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R32C/118

Six Step Method for Brushless DC Motor based on R32C/118

Introduction

This application note is shown how to use R32C/118 to drive brushless DC motor (BLDCM) using the three phase timer module. The characteristic of the driver is using six step method and balanced PWM outputs to make the same switching loss on high side and low side IGBTs. The LED will be flashing to show that the system is working.

Target Device

The target device is the R32C/100 Series.

Contents

1. The Configuration of the system	2
2. Hardware Specification	3
3. Software Specification.....	8
4. Program follow chart	9
5. Reference Hardware	13
6. Website and Support.....	15

1. The Configuration of the system

The configuration of the system is shown in Figure 1. R32C/118 EVB generates the different PWM output according to the hall sensor signals. The three phase signals are generated after the PWM signals drive the Inverter board.

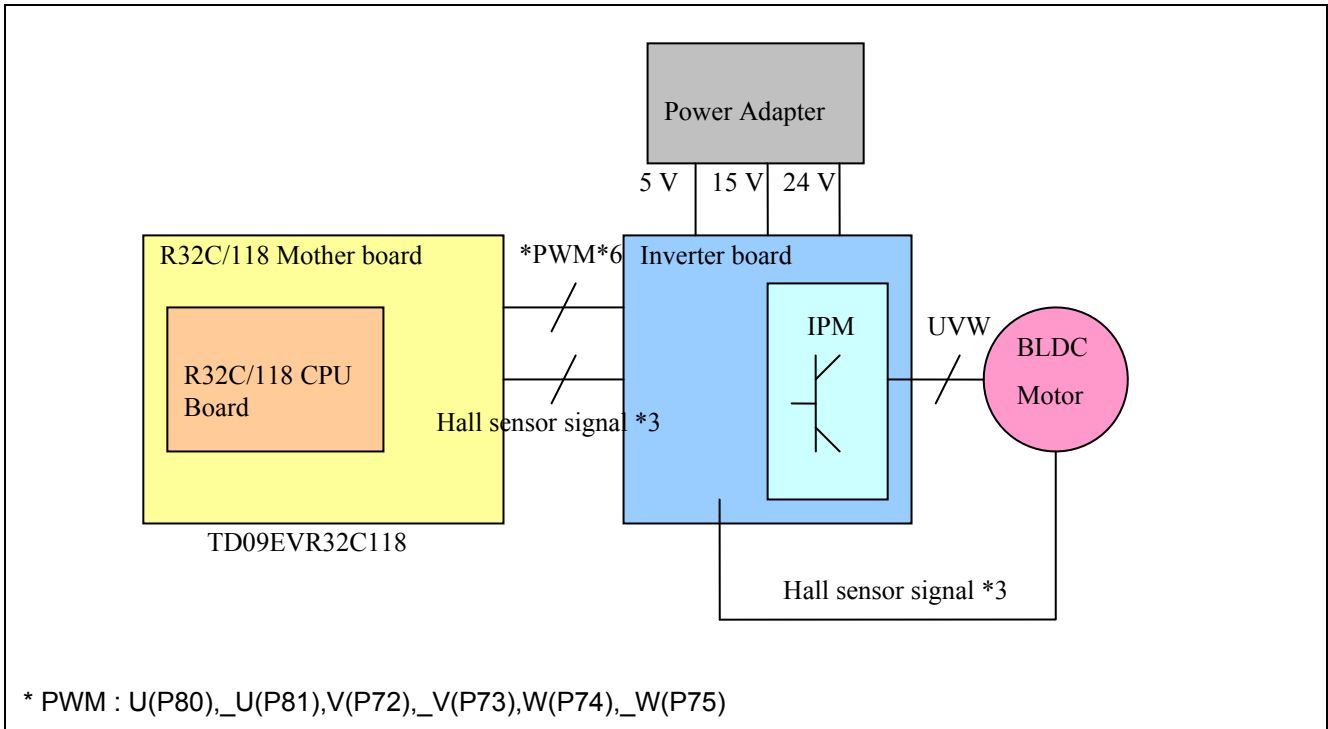


Figure 1 the configuration of the system

2. Hardware Specification

2.1 Brushless DC Motor Specification

The specification of the motor used in this document is shown in Table1 and Figure1. The shape of N2341S014 is shown in Figure 2. The number of the pole in this motor is 2. It is a kind motor of distributed wiring so that the back-EMF waveform of the motor is sinusoidal wave. However, the trapezoidal waveform still can be used to drive this motor and the method is shown in the following.

Table1 Motor specification

Model	N2341S014(PITTMAN® ELCOM ST™)
Number of Pole	2(1 pair)
Sensor:	Hall sensor : U V W Encoder : 500 CPR, A , B , Z
Voltage	24 Volt

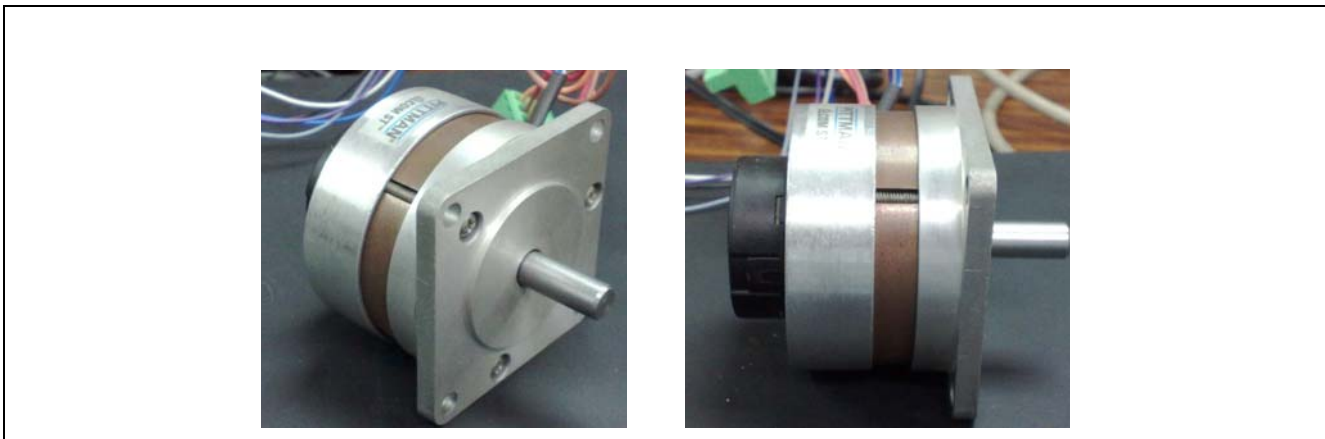


Figure 2 N2341S014

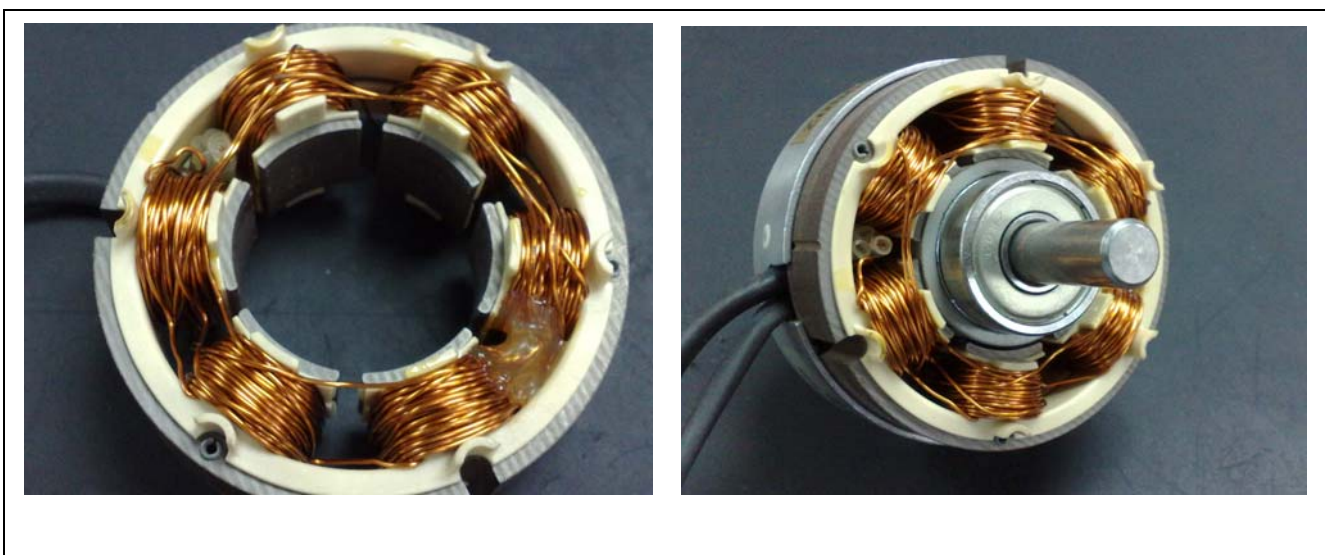


Figure 3 The structure of N2341S014

2.2 The relationship of the hall sensor signals and the Back-EMF

We need to know the relationship between hall sensor signals and the Back-EMF at each phase. A simple method to get this relationship is shown in this paper.

