

# **Application Note**

# Selecting Input and Output Capacitors with GreenFET Load Switches

**AN-CM-251** 

#### **Abstract**

This application note is devoted to the topic of selecting proper capacitors to use with GreenFET Load Switches. Important considerations include desired electrical performance, system transient requirements, load parameters, and voltage deviation specifications.



## **Contents**

Αb	ostract	1
Contents Figures		2
		2
1	Terms and Definitions	
2	References	_
3	Introduction	
4	Capacitor Parasitic Effects	4
5	How to Select Capacitors	5
6	Conclusions	13
	gures	
Fig	gure 1. Capacitor Equivalent Circuit	4
	gure 2. A Typical Power Up Behavior of a SLG59M1714Vgure 3. Powering Up a Load Switch with Fast V <sub>DD</sub> and V <sub>IN</sub> Rise Times	
	gure 4. Powering Up a Load Switch when V <sub>DD</sub> and V <sub>IN</sub> Are Applied Simultaneously	
	gure 5. Powering Up a Load Switch when V <sub>IN</sub> is Applied Before V <sub>DD</sub>	
	gure 6. A Load Switch Powering Up a 1 $\mu$ F Load Capacitor with Fast $V_{DD}$ and $V_{IN}$ rise Times gure 7. The GreenFET Load Switch Powering Up a 1 $\mu$ F Load Capacitor when $V_{DD}$ and $V_{IN}$ are	7
Αp	plied Simultaneously.	8
	gure 8. The GreenFÉT Load Switch Powering Up a 1 μF Load Capacitor when V <sub>IN</sub> is Applied fore V <sub>DD</sub>	0
Fic	gure 9. A Fast, Nanopower Load Switch Powering Up a 10 µF Load Capacitor	9
Fig	gure 10. A Fast, Nanopower Load Switch Powering Up a 10 μF Load Capacitor with a 1000 μF	
	uminum Electrolytic Capacitor at VIN (C <sub>IN</sub> ).	
	gure 11. A Fast, Nanopower Load Switch Powering up a 10 μF Load Capacitor with a 270 μF Lo SR/ESL Aluminum Polymer Input Capacitor at VIN (C <sub>IN</sub> )	
	gure 12. The Parasitic Inductance of 1.2-m Length AWG13 Wires from Power Supply Causes	. 10
Vo	Itage Spikes at V <sub>IN</sub> during Active Current Limit Operation	
	gure 13. Shorter, 0.2-m Length of AWG13 Wires from Power Supply Do Not Cause Voltage Spik	
Fig	V <sub>IN</sub> during Active Current Limit Operation gure 14. Adding 470 μF Low ESR/ESL Aluminum Polymer Capacitor at VIN Eliminates Voltage ikes Caused by the Parasitic Inductance of the 1.2-m Length AWG13 Wires from Power Supply	
	ring Active Current Limit Operation.	



#### 1 Terms and Definitions

ACL Active Current Limit

ESL Equivalent Series Inductance
ESR Equivalent Series Resistance
SCL Short-circuit Current Limit

#### 2 References

- [1] SLG59M1714V, Datasheet, Renesas Electronics
- [2] AN-1207, Load Switch Considerations for Inductive Loads, Renesas Electronics
- [3] AN-1068, GreenFET and High Voltage GreenFET Load Switch Basics, Renesas Electronics
- [4] AN-CM-246 Using GreenFET Load Switches in Super Capacitor Applications

Author: Andrii Hrypa and Petro Zeykan



#### 3 Introduction

Capacitors are used in almost all electronic products in a variety of ways. Capacitors provide a number of essential functions in circuit design, such as providing flexible filter options, noise reduction, power storage, and sensing capabilities. To effectively use capacitors in load switch designs, some key details need to be taken into account. Desired electrical performance, system transient requirements, load parameters, and voltage deviation are very important to consider when selecting proper input and output capacitors in a given application.

