

## Description

The IDT8S89871I is a high speed Differential-to-LVPECL Buffer/Divider. The IDT8S89871I has a selectable  $\div 2$ ,  $\div 4$ ,  $\div 8$ ,  $\div 16$  output dividers. The clock input has internal termination resistors, allowing it to interface with several differential signal types while minimizing the number of required external components.

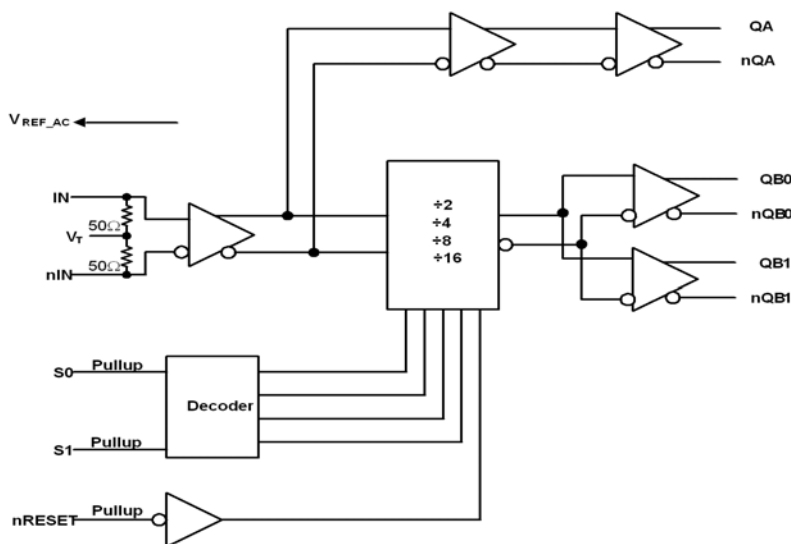
The QA output provides an extra copy of the input frequency, while QB is a divided down output of the input frequency. Both output banks are phase matched and maintain a matched delay independent of the divider setting.

The device is packaged in a small, 3 x 3 mm VFQFN package, making it ideal for use on space-constrained boards.

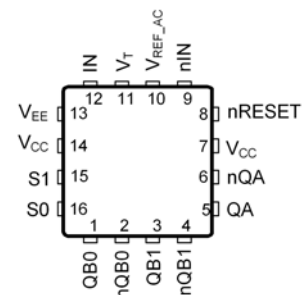
## Features

- Three LVPECL output pairs
- Frequency divide select options:  $\div 2$ ,  $\div 4$ ,  $\div 8$ ,  $\div 16$  (Bank B)
- Pass-through output (Bank A)
- IN, nIN input can accept the following differential input levels: LVPECL, LVDS, CML
- Output frequency: 2.5GHz
- Bank skew: 40ps (maximum), Bank B
- Part-to-part skew: 230ps (maximum)
- Additive phase jitter, RMS: 0.15ps (typical)
- Voltage supply range: 2.375V to 3.6V
- $-40^{\circ}\text{C}$  to  $85^{\circ}\text{C}$  ambient operating temperature
- Available in lead-free (RoHS 6) package

## Block Diagram



## Pin Assignment



### IDT8S89871I

16-Lead VFQFN

3mm x 3mm x 0.925mm package body

NL Package

Top View

### Table 1. Pin Descriptions

Number	Name	Type		Description
1, 2	QB0, nQB0	Output		Differential output pair. LVPECL interface levels.
3, 4	QB1, nQB1	Output		Differential output pair. LVPECL interface levels.
5, 6	QA, nQA	Output		Differential output pair. LVPECL interface levels.
7, 14	V <sub>CC</sub>	Power		Power supply pins.
8	nRESET	Input	Pullup	QB [1:0] output reset. Input threshold is V <sub>CC</sub> /2. Includes a 37kΩ pull-up resistor. LVTTTL/LVCMOS interface levels. See Table 3.
9	nIN	Input		Inverting differential LVPECL clock input. R <sub>IN</sub> = 50Ω termination to V <sub>T</sub> .
10	V <sub>REF_AC</sub>	Output		Reference voltage for AC-coupled applications.
11	V <sub>T</sub>	Input		Termination input.
12	IN	Input		Non-inverting LVPECL differential clock input. R <sub>IN</sub> = 50Ω termination to V <sub>T</sub> .
13	V <sub>EE</sub>	Power		Negative supply pin.
15, 16	S1, S0	Input	Pullup	Select pins. LVCMOS/LVTTTL interface levels. See Table 3.

NOTE: *Pullup* refers to internal input resistors. See Table 2, *Pin Characteristics*, for typical values.

### Table 2. Pin Characteristics

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
R <sub>PULLUP</sub>	Input Pullup Resistor			37		kΩ

## Function Table

Table 3. Control Input and Divider Truth Table

Inputs			Outputs	
nRESET	S1	S0	QA Output	QB Output
1 (default)	0	0	Reference Clock	Reference Clock ÷2
1	0	1	Reference Clock	Reference Clock ÷4
1	1	0	Reference Clock	Reference Clock ÷8
1	1 (default)	1 (default)	Reference Clock	Reference Clock ÷16
0	X	X	Reference Clock	Q = LOW, nQ = HIGH; Clock Disabled