The direction of torque will be decided by the phase delay between Back-EMF and the current into motor. The phase delay should be zero if the rotor rotates in clockwise. However the phase delay should be 180 degree if the rotor will rotate in counter clockwise. Figure 4 shows the phase between inputs current I and Back-EMF.

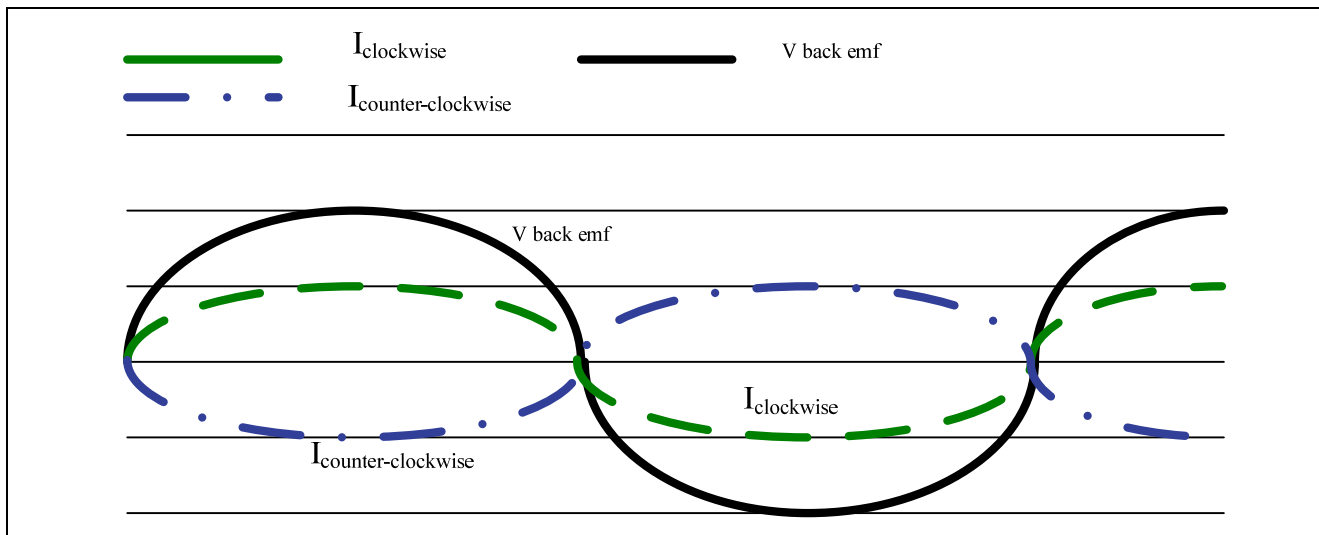


Figure 4 Current phase between Back-EMF and input current

Figure 5 shows the example how to measure the Back-EMF from motor. The voltage of the Back-EMF will be scaled by the rotating speed of the rotor. It will be larger in high speed than in the low speed. We can get the relationship between Back-EMF and sensor signals if we measure those signals simultaneously. Then we can make the switching table to drive the motor according to the measurement result.

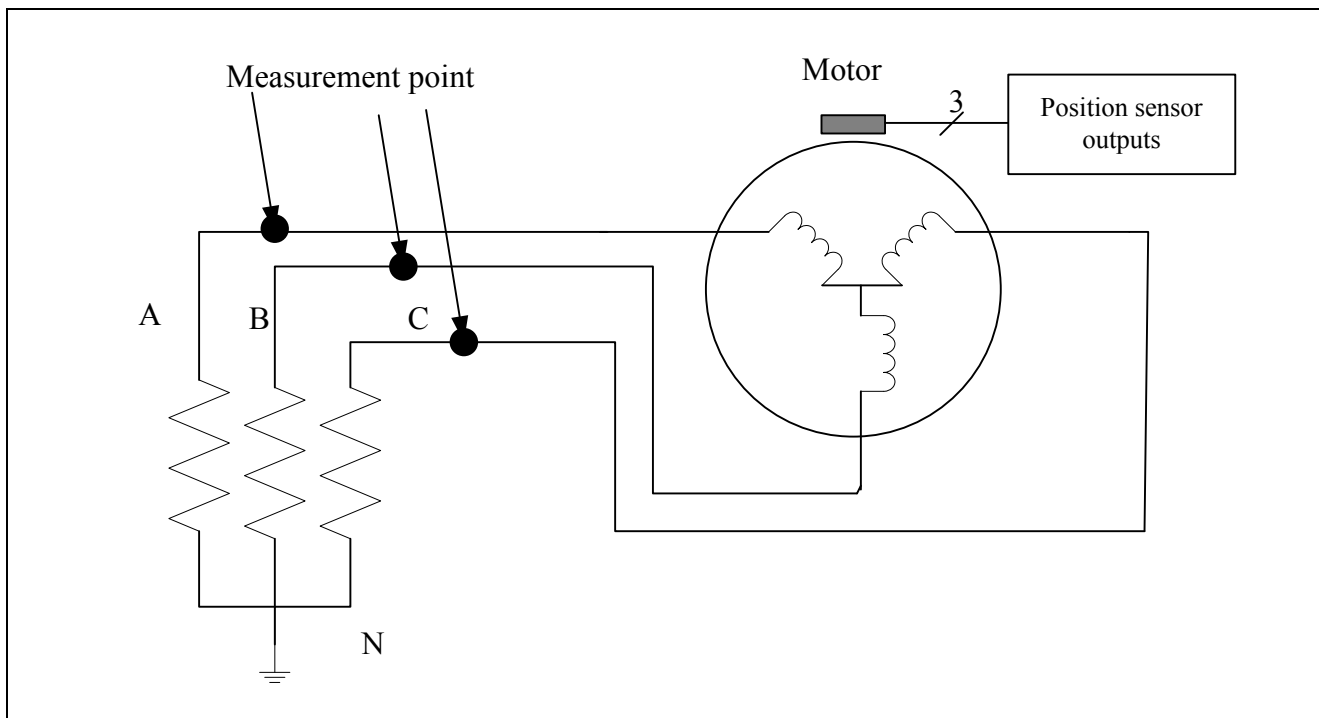


Figure 5 Measurement for the Back-EMF of the motor

Figure 6 can be got by rotating the rotor smoothly. It will be unstable waveform if the rotating speed is not constant. Figure 7 can be got according to Figure6. The unit of X axis is electrical phase. The switching table can be made by the result of Figure 7.