Renesas Electronics offers a wide range of GreenFET Load Switches for a variety of applications. For more information, please visit <a href="https://www.dialog-semiconductor.com/products/load-switches">https://www.dialog-semiconductor.com/products/load-switches</a>.

## 4 Capacitor Parasitic Effects

While an ideal capacitor is capable of transferring all its stored energy to a load instantaneously, a real capacitor has parasitic components that prevent this behavior. An equivalent circuit model of a real capacitor is illustrated in Figure 1. As shown, the capacitor equivalent circuit comprises four elements: capacitance, an equivalent series inductance (ESL), a high-resistance DC path (Rp) in parallel with the capacitance, and an equivalent series resistance (ESR).

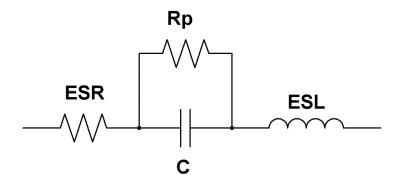


Figure 1. Capacitor Equivalent Circuit

The electrodes and the leads of a capacitor contributes the resistive and inductive components while its dielectric material and its construction contribute to the insulation resistance.

High ESR degrades performance due to I<sup>2</sup>R losses, noise, and larger voltage drop. On the other hand, ESL causes a magnetic field to buildup in capacitor. The buildup of magnetic field interferes with how current peaks and recovers. Both ESR and ESL depend on the type of capacitor and its construction.

Wet aluminum electrolytic capacitors are used primarily for bulk decoupling applications. However, their relatively high ESR and ESL slow response times and reduce performance.

Aluminum polymer capacitors have better performance characteristics, and they are increasingly replacing wet aluminum capacitors in bulk decoupling applications. Aluminum polymer capacitors exhibit much lower parasitic ESR and ESL.

Tantalum capacitors are a subclass of electrolytic capacitors. They are made of tantalum metal which acts as an anode, covered by a layer of oxide which acts as the dielectric, then surrounded by a conductive cathode. Tantalum capacitors have an equivalent series resistance (ESR) ten times smaller than the ESR of aluminum electrolytic capacitors, which allows for larger currents to pass through with less heat generated and, in addition, smaller, parasitic IR voltage drops. Tantalum



capacitors are very stable over time and their capacitance doesn't change with age significantly, especially when compared to aluminum electrolytic capacitors.

Ceramic capacitors are most commonly used in electronic circuits for decoupling applications. They have relatively low equivalent series resistance, but their ESL is greatly determined by the distance between terminations (its construction).

If PCB space is not an issue, capacitors can be connected in parallel to reduce both ESR and ESL while beneficially increasing the effective capacitance.

## 5 How to Select Capacitors

In a general sense, GreenFET load switches don't require any input or output capacitors. The use of input and output capacitors is determined by the usage scenario (application) Typically, to turn on a Renesas load switch,  $V_{DD}$ ,  $V_{IN}$ , and ON signals are applied. Every GreenFET load switch datasheet contains an information on the proper sequencing of these three signals. In short  $V_{DD}$  should be applied first, then  $V_{IN}$ , and finally the ON signal can be toggled low-to-high (for asserted-HIGH ON signals or high-to-low for asserted-LOW ON signals) to close the switch. Also, it is recommended that  $V_{DD}$  and  $V_{IN}$  rise times should be longer than 2 ms. As an example a typical power up operation of SLG59M1714V [1] is illustrated in Figure 2.

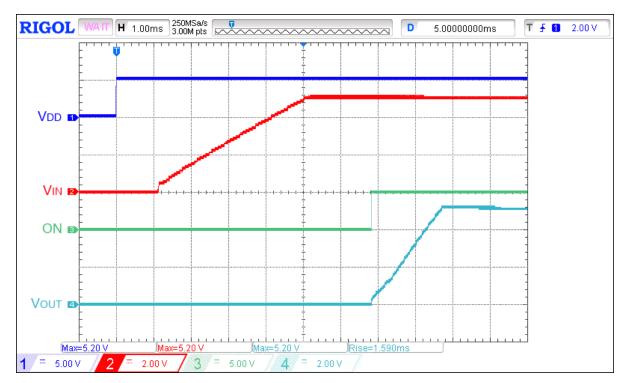


Figure 2. A Typical Power Up Behavior of a SLG59M1714V.