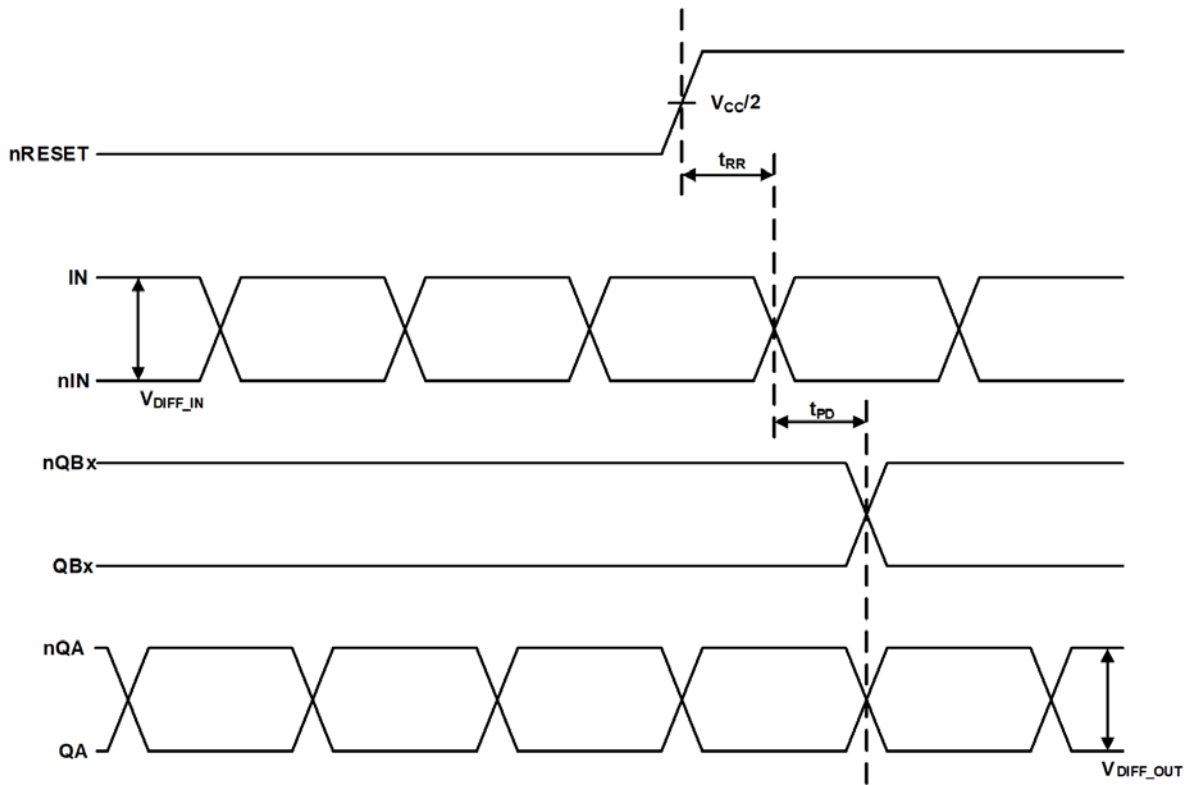


Figure 1. nRESET Timing Diagram

## Absolute Maximum Ratings

Stresses beyond those listed under *Absolute Maximum Ratings* may cause permanent damage to the device. These ratings are stress specifications only. Functional operation of product at these conditions or any conditions beyond those listed in the *DC Characteristics* or *AC Characteristics* is not implied. Exposure to absolute maximum rating conditions for extended periods may affect product reliability.

Item	Rating
Supply Voltage, $V_{CC}$	4.6V
Inputs, $V_I$	-0.5V to $V_{CC} + 0.5V$
Outputs, $I_O$ Continuous Current Surge Current	50mA 100mA
Input Current, $I_{IN}$ , nIN	$\pm 50mA$
$V_T$ Current, $I_{VT}$	$\pm 100mA$
$V_{REF\_AC}$ Input Sink/Source, $I_{REF\_AC}$	$\pm 2mA$
Operating Temperature Range, $T_A$	-40°C to +85°C
Package Thermal Impedance, $\theta_{JA}$ , (Junction-to-Ambient)	74.7°C/W (0 mps)
Storage Temperature, $T_{STG}$	-65°C to 150°C

## DC Electrical Characteristics

**Table 4A. Power Supply DC Characteristics**,  $V_{CC} = 3.3V \pm 0.3V$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{CC}$	Power Supply Voltage		2.375	3.3	3.6	V
$I_{CC}$	Power Supply Current	Outputs Unterminated			57	mA

**Table 4B. LVCMOS/LVTTL DC Characteristics**,  $V_{CC} = 3.3V \pm 0.3V$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{IH}$	Input High Voltage		2.2		$V_{CC} + 0.3$	V
$V_{IL}$	Input Low Voltage		0		0.8	V
$I_{IH}$	Input High Current	S[1:0], nRESET $V_{CC} = V_{IN} = 3.6V$ or $2.625V$			10	$\mu A$
$I_{IL}$	Input Low Current	S[1:0], nRESET $V_{CC} = 3.6V$ or $2.625V$ , $V_{IN} = 0V$	-150			$\mu A$

**Table 4C. Differential DC Characteristics,  $V_{CC} = 3.3V \pm 0.3V$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$** 

Symbol	Parameter		Test Conditions	Minimum	Typical	Maximum	Units
$R_{IN}$	Input Resistance	IN, nIN	IN to $V_T$	40	50	60	$\Omega$
$V_{IH}$	Input High Voltage	IN, nIN		0.15		$V_{CC} + 0.3$	V
$V_{IL}$	Input Low Voltage	IN, nIN		0		$V_{IH} - 0.15$	V
$V_{IN}$	Input Voltage Swing; NOTE 1			0.15		$V_{CC}$	V
$V_{DIFF\_IN}$	Differential Input Voltage Swing			0.3			V
$I_{IN}$	Input Current; NOTE 2, 3	IN, nIN				45	mA
$V_{REF\_AC}$	Bias Voltage			$V_{CC} - 1.46$	$V_{CC} - 1.38$	$V_{CC} - 1.31$	V

NOTE 1: Refer to Parameter Measurement Information, *Input Voltage Swing* diagram.

NOTE 2: Guaranteed by design.

NOTE 3: Because of the internal termination  $R_{IN}$ , the input current  $I_{IN}$  will be determined by the voltages applied at IN, nIN and  $V_T$ . Observe the voltages applied to those pins so the input current does not exceed the maximum limit.

**Table 4D. LVPECL DC Characteristics,  $V_{CC} = 3.3V \pm 0.3V$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$** 

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$V_{OH}$	Output High Voltage; NOTE 1		$V_{CC} - 1.175$		$V_{CC} - 0.82$	V
$V_{OL}$	Output Low Voltage; NOTE 1		$V_{CC} - 2.0$		$V_{CC} - 1.575$	V
$V_{OUT}$	Output Voltage Swing		0.6		1.0	V
$V_{DIFF\_OUT}$	Differential Output Voltage Swing		1.2		2.0	V

NOTE: Input and output parameters vary 1:1 with  $V_{CC}$ .

NOTE 1: Outputs terminated with  $50\Omega$  to  $V_{CC} - 2V$ .

## AC Electrical Characteristics

**Table 5. AC Characteristics,  $V_{CC} = 3.3V \pm 0.3V$  or  $2.5V \pm 5\%$ ,  $V_{EE} = 0V$ ,  $T_A = -40^\circ C$  to  $85^\circ C$**

Symbol	Parameter	Test Conditions	Minimum	Typical	Maximum	Units
$f_{OUT}$	Output Frequency	Output Swing $\geq 450mV$			2.5	GHz
$f_{IN}$	Input Frequency	$\pm 2, \pm 4, \pm 8$			3.2	GHz
		$\pm 1, \pm 16$			2.5	GHz
$t_{PD}$	Propagation Delay; (Differential); NOTE 1	Input Swing: $< 400mV$	520	665	805	ps
		Input Swing: $\geq 400mV$	510	650	790	ps
$t_{sk(o)}$	Output Skew; NOTE 2, 5			15	55	ps
$t_{sk(b)}$	Bank Skew; NOTE 3, 5	Bank B		13	40	ps
$t_{sk(pp)}$	Part-to-Part Skew; NOTE 4, 5				230	ps
$f_{jit}$	Buffer Additive Jitter; RMS; refer to Additive Phase Jitter Section;	Bank A. 155.52MHz, Integration Range: 12kHz - 20MHz		0.15	0.18	ps
$t_{RR}$	Reset Recovery Time		600			ps
$t_R / t_F$	Output Rise/Fall Time	20% to 80%	40		250	ps

NOTE: Electrical parameters are guaranteed over the specified ambient operating temperature range, which is established when the device is mounted in a test socket with maintained transverse airflow greater than 500 lfm. The device will meet specifications after thermal equilibrium has been reached under these conditions.

