

DATASHEET

General Description

The IDT8T49N244I is a dual PLL using FemtoClock® NG technology. It integrates low phase noise Frequency Translation / Synthesis and Jitter attenuation. It includes alarm and monitoring functions suitable for networking and communications applications. The device has two fully independent PLLs. Each PLL is able to generate any output frequency in the 0.98MHz - 312.5MHz range and most output frequencies in the 312.5MHz - 1,300MHz range (see Table 3 for details). A wide range of input reference clocks may be used as the source for the output frequencies.

Each PLL of IDT8T49N244I has three operating modes to support a very broad spectrum of applications:

1) Frequency Synthesizer

- Synthesizes output frequencies from an external reference clock REFCLK.
- Fractional feedback division is used, so there are no requirements for any specific input reference clock frequency to produce the desired output frequency with a high degree of accuracy.

2) High-Bandwidth Frequency Translator

- Applications: PCI Express, Computing, General Purpose
- Translates any input clock in the 16MHz - 710MHz frequency range into any supported output frequency.
- This mode has a high PLL loop bandwidth in order to track input reference changes, such as Spread-Spectrum Clock modulation.

3) Low-Bandwidth Frequency Translator

- Applications: Networking & Communications.
- Translates any input clock in the 8kHz - 710MHz frequency range into any supported output frequency.
- This mode supports PLL loop bandwidths in the 10Hz - 580Hz range and makes use of an external REFCLK to provide significant jitter attenuation.

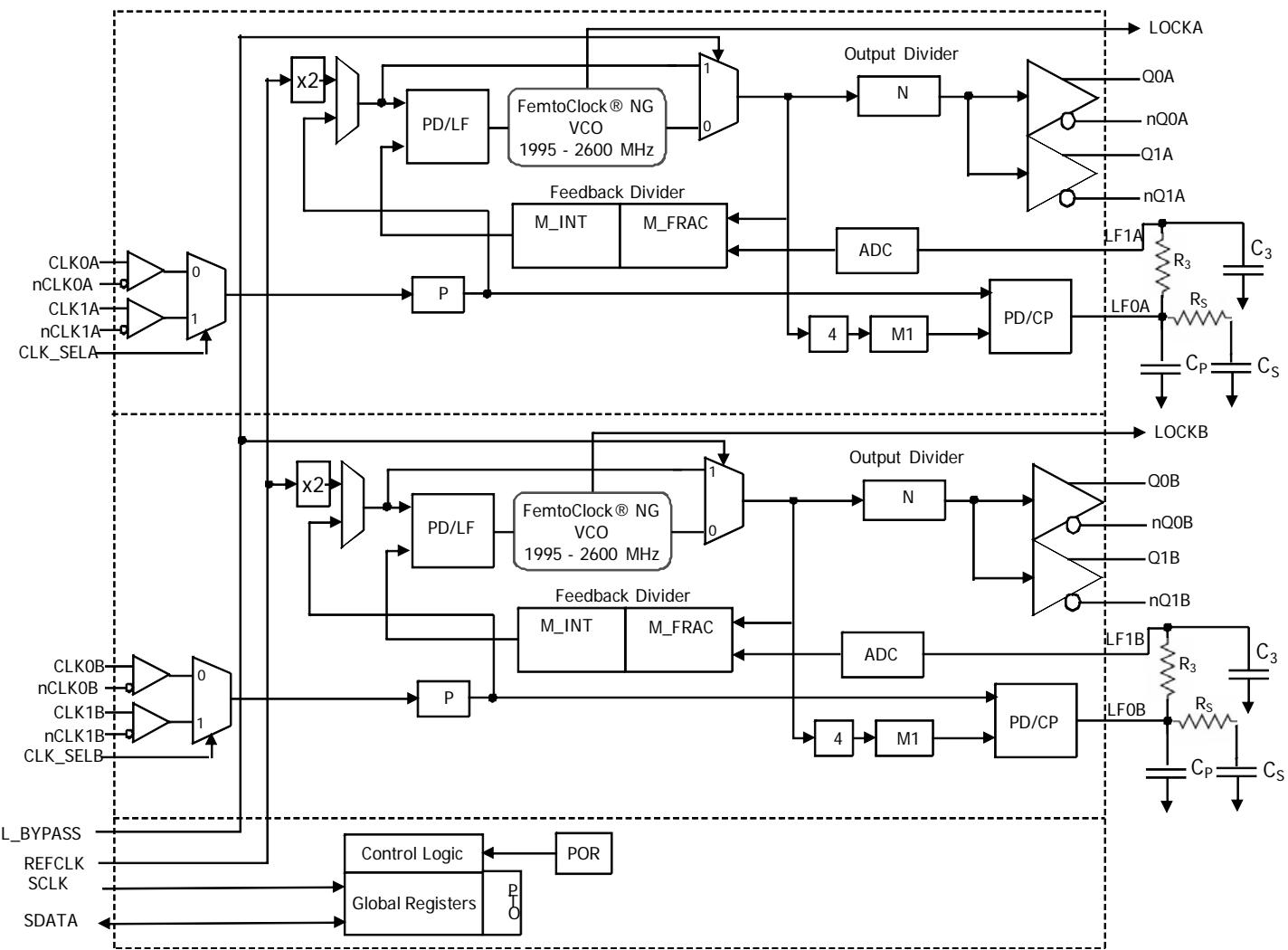
Each PLL provides factory-programmed default power-up configuration burned into One-Time Programmable (OTP) memory. The configuration is specified by customer and are programmed by IDT during the final test phase from an on-hand stock of blank devices.

To implement other configurations, these power-up default settings can be overwritten after power-up using the I²C interface and the device can be completely reconfigured.

Features

- Fourth generation FemtoClock® NG technology
- Two fully independent PLLs
- Universal Frequency Translator™/Frequency Synthesizer and Jitter attenuator
- Outputs are programmable as LVPECL or LVDS
 - Programmable output frequency: 0.98MHz up to 1,300MHz
- Two differential inputs per PLL support the following input types: LVPECL, LVDS, LVHSTL, HCSL
- Input frequency range: 8kHz - 710MHz (Low-Bandwidth mode)
- Input frequency range: 16MHz - 710MHz (High-Bandwidth mode)
- REFCLK frequency range: 16MHz - 40MHz
- Input clock monitor on each PLL will smoothly switch between redundant input references
- Factory-set register configuration for power-up default state
 - Power-up default configuration
 - Configuration customized via One-Time Programmable ROM
 - Settings may be overwritten after power-up via I²C
 - I²C Serial interface for register programming
- RMS phase jitter at 161.1328125MHz, using a 40MHz REFCLK (12kHz - 20MHz): 486fs (typical), Low Bandwidth Mode (FracN)
- RMS phase jitter at 400MHz, using a 40MHz REFCLK (12kHz - 20MHz): 326fs (typical), Synthesizer Mode (Integer FB)
- Supply modes:
V_{CC} / V_{CCA} / V_{CCO}
3.3V / 3.3V / 3.3V
3.3V / 3.3V / 2.5V (LVPECL only)
2.5V / 2.5V / 2.5V
- -40°C to 85°C ambient operating temperature
- 10mm X 10mm CABGA package
- Lead-free (RoHS 6) packaging

Complete Block Diagram



Pin Description and Pin Characteristic Tables

Table 1. Pin Descriptions

| Number | Name | Type | | Description |
|--------|----------------|------------|------------------|--|
| E5 | REFCLK | Input | Pulldown | Reference clock for device operation. |
| G8, G7 | CLK0A, CLK1A | Input | Pulldown | Non-inverting differential clock input. |
| F8, F7 | nCLK0A, nCLK1A | Input | Pullup/ Pulldown | Inverting differential clock input. $V_{CC}/2$ default when left floating (set by the internal pullup and pulldown resistors). |
| H3, G3 | CLK0B, CLK1B | Input | Pulldown | Non-inverting differential clock input. |
| H4, G4 | nCLK0B, nCLK1B | Input | Pullup/ Pulldown | Inverting differential clock input. $V_{CC}/2$ default when left floating (set by the internal pullup and pulldown resistors). |
| J9, J8 | Q0A, nQ0A | Output | | Differential output. Output type is programmable to LVDS or LVPECL interface levels. |
| H6, J4 | Q1A, nQ1A | Output | | Differential output. Output type is programmable to LVDS or LVPECL interface levels. |
| J1, H1 | Q0B, nQ0B | Output | | Differential output. Output type is programmable to LVDS or LVPECL interface levels. |
| F2, D1 | Q1B, nQ1B | Output | | Differential output. Output type is programmable to LVDS or LVPECL interface levels. |
| J6, J5 | LF0A, LF1A | Analog I/O | | Loop filter connection node pins. LF0A is the output, LF1A is the input. |
| F1, E1 | LF0B, LF1B | Analog I/O | | Loop filter connection node pins. LF0B is the output, LF1B is the input. |
| H7 | Rsvd | Input | | Reserved, connect to V_{EE} |
| G2 | Rsvd | Input | | Reserved, connect to V_{EE} |
| E8 | LOCKA | Output | | Lock Indicator - indicates that PLLA is in a locked condition. LVC MOS/LVTTL interface levels. |
| H5 | LOCKB | Output | | Lock Indicator - indicates that PLLB is in a locked condition. LVC MOS/LVTTL interface levels. |
| F6 | CLK_SELA | Input | Pulldown | Input clock select. Selects the active differential clock input. 0 = CLK0A, nCLK0A (default) 1 = CLK1A, nCLK1A |
| F4 | CLK_SELB | Input | Pulldown | Input clock select. Selects the active differential clock input. 0 = CLK0B, nCLK0B (default) 1 = CLK1B, nCLK1B |
| E6 | PLL_BYPASS | Input | Pulldown | Bypasses the VCXO PLL. 0 = PLL Mode (default) 1 = PLL Bypassed |
| G6 | SDATA | I/O | Pullup | I^2C Data Input/Output. Open drain. |
| G5 | SCLK | Input | Pullup | I^2C Clock Input. LVC MOS/LVTTL interface levels. |
| E7 | V_{CCA_A} | Power | | Analog power supply for PLLA. |
| G9 | V_{CCO_A} | Power | | Output power supply for PLLA. |
| F5 | V_{CC_A} | Power | | Core power supply for PLLA. |
| E3 | V_{CCA_B} | Power | | Analog power supply for PLLB. |
| J3 | V_{CCO_B} | Power | | Output power supply for PLLB. |

Table 1. Pin Descriptions

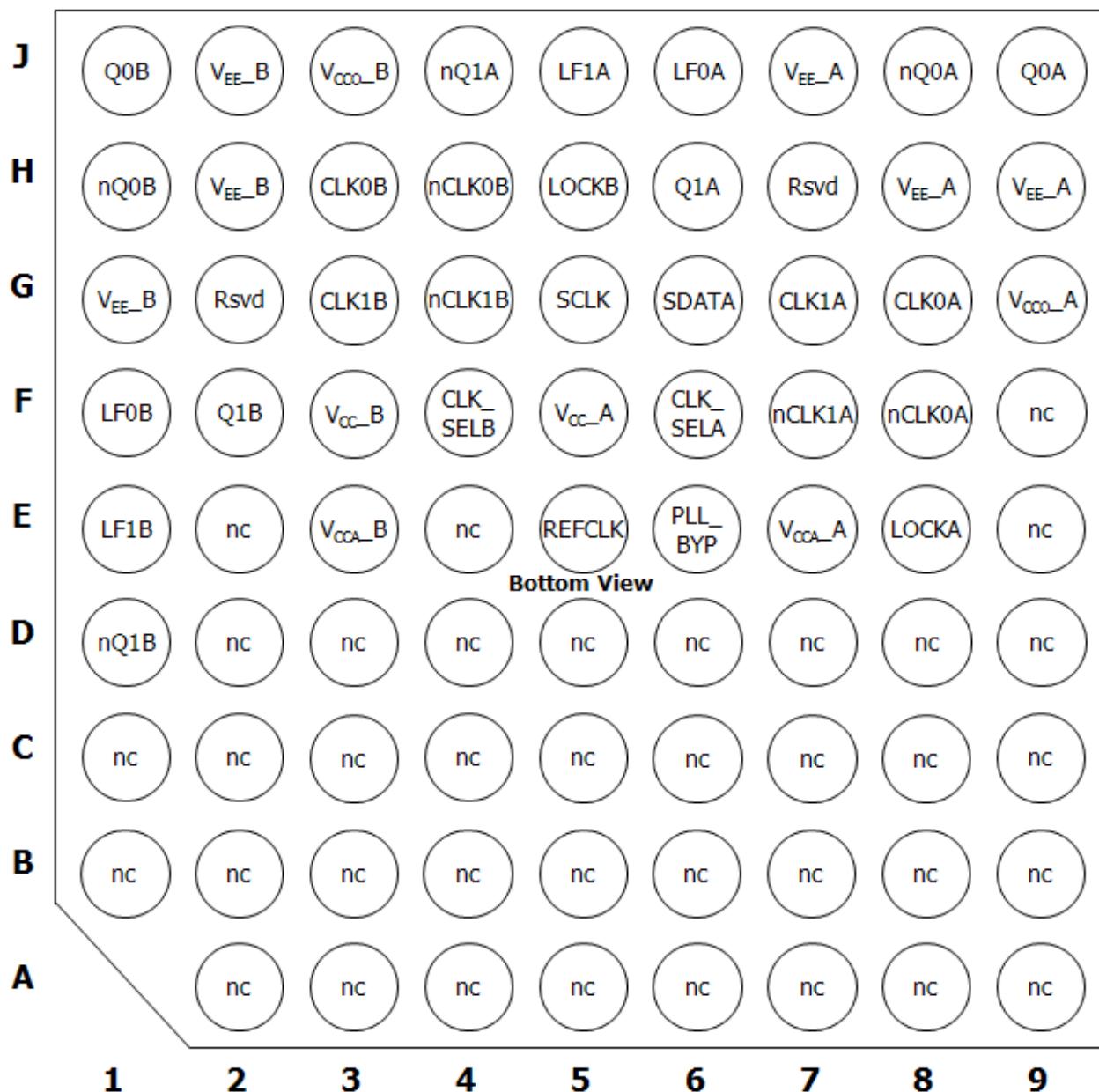
| Number | Name | Type | | Description |
|----------------------------|-------------------|-------|--|---|
| F3 | V _{CC_B} | Power | | Core power supply for PLLB |
| H8, H9, J7 | V _{EE_A} | Power | | Negative supply for PLLA |
| G1, H2, J2 | V _{EE_B} | Power | | Negative supply for PLLB |
| A2-A9, B1-B9, C1-C9, | nc | - | | No connect. These pins may be left unconnected. |

NOTE: *Pullup* and *Pulldown* refer to internal input resistors. See Table 2, *Pin Characteristics*, for typical values.

Table 2. Pin Characteristics

| Symbol | Parameter | | Test Conditions | Minimum | Typical | Maximum | Units |
|-----------------------|-------------------------|-----------------------|-----------------|---------|---------|---------|-------|
| C _{IN} | Input Capacitance | | | | 3.5 | | pF |
| R _{PULLUP} | Input Pullup Resistor | | | | 51 | | kΩ |
| R _{PULLDOWN} | Input Pulldown Resistor | | | | 51 | | kΩ |
| R _{PULLDOWN} | Input Pulldown Resistor | REFCLK, PLL_BYPASS | | | 25 | | kΩ |
| R _{PULLUP} | Input Pullup Resistor | SDATA, SCLK | | | 25 | | kΩ |

Pin Assignment



IDT8T49N244I Pin Map

80-Ball Lead

10mm x 10mm x 1mm package body

CABGA Package

(bottom view)

Functional Description

The IDT8T49N244I is a two PLL device. The two PLLs are fully independent and identical. Each PLL can generate desired outputs frequency (0.98MHz - 1300MHz) from any input source in the operating range (8kHz - 710MHz). It is capable of synthesizing frequencies from REFCLK source. The output frequency is generated regardless of the relationship to the input frequency. Each PLL of IDT8T49N244I can translate the desired output frequency from input clock. In this frequency translation mode, a low-bandwidth, jitter attenuation option is available that makes use of an external fixed-frequency REFCLK to translate from a noisy input source. If the input clock is known to be fairly clean or if some modulation on the input needs to be tracked, then the high-bandwidth frequency translation mode can be used, without the need for the external clock source.