However, there are some applications when  $V_{DD}$  and  $V_{IN}$  have fast rise times (Figure 3), or when  $V_{DD}$  and  $V_{IN}$  are applied simultaneously (Figure 4), or when  $V_{IN}$  is applied before  $V_{DD}$  (Figure 5). In each of these cases, a voltage glitch may appear at the output even when ON = GND (for asserted-HIGH ON signals).



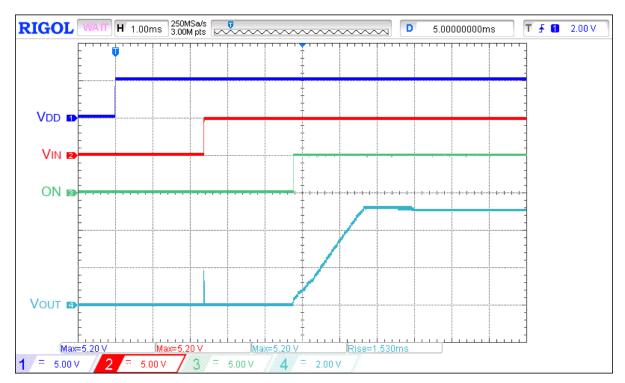


Figure 3. Powering Up a Load Switch with Fast  $V_{\text{DD}}$  and  $V_{\text{IN}}$  Rise Times.

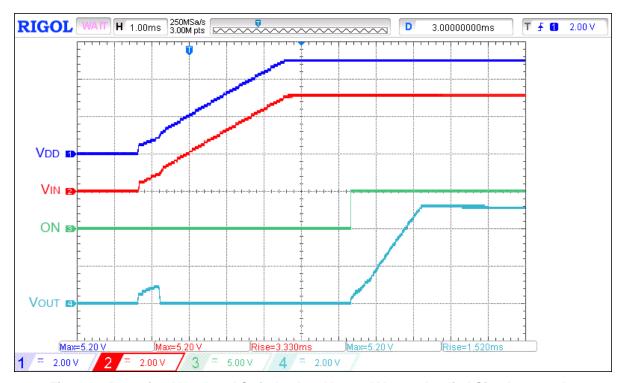


Figure 4. Powering Up a Load Switch when  $V_{DD}$  and  $V_{IN}$  are Applied Simultaneously.



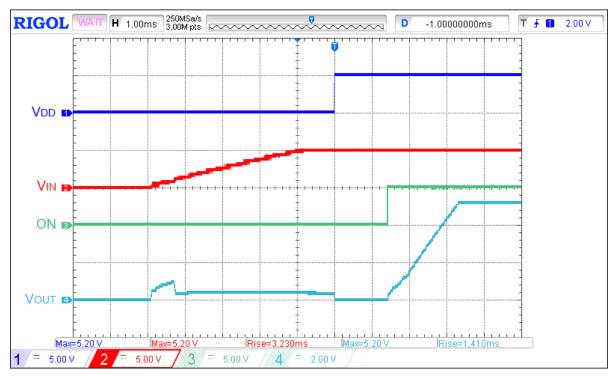


Figure 5. Powering Up a Load Switch when V<sub>IN</sub> is Applied Before V<sub>DD</sub>.

To avoid such glitches, an output capacitor (C<sub>LOAD</sub>) or a resistor (R<sub>LOAD</sub>) should be added at the downstream side of the GreenFET load switch. In this particular case, a 1µF ceramic capacitor at VOUT helps to get rid of that glitch (Please see Figure 6, Figure 7, and Figure 8, inclusive).