NOTE: All parameters characterized at  $\leq 2.5GHz$ , 800mV input signal, unless otherwise noted.

NOTE 1: Measured from the differential input crossing point to the differential output crossing point.

NOTE 2: Defined as skew between outputs at the same supply voltage and with equal load conditions. Measured at the output differential cross points.

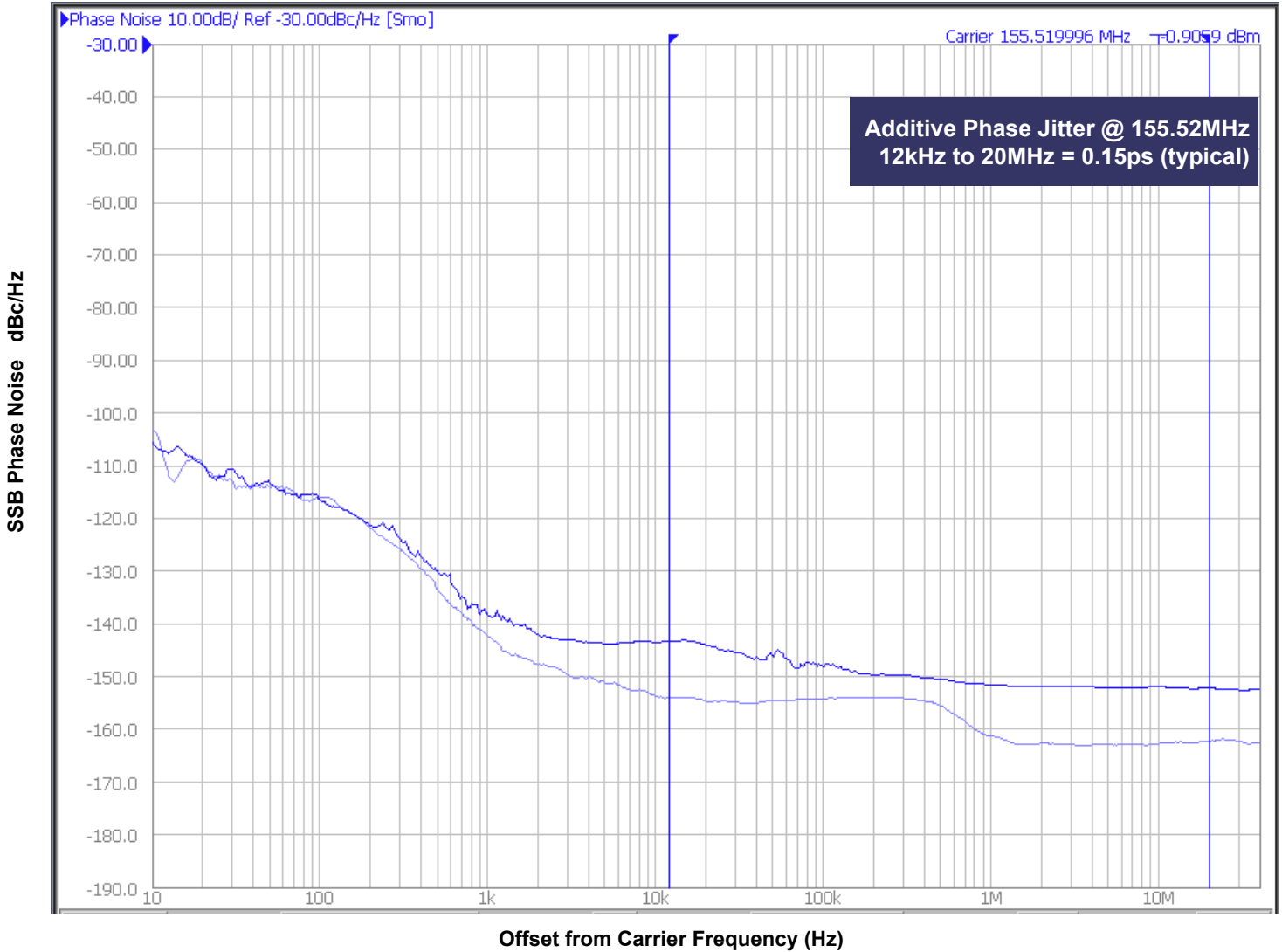
NOTE 3: Defined as skew within a bank of outputs at the same supply voltage and with equal load conditions.

NOTE 4: Defined as skew between outputs on different devices operating at the same supply voltage, same temperature, same frequency and with equal load conditions. Using the same type of inputs on each device, the outputs are measured at the differential cross points.

NOTE 5: This parameter is defined in accordance with JEDEC Standard 65.

## Additive Phase Jitter

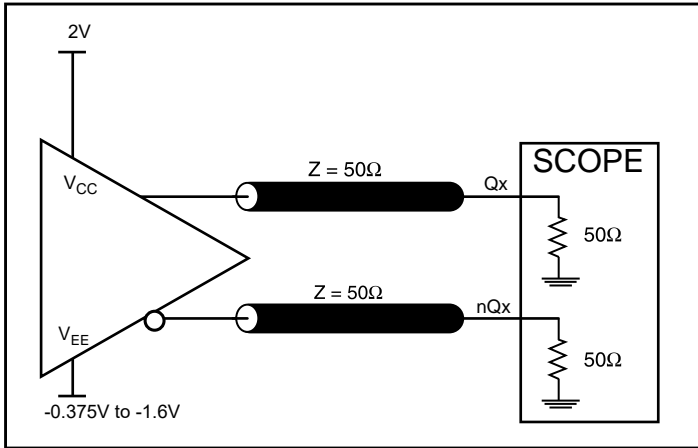
The spectral purity in a band at a specific offset from the fundamental compared to the power of the fundamental is called the **dBc Phase Noise**. This value is normally expressed using a Phase Noise plot and is most often the specified plot in many applications. Phase noise is defined as the ratio of the noise power present in a 1Hz band at a specified offset from the fundamental frequency to the power value of the fundamental. This ratio is expressed in decibels (dBm) or a ratio of the power in the 1Hz band to the power in the fundamental. When the required offset is specified, the phase noise is called a **dBc** value, which simply means dBm at a specified offset from the fundamental. By investigating jitter in the frequency domain, we get a better understanding of its effects on the desired application over the entire time record of the signal. It is mathematically possible to calculate an expected bit error rate given a phase noise plot.