The input clock references and REFCLK input are monitored continuously and appropriate alarm outputs are raised by register bits in the event of any out-of-specification conditions arising. Clock switching is supported in manual, revertive & non-revertive modes.

Each PLL of IDT8T49N244I has a factory-programmed configuration as the default operating state after reset. This default may be over-written by I²C register access at any time, but those over-written settings will be lost on power-down. Please contact IDT if a specific set of power-up default settings is desired.

The following sections apply individually to each PLL. Signal and register bit names have a lowercase 'x' on the end where 'x' should be either 'A' or 'B' as appropriate for the PLL being controlled.

Operating Modes

Each PLL of IDT8T49N244I has three operating modes which are set by the MODE_SELx[1:0] bits. There are two frequency translator modes - low bandwidth and high bandwidth and a frequency synthesizer mode.

Please make use of IDT-provided configuration applications to determine the best operating settings for the desired configuration of the device.

Output Dividers & Supported Output Frequencies

The internal VCO is capable of operating in a range anywhere from 1.995GHz - 2.6GHz. It is necessary to choose an integer multiplier of the desired output frequency that results in a VCO operating frequency within that range. The output divider stage N[10:0] is limited to selection of integers from 2 to 2046. Please refer to Table 3 for the values of N applicable to the desired output frequency.

Table 3. Output Divider Settings & Frequency Ranges

| Register Setting | Frequency Divider | Minimum f _{OUT} | Maximum f _{OUT} |
|------------------|-------------------|--------------------------|--------------------------|
| Nx[10:0] | N | (MHz) | (MHz) |
| 0000000000x | 2 | 997.5 | 1300 |
| 00000000010 | 2 | 997.5 | 1300 |
| 00000000011 | 3 | 665 | 866.7 |
| 00000000100 | 4 | 498.75 | 650 |
| 00000000101 | 5 | 399 | 520 |
| 0000000011x | 6 | 332.5 | 433.3 |
| 0000000100x | 8 | 249.4 | 325 |
| 0000000101x | 10 | 199.5 | 260 |
| 0000000110x | 12 | 166.3 | 216.7 |
| 0000000111x | 14 | 142.5 | 185.7 |
| 0000001000x | 16 | 124.7 | 162.5 |
| 0000001001x | 18 | 110.8 | 144.4 |
| ... | Even N | 1995 / N | 2600 / N |
| 1111111111x | 2046 | 0.98 | 1.27 |

Frequency Synthesizer Mode

This mode of operation allows an arbitrary output frequency to be generated from external REFCLK input. For improved phase noise performance, the REFCLK input frequency is doubled. As can be seen from the block diagram in *Figure 1*, only the upper feedback loop is used in this mode of operation. It is recommended that CLK0 and CLK1 be left unused in this mode of operation.

The upper feedback loop supports a delta-sigma fractional feedback divider. This allows the VCO operating frequency to be a non-integer multiple of the REFCLK frequency. By using an integer multiple only, lower phase noise jitter on the output can be achieved, however the use of the delta-sigma divider logic will provide excellent performance on the output if a fractional divisor is used.

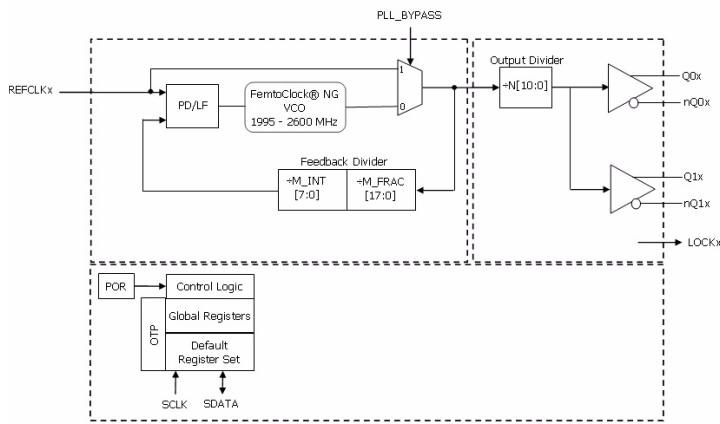


Figure 1. Frequency Synthesizer Mode Block Diagram

High-Bandwidth Frequency Translator Mode

This mode of operation is used to translate one of two input clocks of the same nominal frequency into an output frequency. As can be seen from the block diagram in Figure 2, similarly to the Frequency Synthesizer mode, only the upper feedback loop is used.

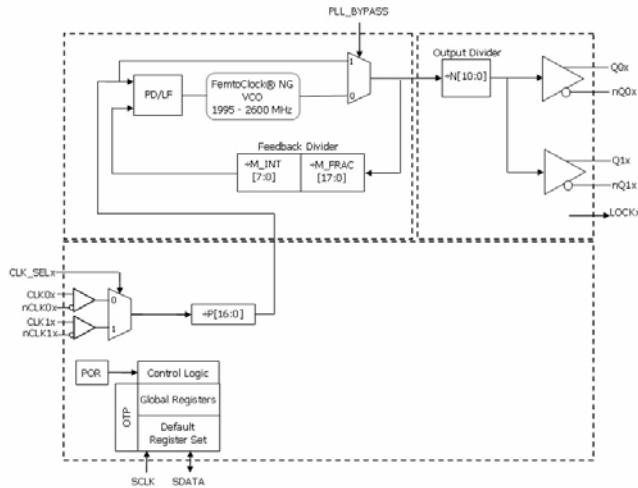


Figure 2. High Bandwidth Frequency Translator Mode Block Diagram

The input reference frequency range is now extended up to 710MHz. A pre-divider stage P is needed to keep the operating frequencies at the phase detector within limits.

Low-Bandwidth Frequency Translator Mode

As can be seen from the block diagram in Figure 3, this mode involves two PLL loops. The lower loop with the large integer dividers is the low bandwidth loop and it sets the output-to-input frequency translation ratio. This loop drives the upper DCXO loop (digitally controlled crystal oscillator) via an analog-digital converter.

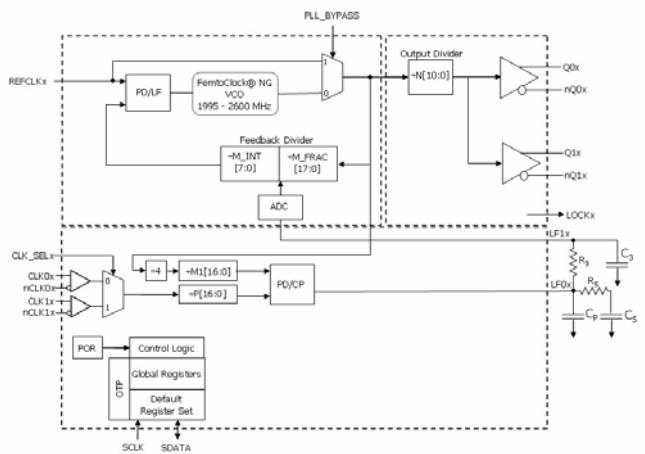


Figure 3. Low Bandwidth Frequency Translator Mode Block Diagram

The phase detector of the lower loop is designed to work with frequencies in the 8kHz - 16kHz range. The pre-divider stage is used to scale down the input frequency by an integer value to achieve a frequency in this range. By dividing down the fed-back VCO operating frequency by the integer divider M1x[18:0] to as close as possible to the same frequency, exact output frequency translations can be achieved.

Alarm Conditions & Status Bits

Each PLL of IDT8T49N244I monitors a number of conditions and reports their status via output pins and/or register bits. All alarms will behave as indicated below in all modes of operation, but some of the conditions monitored have no valid meaning in some operating modes. For example, the status of CLK0BADx, CLK1BADx and CLK_ACTIVEEx are not relevant in Frequency Synthesizer mode. The outputs will still be active and it is left to the user to determine which to monitor and how to respond to them based on the known operating mode.

CLK_ACTIVEEx - indicates which input clock reference is being used to derive the output frequency.

LOCKx - This status is asserted on the pin & register bit when the PLL is locked to the appropriate input reference for the chosen mode of operation. The status bit will not assert until frequency lock has been achieved, but will de-assert once lock is lost.

REFBAD - indicates if valid edges are being received on the REFCLK

input. Detection is performed by comparing the input to the feedback signal at the upper loop's Phase / Frequency Detector (PFD). If three edges are received on the feedback without an edge on the REFCLK input, the REFBAD alarm is asserted in register bit. Once an edge is detected on the REFCLK input, the alarm is immediately deasserted.

CLK0BADx - indicates if valid edges are being received on the CLK0 reference input. Detection is performed by comparing the input to the feedback signal at the appropriate Phase / Frequency Detector (PFD). When operating in high-bandwidth mode, the feedback at the upper PFD is used. In low-bandwidth mode, the feedback at the lower PFD is used. If three edges are received on the feedback without an edge on the divided down ($\div P$) CLK0x reference input, the CLK0BADx alarm is asserted on register bit. Once an edge is detected on the CLK0x reference input, the alarm is deasserted.

CLK1BADx - indicates if valid edges are being received on the CLK1x reference input. Behavior is as indicated for the CLK0BADx alarm, but with the CLK1x input being monitored and the CLK1BADx register bits being affected.

HOLDOVERx - indicates that the device is not locked to a valid input reference clock. This can occur in Manual switchover mode if the selected reference input has gone bad, even if the other reference input is still good. In automatic mode, this will only assert if both input references are bad.

Input Reference Selection and Switching

When operating in Frequency Synthesizer mode, the CLK0x and CLK1x inputs are not used and the contents of this section do not apply. Except as noted below, when operating in either High or Low Bandwidth Frequency Translator mode, the contents of this section apply equally when in either of those modes.

Both input references CLK0x and CLK1x must be the same nominal frequency. These may be driven by any type of clock source. A difference in frequency may cause the PLL to lose lock when switching between input references. Please contact IDT for the exact limits for your situation.

Each PLL has global control bits AUTO_MANx[1:0] to dictate the order of priority and switching mode to be used between the CLK0x and CLK1x inputs.

Manual Switching Mode

When the AUTO_MANx[1:0] field is set to Manual via Pin, then each PLL of IDT8T49N244I will use the CLK_SELx input pin to determine which input to use as a reference. Similarly, if set to Manual via Register, then the device will use the CLK_SELx register bit to determine the input reference. In either case, the PLL will lock to the selected reference if there is a valid clock present on that input.

If there is not a valid clock present on the selected input, each PLL of IDT8T49N244I will go into holdover (Low Bandwidth Frequency Translator mode) or free-run (High Bandwidth Frequency Translator mode) state. In either case, the HOLDOVERx alarm will be raised. This will occur even if there is a valid clock on the non-selected reference input. The device will recover from holdover / free-run state once a valid clock is re-established on the selected reference input.

Each PLL of IDT8T49N244I will only switch input references on command from the user. The user must either change the CLK_SELx register bit (if in Manual via Register) or CLK_SELx input pin (if in Manual via Pin).

Automatic Switching Mode

When the AUTO_MANx[1:0] field is set to either of the automatic selection modes (Revertive or Non-Revertive), each PLL of IDT8T49N244I determines which input reference it prefers / starts from by the state of the CLK_SELx register bit only. The CLK_SELx input pin is not used in either Automatic switching mode.

When starting from an unlocked condition, the device will lock to the input reference indicated by the CLK_SELx register bit. It will not pay attention to the non-selected input reference until a locked state has been achieved. This is necessary to prevent 'hunting' behavior during the locking phase.

Once the PLL of IDT8T49N244I has achieved a stable lock, it will remain locked to the preferred input reference as long as there is a valid clock on it. If at some point, that clock fails, then the device will automatically switch to the other input reference as long as there is a valid clock there. If there is not a valid clock on either input reference, each PLL of IDT8T49N244I will go into holdover (Low Bandwidth Frequency Translator mode) or free-run (High Bandwidth Frequency Translator mode) state. In either case, the HOLDOVER alarm will be raised.

The PLL will recover from holdover / free-run state once a valid clock is re-established on either reference input. If clocks are valid on both input references, the device will choose the reference indicated by the CLK_SELx register bit.

If running from the non-preferred input reference and a valid clock returns, there is a difference in behavior between Revertive and Non-revertive modes. In Revertive mode, the device will switch back to the reference indicated by the CLK_SELx register bit even if there is still a valid clock on the non-preferred reference input. In Non-revertive mode, each PLL IDT8T49N244I will not switch back as long as the non-preferred input reference still has a valid clock on it.

Switchover Behavior of the PLL

Even though the two input references have the same nominal frequency, there may be minor differences in frequency and potentially large differences in phase between them. Each PLL of IDT8T49N244I will adjust its output to the new input reference. It will use Phase Slope Limiting to adjust the output phase at a fixed maximum rate until the output phase and frequency are now aligned to the new input reference. Phase will always be adjusted by extending the clock period of the output so that no unacceptably short clock periods are generated on the output.

Holdover / Free-run Behavior

When both input references have failed (Automatic mode) or the selected input has failed (Manual mode), each PLL of IDT8T49N244I will enter holdover (Low Bandwidth Frequency Translator mode) or free-run (High Bandwidth Frequency Translator mode) state. In both cases, once the input reference is lost, the PLL will stop making adjustments to the output phase.

If operating in Low Bandwidth Frequency Translation mode, the PLL will continue to reference itself to the local reference and will hold its output phase and frequency in relation to that source. Output stability is determined by the stability of the local reference REFCLK in this case.

However, if operating in High Bandwidth Frequency Translation mode, the PLL no longer has any frequency reference to use and output stability is now determined by the stability of the internal VCO.

If the device is programmed to perform Manual switching, once the selected input reference recovers, the IDT8T49N244I will switch back to that input reference. If programmed for either Automatic mode, the device will switch back to whichever input reference has a valid clock first.

Output Configuration

The two outputs of each PLL both provide the same clock frequency. Both outputs must operate from the same output voltage level of 2.5V or 3.3V. This output voltage must be the same as or less than the core voltage the rest of the device is operating from. The output voltage level used on the two outputs is supplied on the V_{CCOX} pin.

The two outputs are individually selectable as LVDS or LVPECL output types via the Q0_TYPEEx and Q1_TYPEEx register bits. When the output is disabled, it will show a high impedance condition.

The two outputs can be enabled individually also via register control bit OE0x and OE1x respectively. When the differential output is in the disabled state, it will show a high impedance condition.

Serial Interface Configuration Description

The IDT8T49N244I has an I²C-compatible configuration interface to access any of the internal registers (Table 4D) for frequency and PLL parameter programming. Each PLL acts as a slave device on the I²C bus and has the address 0b11011xx, where xx is set to fixed value by A0 & A1 (see Table 4A for details). The interface accepts byte-oriented block write and block read operations. An address byte (P) specifies the register address (Table 4D) as the byte position of the first register to write or read. Data bytes (registers) are accessed in sequential order from the lowest to the highest byte (most significant bit first, see table 4B, 4C). Read and write block transfers can be stopped after any complete byte transfer. It is recommended to terminate I²C the read or write transfer after accessing byte #23 of each PLL.