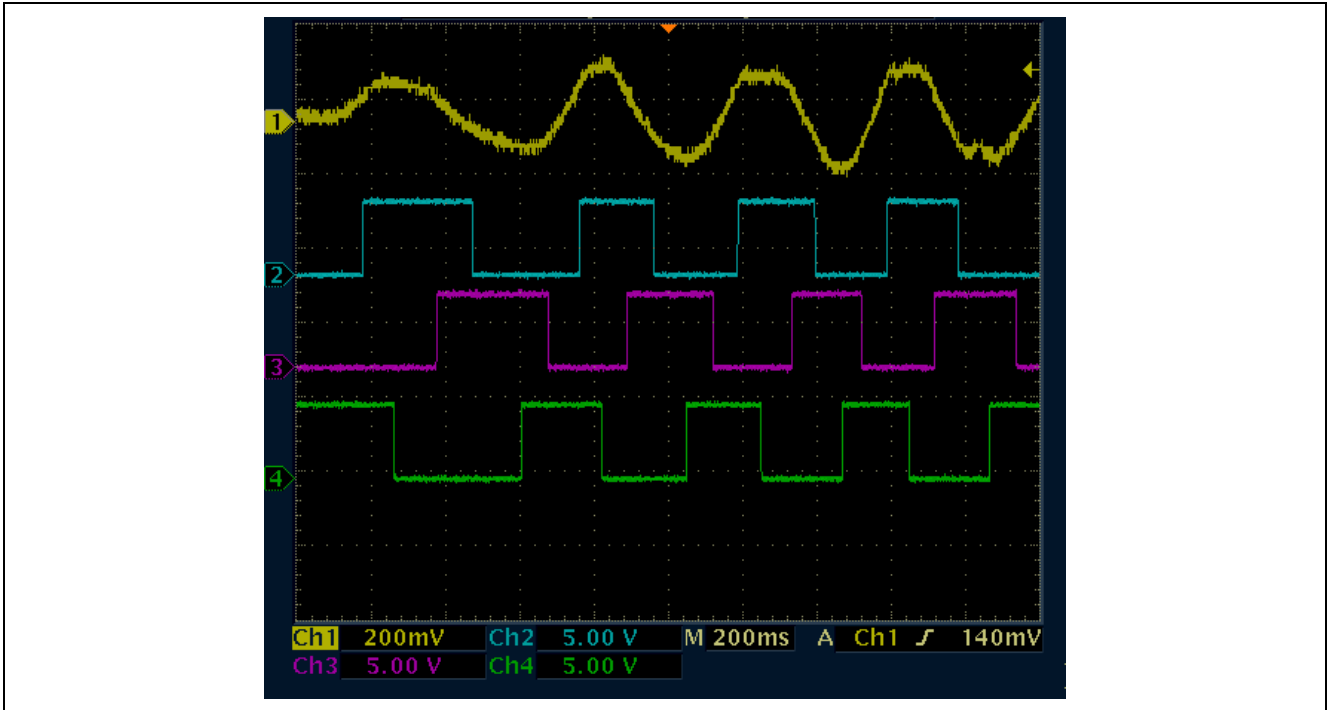


Figure 6 The Back-EMF and Hall sensor signals from scope

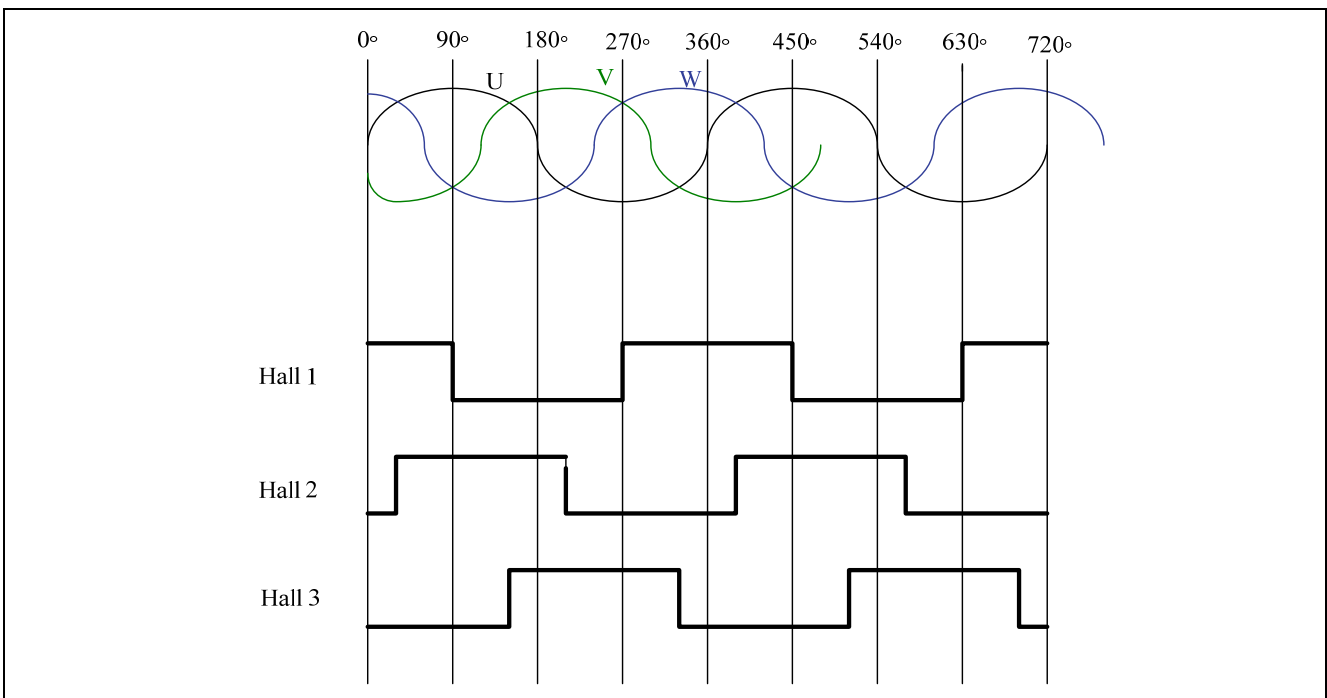


Figure 7 The signal follow chart between position sensor and the Back-EMF

2.3 Construct the switching table

How to use trapezoidal waveform to drive the motor which is the type of the distributed winding is shown in Figure 8. The blue line is the Back-EMF of the motor and the red line is the driving voltage. The simple model of the system is shown in Figure 9. In the Eq1 it explains how to get the value of the input current *i*. However it is just a brief expansion because of ignoring the real complex model. The trapezoidal can not fully match to sinusoidal waveform so that the torque will not be a constant. The noise and torque ripper will be larger in using trapezoidal waveform than sinusoidal waveform for distributed winding.