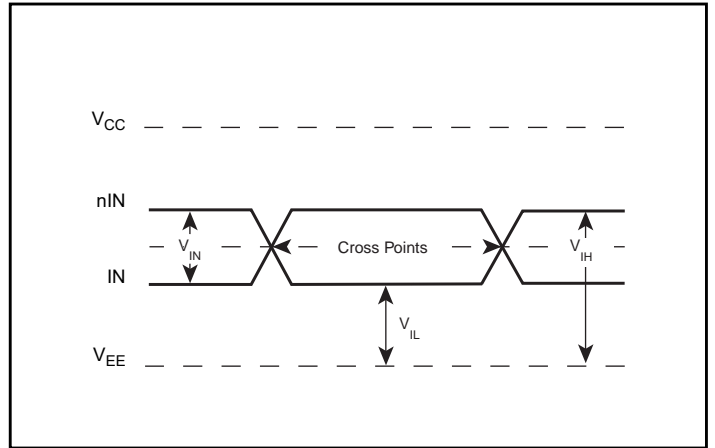
As with most timing specifications, phase noise measurements has issues relating to the limitations of the equipment. Often the noise floor of the equipment is higher than the noise floor of the device. This is illustrated above. The device meets the noise floor of what is shown, but can actually be lower. The phase noise is dependent on the input source and measurement equipment.

Measured using a Rohde & Schwarz SMA100A as the input source.

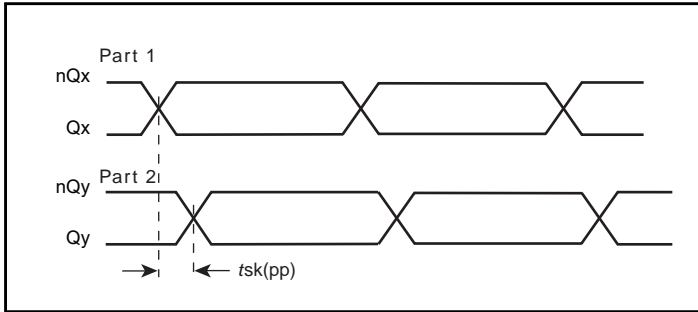
## Parameter Measurement Information



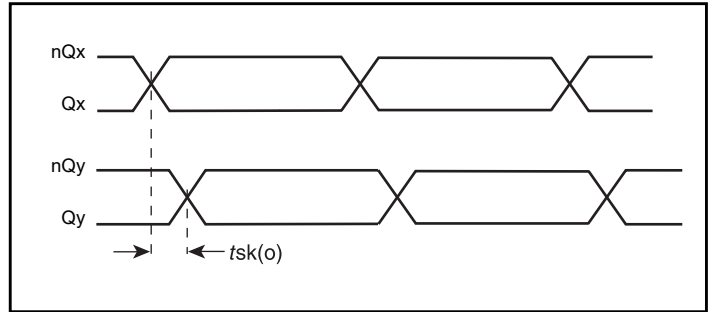
Output Load AC Test Circuit



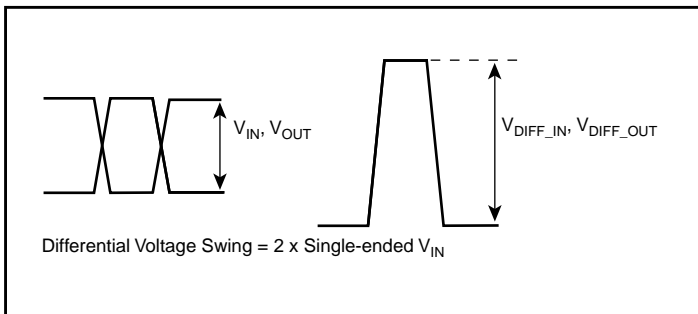
Differential Input Level



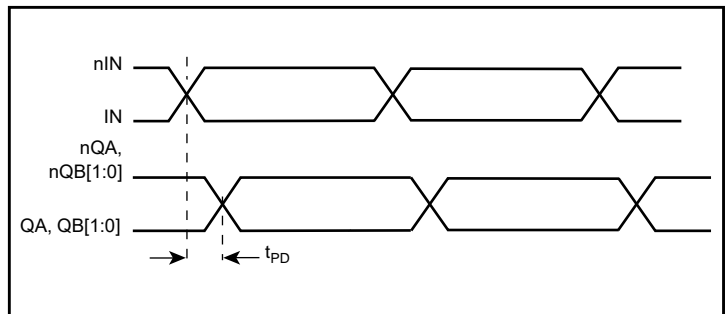
Part-to-Part Skew



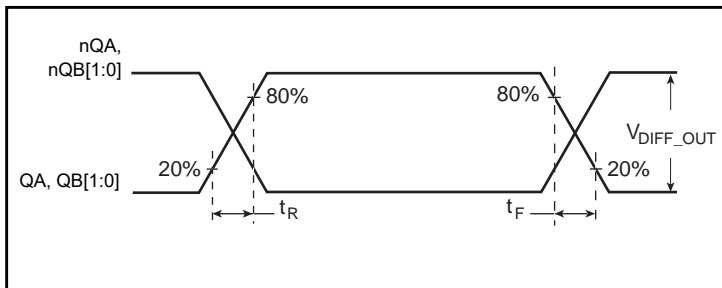
Output Skew



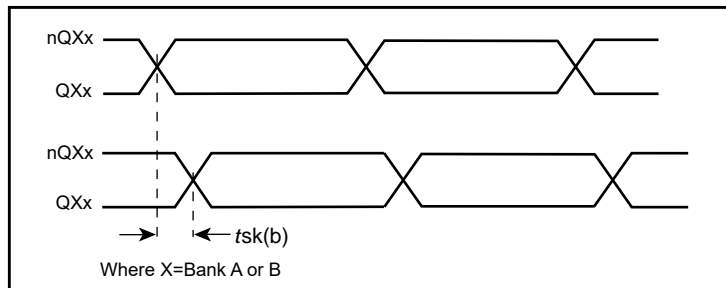
Single-ended & Differential Input Voltage Swing



Propagation Delay



Output Rise/Fall Time



Bank Skew

## Applications Information

### 3.3V Differential Input with Built-In 50Ω Termination Interface

The IN/nIN with built-in 50Ω terminations accept LVDS, LVPECL, CML and other differential signals. Both signals must meet the  $V_{IN}$  and  $V_{IH}$  input requirements. Figures 2A to 2D show interface examples for the IN/nIN input with built-in 50Ω terminations driven by

the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.