For full electrical I²C compliance, it is recommended to use external pull-up resistors for SDATA and SCLK.

Note: if a different device slave address is desired, please contact IDT.

Table 4A. I²C Device Slave Address

| | A6 | A5 | A4 | A3 | A2 | A1 | A0 | R/W |
|-------|----|----|----|----|----|----|----|-----|
| PLL-A | 1 | 1 | 0 | 1 | 1 | 1 | 0 | R/W |
| PLL-B | 1 | 1 | 0 | 1 | 1 | 1 | 1 | R/W |

Table 4B. Block Write Operation

| Bit | 1 | 2:8 | 9 | 10 | 11:18 | 19 | 20:27 | 28 | 29-36 | 37 | ... | ... | ... |
|---------------|-------|---------------|-------|-----|------------------|-----|---------------|-----|-----------------|-----|---------------|-----|------|
| Description | START | Slave Address | W (0) | ACK | Address Byte (P) | ACK | Data Byte (P) | ACK | Data Byte (P+1) | ACK | Data Byte ... | ACK | STOP |
| Length (bits) | 1 | 7 | 1 | 1 | 8 | 1 | 8 | 1 | 8 | 1 | 8 | 1 | 1 |

Table 4C. Block Read Operation

| Bit | 1 | 2:8 | 9 | 10 | 11:18 | 19 | 20 | 21:27 | 28 | 29 | 30:37 | 38 | 39-46 | 47 | ... | ... | ... |
|---------------|-------|---------------|-------|-------|------------------|-------|----------------|---------------|-------|-------|---------------|-------|-----------------|-------|---------------|-------|------|
| Description | START | Slave Address | W (0) | A C K | Address Byte (P) | A C K | Repeated START | Slave Address | R (1) | A C K | Data Byte (P) | A C K | Data Byte (P+1) | A C K | Data Byte ... | A C K | STOP |
| Length (bits) | 1 | 7 | 1 | 1 | 8 | 1 | 1 | 7 | 1 | 1 | 8 | 1 | 8 | 1 | 8 | 1 | 1 |

Register Descriptions

Please consult IDT fro configuration software and/or guides to assist in selection of optimal register settings for the desired configurations. The below register map applies to each PLL. Register bit names have

'x' at the end of the name. This should be replaced by 'A' or 'B' as appropriate to the PLL being addressed at the time.

Table 4D. I²C Register Map

| Reg | Binary Register Address | Register Bit | | | | | | | |
|-----|-------------------------|--------------|--------------|--------------|------------|---------------|---------------|-------------|-------------|
| | | D7 | D6 | D5 | D4 | D3 | D2 | D1 | D0 |
| 0 | 00000 | MFRACx[17] | MFRACx[16] | MFRACx[15] | MFRACx[14] | MFRACx[13] | MFRACx[12] | MFRACx[11] | MFRACx[10] |
| 1 | 00001 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 2 | 00010 | MFRACx[9] | MFRACx[8] | MFRACx[7] | MFRACx[6] | MFRACx[5] | MFRACx[4] | MFRACx[3] | MFRACx[2] |
| 3 | 00011 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 4 | 00100 | MFRACx[1] | MFRACx[0] | MINTx[7] | MINTx[6] | MINTx[5] | MINTx[4] | MINTx[3] | MINTx[2] |
| 5 | 00101 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 6 | 00110 | MINTx[1] | MINTx[0] | Px[16] | Px[15] | Px[14] | Px[13] | Px[12] | Px[11] |
| 7 | 00111 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 8 | 01000 | Px[10] | Px[9] | Px[8] | Px[7] | Px[6] | Px[5] | Px[4] | Px[3] |
| 9 | 01001 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 10 | 01010 | Px[2] | Px[1] | Px[0] | M1_x[16] | M1_x[15] | M1_x[14] | M1_x[13] | M1_x[12] |
| 11 | 01011 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 12 | 01100 | M1_x[11] | M1_x[10] | M1_x[9] | M1_x[8] | M1_x[7] | M1_x[6] | M1_x[5] | M1_x[4] |
| 13 | 01101 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 14 | 01110 | M1_x[3] | M1_x[2] | M1_x[1] | M1_x[0] | Nx[10] | Nx[9] | Nx[8] | Nx[7] |
| 15 | 01111 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 16 | 10000 | Nx[6] | Nx[5] | Nx[4] | Nx[3] | Nx[2] | Nx[1] | Nx[0] | BWx[6] |
| 17 | 10001 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 18 | 10010 | BWx[5] | BWx[4] | BWx[3] | BWx[2] | BWx[1] | BWx[0] | Q1_TYPEEx | Q0_TYPEEx |
| 19 | 10011 | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd | Rsvd |
| 20 | 10100 | MODE_SELx[1] | MODE_SELx[0] | 0 | 1 | OE1x | OE0x | Rsvd | Rsvd |
| 21 | 10101 | CLK_SELx | AUTO_MANx[1] | AUTO_MANx[0] | 0 | ADC_RATEEx[1] | ADC_RATEEx[0] | LCK_WINx[1] | LCK_WINx[0] |
| 22 | 10110 | 1 | 0 | 1 | 0 | DBL_REFCLKx | 0 | 0 | 1 |
| 23 | 10111 | CLK_ACTIVEx | HOLDOVERx | CLK1BADx | CLK0BADx | REFBADx | LOCKx | Rsvd | Rsvd |

Table 4E. Configuration-Specific Control Bits

| Register Bits | Function |
|---------------|--|
| Q0_TYPEEx | Determines the output type for output pair Q0, nQ0 for PLLx 0 = LVPECL 1 = LVDS |
| Q1_TYPEEx | Determines the output type for output pair Q1, nQ1 for PLLx. 0 = LVPECL 1 = LVDS |
| Px[16:0] | Reference Pre-Divider for PLLx. |

| Register Bits | Function |
|---------------|--|
| M1_x[16:0] | Integer Feedback Divider in Lower Feedback Loop for PLLx. |
| M_INTx[7:0] | Feedback Divider, Integer Value in Upper Feedback Loop for PLLx. |
| M_FRACx[17:0] | Feedback Divider, Fractional Value in Upper Feedback Loop for PLLx. |
| Nx[10:0] | Output Divider for PLLx. |
| BWx[6:0] | Internal Operation Settings for PLLx. Please use IDT IDT8T49N244I Configuration Software to determine the correct settings for these bits for the specific configuration. Alternatively, please consult with IDT directly for further information on the functions of these bits. The function of these bits are explained in Tables 4H and 4I. |

Table 4F. Global Control Bits

| Register Bits | Function |
|----------------|--|
| MODE_SELx[1:0] | PLL Mode Select for PLLx 00 = Low Bandwidth Frequency Translator 01 = Frequency Synthesizer 10 = High Bandwidth Frequency Translator 11 = High Bandwidth Frequency Translator |
| OE0x | Output Enable Control for Output 0 for PLLx. 0 = Output Q0, nQ0 disabled 1 = Output Q0, nQ0 enabled |
| OE1x | Output Enable Control for Output 1 for PLLx. 0 = Output Q1, nQ1 disabled 1 = Output Q1, nQ1 enabled |
| Rsvd | Reserved bits - user should write a '0' to these bit positions if a write to these registers is needed |
| AUTO_MANx[1:0] | Selects how input clock selection is performed for PLLx. 00 = Manual Selection via pin only 01 = Automatic, non-revertive 10 = Automatic, revertive 11 = Manual Selection via register only |
| CLK_SELx | In manual clock selection via register mode for PLLx, this bit will command which input clock is selected. In the automatic modes, this indicates the primary clock input. In manual selection via pin mode, this bit has no effect. 0 = CLK0 1 = CLK1 |
| ADC_RATEx[1:0] | Sets the ADC sampling rate in Low-Bandwidth Mode as a fraction of the REFCLK input frequency for PLLx 00 = REFCLK Frequency / 16 when doubler is disabled 01 = REFCLK Frequency / 8 when doubler is disabled 10 = REFCLK Frequency / 4 (recommended) when doubler is disabled 11 = REFCLK Frequency / 2 when doubler is disabled |
| LCK_WINx[1:0] | Sets the width of the window in which a new reference edge must fall relative to the feedback edge for PLLx: 00 = 2usec (recommended), 01 = 4usec, 10 = 8usec, 11 = 16usec |
| DBL_REFCLKx | When set, this bit will double the frequency of the REFCLK input before applying it to the Phase-Frequency Detector for PLLx |

Table 4G. Global Status Bits

| Register Bits | Function |
|---------------|--|
| CLK0BADx | Status Bit for input clock 0 for PLLx. 0 = input 0 good 1 = input 0 bad. Self clears when input clock returns to good status |
| CLK1BADx | Status Bit for input clock 1 for PLLx. 0 = input 0 good 1 = input 0 bad. Self clears when input clock returns to good status |
| REFBADx | Status Bit for PLLx. 0 = REFCLK input good 1 = REFCLK input bad. Self-clears when the REFCLK clock returns to good status |
| LOCKx | Status bit for PLLx. This bit is mirrored on LOCKx pin. 0 = PLL unlocked 1 = PLL locked |
| HOLDOVERx | Status Bit for PLLx. 0 = Input to phase detector is within specifications and device is tracking to it 1 = Phase detector input not within specifications and DCXO is frozen at last value |
| CLK_ACTIVE | Status Bit for PLLx. Indicates which input clock is active. Automatically updates during fail-over switching. |

Table 4H. BW[6:0] Bits

| Mode | BWx[6] | BWx[5] | BWx[4] | BWx[3] | BWx[2] | BWx[1] | BWx[0] |
|---------------------|---------------|---------------|---------------|---------------|--------------|--------------|---------------|
| Synthesizer Mode | PLL2_LF x[1] | PLL2_LF x[0] | DSM_ORDx | DSM_ENx | PLL2_CP x[1] | PLL2_CP x[0] | PLL2_LOW_ICPx |
| High-Bandwidth Mode | PLL2_LF x[1] | PLL2_LF x[0] | DSM_ORDx | DSM_ENx | PLL2_CP x[1] | PLL2_CP x[0] | PLL2_LOW_ICPx |
| Low-Bandwidth Mode | ADC_GAIN x[3] | ADC_GAIN x[2] | ADC_GAIN x[1] | ADC_GAIN x[0] | PLL1_CP x[1] | PLL1_CP x[0] | PLL2_LOW_ICPx |

Table 4I. Functions of Fields in BW[6:0]

| Register Bits | Function |
|---------------|--|
| PLL2_LFx[1:0] | Sets loop filter values for upper loop PLL in Frequency Synthesizer & High-Bandwidth modes for PLLx. Defaults to setting of 00 when in Low Bandwidth Mode. See Table 4J for settings. |
| DSM_ORDx | Sets Delta-Sigma Modulation to 2nd (0) or 3rd order (1) operation for PLLx. |
| DSM_ENx | Enables Delta-Sigma Modulator for PLLx: 0 = Disabled - feedback in integer mode only 1 = Enabled - feedback in fractional mode |
| PLL2_CPx[1:0] | Upper loop PLL charge pump current settings for PLLx: 00 = 173 μ A (defaults to this setting in Low Bandwidth Mode) 01 = 346 μ A 10 = 692 μ A 11 = reserved |

| | |
|----------------|---|
| PLL2_LOW_ICPx | Reduces Charge Pump current by 1/3 to reduce bandwidth variations resulting from higher feedback register settings or high VCO operating frequency (>2.4GHz) for PLLx. |
| ADC_GAINx[3:0] | Gain setting for ADC in Low Bandwidth Mode. |
| PLL1_CPx[1:0] | Lower loop PLL charge pump current settings (lower loop is only used in Low Bandwidth Mode) for PLLx: 00 = 800 μ A 01 = 400 μ A 10 = 200 μ A 11 = 100 μ A |

Table 4J. Upper Loop (PLL2) Bandwidth Settings

| Desired Bandwidth | PLL2_CP | PLL2ICP | PLL2_LF |
|---|---------|---------|---------|
| Frequency Synthesizer Mode | | | |
| 200kHz | 00 | 1 | 00 |
| 400kHz | 01 | 1 | 01 |
| 800kHz | 10 | 1 | 10 |
| 2MHz | 10 | 1 | 11 |
| High Bandwidth Frequency Translator Mode | | | |
| 200kHz | 00 | 1 | 00 |
| 400kHz | 01 | 1 | 01 |
| 800kHz | 10 | 1 | 10 |
| 4MHz | 10 | 0 | 11 |

NOTE: To achieve 4MHz bandwidth, reference to the phase detector should be 80MHz.

Absolute Maximum Ratings

NOTE: Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

| Item | Rating |
|--|------------------------------|
| Supply Voltage, V_{CC} | 3.6V |
| Inputs, V_I | -0.5V to $V_{CC_X} + 0.5V$ |
| Outputs, V_O (LVCMOS) | -0.5V to $V_{CCO_X} + 0.5V$ |
| Outputs, I_O (LVPECL) Continuous Current Surge Current | 50mA 100mA |
| Outputs, I_O (LVDS) Continuous Current Surge Current | 10mA 15mA |
| Package Thermal Impedance, θ_{JA} | 12.4°C/W (0 mps) |
| Package Thermal Impedance, θ_{JC} | 12.5°C/W (0 mps) |
| Storage Temperature, T_{STG} | -65°C to 150°C |

DC Electrical Characteristics

Table 5A. LVPECL Power Supply DC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------------|-----------------------------|-----------------|--------------------|---------|-------------|-------|
| V_{CC_X} | Core Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| V_{CCA_X} | Analog Power Supply Voltage | | $V_{CC_X} - 0.32$ | 3.3 | V_{CC_X} | V |
| V_{CCO_X} | Output Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| I_{EE} | Power Supply Current | | | | 624 | mA |
| I_{CCA} | Analog Power Supply Current | | | | 64 | mA |

NOTE: X denotes: A or B.

Table 5B. LVPECL Power Supply DC Characteristics, $V_{CC_X} = 3.3V \pm 5\%$, $V_{CCO_X} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------------|-----------------------------|-----------------|--------------------|---------|-------------|-------|
| V_{CC_X} | Core Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| V_{CCA_X} | Analog Power Supply Voltage | | $V_{CC_X} - 0.32$ | 3.3 | V_{CC_X} | V |
| V_{CCO_X} | Output Supply Voltage | | 2.375 | 2.5 | 2.625 | V |
| I_{EE} | Power Supply Current | | | | 624 | mA |
| I_{CCA} | Analog Power Supply Current | | | | 64 | mA |

NOTE: X denotes: A or B.

Table 5C. LVPECL Power Supply DC Characteristics, $V_{CC_X} = V_{CCO_X} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------------|-----------------------------|-----------------|--------------------|---------|-------------|-------|
| V_{CC_X} | Core Supply Voltage | | 2.375 | 2.5 | 2.625 | V |
| V_{CCA_X} | Analog Power Supply Voltage | | $V_{CC_X} - 0.29$ | 2.5 | V_{CC_X} | V |
| V_{CCO_X} | Output Supply Voltage | | 2.375 | 2.5 | 2.625 | V |
| I_{EE} | Power Supply Current | | | | 589 | mA |
| I_{CCA} | Analog Power Supply Current | | | | 58 | mA |

NOTE: X denotes: A or B.

Table 5D. LVDS Power Supply DC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------------|-----------------------------|-----------------|--------------------|---------|-------------|-------|
| V_{CC_X} | Core Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| V_{CCA_X} | Analog Power Supply Voltage | | $V_{CC_X} - 0.32$ | 3.3 | V_{CC_X} | V |
| V_{CCO_X} | Output Supply Voltage | | 3.135 | 3.3 | 3.465 | V |
| I_{EE} | Power Supply Current | | | | 526 | mA |
| I_{CCA} | Analog Power Supply Current | | | | 64 | mA |
| I_{CCO} | Output Power Supply Current | | | | 88 | mA |

NOTE: X denotes: A or B.