$$i = \frac{V_{out} - E}{r} \text{----- Eq.1}$$

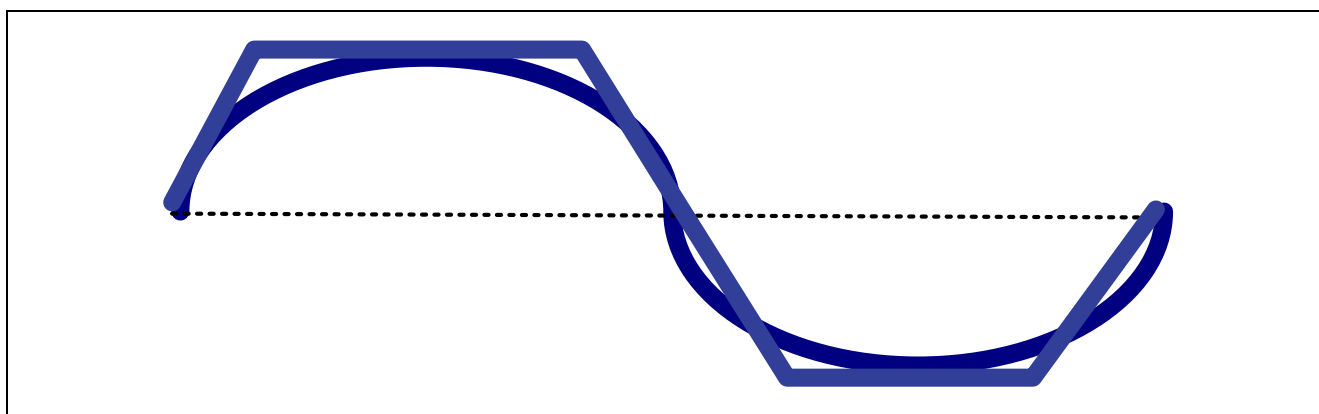


Figure 8 Use trapezoidal waveform to replace sinusoidal waveform

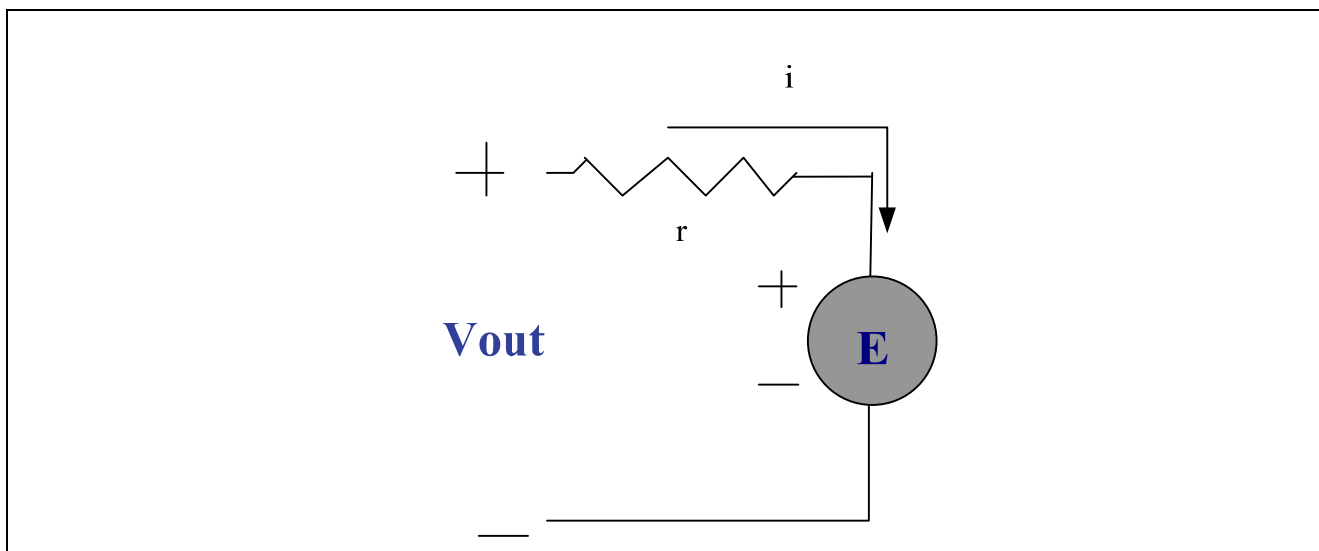
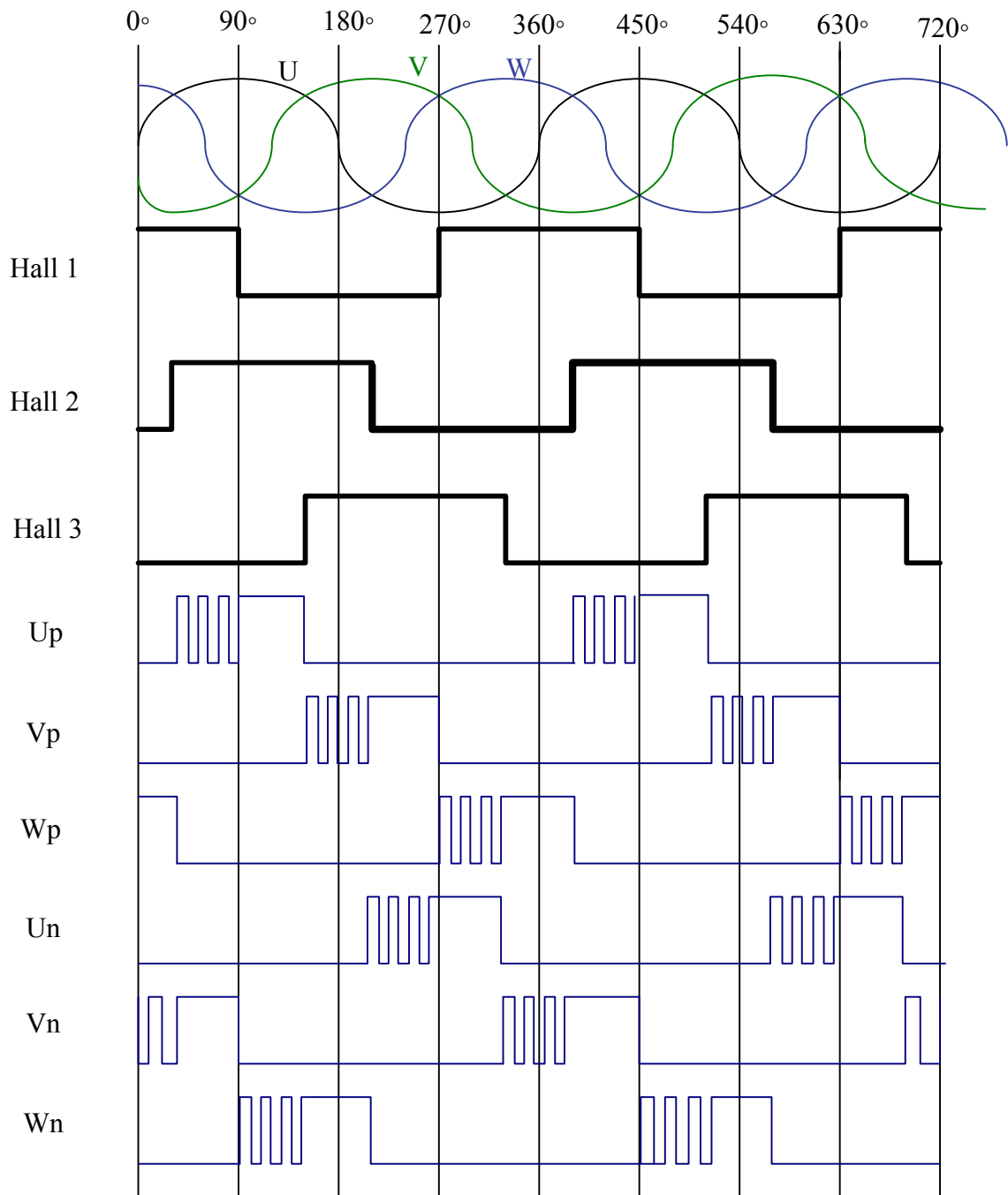


Figure 9 The driving voltage and Back-EMF

By turning on the switch during 120 degrees of a half electrical cycle, the trapezoidal waveform can be constructed. A total electrical phase can be divided into six zones. Each zone is 60 degrees. There are always two switches turned on at each zone. This method is called “six step method”, “120 degrees driving method” or “trapezoidal waveform driving” etc.

Figure 10 shows the switching table according to Figure 7. Each PWM signal is output during 120 degrees at a electrical period. The signal is chopping at the first 60 degrees and then full opening at the last 60 degrees. This method is called balanced PWM output which means the both positive and negative signal are chopping. The IGBTs which are for the one phase will consume the same switching loss. The life cycle of the IGBT can be extended.



U : U phase back-emf V : V phase back-emf W : W phase back-emf

Hall1 : Hall sensor 1 Hall2 : Hall sensor 2 Hall3: Hall sensor 3

Up : PWM output for positive phase of U Wp : PWM output for the positive phase of W

Un : PWM output for negative phase of U Wn : PWM output for the negative phase of W

Vp : PWM output for positive phase of V

Vn : PWM output for negative phase of V

Figure 10 six-step switching table

3. Software Specification

The software specification of the sample code is shown on the Table 2.

Table 2 Software specification

Control method	Six-Step method (120-degree commutation using trapezoidal waves)
Rotor position detection	Detecting by hall sensors
Carrier frequency	20KHz
PWM method	Balanced PWM method (chopping at high side and low side)

The positive phase and negative phase will be chopping at each turned-on period. We called this balanced pwm method in Figure 11. There is the same switching loss for the IGBTs or MOSFETs of the inverter at high side and low side. The method is also useful to reduce the ripper of the torque.