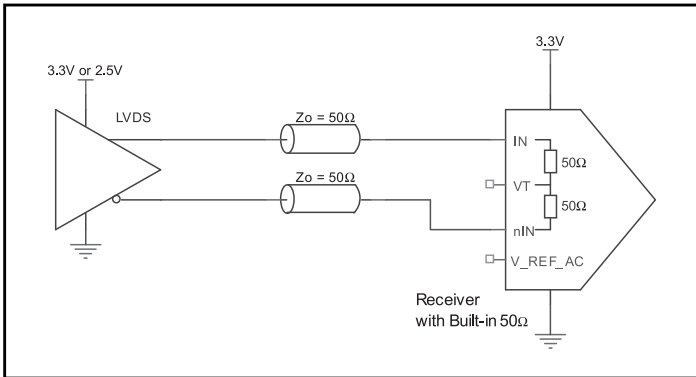


Figure 2A. N/nIN Input with Built-In 50Ω Driven by an LVDS Driver

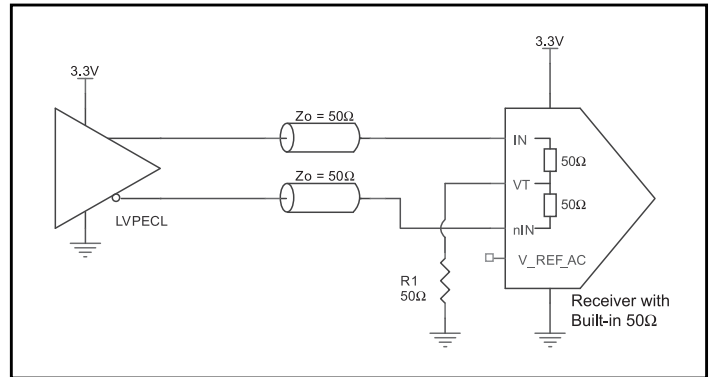


Figure 2B. IN/nIN Input with Built-In 50Ω Driven by an LVPECL Driver

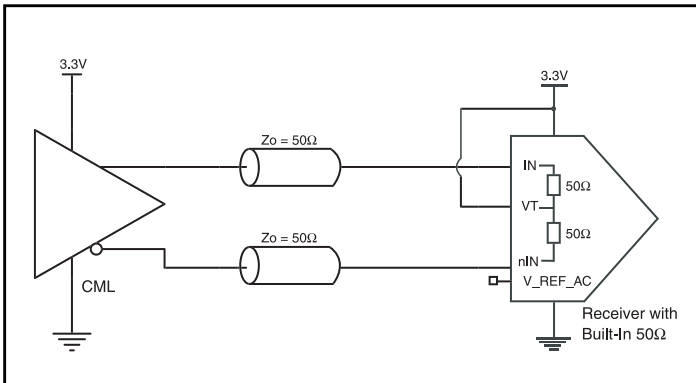


Figure 2C. IN/nIN Input with Built-In 50Ω Driven by a CML Driver

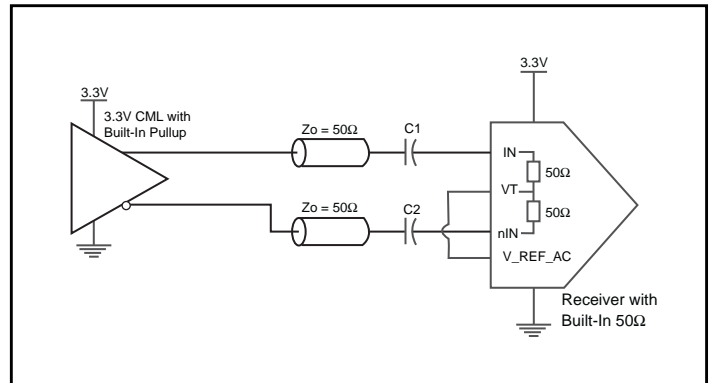
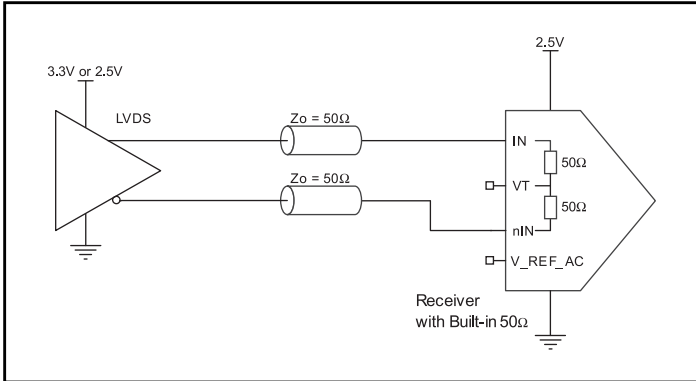


Figure 2D. IN/nIN Input with Built-In 50Ω Driven by a CML Driver with Built-In 50Ω Pullup

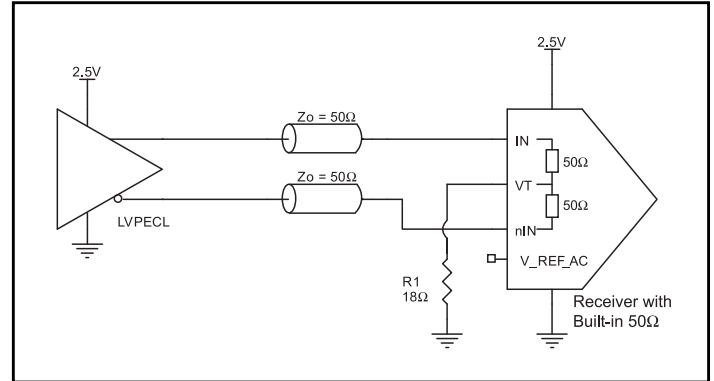
## 2.5V LVPECL Input with Built-In 50Ω Termination Interface

The IN /nIN with built-in 50Ω terminations accept LVDS, LVPECL, CML and other differential signals. Both signals must meet the  $V_{IN}$  and  $V_{IH}$  input requirements. Figures 3A to 3D show interface examples for the IN/nIN with built-in 50Ω termination input driven by

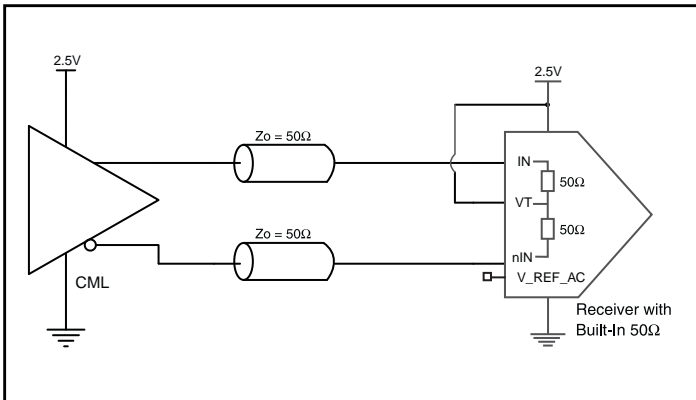
the most common driver types. The input interfaces suggested here are examples only. If the driver is from another vendor, use their termination recommendation. Please consult with the vendor of the driver component to confirm the driver termination requirements.