Table 5E. LVDS Power Supply DC Characteristics, $V_{CC_X} = V_{CCO_X} = 2.5V \pm 5\%$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|--------------|-----------------------------|-----------------|--------------------|---------|-------------|-------|
| V_{CC_X} | Core Supply Voltage | | 2.375 | 2.5 | 2.625 | V |
| V_{CCA_X} | Analog Power Supply Voltage | | $V_{CC_X} - 0.29$ | 2.5 | V_{CC_X} | V |
| V_{CCO_X} | Output Supply Voltage | | 2.375 | 2.5 | 2.625 | V |
| I_{EE} | Power Supply Current | | | | 501 | mA |
| I_{CCA} | Analog Power Supply Current | | | | 58 | mA |
| I_{CCO} | Output Power Supply Current | | | | 88 | mA |

NOTE: X denotes: A or B.

Table 5F. LVCMS/LVTTL DC Characteristics, $T_A = -40^\circ\text{C}$ to 85°C

| Symbol | Parameter | Test Conditions | | Minimum | Typical | Maximum | Units |
|----------|----------------------------|------------------------------|---|---------|---------|----------------|---------------|
| V_{IH} | Input High Voltage | | $V_{CC_X} = 3.3\text{V}$ | 2 | | $V_{CC} + 0.3$ | V |
| | | | $V_{CC_X} = 2.5\text{V}$ | 1.7 | | $V_{CC} + 0.3$ | V |
| V_{IL} | Input Low Voltage | | $V_{CC_X} = 3.3\text{V}$ | -0.3 | | 0.8 | V |
| | | | $V_{CC_X} = 2.5\text{V}$ | -0.3 | | 0.7 | V |
| I_{IH} | Input High Current; NOTE 1 | PLL_BYPASS, REFCLK, CLK_SELx | $V_{CC_X} = V_{IN} = 3.465\text{V}$ or 2.625V | | | 150 | μA |
| | | SCLK, SDATA | $V_{CC_X} = V_{IN} = 3.465\text{V}$ or 2.625V | | | 5 | μA |
| I_{IL} | Input Low Current; NOTE 1 | PLL_BYPASS, REFCLK, CLK_SELx | $V_{CC_X} = 3.465\text{V}$ or 2.625V , $V_{IN} = 0\text{V}$ | -5 | | | μA |
| | | SCLK, SDATA | $V_{CC_X} = 3.465\text{V}$ or 2.625V , $V_{IN} = 0\text{V}$ | -150 | | | μA |
| V_{OH} | Output High Voltage | SDATA, LOCKx | $V_{CCO_X} = 3.3\text{V} \pm 5\%$, $I_{OH} = -8\text{mA}$ | 2.6 | | | V |
| | | | $V_{CCO_X} = 2.5\text{V} \pm 5\%$, $I_{OH} = -8\text{mA}$ | 1.8 | | | V |
| V_{OL} | Output Low Voltage | SDATA, LOCKx | $V_{CCO_X} = 3.3\text{V} \pm 5\%$ or $2.5\text{V} \pm 5\%$, $I_{OL} = 8\text{mA}$ | | | 0.5 | V |

NOTE: X denotes: A or B.

NOTE1: For PLL_BYPASS/REFCLK/SCLK/SDATA pins, specification is for each individual PLL and is guaranteed by design. Production Test is performed all PLLs combined

Table 5G. Differential DC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3\text{V} \pm 5\%$ or $2.5\text{V} \pm 5\%$, $V_{EE} = 0\text{V}$, $T_A = -40^\circ\text{C}$ to 85°C

| Symbol | Parameter | Test Conditions | | Minimum | Typical | Maximum | Units |
|-----------|-----------------------------------|---|---|----------------|---------|----------------|---------------|
| I_{IH} | Input High Current | CLK0[A:B], nCLK0[A:B], CLK1[A:B], nCLK1[A:B], $V_{CC_X} = V_{IN} = 3.465\text{V}$ or 2.625V | | | | 150 | μA |
| I_{IL} | Input Low Current | CLK0[A:B], CLK1[A:B], | $V_{CC_X} = 3.465\text{V}$ or 2.625V , $V_{IN} = 0\text{V}$ | | -5 | | μA |
| | | nCLK0[A:B], nCLK1[A:B] | $V_{CC_X} = 3.465\text{V}$ or 2.625V , $V_{IN} = 0\text{V}$ | | -150 | | μA |
| V_{PP} | Peak-to-Peak Voltage | | | 0.15 | | 1.3 | V |
| V_{CMR} | Common Mode Input Voltage; NOTE 1 | | | $V_{EE} + 0.5$ | | $V_{CC} - 1.0$ | V |

NOTE: X denotes: A or B.

NOTE 1: Common mode input voltage is defined as the cross point.

Table 5H. LVPECL DC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-------------|-----------------------------------|-----------------|--------------------|---------|--------------------|-------|
| V_{OH} | Output High Voltage; NOTE 1 | | $V_{CCO_X} - 1.1$ | | $V_{CCO_X} - 0.7$ | V |
| V_{OL} | Output Low Voltage NOTE 1 | | $V_{CCO_X} - 2.0$ | | $V_{CCO_X} - 1.5$ | V |
| V_{SWING} | Peak-to-Peak Output Voltage Swing | | 0.6 | | 1.0 | V |

NOTE: X denotes: A or B.

NOTE 1: Outputs terminated with 50Ω to $V_{CCO_X} - 2V$.**Table 5I. LVPECL DC Characteristics, $V_{CC_X} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, $V_{CCO_X} = 2.5V \pm 5\%$, $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C$**

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-------------|-----------------------------------|-----------------|--------------------|---------|--------------------|-------|
| V_{OH} | Output High Voltage; NOTE 1 | | $V_{CCO_X} - 1.1$ | | $V_{CCO_X} - 0.8$ | V |
| V_{OL} | Output Low Voltage NOTE 1 | | $V_{CCO_X} - 2.0$ | | $V_{CCO_X} - 1.5$ | V |
| V_{SWING} | Peak-to-Peak Output Voltage Swing | | 0.6 | | 1.0 | V |

NOTE: X denotes: A or B.

NOTE 1: Outputs terminated with 50Ω to $V_{CCO_X} - 2V$.**Table 5J. LVDS DC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $85^\circ C$**

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-----------------|-----------------------------|-----------------|---------|---------|---------|-------|
| V_{OD} | Differential Output Voltage | | 247 | | 454 | mV |
| ΔV_{OD} | V_{OD} Magnitude Change | | | | 50 | mV |
| V_{OS} | Offset Voltage | | 1.125 | | 1.375 | V |
| ΔV_{OS} | V_{OS} Magnitude Change | | | | 50 | mV |

NOTE: X denotes: A or B.

Table 5K. LVDS DC Characteristics, $V_{CC_X} = V_{CCO_X} = 2.5V \pm 5\%$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|-----------------|-----------------------------|-----------------|---------|---------|---------|-------|
| V_{OD} | Differential Output Voltage | | 247 | | 454 | mV |
| ΔV_{OD} | V_{OD} Magnitude Change | | | | 50 | mV |
| V_{OS} | Offset Voltage | | 1.125 | | 1.375 | V |
| ΔV_{OS} | V_{OS} Magnitude Change | | | | 50 | mV |

NOTE: X denotes: A or B.

Table 6. Input Frequency Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$, $T_A = -40^\circ C$ to $85^\circ C$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|----------|-----------------|--|---------------------|---------|---------|-------|
| f_{IN} | Input Frequency | REFCLK; NOTE 1 | 16 | | 40 | MHz |
| | | CLK0[B:D], nCLK0[B:D], CLK1[B:D], nCLK1[B:D], | High Bandwidth Mode | 16 | 710 | MHz |
| | | | Low Bandwidth Mode | 0.008 | 710 | MHz |
| | SCLK | | | | 5 | MHz |

NOTE: X denotes: A or B.

NOTE 1: For the input REFCLK and CLK0x, nCLK0x, CLK1x, cCLK1x frequency range, the M value must be set for the VCO to operate within the 1995MHz to 2600MHz range.

AC Electrical Characteristics

Table 7. AC Characteristics, $V_{CC_X} = V_{CCO_X} = 3.3V \pm 5\%$ or $2.5V \pm 5\%$, or $V_{CC_X} = 3.3V \pm 5\%$, $V_{CCO_X} = 2.5V \pm 5\%$ (LVPECL Only), $V_{EE} = 0V$, $T_A = -40^\circ C$ to $85^\circ C^*$

| Symbol | Parameter | Test Conditions | Minimum | Typical | Maximum | Units |
|------------------------------------|---|---|------------|---------|---------|--------|
| f_{OUT} | Output Frequency | | 0.98 | | 1300 | MHz |
| f_{VCO} | VCO Frequency | | 1995 | | 2600 | MHz |
| $\text{f}_{\text{jit}}(\emptyset)$ | RMS Phase Jitter (Random), Integer Divide Ratio; NOTE 1 | Synth Mode (Integer FB), $f_{OUT} = 400\text{MHz}$, 40MHz REFCLK, Integration Range: 12kHz – 20MHz | | 326 | 405 | fs |
| | | Synth Mode (FracN FB), $f_{OUT} = 698.81\text{MHz}$, 40MHz REFCLK, Integration Range: 12kHz – 20MHz | | 403 | 640 | fs |
| | | HBW Mode, $f_{IN} = 133.33\text{MHz}$, $f_{OUT} = 400\text{MHz}$, Integration Range: 12kHz – 20MHz | | 311 | 490 | fs |
| | | LBW Mode (near integer), 40MHz REFCLK, $f_{IN} = 19.44\text{MHz}$, $f_{OUT} = 622.08\text{MHz}$, Integration Range: 12kHz – 20MHz | | 453 | 685 | fs |
| | | LBW Mode (FracN), 40MHz REFCLK, $f_{IN} = 25\text{MHz}$, $f_{OUT} = 161.138125\text{MHz}$, Integration Range: 12kHz – 20MHz | | 486 | 685 | fs |
| $\text{f}_{\text{jit}}(\text{cc})$ | Cycle-to-Cycle Jitter; NOTE 2, 3 | Frequency Synthesizer Mode | | | 40 | ps |
| | | Frequency Translator Mode | | | 40 | ps |
| t_R / t_F | Output Rise/Fall Time; NOTE 3 | LVPECL Outputs | 20% to 80% | 40 | | 530 ps |
| | | LVDS Outputs | 20% to 80% | 80 | | 530 ps |
| odc | Output Duty Cycle; NOTE 3 | $f_{OUT} < 600\text{MHz}$ | 45 | | 55 | % |
| | | $f_{OUT} \geq 600\text{MHz}$ | 40 | | 60 | % |

NOTE: X denotes: A or B.

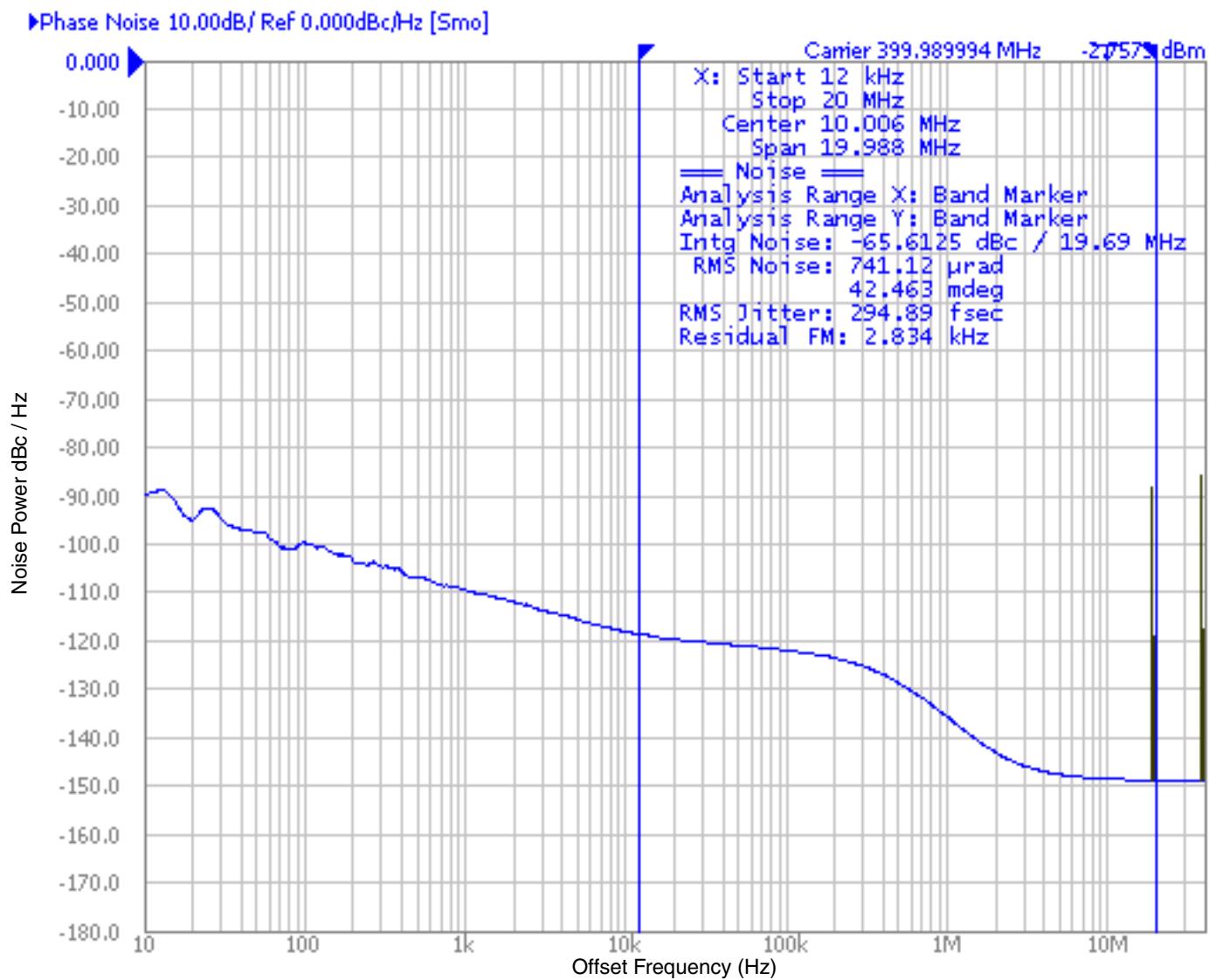
NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE 1: For RMS phase jitter measurement in Synth and HBW modes, both PLLA and PLLB are programmed with the same configuration. For LBW mode, only the PLL under test is programmed with LBW configuration. The other PLL is programmed in Synth mode with the same output frequency. A Rohde & Schwartz SMA-100 Signal generator, 9kHz to 6GHz is used as the REFCLK source. All configurations are with DBL_REFCLK bit set to 1.

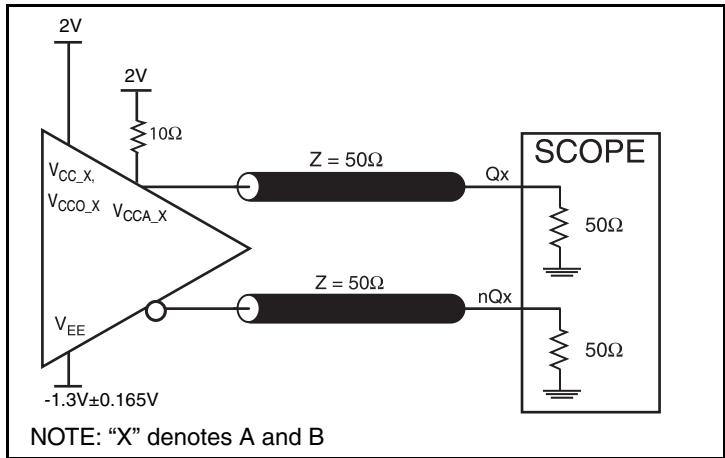
NOTE 2: This parameter is defined in accordance with JEDEC Standard 65.