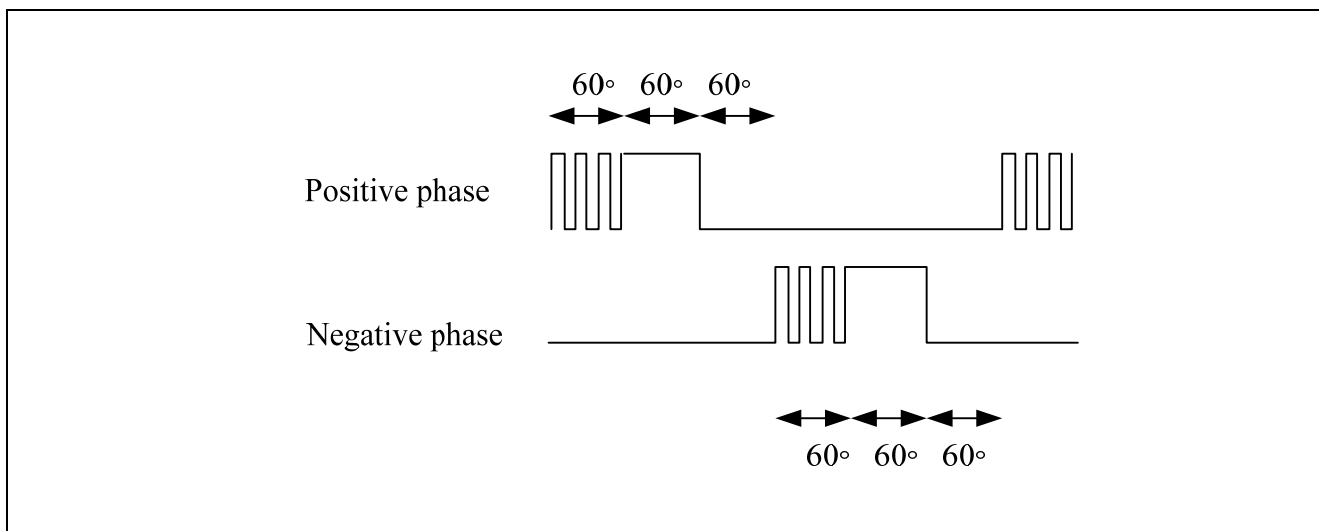


Figure 11 Balanced PWM method

4. Program follow chart

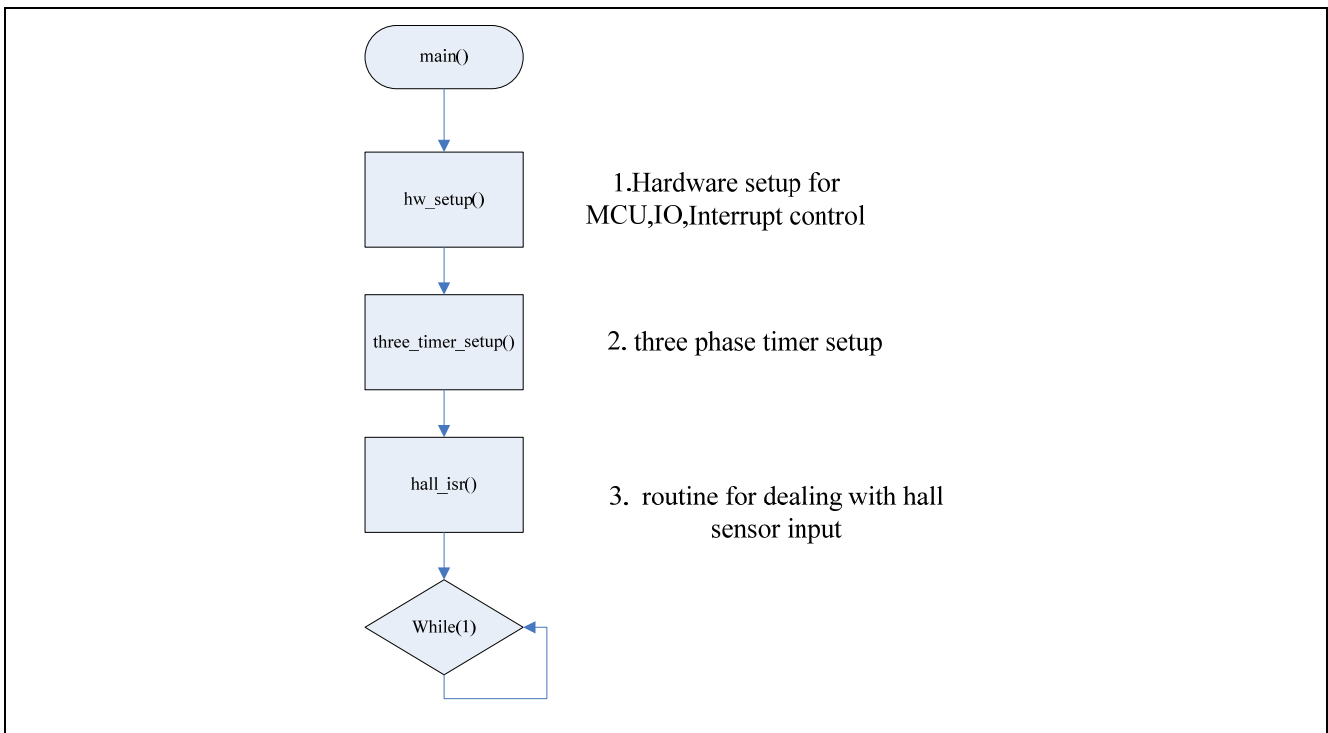


Figure 12 Program follow chart of main()

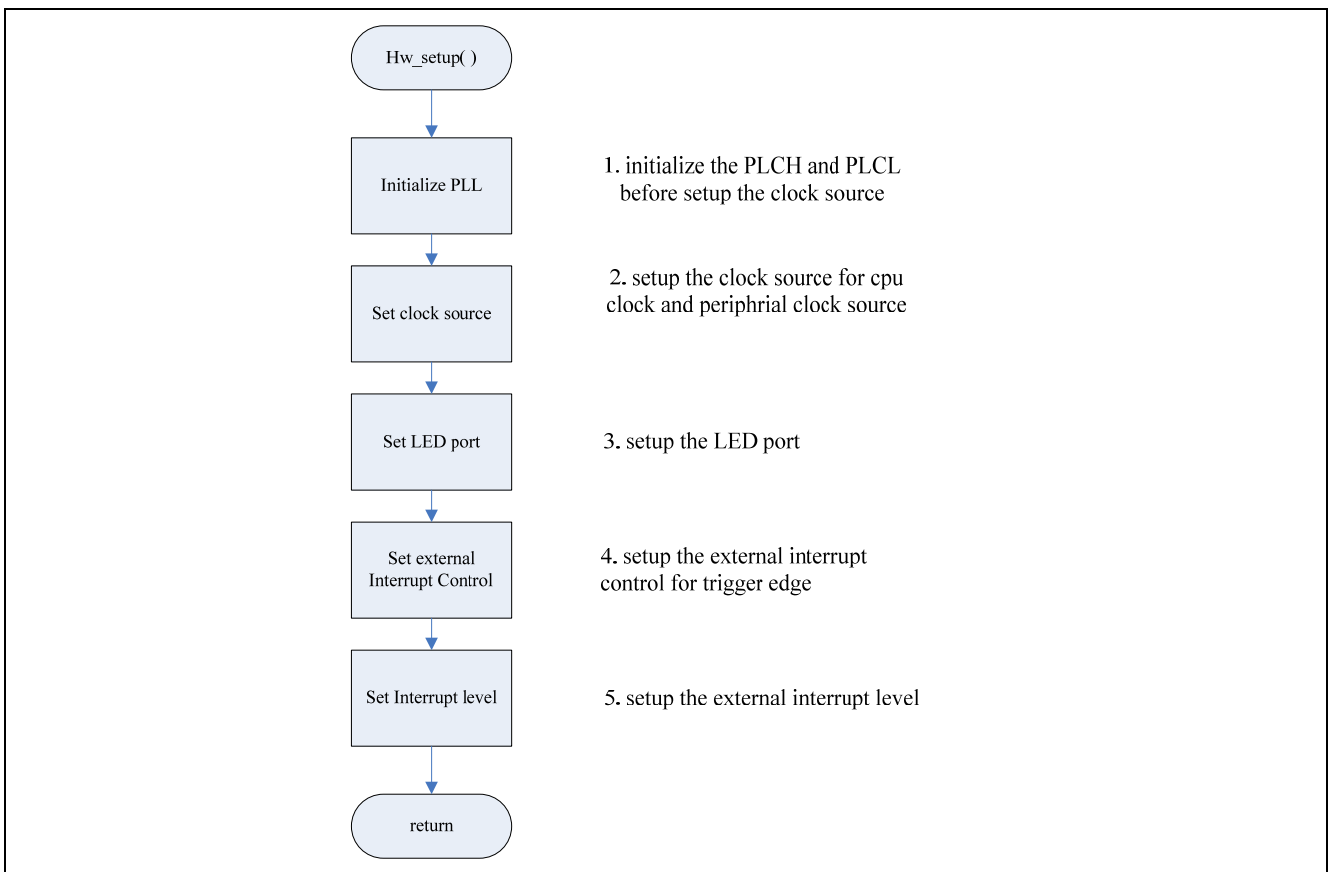


Figure 13 Program follow chart of hw_setup()

