D ₁₀ to D ₁₅	D ₂₀ to D ₂₅

R,/L = L (left shift)

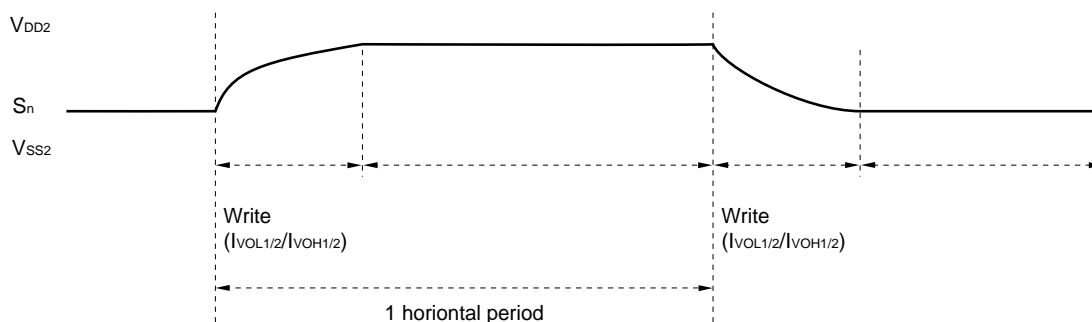
Output	S ₁	S ₂	S ₃	...	S ₃₀₈	S ₃₀₉
Data	D ₀₀ to D ₀₅	D ₁₀ to D ₁₅	D ₂₀ to D ₂₅	...	D ₁₀ to D ₁₅	D ₂₀ to D ₂₅

6. OPERATION OF OUTPUT BUFFER

The output buffer consists of an operational amplifier circuit that does not perform precharge operation.

Therefore, driver output current $I_{VOH1/2}$ is the charging current to the LCD, and $I_{VOL1/2}$ is the discharging current.

<LCD panel driving waveform of μ PD16640B>



7. ELECTRICAL SPECIFICATIONS

Absolute Maximum Ratings (T_A = 25 °C, V_{SS1} = V_{SS2} = 0 V)

Parameter	Symbol	Ratings	Unit
Logic Power Supply	V _{DD1}	-0.3 to +4.5	V
Driver Power Supply	V _{DD2}	-0.3 to +6.0	V
Input Voltage	V _I	-0.3 to V _{DD1,2} + 0.3	V
Output Voltage	V _O	-0.3 to V _{DD1,2} + 0.3	V
Operating Ambient Temperature	T _A	-10 to +75	°C
Storage Temperature	T _{stg}	-55 to +125	°C

Caution If the absolute maximum rating of even one of the above parameters is exceeded even momentarily, the quality of the product may be degraded. Absolute maximum ratings, therefore, specify the values exceeding which the product may be physically damaged. Be sure to use the product within the range of the absolute maximum ratings.

Recommended Operating Range (T_A = -10 to +75 °C, V_{SS1} = V_{SS2} = 0 V)

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Logic Supply Voltage	V _{DD1}		3.0	3.3	3.6	V
Driver Supply Voltage	V _{DD2}	V _{sel} = H	3.0	3.3	3.6	V
		V _{sel} = L	4.5	5.0	5.5	V
High-Level Input Voltage	V _{IH}		0.7 V _{DD1}		V _{DD1}	V
Low-Level Input Voltage	V _{IL}		0		0.3 V _{DD1}	V
γ-Corrected Power Voltage	V _O to V _{I0}		V _{SS2} + 0.1		V _{DD2} - 0.1	V
Maximum Clock Frequency	f _{MAX.}		55			MHz

Electrical Characteristics ($T_A = -10$ to $+75$ °C, $V_{DD1} = 3.3$ V \pm 0.3 V, $V_{DD2} = 3.3$ V \pm 0.3 V or 5.0 V \pm 0.5 V, $V_{SS1} = V_{SS2} = 0$ V)