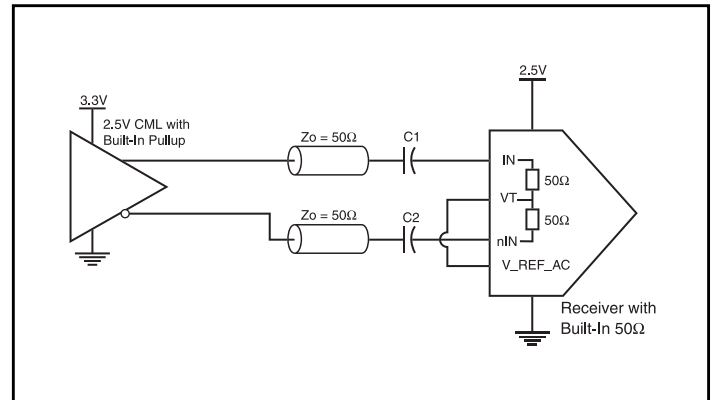
**Figure 3A. IN/nIN Input with Built-In 50Ω Driven by an LVDS Driver**



**Figure 3B. IN/nIN Input with Built-In 50Ω Driven by an LVPECL Driver**



**Figure 3C. IN/nIN Input with Built-In 50Ω Driven by a CML Driver**



**Figure 3D. IN/nIN Input with Built-In 50Ω Driven by a CML Driver with Built-In 50Ω Pullup**

## Recommendations for Unused Input and Output Pins

### Inputs:

#### LVC MOS Control Pins

All control pins have internal pullups; additional resistance is not required but can be added for additional protection. A 1kΩ resistor can be used.

### Termination for 3.3V LVPECL Outputs

The clock layout topology shown below is a typical termination for LVPECL outputs. The two different layouts mentioned are recommended only as guidelines.

The differential outputs are low impedance follower outputs that generate ECL/LVPECL compatible outputs. Therefore, terminating resistors (DC current path to ground) or current sources must be used for functionality. These outputs are designed to drive 50Ω

### Outputs:

#### LVPECL Outputs

All unused LVPECL outputs can be left floating. We recommend that there is no trace attached. Both sides of the differential output pair should either be left floating or terminated.

transmission lines. Matched impedance techniques should be used to maximize operating frequency and minimize signal distortion. *Figures 4A and 4B* show two different layouts which are recommended only as guidelines. Other suitable clock layouts may exist and it would be recommended that the board designers simulate to guarantee compatibility across all printed circuit and clock component process variations.

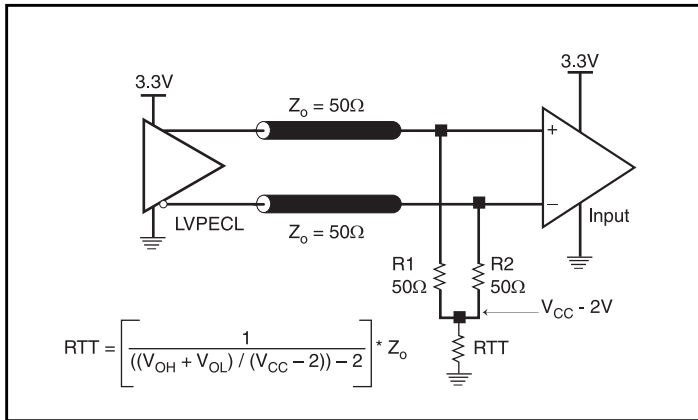


Figure 4A. 3.3V LVPECL Output Termination

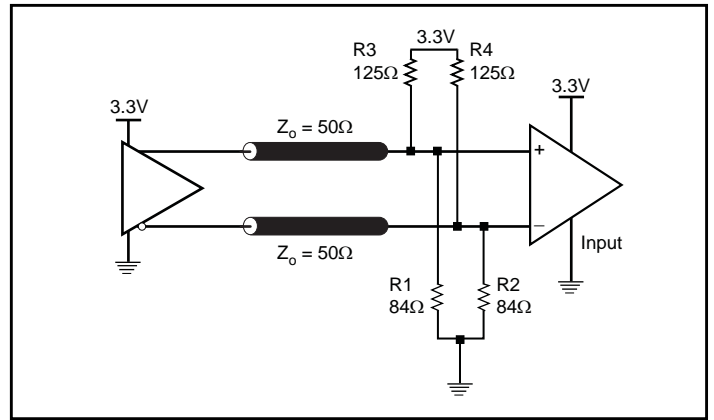


Figure 4B. 3.3V LVPECL Output Termination

## Termination for 2.5V LVPECL Outputs

Figure 5A and Figure 5B show examples of termination for 2.5V LVPECL driver. These terminations are equivalent to terminating  $50\Omega$  to  $V_{CC} - 2V$ . For  $V_{CC} = 2.5V$ , the  $V_{CC} - 2V$  is very close to ground level. The R3 in Figure 5B can be eliminated and the termination is shown in Figure 5C.