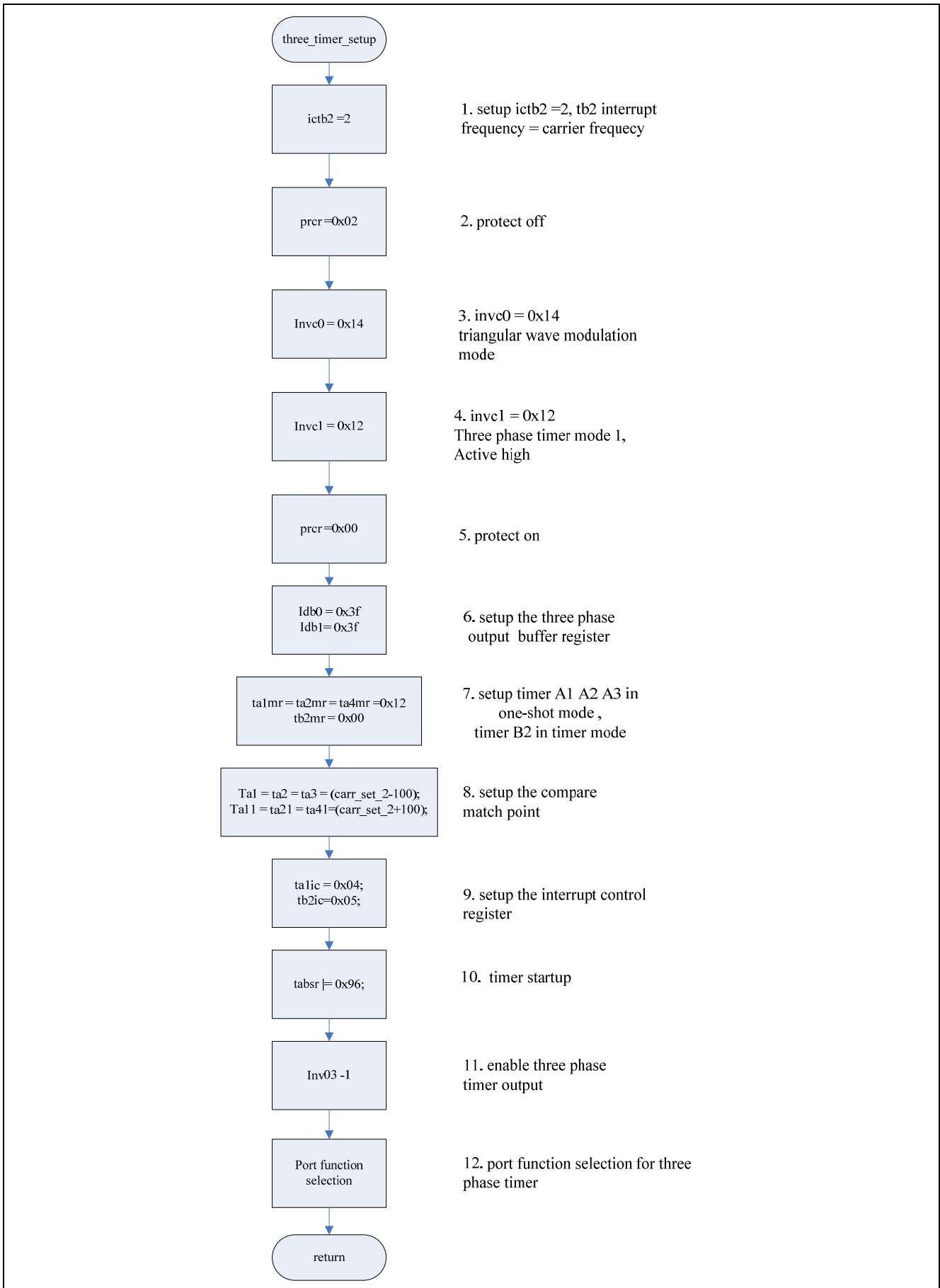


Figure 14 Program follow chart of three phase timer setup

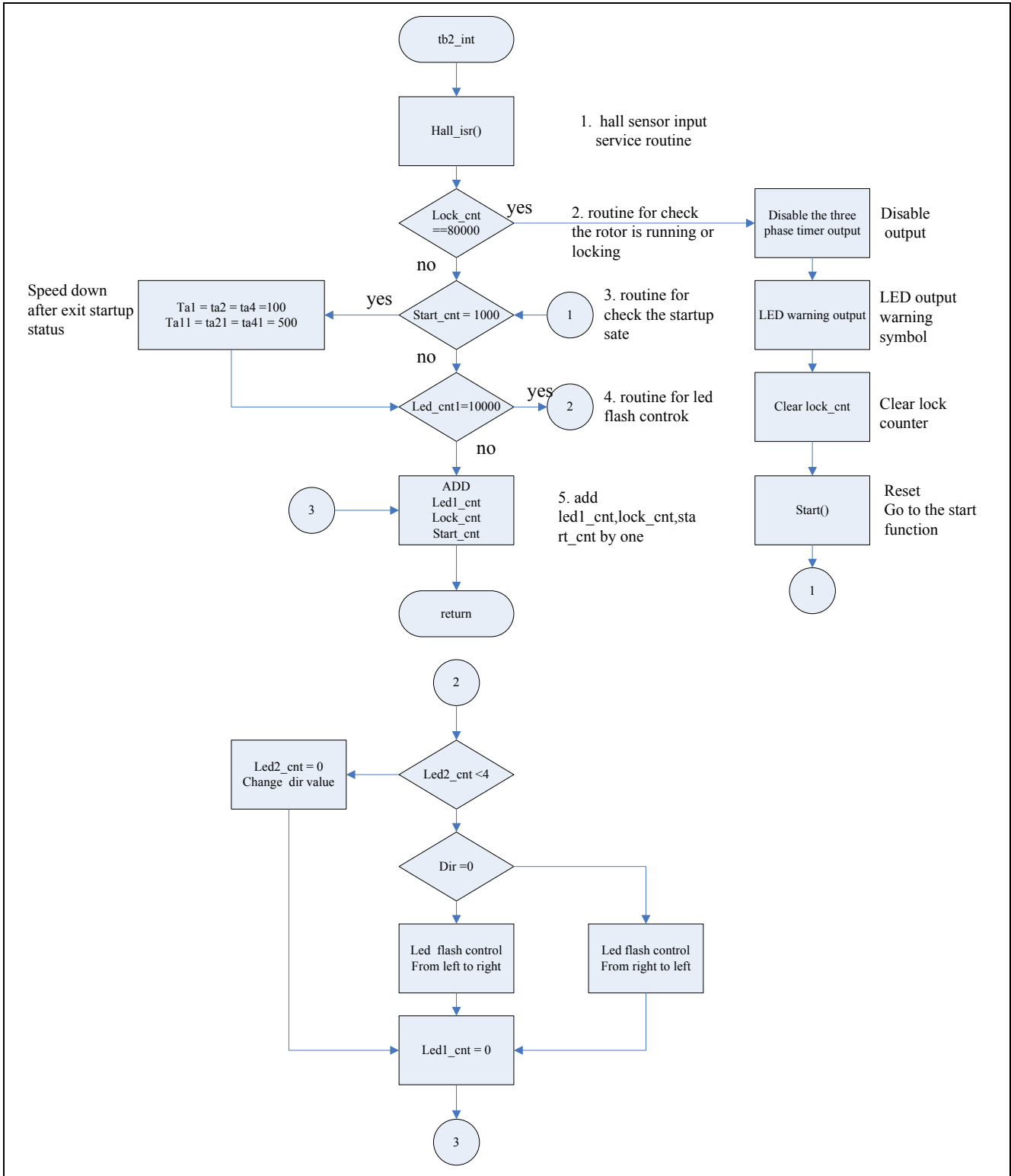


Figure 15 Program follow chart of tb2_int()