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit	
Input Leakage Current	I_{IL}	R,/L, CLK, STB, INV, STHR(STHL), D ₀₀ to D ₀₅ , D ₁₀ to D ₁₅ , D ₂₀ to D ₂₅			±1.0	μ A	
Pull-up Resistor	R_{PU}	$V_{DD1} = 3.3$ V, O_{sel} , V_{sel}	40	100	250	kΩ	
High-Level Output Voltage	V_{OH}	STHR (STHL), $I_o = -1.0$ mA	$V_{DD1} - 0.5$			V	
Low-Level Output Voltage	V_{OL}	STHR (STHL), $I_o = +1.0$ mA			0.5	V	
Static Current Consumption of γ -Corrected Power ($V_{DD2} = 3.3$ V or 5.5 V)	I_{vn1}	$V_{DD1} = 3.3$ V, $V_{DD2} = 3.3$ V	V_0 to V_1	100	200	400	μ A
			V_1 to V_2	54	109	218	μ A
			V_2 to V_3	39	79	158	μ A
			V_3 to V_4	68	137	274	μ A
			V_4 to V_5	109	219	438	μ A
			V_5 to V_6	116	232	464	μ A
			V_6 to V_7	144	288	576	μ A
			V_7 to V_8	116	232	464	μ A
			V_8 to V_9	72	145	290	μ A
Driver Output Current ($V_{DD2} = 3.3$ V)	I_{VOH1}	STB = 3.3 V, $V_{OUT} = 2.7$ V, $V_X = 3.2$ V ^{Note1} , $V_{DD1} = V_{DD2} = 3.3$ V		-46	-20	μ A	
	I_{VOL1}	STB = 3.3 V, $V_{OUT} = 0.6$ V, $V_X = 0.1$ V ^{Note1} , $V_{DD1} = V_{DD2} = 3.3$ V	30	62		μ A	
Driver Output Current ($V_{DD2} = 5.0$ V)	I_{VOH2}	STB = 5.0 V, $V_{OUT} = 4.4$ V, $V_X = 4.9$ V ^{Note1} , $V_{DD1} = 3.3$ V, $V_{DD2} = 5.0$ V		-69	-30	μ A	
	I_{VOL2}	STB = 5.0 V, $V_{OUT} = 0.6$ V, $V_X = 0.1$ V ^{Note1} , $V_{DD1} = 3.3$ V, $V_{DD2} = 5.0$ V	40	81		μ A	
Output Voltage Deviation	ΔV_o	$V_{DD1} = 3.3$ V, $V_{DD2} = 3.3$ V or 5.0 V, $V_{OUT} = 0.5$ V ^{Note1} , $0.5 V_{DD2}$, $V_{DD2} - 0.5$ V		±10	±20	mV	
Output Voltage Range	V_o	Input data: 00H to 3FH	$V_{SS2} + 0.1$		$V_{DD2} - 0.1$	V	
Dynamic Logic Current Consumption	I_{DD1}	No load ^{Note2}		0.5	2.5	mA	
Dynamic Driver Current Consumption	I_{DD21}	No load, $V_{DD2} = 3.3$ V ^{Note2}			9.0	mA	
Dynamic Driver Current Consumption	I_{DD22}	No load, $V_{DD2} = 5.0$ V ^{Note2}			9.0	mA	

Note1. V_X is output voltage of analog output pin S_1 to S_{309} .
 V_{OUT} is the voltage applied to analog output pin S_1 to S_{309} .
2. The STB cycle is specified at 31 μs and $f_{CLK} = 16$ MHz.

Switching Characteristics ($T_A = -10$ to $+75$ °C, $V_{DD1} = 3.3$ V \pm 0.3 V, $V_{DD2} = 3.3$ V \pm 0.3 V or 5.0 V \pm 0.5 V, $V_{SS1} = V_{SS2} = 0$ V)