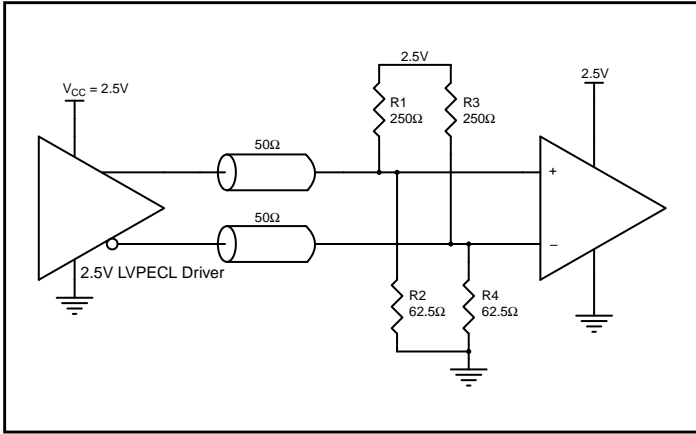


Figure 5A. 2.5V LVPECL Driver Termination Example

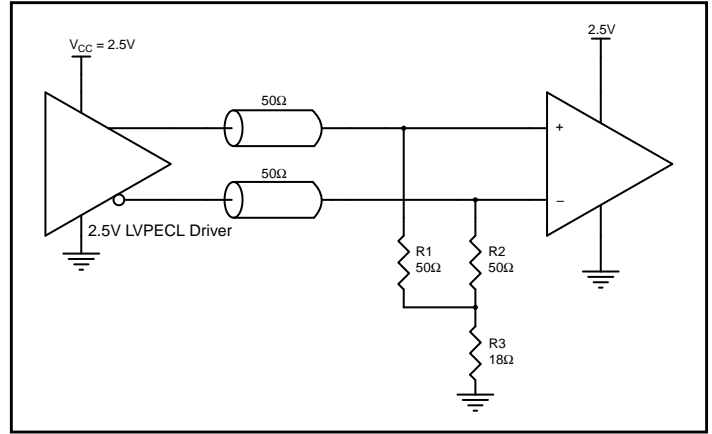


Figure 5B. 2.5V LVPECL Driver Termination Example

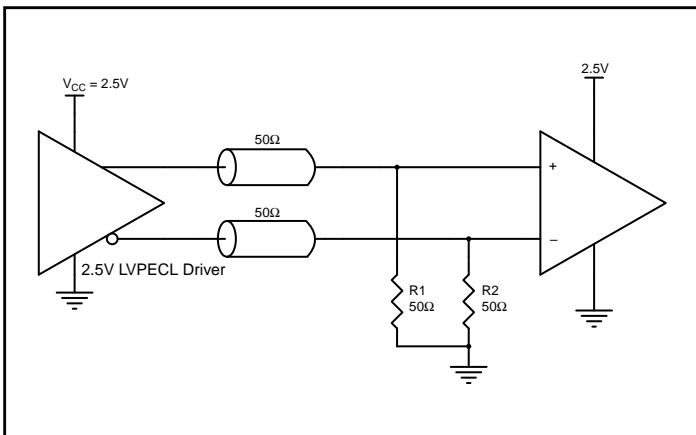
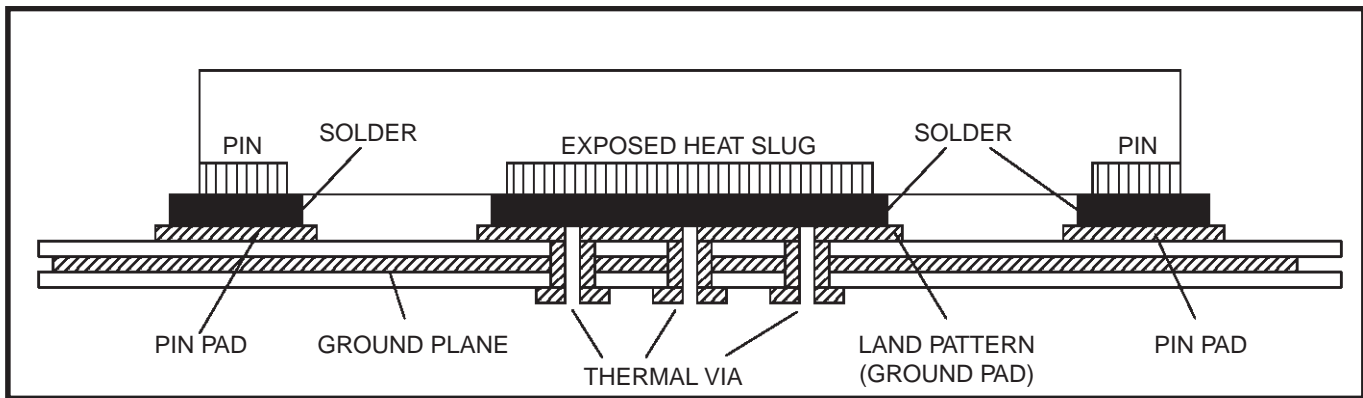


Figure 5C. 2.5V LVPECL Driver Termination Example

## VFQFN EPAD Thermal Release Path

In order to maximize both the removal of heat from the package and the electrical performance, a land pattern must be incorporated on the Printed Circuit Board (PCB) within the footprint of the package corresponding to the exposed metal pad or exposed heat slug on the package, as shown in *Figure 6*. The solderable area on the PCB, as defined by the solder mask, should be at least the same size/shape as the exposed pad/slug area on the package to maximize the thermal/electrical performance. Sufficient clearance should be designed on the PCB between the outer edges of the land pattern and the inner edges of pad pattern for the leads to avoid any shorts.

While the land pattern on the PCB provides a means of heat transfer and electrical grounding from the package to the board through a solder joint, thermal vias are necessary to effectively conduct from the surface of the PCB to the ground plane(s). The land pattern must be connected to ground through these vias. The vias act as “heat pipes”. The number of vias (i.e. “heat pipes”) are application specific and dependent upon the package power dissipation as well as electrical conductivity requirements. Thus, thermal and electrical analysis and/or testing are recommended to determine the minimum number needed. Maximum thermal and electrical performance is achieved when an array of vias is incorporated in the land pattern. It is recommended to use as many vias connected to ground as possible. It is also recommended that the via diameter should be 12 to 13mils (0.30 to 0.33mm) with 1oz copper via barrel plating. This is desirable to avoid any solder wicking inside the via during the soldering process which may result in voids in solder between the exposed pad/slug and the thermal land. Precautions should be taken to eliminate any solder voids between the exposed heat slug and the land pattern. Note: These recommendations are to be used as a guideline only. For further information, please refer to the Application Note on the Surface Mount Assembly of Amkor’s Thermally/ Electrically Enhance Leadframe Base Package, Amkor Technology.



**Figure 6. P.C. Assembly for Exposed Pad Thermal Release Path – Side View (drawing not to scale)**

## Power Considerations

This section provides information on power dissipation and junction temperature for the IDT8S89871I. Equations and example calculations are also provided.

### 1. Power Dissipation.

The total power dissipation for the IDT8S89871I is the sum of the core power plus the power dissipated at the output(s). The following is the power dissipation for  $V_{CC} = 3.6V$ , which gives worst case results.

NOTE: Please refer to Section 3 for details on calculating power dissipated at the output(s).

- Power (core)<sub>MAX</sub> =  $V_{CC\_MAX} * I_{EE\_MAX} = 3.6V * 57mA = \mathbf{205.2mW}$
- Power (outputs)<sub>MAX</sub> = **32.62mW/Loaded Output pair**  
If all outputs are loaded, the total power is  $3 * 32.62mW = \mathbf{97.86mW}$
- Power Dissipation for internal termination  $R_{IN}$   
Power  $(R_T)_{MAX} = (V_{IN\_MAX})^2 / R_{T\_MIN} * 2 = (3.6V)^2 / 80\Omega = \mathbf{162mW}$

**Total Power<sub>MAX</sub>** = (3.6V, with all outputs switching) = 205.2mW + 97.86mW + 162mW = **465.06mW**

### 2. Junction Temperature.

Junction temperature,  $T_j$ , is the temperature at the junction of the bond wire and bond pad, and directly affects the reliability of the device. The maximum recommended junction temperature is 125°C. Limiting the internal transistor junction temperature,  $T_j$ , to 125°C ensures that the bond wire and bond pad temperature remains below 125°C.

The equation for  $T_j$  is as follows:  $T_j = \theta_{JA} * Pd\_total + T_A$

$T_j$  = Junction Temperature

$\theta_{JA}$  = Junction-to-Ambient Thermal Resistance

$Pd\_total$  = Total Device Power Dissipation (example calculation is in section 1 above)

$T_A$  = Ambient Temperature

In order to calculate junction temperature, the appropriate junction-to-ambient thermal resistance  $\theta_{JA}$  must be used. Assuming no air flow and a multi-layer board, the appropriate value is 74.7°C/W per Table 6 below.