NOTE 3: Measurements are collected with the following output frequency: 10MHz, 66.6667MHz, 125MHz, 156.25MHz, 161.138125MHz, 400MHz, 622.08MHz, 698.81MHz, 1300MHz.

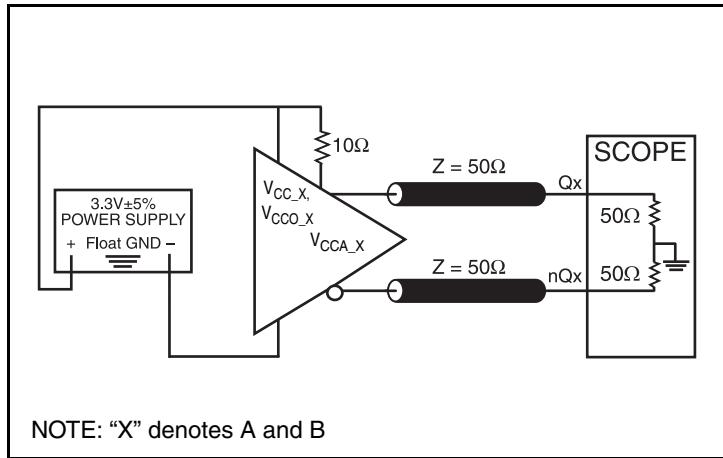
Typical Phase Noise at 400MHz (3.3V)



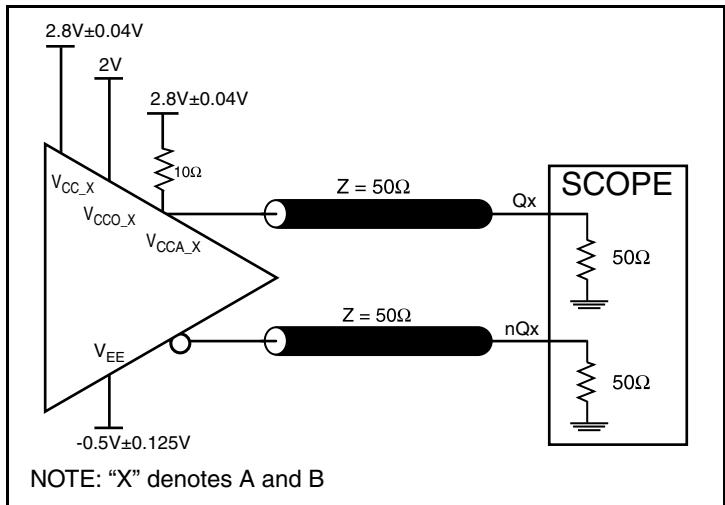
Parameter Measurement Information



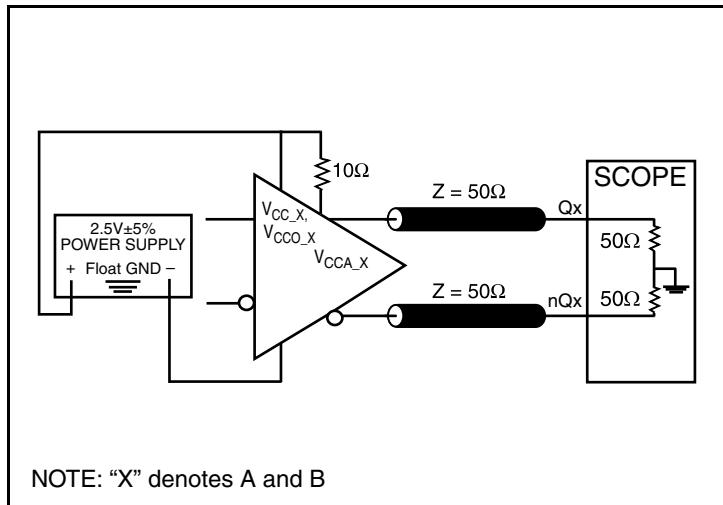
3.3 Core/3.3V LVPECL Output Load Test Circuit



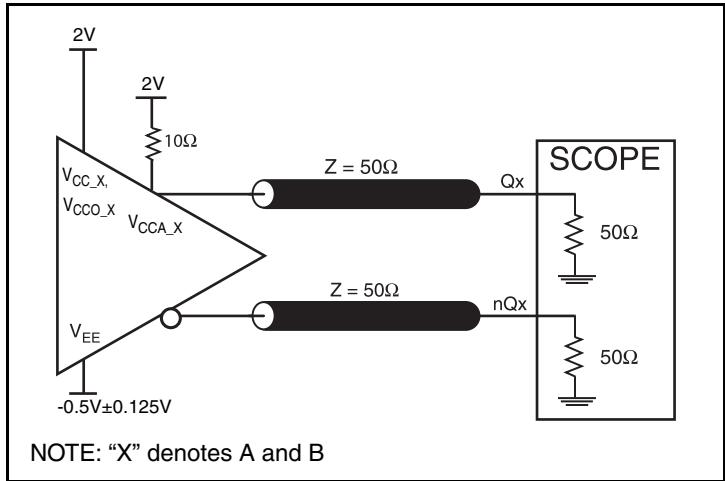
3.3 Core/3.3V LVDS Output Load Test Circuit



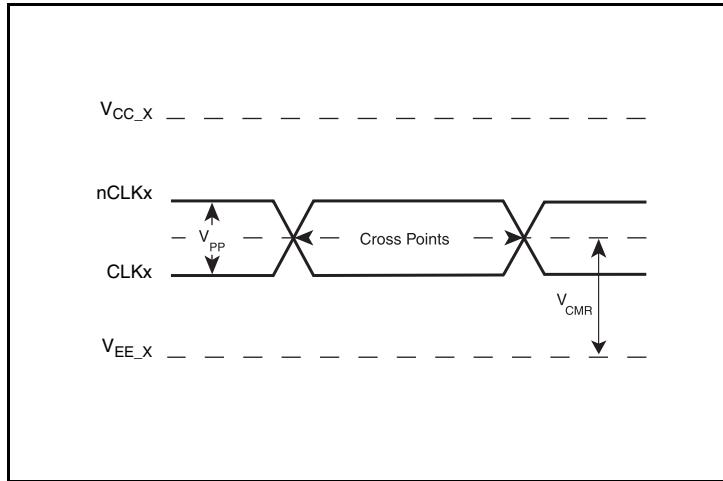
3.3 Core/2.5V LVPECL Output Load Test Circuit



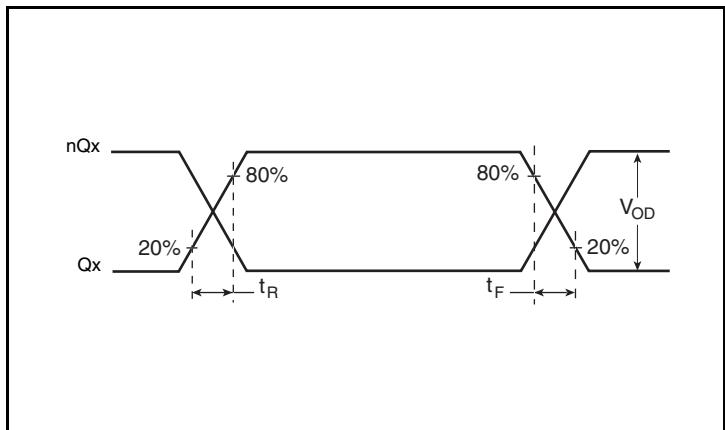
2.5 Core/2.5V LVDS Output Load Test Circuit



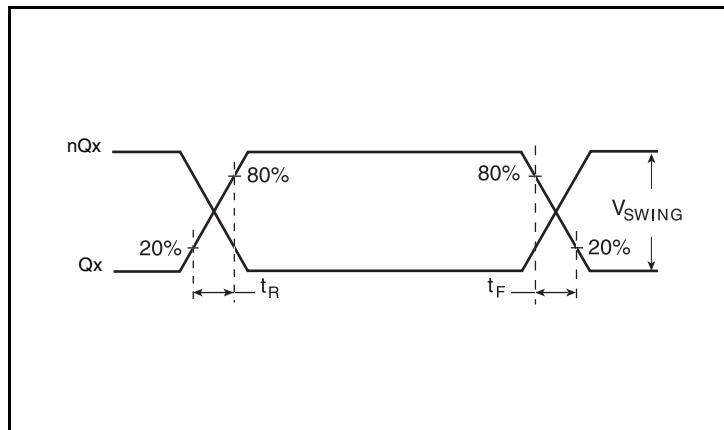
2.5 Core/2.5V LVPECL Output Load Test Circuit



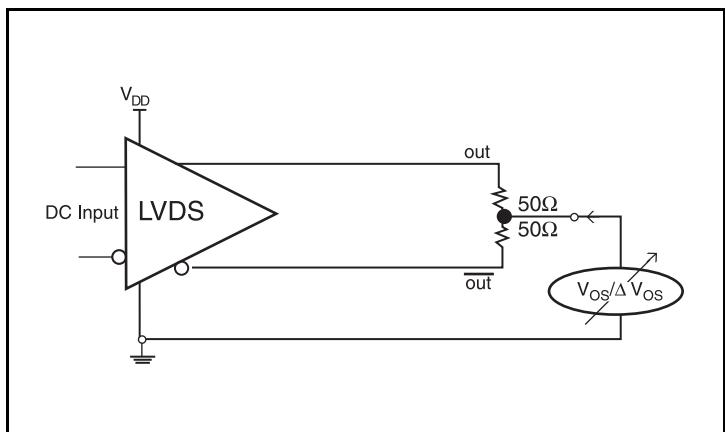
Parameter Measurement Information, continued



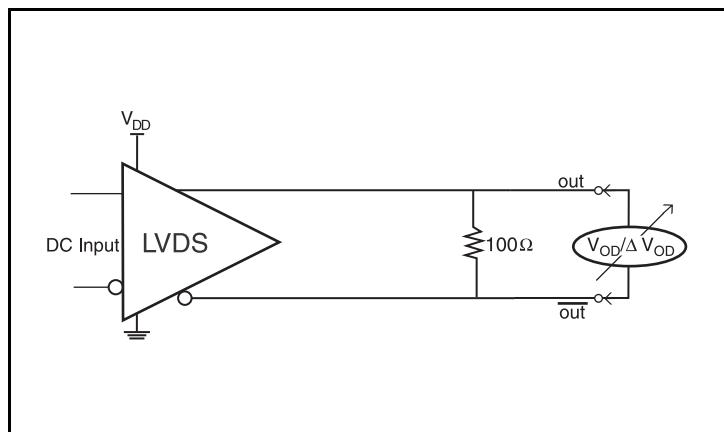
LVDS Output Rise/Fall Time



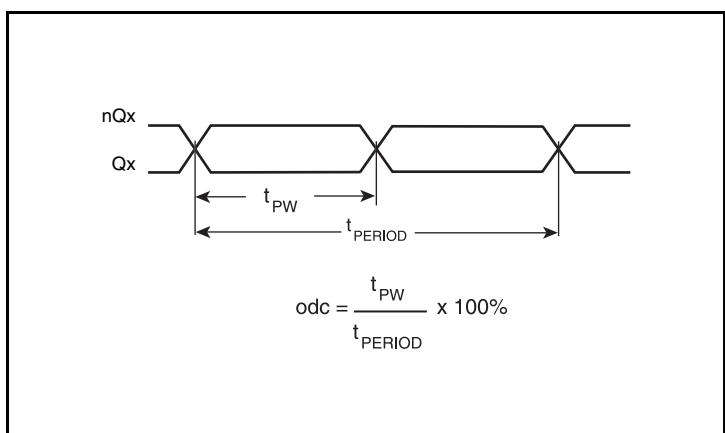
LVPECL Output Rise/Fall Time



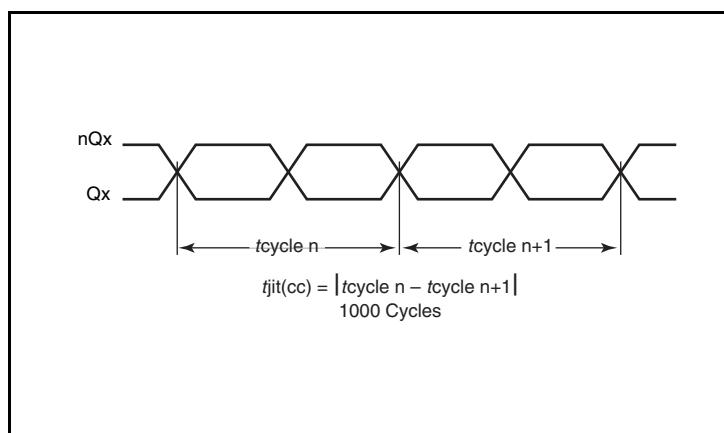
Offset Voltage Setup



Differential Output Voltage Setup



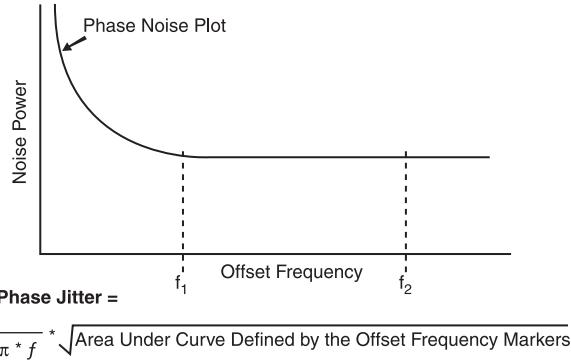
Differential Output Duty Cycle/Output Pulse Width/Period



Cycle-to-Cycle Jitter

Parameter Measurement Information, continued

RMS Phase Jitter



Applications Information

Recommendations for Unused Input and Output Pins

Inputs:

CLKx/nCLKx Inputs

For applications not requiring the use of either differential input, both CLKx and nCLKx can be left floating. Though not required, but for additional protection, a $1\text{k}\Omega$ resistor can be tied from CLKx to ground. It is recommended that CLKx, nCLKx be left unconnected in frequency synthesizer mode.

LVC MOS Control Pins

All control pins have internal pullups or pulldowns; additional resistance is not required but can be added for additional protection. A $1\text{k}\Omega$ resistor can be used.

Outputs:

LVPECL Outputs

All unused LVPECL output pairs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

LVDS Outputs

All unused LVDS output pairs can be either left floating or terminated with 100Ω across. If they are left floating there should be no trace attached.

LVC MOS Outputs

All unused LVC MOS output can be left floating. There should be no trace attached.

Recommended Values for Low-Bandwidth Mode Loop Filter

External loop filter components are not needed in Frequency Synthesizer or High-Bandwidth modes. In Low-Bandwidth mode, the loop filter structure and components are recommended, refer to the Application Schematic. Please consult IDT if other values are needed.