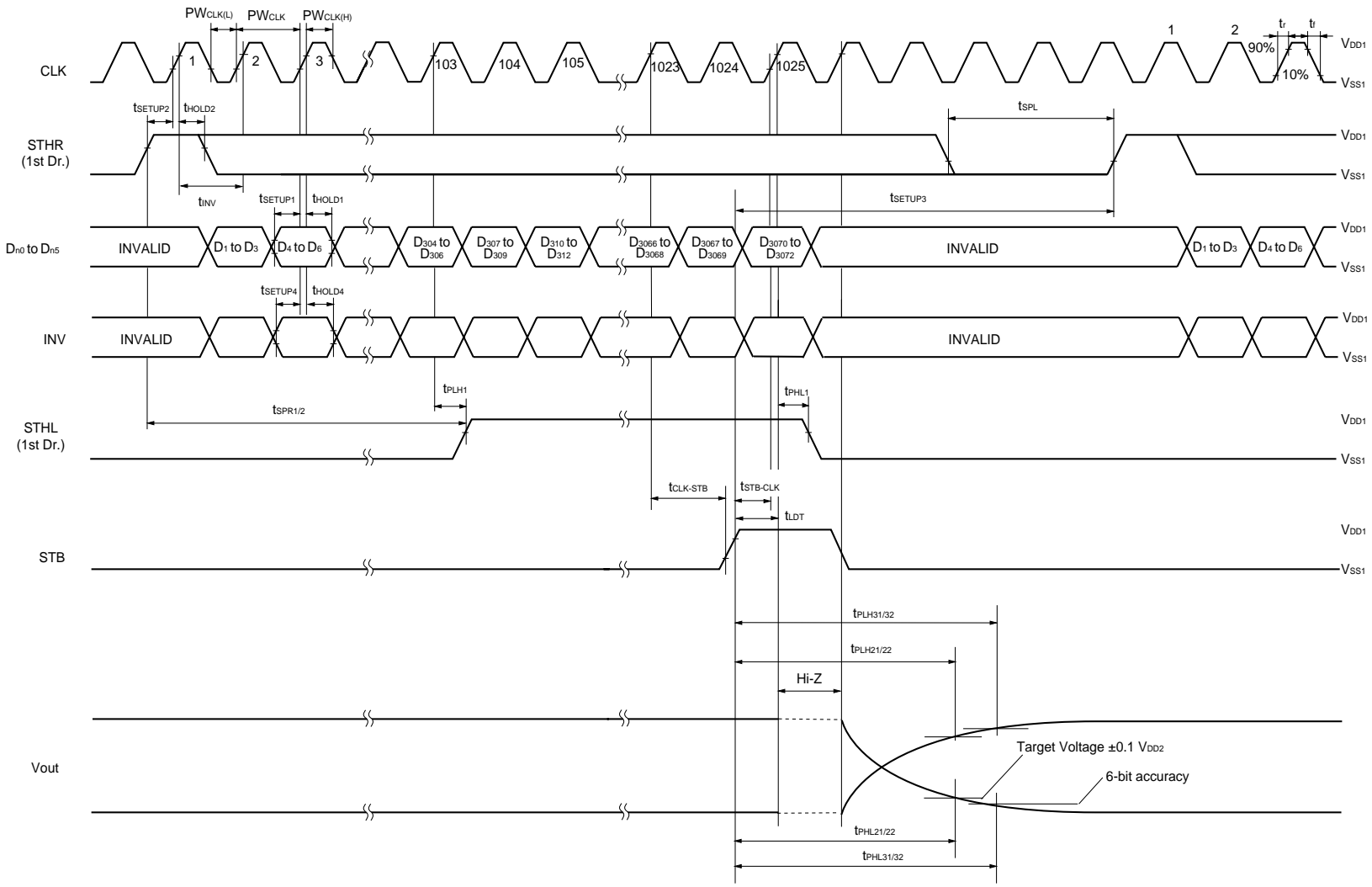
Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit		
Start Pulse Delay time	t _{PLH1}	C _L = 15 pF			12	ns		
	t _{PHL1}				12	ns		
Driver Output Delay Time	t _{PLH21}	V _{DD2} = 3.3 V, 4 kΩ +24 pF × 2	V _O : 0.1 V → 3.2 V		3.3		μ s	
	t _{PLH31}				5.1	10	μ s	
	t _{PHL21}			V _O : 3.2 V → 0.1 V		4.3		μ s
	t _{PHL31}					5.4	10	μ s
	t _{PLH22}	V _{DD2} = 5.0 V, 4 kΩ +24 pF × 2	V _O : 0.1 V → 4.9 V		4.2		μ s	
	t _{PLH32}				5.4	10	μ s	
	t _{PHL22}		V _O : 4.9 V → 0.1 V		4.9		μ s	
	t _{PHL32}				6.2	10	μ s	
Input capacitance	C ₁₁	STHR (STHL), T _A = 25 °C		10	15	pF		
	C ₁₂	V ₀ to V ₁₀ , T _A = 25 °C		100		pF		
	C ₁₃	STHR (STHL), other than V ₀ to V ₁₀ , T _A = 25 °C		5	10	pF		

Timing Requirements ($T_A = -10$ to $+75$ °C, $V_{DD1} = 3.3$ V \pm 0.3 V, $V_{DD2} = 3.3$ V \pm 0.3 V or 5.0 V \pm 0.5 V, $V_{SS1} = V_{SS2} = 0$ V)

Parameter	Symbol	Condition	MIN.	TYP.	MAX.	Unit
Clock Pulse Width	PW _{CLK}		18			ns
Clock Pulse High Period	PW _{CLK (H)}		4			ns
Clock Pulse Low Period	PW _{CLK (L)}		4			ns
Data Setup Time	t _{SETUP1}		4			ns
Data Hold Time	t _{HOLD1}		0			ns
Start Pulse Setup Time	t _{SETUP2}		4			ns
Start Pulse Hold Time	t _{HOLD2}		0			ns
INV Setup time	t _{SETUP4}		4			ns
INV Hold time	t _{HOLD4}		0			ns
Start Pulse Low period	t _{SPL}		2			CLK
Start Pulse Rise Time	t _{SPR1}	O _{sel} = H		100		CLK
Start Pulse Rise Time	t _{SPR2}	O _{sel} = L		103		CLK
STB setup time	t _{SETUP3}		2			CLK
Data Invalid Period	t _{INV}			1		CLK
Last Data Timing	t _{LDT}				1	CLK
CLK-STB Time	t _{CLK-STB}	CLK ↑ → STB ↑ or ↓	7			ns
STB-CLK Time	t _{STB-CLK}	STB ↑ or ↓ → CLK ↑	7			ns