Therefore,  $T_j$  for an ambient temperature of 85°C with all outputs switching is:

$$85^\circ\text{C} + 0.465\text{W} * 74.7^\circ\text{C/W} = 119.7^\circ\text{C}. \text{ This is below the limit of } 125^\circ\text{C}.$$

This calculation is only an example.  $T_j$  will obviously vary depending on the number of loaded outputs, supply voltage, air flow and the type of board (multi-layer).

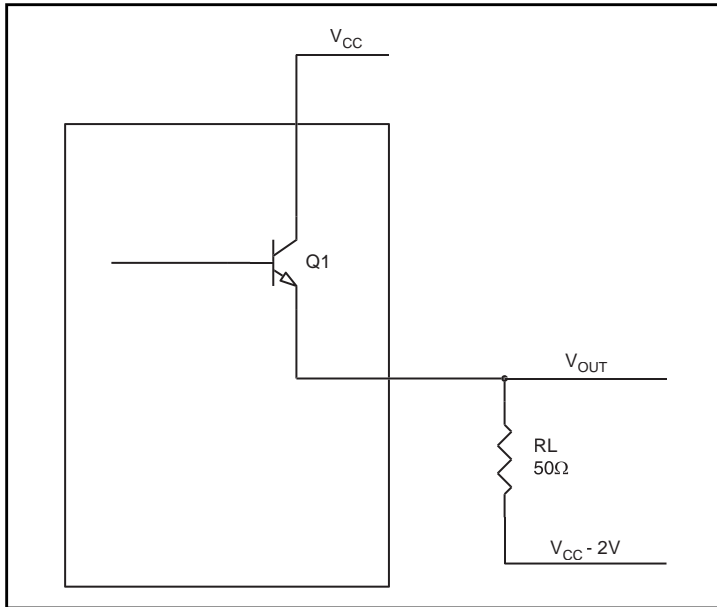
**Table 6. Thermal Resistance  $\theta_{JA}$  for 16 Lead VFQFN, Forced Convection**

Meters per Second	$\theta_{JA}$ vs. Air Flow		
	0	1	2.5
Multi-Layer PCB, JEDEC Standard Test Boards	74.7°C/W	65.3°C/W	58.5°C/W

### 3. Calculations and Equations.

The purpose of this section is to calculate the power dissipation for the LVPECL output pairs.

LVPECL output driver circuit and termination are shown in *Figure 7*.



**Figure 7. LVPECL Driver Circuit and Termination**

To calculate worst case power dissipation at the output(s), use the following equations which assume a 50Ω load, and a termination voltage of  $V_{CC} - 2V$ .

- For logic high,  $V_{OUT} = V_{OH\_MAX} = V_{CC\_MAX} - 0.82V$   
 $(V_{CC\_MAX} - V_{OH\_MAX}) = 0.82V$
- For logic low,  $V_{OUT} = V_{OL\_MAX} = V_{CC\_MAX} - 1.58V$   
 $(V_{CC\_MAX} - V_{OL\_MAX}) = 1.58V$

$Pd\_H$  is the power dissipation when the output drives high.

$Pd\_L$  is the power dissipation when the output drives low.

$$Pd\_H = [(V_{OH\_MAX} - (V_{CC\_MAX} - 2V)) / R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OH\_MAX})) / R_L] * (V_{CC\_MAX} - V_{OH\_MAX}) = [(2V - 0.82V) / 50\Omega] * 0.82V = \mathbf{19.35mW}$$

$$Pd\_L = [(V_{OL\_MAX} - (V_{CC\_MAX} - 2V)) / R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - (V_{CC\_MAX} - V_{OL\_MAX})) / R_L] * (V_{CC\_MAX} - V_{OL\_MAX}) = [(2V - 1.58V) / 50\Omega] * 1.58V = \mathbf{13.27mW}$$

$$\text{Total Power Dissipation per output pair} = Pd\_H + Pd\_L = \mathbf{32.62mW}$$

## Reliability Information

**Table 7.  $\theta_{JA}$  vs. Air Flow Table for a 16 Lead VFQFN**

Meters per Second	$\theta_{JA}$ by Velocity		
	0	1	2.5
Multi-Layer PCB, JEDEC Standard Test Boards	74.7°C/W	65.3°C/W	58.5°C/W

### Transistor Count

The transistor count for IDT8S898711 is: 239  
 Pin compatible with ICS889871

## Package Outline Drawings

The package outline drawings are located at the end of this document. The package information is the most current data available and is subject to change without notice or revision of this document.

## Ordering Information

**Table 8. Ordering Information**

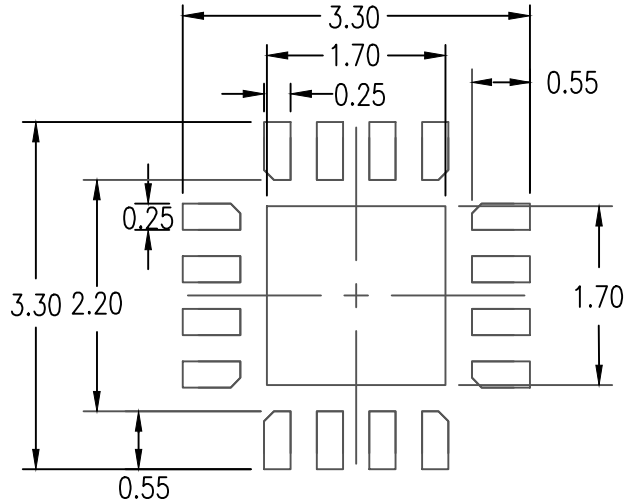
Part/Order Number	Marking	Package	Shipping Packaging	Temperature
8S89871ANLGI	871AI	"Lead-Free" 16 Lead VFQFN	Tube	-40°C to 85°C
8S89871ANLGI8	871AI	"Lead-Free" 16 Lead VFQFN	2500 Tape & Reel	-40°C to 85°C

NOTE: Parts that are ordered with an "G" suffix to the part number are the Pb-Free configuration and are RoHS compliant.

## Revision History

Revision Date	Description of Change
November 28, 2017	Updated the package outline drawings; however, no technical changes Completed other minor changes
April 20, 2012	Initial release.





RECOMMENDED LAND PATTERN DIMENSION

NOTES:

1. ALL DIMENSIONS ARE IN mm. ANGLES IN DEGREES
2. TOP DOWN VIEW-AS VIEWED ON PCB
3. LAND PATTERN RECOMMENDATION IS PER IPC-7351B GENERIC REQUIREMENT FOR SURFACE MOUNT DESIGN AND LAND PATTERN

Package Revision History		
Date Created	Rev No.	Description
Oct 25, 2017	Rev 04	Remove Bookmak at Pdf Format & Update Thickness Tolerance
Aug 15, 2017	Rev 03	Update Epad Range

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