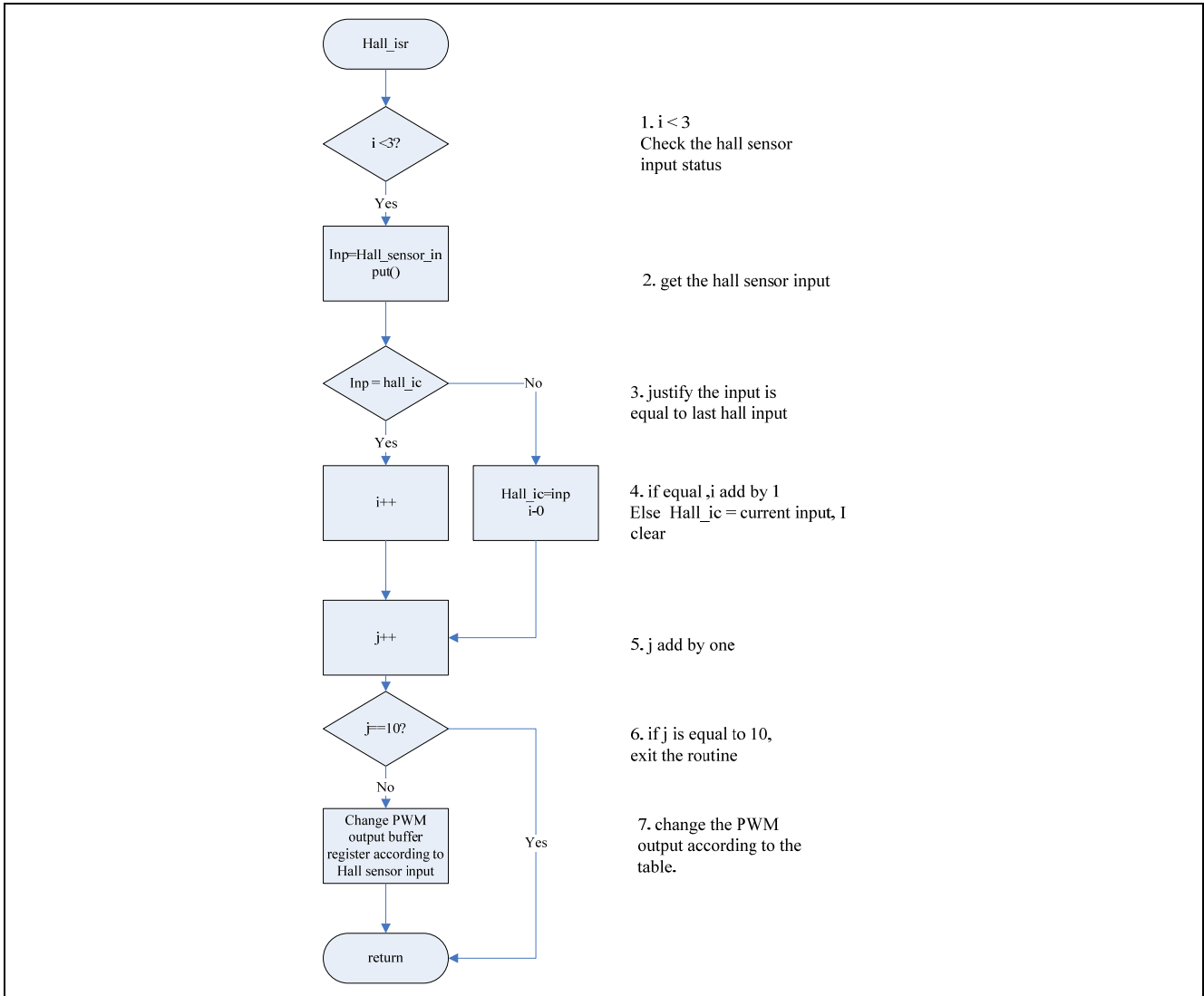


Figure 16 Program follow chart of hall_isr()

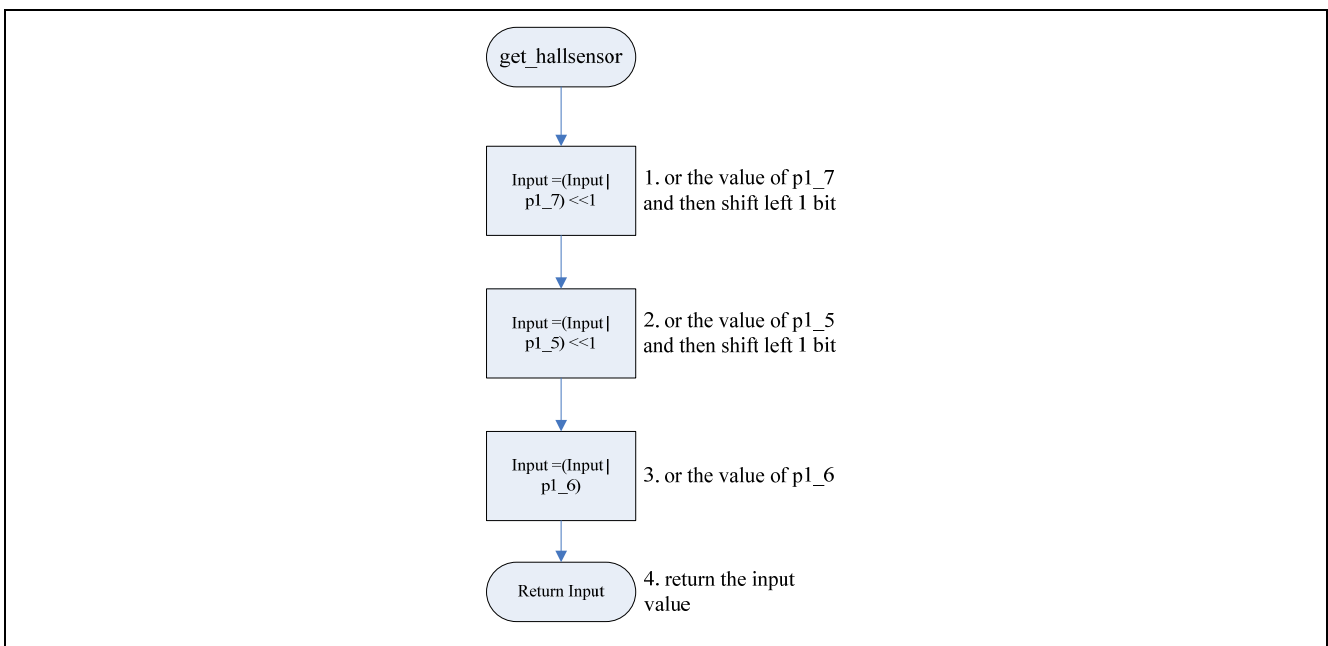


Figure 17 Program follow chart of get_hallsensor()

5. Reference Hardware

The reference circuit of the inverter will be shown in this chapter. The connections are shown in Figure 18.

The six PWM outputs are connected to IPM and the hall sensor signals are feedback through the pulling-high circuit.