Wiring the Differential Input to Accept Single-Ended Levels

Figure 4 shows how a differential input can be wired to accept single ended levels. The reference voltage $V_1 = V_{CC}/2$ is generated by the bias resistors R1 and R2. The bypass capacitor (C1) is used to help filter noise on the DC bias. This bias circuit should be located as close to the input pin as possible. The ratio of R1 and R2 might need to be adjusted to position the V_1 in the center of the input voltage swing. For example, if the input clock swing is 2.5V and $V_{CC} = 3.3V$, R1 and R2 value should be adjusted to set V_1 at 1.25V. The values below are for when both the single ended swing and V_{CC} are at the same voltage. This configuration requires that the sum of the output impedance of the driver (R_o) and the series resistance (R_s) equals the transmission line impedance. In addition, matched termination at the input will attenuate the signal in half. This can be done in one of two ways. First, R3 and R4 in parallel should equal the transmission line

impedance. For most 50Ω applications, R3 and R4 can be 100Ω . The values of the resistors can be increased to reduce the loading for slower and weaker LVCMOS driver. When using single-ended signaling, the noise rejection benefits of differential signaling are reduced. Even though the differential input can handle full rail LVCMOS signaling, it is recommended that the amplitude be reduced. The datasheet specifies a lower differential amplitude, however this only applies to differential signals. For single-ended applications, the swing can be larger, however V_{IL} cannot be less than $-0.3V$ and V_{IH} cannot be more than $V_{CC} + 0.3V$. Though some of the recommended components might not be used, the pads should be placed in the layout. They can be utilized for debugging purposes. The datasheet specifications are characterized and guaranteed by using a differential signal.

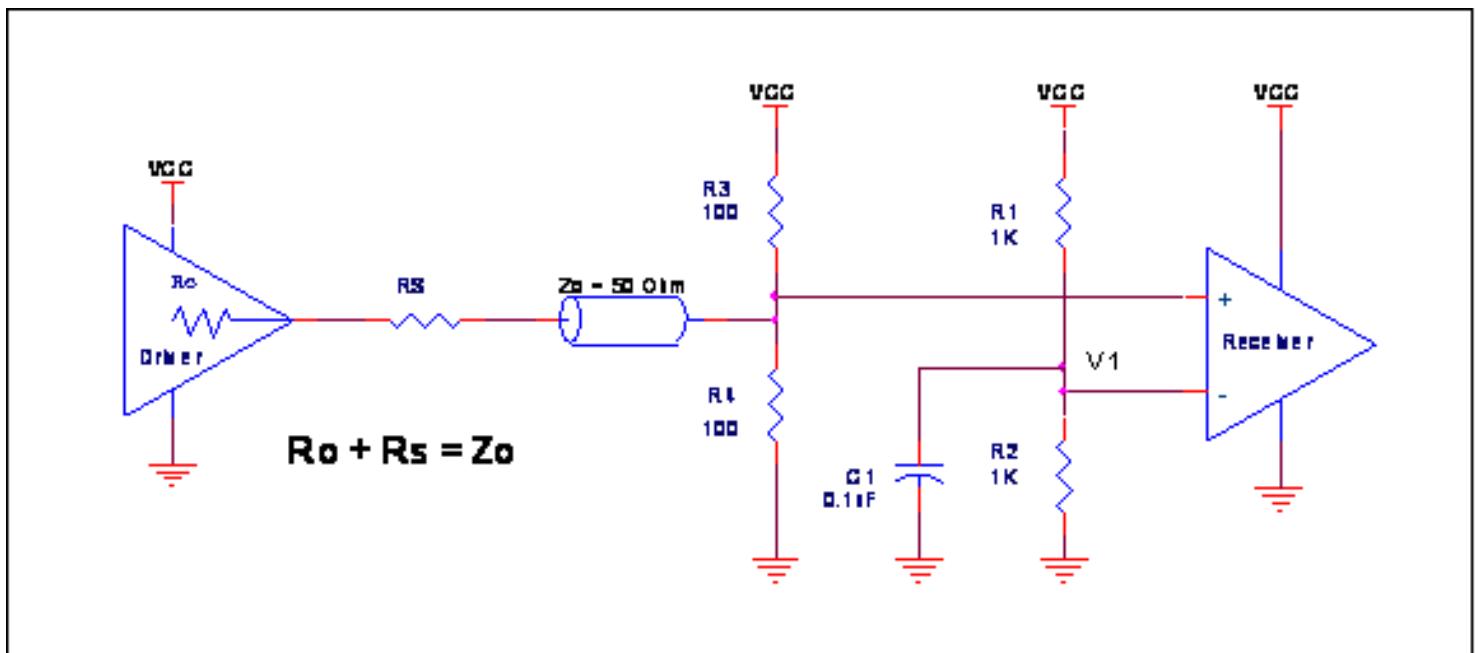


Figure 4. Recommended Schematic for Wiring a Differential Input to Accept Single-ended Levels

3.3V Differential Clock Input Interface

The CLKx /nCLKx inputs accept LVDS, LVPECL, LVHSTL, HCSL and other differential signals. Both differential signals must meet the V_{PP} and V_{CMR} input requirements. *Figures 5A to 5E* show interface examples for the CLK/nCLK input driven by the most common driver types. The input interfaces suggested here are examples only.

Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in Figure 5A, the input termination applies for IDT open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

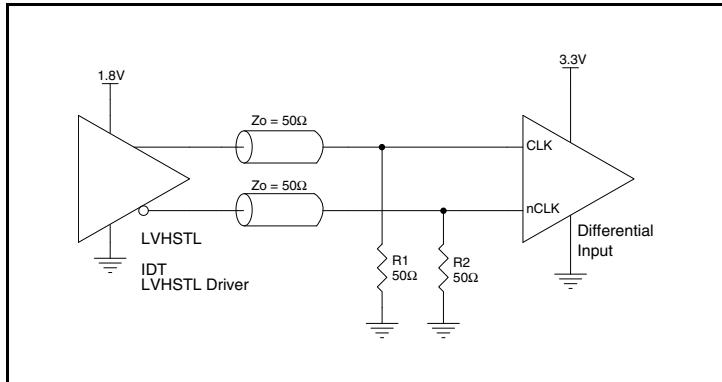


Figure 5A. CLK/nCLK Input driven by an IDT Open Emitter LVHSTL driver

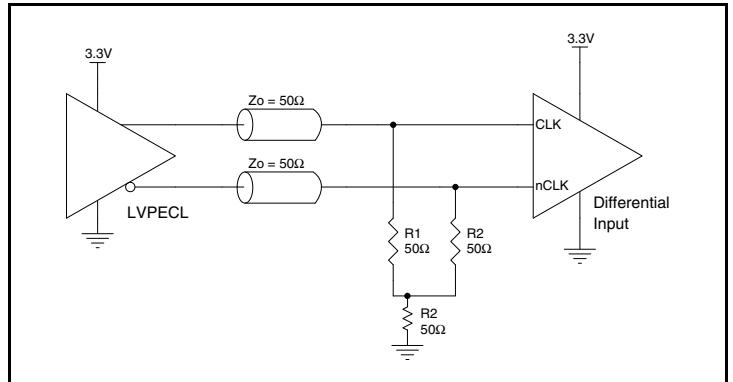


Figure 5B. CLK/nCLK Input driven by a 3.3V LVPECL driver

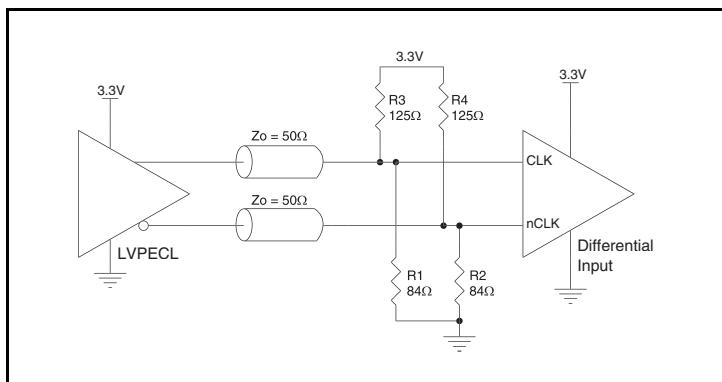


Figure 5C. CLK/nCLK Input driven by a 3.3V LVPECL driver

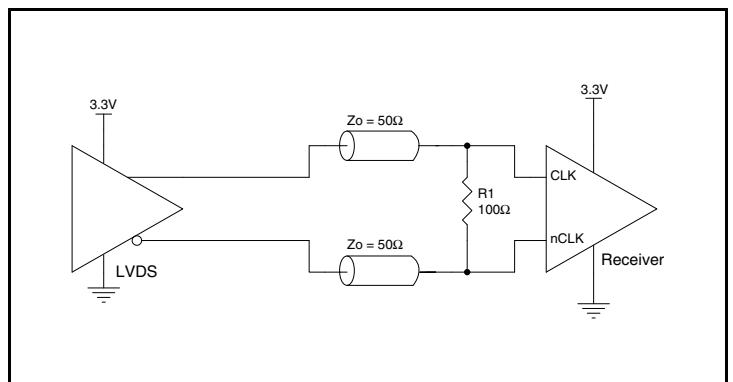


Figure 5D. CLK/nCLK Input driven by a 3.3V LVDS driver

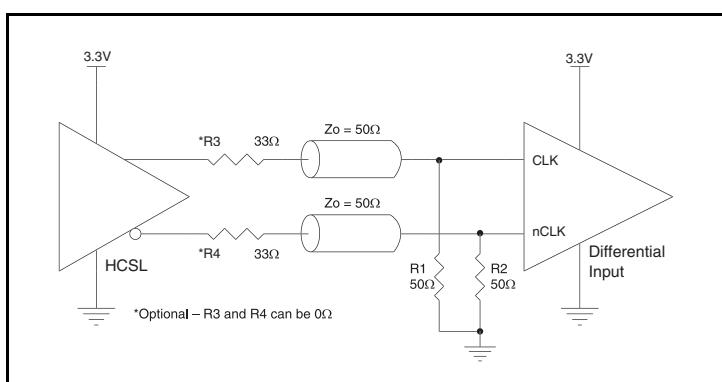


Figure 5E. CLK/nCLK Input driven by a 3.3V HCSL driver

2.5V Differential Clock Input Interface

Each CLKnx / nCLKnx input accepts LVDS, LVPECL, LVHSTL, HCSL and other differential signals. Both V_{SWING} and V_{OH} must meet the V_{PP} and V_{CMR} input requirements. *Figures 6A to 6E* show interface examples for the CLKnx / nCLKnx input driven by the most common driver types. The input interfaces suggested here are examples only.

Please consult with the vendor of the driver component to confirm the driver termination requirements. For example, in Figure 6A, the input termination applies for IDT open emitter LVHSTL drivers. If you are using an LVHSTL driver from another vendor, use their termination recommendation.