Phase-out/Discontinued

8. SWITCHING CHARACTERISTIC WAVEFORM (R_L/L = H)
 (Unless otherwise specified, the input level is defined to be V_{IH} = 0.7 V_{DD1}, V_{IL} = 0.3 V_{DD1}.)



9. RECOMMENDED SOLDERING CONDITIONS

The following conditions must be met for mounting conditions of the μ PD16640B.

For more details, refer to the **Semiconductor Device Mounting Technology Manual (C10535E)**.

Please consult with our sales offices in case other mounting process is used, or in case the mounting is done under different conditions.

μ PD16640BN-xxx : TCP(TAB package)

Mounting Condition	Mounting Method	Condition
Thermocompression	Soldering	Heating tool 300 to 350°C, heating for 2 to 3 seconds: pressure 100 g (per solder)
	ACF (Adhesive Conductive Film)	Temporary bonding 70 to 100°C; pressure 3 to 8 kg/cm ² ; time 3 to 5 secs. Real bonding 165 to 180°C; pressure 25 to 45 kg/cm ² , time 30 to 40 secs. (When using the anisotropy conductive film SUMIZAC1003 of Sumitomo Bakelite, Ltd.)

Caution To find out the detailed conditions for mounting the ACF part, please contact the ACF manufacturing company. Be sure to avoid using two or more mounting methods at a time.

[MEMO]

NOTES FOR CMOS DEVICES

① PRECAUTION AGAINST ESD FOR SEMICONDUCTORS

Note:

Strong electric field, when exposed to a MOS device, can cause destruction of the gate oxide and ultimately degrade the device operation. Steps must be taken to stop generation of static electricity as much as possible, and quickly dissipate it once, when it has occurred. Environmental control must be adequate. When it is dry, humidifier should be used. It is recommended to avoid using insulators that easily build static electricity. Semiconductor devices must be stored and transported in an anti-static container, static shielding bag or conductive material. All test and measurement tools including work bench and floor should be grounded. The operator should be grounded using wrist strap. Semiconductor devices must not be touched with bare hands. Similar precautions need to be taken for PW boards with semiconductor devices on it.

② HANDLING OF UNUSED INPUT PINS FOR CMOS

Note:

No connection for CMOS device inputs can be cause of malfunction. If no connection is provided to the input pins, it is possible that an internal input level may be generated due to noise, etc., hence causing malfunction. CMOS devices behave differently than Bipolar or NMOS devices. Input levels of CMOS devices must be fixed high or low by using a pull-up or pull-down circuitry. Each unused pin should be connected to V_{DD} or GND with a resistor, if it is considered to have a possibility of being an output pin. All handling related to the unused pins must be judged device by device and related specifications governing the devices.

③ STATUS BEFORE INITIALIZATION OF MOS DEVICES

Note:

Power-on does not necessarily define initial status of MOS device. Production process of MOS does not define the initial operation status of the device. Immediately after the power source is turned ON, the devices with reset function have not yet been initialized. Hence, power-on does not guarantee out-pin levels, I/O settings or contents of registers. Device is not initialized until the reset signal is received. Reset operation must be executed immediately after power-on for devices having reset function.

Reference Documents

NEC Semiconductor Device Reliability / Quality Control System (C10983E)

Quality Grades to NEC's Semiconductor Devices (C11531E)

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 - "Standard", "Special", and "Specific". The Specific quality grade applies only to devices developed based on a customer designated "quality assurance program" for a specific application. The recommended applications of a device depend on its quality grade, as indicated below. Customers must check the quality grade of each device before using it in a particular application.
 - Standard: Computers, office equipment, communications equipment, test and measurement equipment, audio and visual equipment, home electronic appliances, machine tools, personal electronic equipment and industrial robots
 - Special: Transportation equipment (automobiles, trains, ships, etc.), traffic control systems, anti-disaster systems, anti-crime systems, safety equipment and medical equipment (not specifically designed for life support)
 - Specific: Aircraft, aerospace equipment, submersible repeaters, nuclear reactor control systems, life support systems or medical equipment for life support, etc.
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