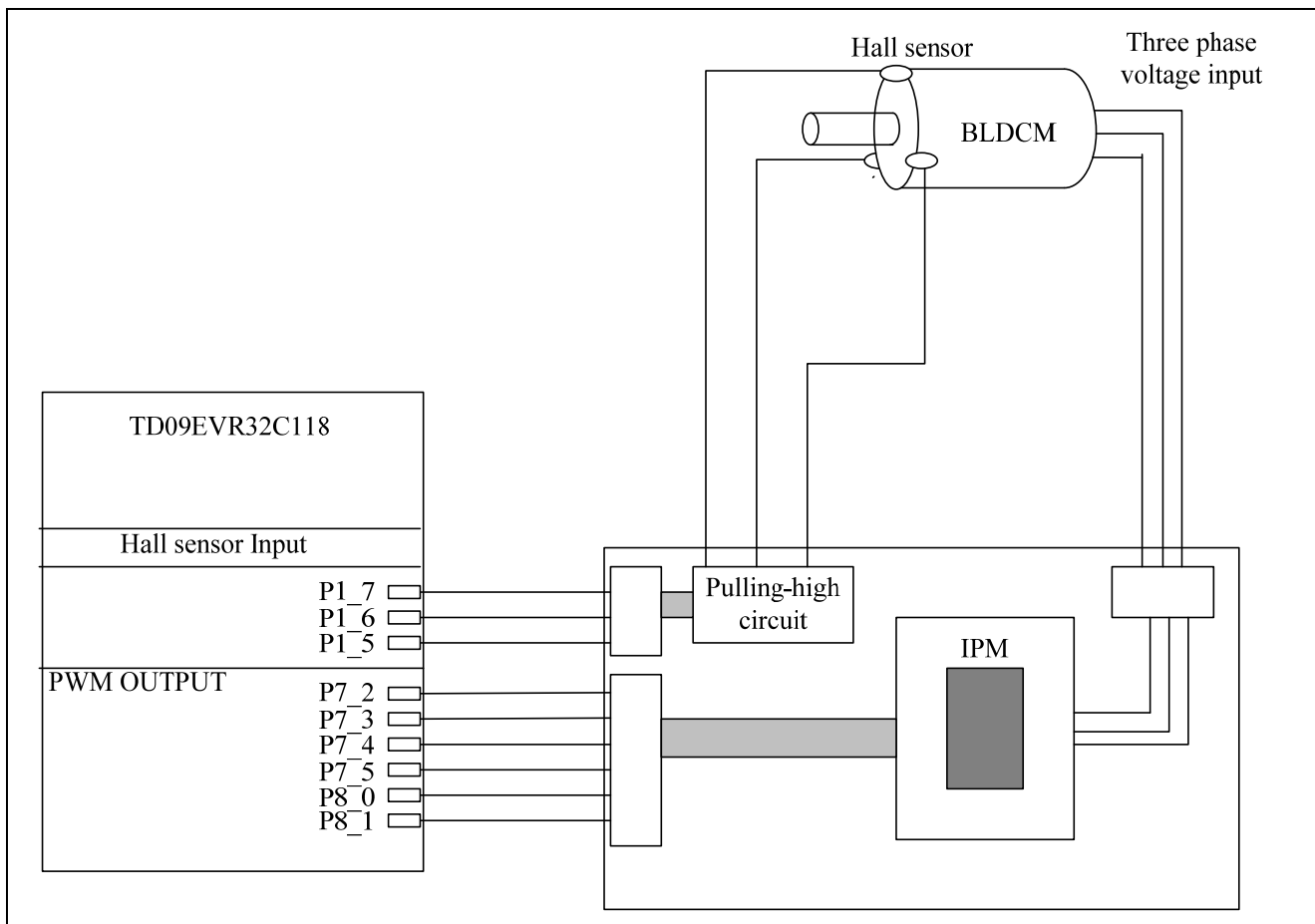


Figure 18 The hardware connection

5.1 Reference circuit of the pulling-high circuit for hall sensor input

The output stage of the hall sensors is open-drain and there are pulling-high circuit to supply the voltage to signals.

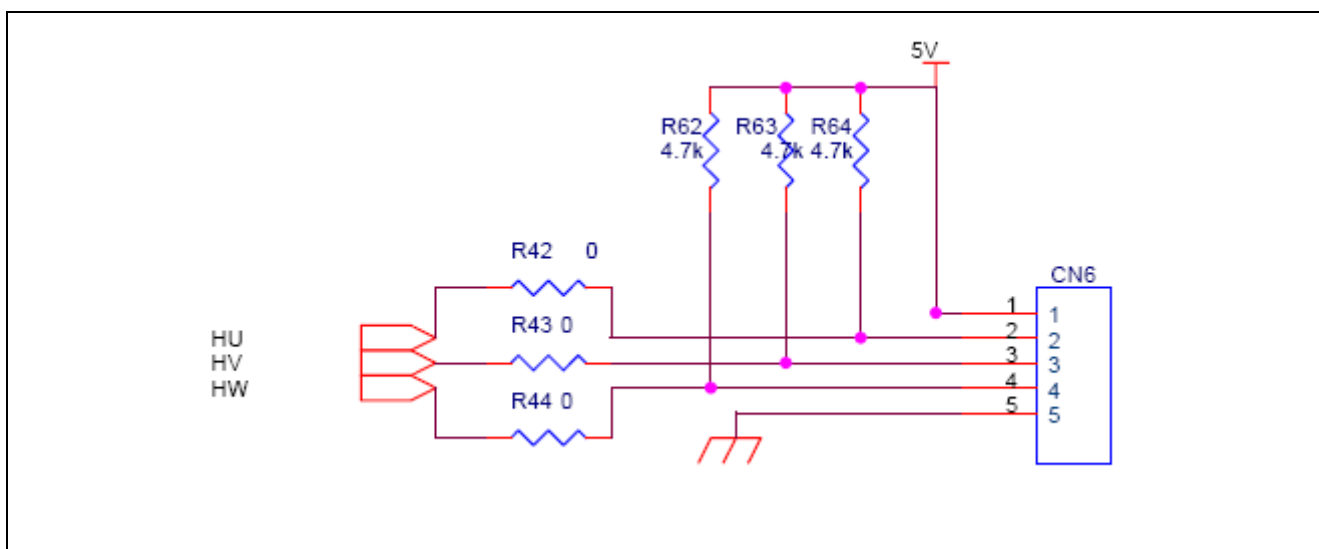


Figure 19 the pulling high circuit

5.2 Reference circuit of inverter circuit

The reference inverter circuit is shown in Figure 20. The IPM module is PS21964.

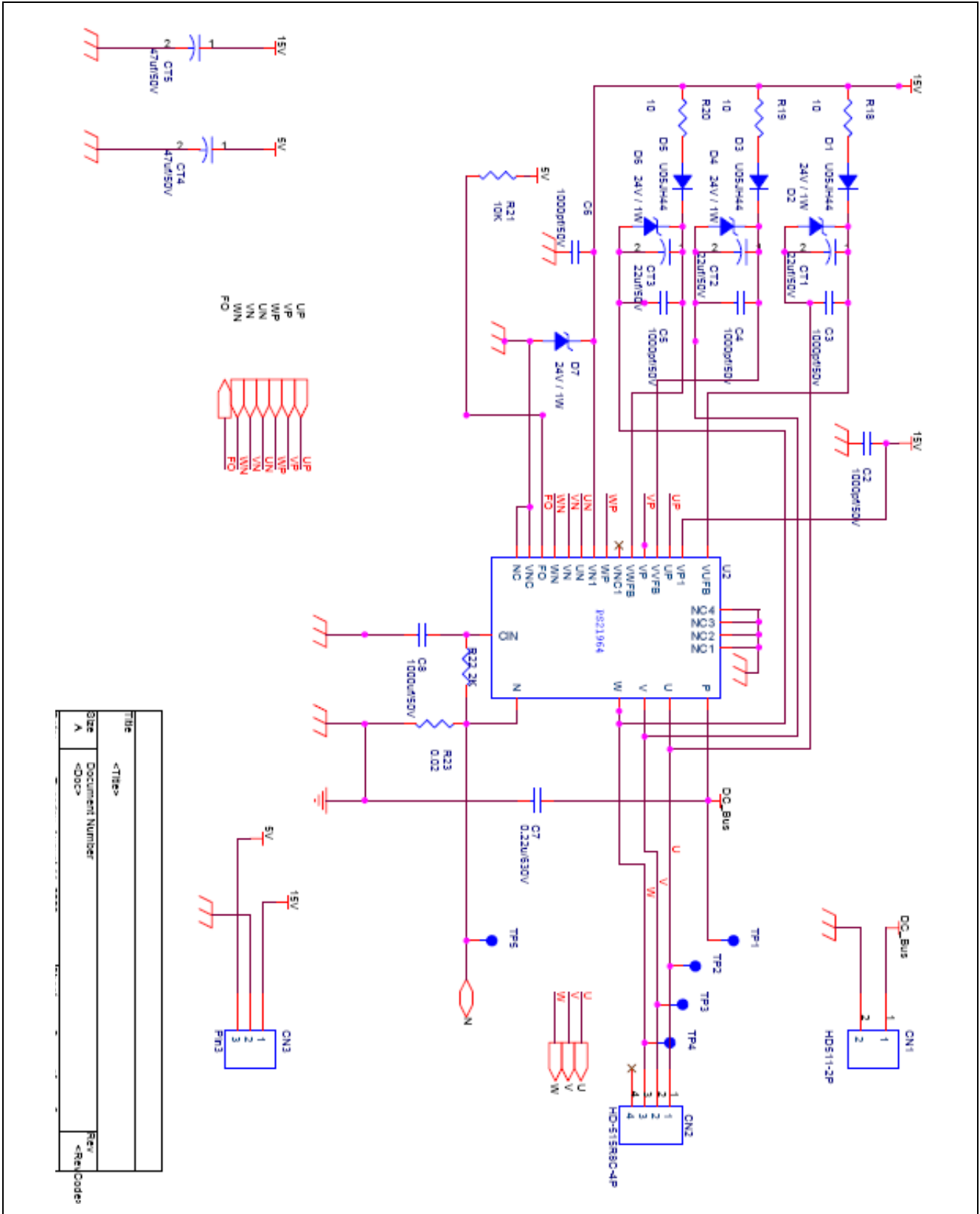


Figure 20 The reference inverter circuit

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		Page	Summary
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