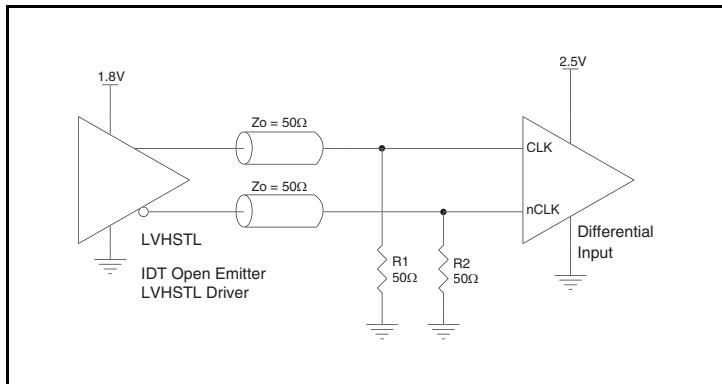


Figure 6A. CLKnx / nCLKnx Input driven by an IDT Open Emitter LVHSTL driver

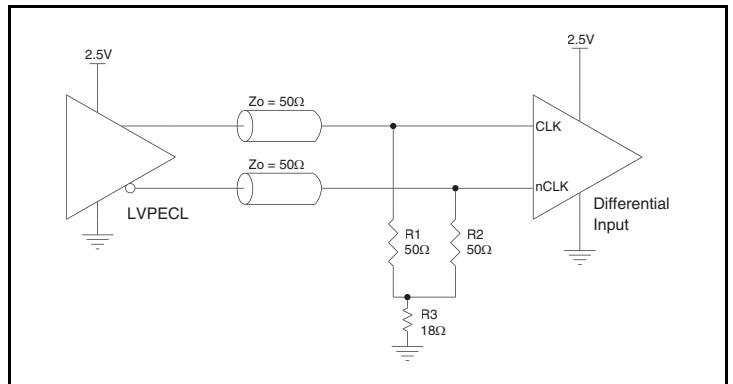


Figure 6B. CLKnx / nCLKnx Input driven by a 2.5V LVPECL driver

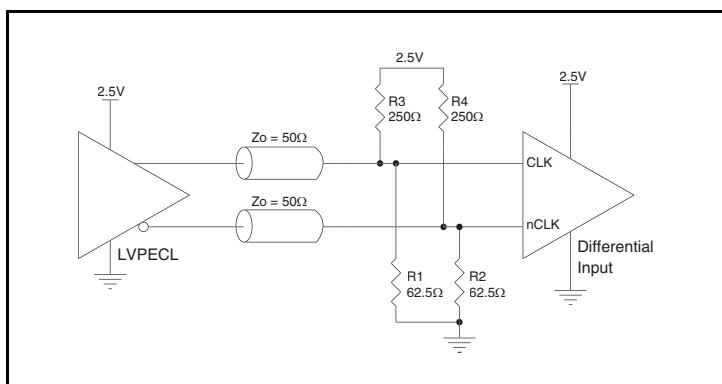


Figure 6C. CLKnx / nCLKnx Input driven by a 2.5V LVPECL driver

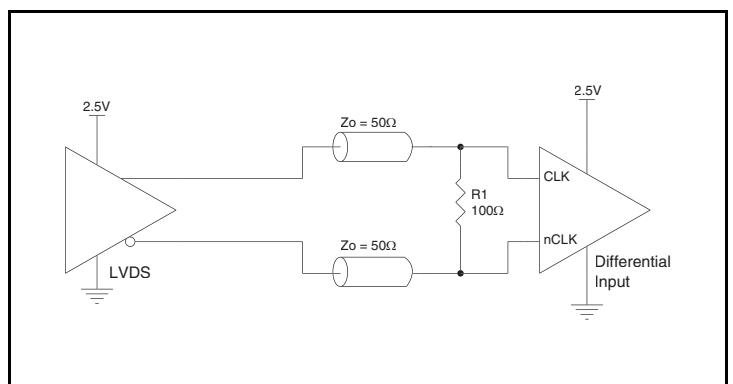


Figure 6D. CLKnx / nCLKnx Input driven by a 2.5V LVDS driver

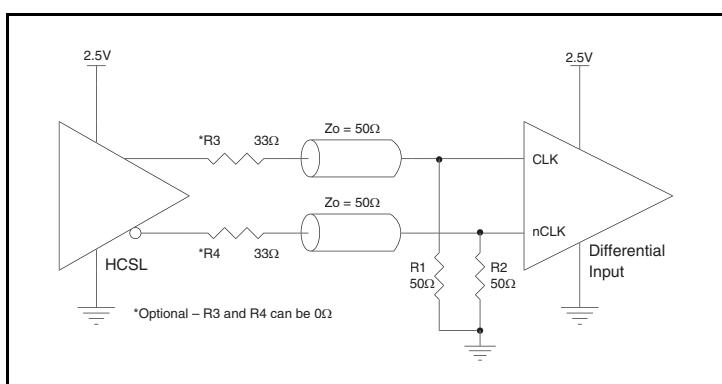
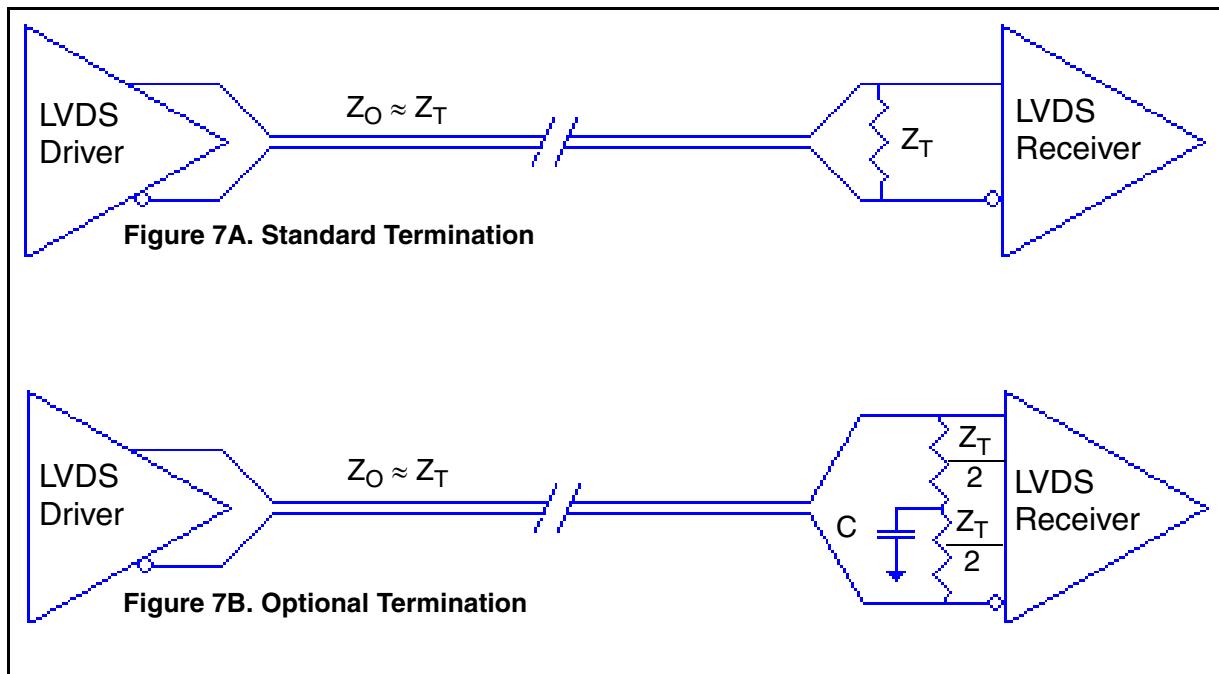


Figure 6E. CLKnx / nCLKnx Input driven by a 2.5V HCSL driver

LVDS Driver Termination

For a general LVDS interface, the recommended value for the termination impedance (Z_T) is between 90Ω and 132Ω . The actual value should be selected to match the differential impedance (Z_0) of your transmission line. A typical point-to-point LVDS design uses a 100Ω parallel resistor at the receiver and a 100Ω differential transmission-line environment. In order to avoid any transmission-line reflection issues, the components should be surface mounted and must be placed as close to the receiver as possible. IDT offers a full line of LVDS compliant devices with two types of output structures: current source and voltage source. The

standard termination schematic as shown in *Figure 7A* can be used with either type of output structure. *Figure 7B*, which can also be used with both output types, is an optional termination with center tap capacitance to help filter common mode noise. The capacitor value should be approximately 50pF . If using a non-standard termination, it is recommended to contact IDT and confirm if the output structure is current source or voltage source type. In addition, since these outputs are LVDS compatible, the input receiver's amplitude and common-mode input range should be verified for compatibility with the output.



Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 8A and 8B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

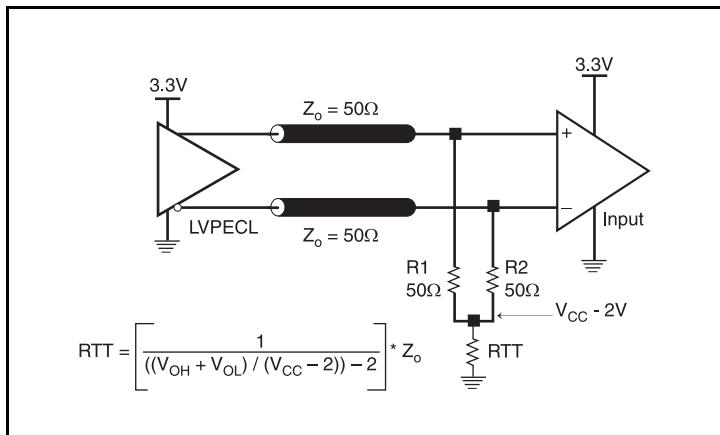


Figure 8A. 3.3V LVPECL Output Termination

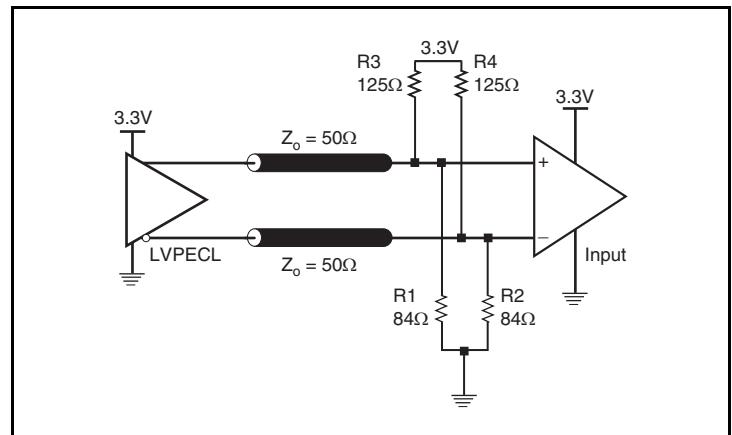


Figure 8B. 3.3V LVPECL Output Termination

Termination for 2.5V LVPECL Outputs

Figure 9A and Figure 9B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating 50Ω to $V_{CCO_X} - 2V$. For $V_{CCO_X} = 2.5V$, the $V_{CCO_X} - 2V$ is very close to

ground level. The R3 in Figure 9B can be eliminated and the termination is shown in Figure 9C.

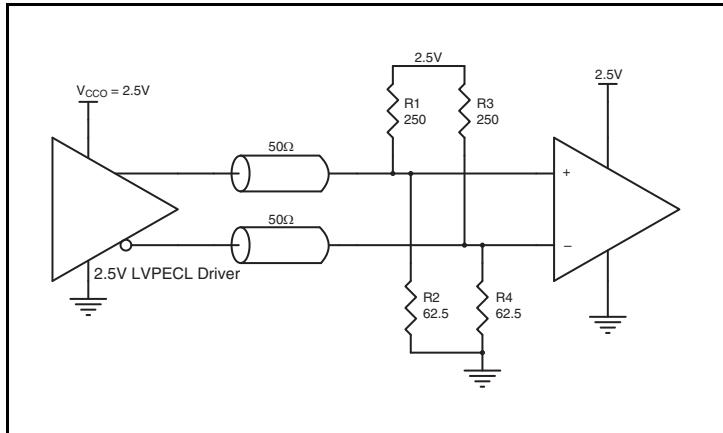


Figure 9A. 2.5V LVPECL Driver Termination Example

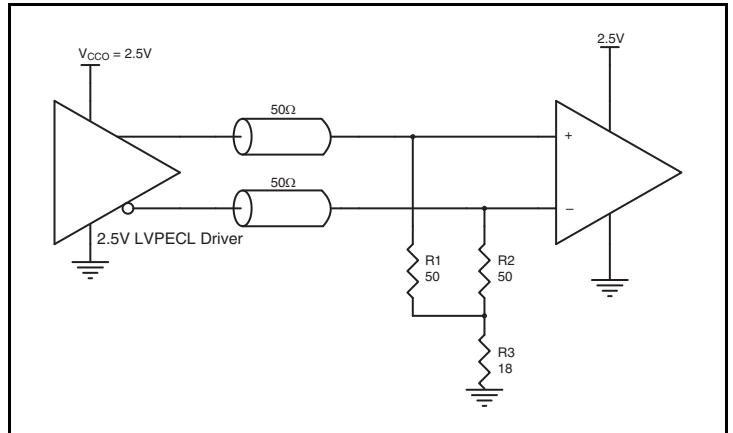


Figure 9B. 2.5V LVPECL Driver Termination Example

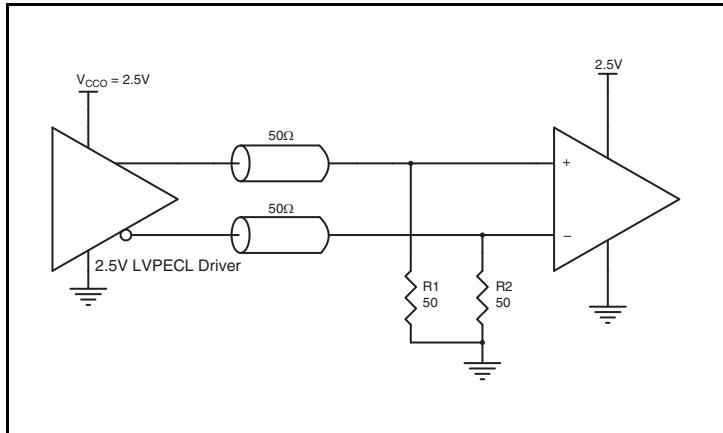


Figure 9C. 2.5V LVPECL Driver Termination Example

Schematic Example

Figure 10 (next page) shows an example IDT8T49N244I application schematic. The schematic example focuses on functional connections and is not configuration specific. Refer to the pin description and functional tables in the datasheet to ensure that PLL_BYPASS and the CLK_SELx pins are properly set. Input and output terminations shown are intended as examples only and may not match the exact user application. To promote readability in this schematic, only Frequency Translator B and the global pins PLL_BYPASS, REF_CLK and SDATA and SCLK are shown connected.

Frequency Translators A and B are fully independent. Therefore use of Frequency Translator A requires that the power filters shown in the schematic for Frequency Translator B are duplicated for A. However different signal path input and output terminations and loop filters may be used for Frequency Translator A than are shown for Frequency Translator B.

As with any high speed analog circuitry, the power supply pins are vulnerable to random noise. To achieve optimum jitter performance, power supply isolation is required. The IDT8T49N244I provides separate V_{CC} , V_{CCA_X} and V_{CCO_X} power supplies for each

Frequency Translator to isolate any high switching noise from coupling into the internal PLLs.

In order to achieve the best possible filtering, it is highly recommended that the 0.1uF capacitors on the device side of the ferrite beads be placed on the device side of the PCB as close to the power pins as possible. This is represented by the placement of these capacitors in the schematic. If space is limited, the ferrite beads, 10uf and 0.1uF capacitor connected to 2.5V can be placed on the opposite side of the PCB. If space permits, place all filter components on the device side of the board.

Power supply filter recommendations are a general guideline to be used for reducing external noise from coupling into the devices. The filter performance is designed for a wide range of noise frequencies. This low-pass filter starts to attenuate noise at approximately 10kHz. If a specific frequency noise component is known, such as switching power supplies frequencies, it is recommended that component values be adjusted and if required, additional filtering be added. Additionally, good general design practices for power plane voltage stability suggests adding bulk capacitance in the local area of all devices.

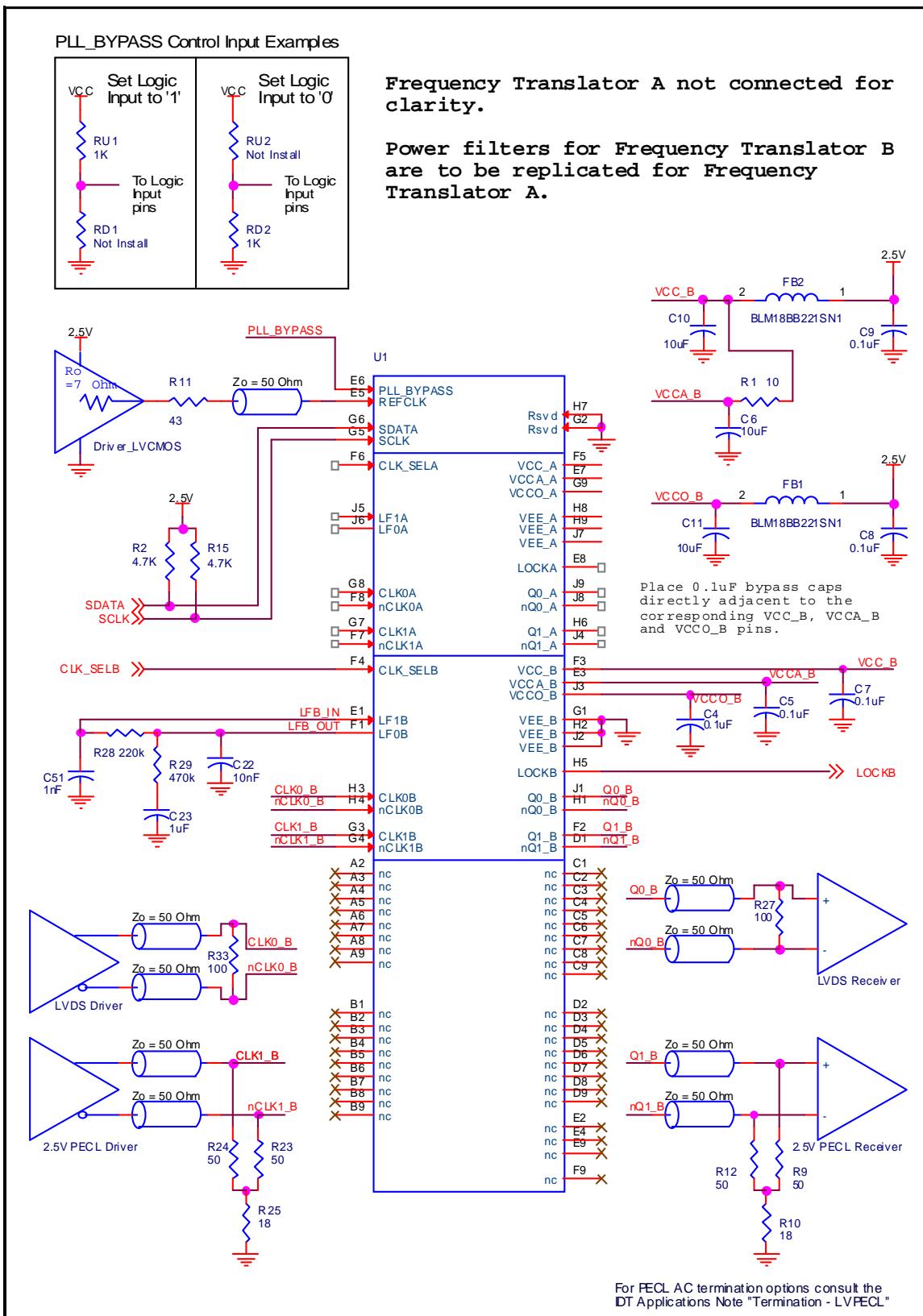


Figure 10. IDT8T49N244I Schematic Layout

Power Considerations

This section provides information on power dissipation and junction temperature for the IDT8T49N244I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the IDT8T49N244I is the sum of the core power plus the output power dissipated due to the load. The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating output power dissipated due to the load.

- Power (core)_{MAX} = $V_{CC_MAX} * I_{EE_MAX} = 3.465V * 624mA = 2162.15mW$
- Power (outputs)_{MAX} = **33.2mW/Loaded Output pair**
If all outputs are loaded, the total power is $4 * 33.2mW = 132.8mW$

Total Power_{MAX} (3.465V, with all outputs switching) = $2162.15mW + 132.8mW = 2227.56mW$

2. Junction Temperature.

Junction temperature, T_j , is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, T_j , to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for T_j is as follows: $T_j = \theta_{JA} * P_{d_total} + T_A$

T_j = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

P_{d_total} = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 12.4°C/W per Table 8 below.

Therefore, T_j for an ambient temperature of 85°C with all outputs switching is:

$85^\circ C + 2.228W * 12.4^\circ C/W = 112.6^\circ C$. This is below the limit of 125°C.

This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 8. θ_{JA} vs. Air FlowTable for an 80-Ball CABGA

| θ_{JA} vs. Air Flow | | | | |
|----------------------------|----------|--------|----------|--------|
| Meters per Second | 0 | 1 | 2 | 3 |
| Multi-Layer PCB, NOTE 1 | 12.4°C/W | 11°C/W | 10.3°C/W | 10°C/W |

NOTE 1: θ_{JA} simulation is performed with 8-layers, 8in. x 8in. PCB for the 10x10, 1mm ball pitch BGA package.

3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

LVPECL output driver circuit and termination are shown in *Figure 11*.

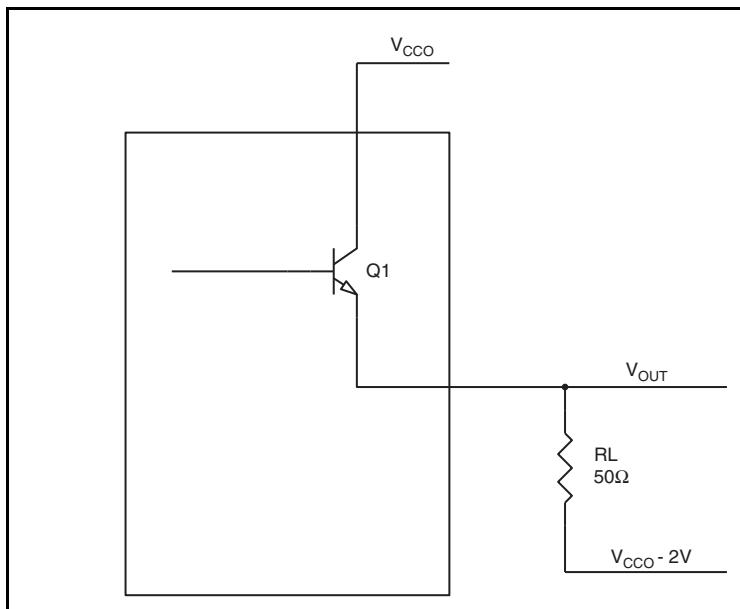


Figure 11. LVPECL Driver Circuit and Termination

To calculate power dissipation per output pair due to the load, use the following equations which assume a 50Ω load, and a termination voltage of $V_{CCO} - 2V$.

- For logic high, $V_{OUT} = V_{OH_MAX} = V_{CCO_MAX} - 0.7V$
 $(V_{CCO_MAX} - V_{OH_MAX}) = 0.7V$
- For logic low, $V_{OUT} = V_{OL_MAX} = V_{CCO_MAX} - 1.5V$
 $(V_{CCO_MAX} - V_{OL_MAX}) = 1.5V$

Pd_H is power dissipation when the output drives high.

Pd_L is the power dissipation when the output drives low.

$$Pd_H = [(V_{OH_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - (V_{CCO_MAX} - V_{OH_MAX}))/R_L] * (V_{CCO_MAX} - V_{OH_MAX}) = [(2V - 0.7V)/50\Omega] * 0.7V = 18.2mW$$

$$Pd_L = [(V_{OL_MAX} - (V_{CCO_MAX} - 2V))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - (V_{CCO_MAX} - V_{OL_MAX}))/R_L] * (V_{CCO_MAX} - V_{OL_MAX}) = [(2V - 1.5V)/50\Omega] * 1.5V = 15mW$$

$$\text{Total Power Dissipation per output pair} = Pd_H + Pd_L = 33.2mW$$

LVDS Power Considerations

This section provides information on power dissipation and junction temperature for the IDT8T49N244I. Equations and example calculations are also provided.

1. Power Dissipation.

The total power dissipation for the IDT8T49N244I is the sum of the core power plus the analog power plus the output power dissipated due to loading. The following is the power dissipation for $V_{CC} = 3.3V + 5\% = 3.465V$, which gives worst case results.

- Power (core)_{MAX} = $V_{CC_MAX} * (I_{CC_X_MAX} + \dots) = 3.465V * (526mA + 64mA) = 2044.35mW$
- Power (outputs)_{MAX} = $V_{CCO_X_MAX} * I_{CCO_X_MAX} = 3.465V * 88mA = 304.92mW$

Total Power_{MAX} = $2044.35mW + 304.92mW = 2349.27mW$

2. Junction Temperature.

Junction temperature, T_j , is the temperature at the junction of the bond wire and bond pad directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature, T_j , to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for T_j is as follows: $T_j = \theta_{JA} * P_{d_total} + T_A$

T_j = Junction Temperature

θ_{JA} = Junction-to-Ambient Thermal Resistance

P_{d_total} = Total Device Power Dissipation (example calculation is in section 1 above)

T_A = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance θ_{JA} must be used. Assuming no air flow and a multi-layer board, the appropriate value is 12.4°C/W per Table 9 below.

Therefore, T_j for an ambient temperature of 85°C with all outputs switching is:

$85^\circ C + 2.349W * 12.4^\circ C/W = 114.1^\circ C$. This is below the limit of 125°C.

This calculation is only an example. T_j will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

Table 9. θ_{JA} vs. Air FlowTable for an 80-Ball CABGA

| θ_{JA} vs. Air Flow | | | | |
|----------------------------|----------|--------|----------|--------|
| Meters per Second | 0 | 1 | 2 | 3 |
| Multi-Layer PCB, NOTE 1 | 12.4°C/W | 11°C/W | 10.3°C/W | 10°C/W |

NOTE 1: θ_{JA} simulation is performed with 8-layers, 8in. x 8in. PCB for the 10x10, 1mm ball pitch BGA package.

Reliability Information

Table 10A. θ_{JA} vs. Air Flow Table for an 80-Ball CABGA

| θ_{JA} vs. Air Flow | | | | |
|----------------------------|----------|--------|----------|--------|
| Meters per Second | 0 | 1 | 2 | 3 |
| Multi-Layer PCB, NOTE 1 | 12.4°C/W | 11°C/W | 10.3°C/W | 10°C/W |

NOTE 1: θ_{JA} simulation is performed with 8-layers, 8in. x 8in. PCB for the 10x10, 1mm ball pitch BGA package.

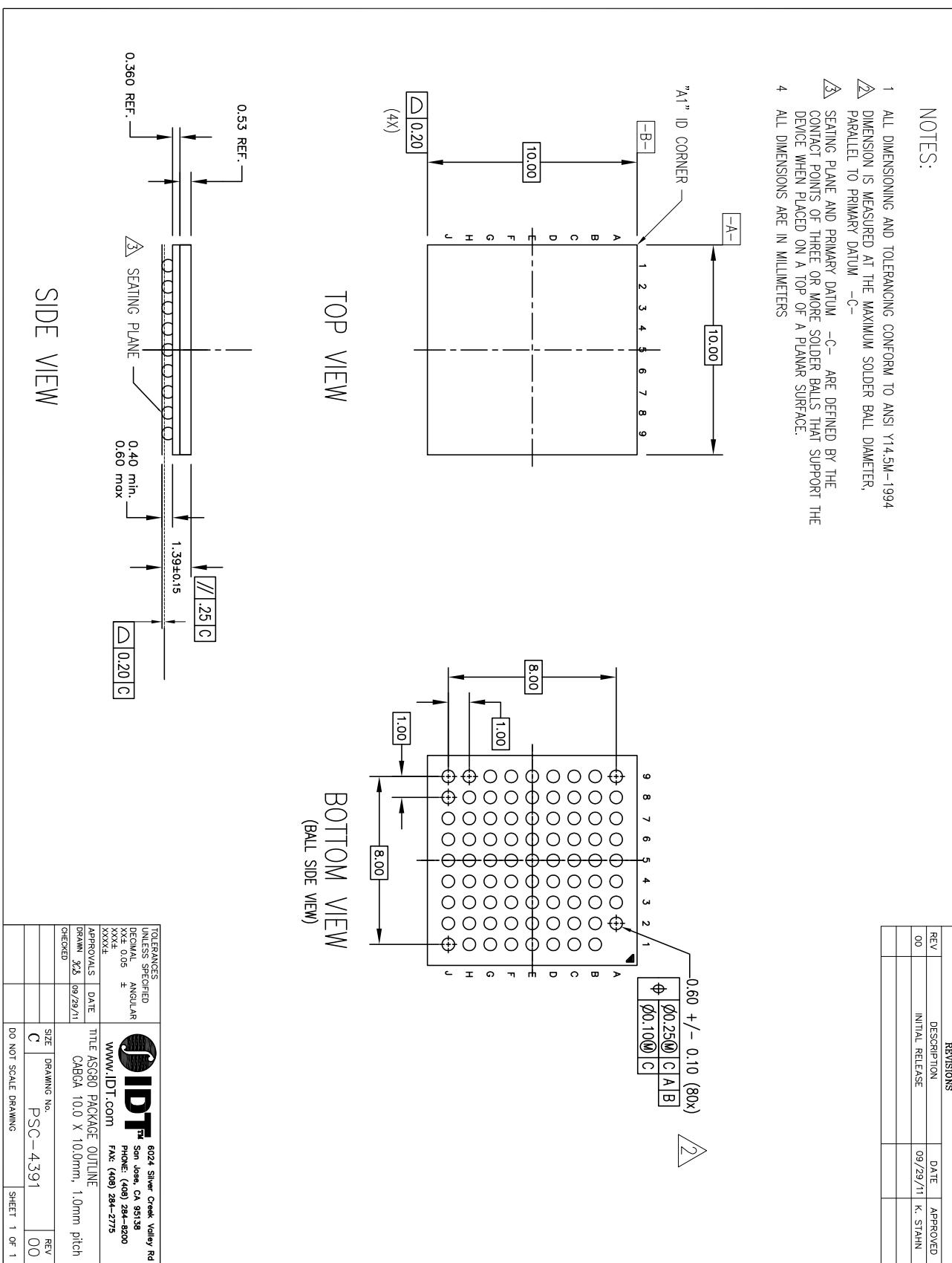
Table 10B. θ_{JC} Table for an 80-Ball CABGA

| θ_{JC} | |
|-------------------------|----------|
| Meters per Second | 0 |
| Multi-Layer PCB, NOTE 1 | 12.5°C/W |

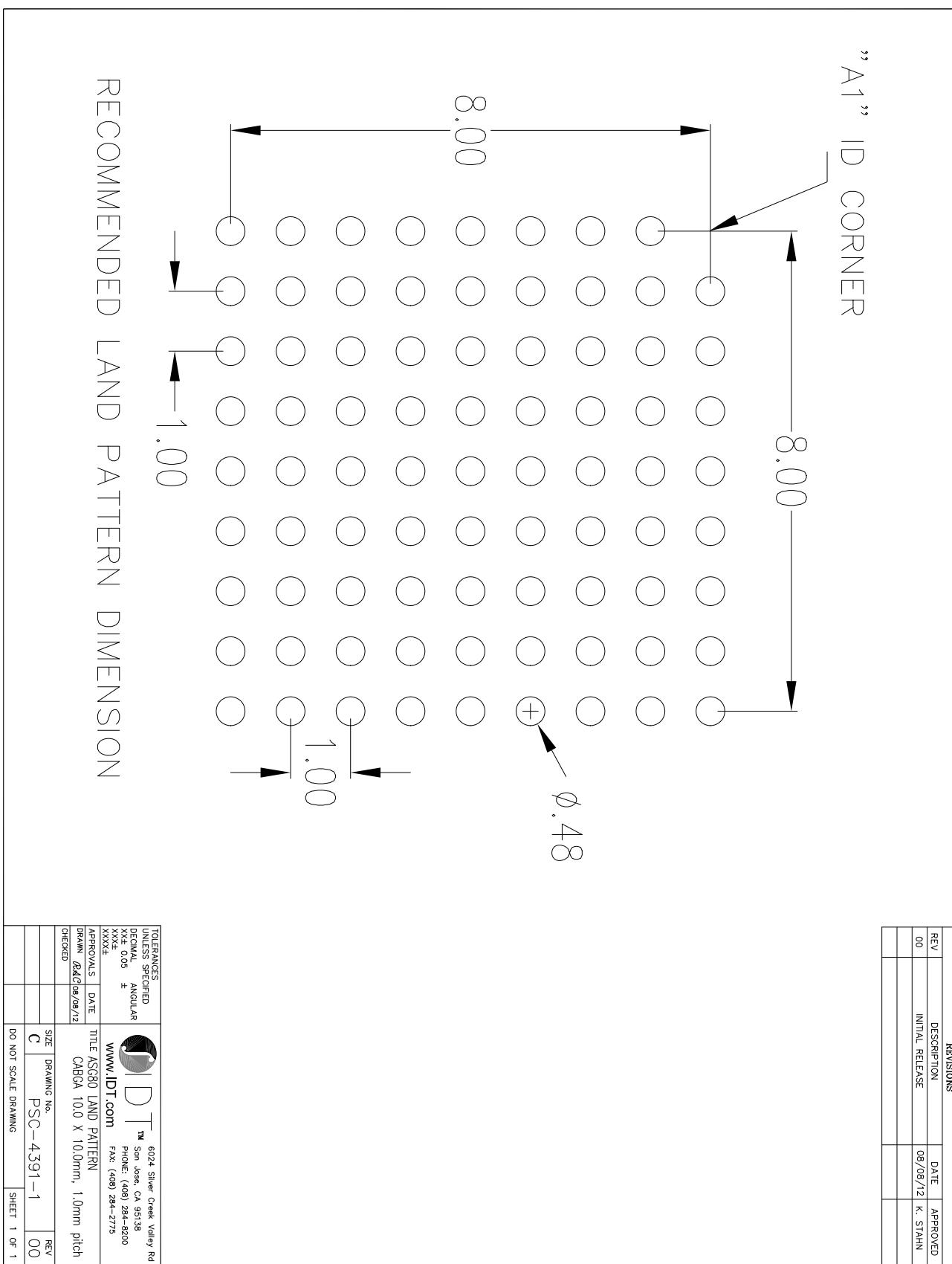
Transistor Count

The transistor count for IDT8T49N244I is: 101,286

80-Ball CABGA (AS) Package Outline and Package Dimensions



80-Ball CABGA (AS) Land Pattern



Ordering Information

Table 11. Ordering Information

| Part/Order Number | Marking | Package | Shipping Packaging | Temperature |
|--------------------|----------------------|-----------------------------------|--------------------|----------------|
| 8T49N244A-dddASGI | IDT8T49N244A-dddASGI | “Lead-Free” Plastic 80-Ball CABGA | Tray | -40°C to +85°C |
| 8T49N244A-dddASGI8 | IDT8T49N244A-dddASGI | “Lead-Free” Plastic 80-Ball CABGA | Tape & Reel | -40°C to +85°C |

NOTE: For the specific -ddd order codes, refer to *FemtoClock NG Universal Frequency Translator Ordering Product Information* document.

Revision History Sheet

| Rev | Table | Page | Description of Change | Date |
|-----|-----------|---------|--|---------|
| A | T2 T5F | 4 16 | Pin Characteristics Table - added PLL_BYPASS to R_{Pulldown} row. LVC MOS DC Characteristics Table - added NOTE 1 to I_{IH} , I_{IL} . | 2/15/13 |
| A | T4E | 11 | Corrected typo: M_INTx[8:0] to M_INTx[7:0] | 6/28/13 |

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