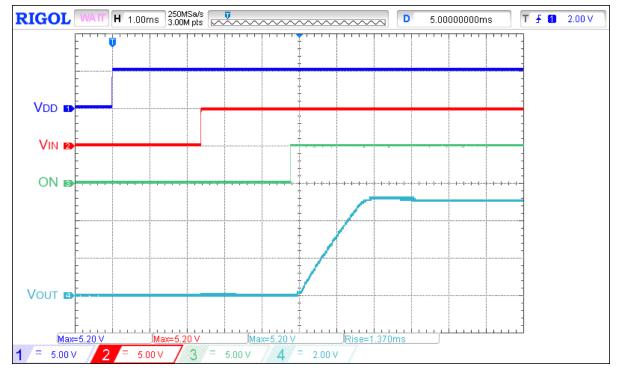


Figure 6. A Load Switch Powering Up a 1 μF Load Capacitor with Fast V<sub>DD</sub> and V<sub>IN</sub> rise Times.



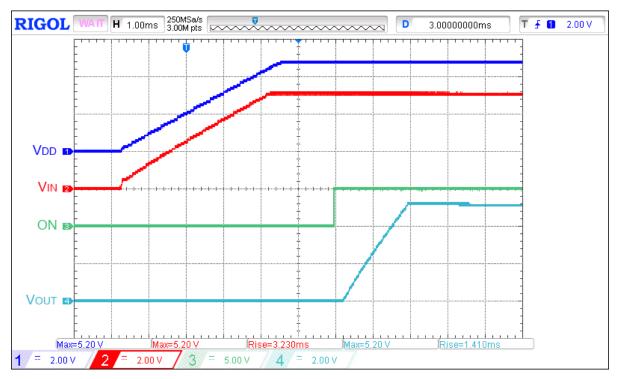


Figure 7. The GreenFET Load Switch Powering Up a 1  $\mu F$  Load Capacitor when  $V_{DD}$  and  $V_{IN}$  are Applied Simultaneously.

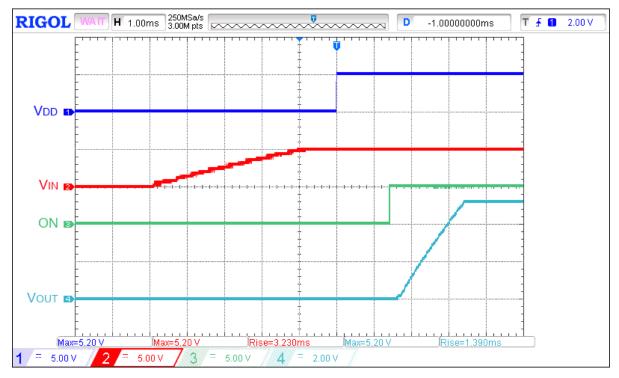


Figure 8. The GreenFET Load Switch Powering Up a 1  $\mu$ F Load Capacitor when  $V_{IN}$  is Applied Before  $V_{DD}$ .



In fast turn-on applications into a capacitive load, some unwanted effects may be observed. One such effect is illustrated in Figure 9, where a Nanopower GreenFET load switch powers up a 10  $\mu$ F load capacitor.

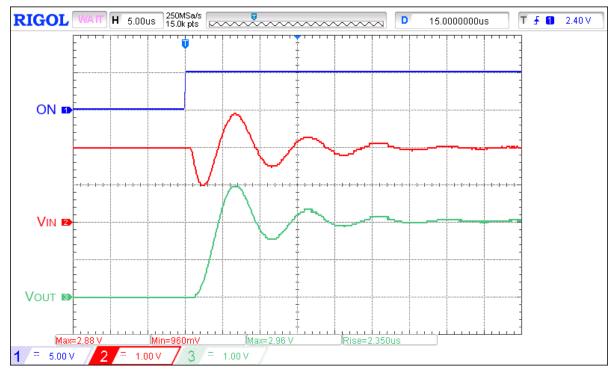


Figure 9. A Fast, Nanopower Load Switch Powering Up a 10 µF Load Capacitor.

This behavior is related to a big inrush current [2], caused by applying a voltage through a load switch across a discharged (or an uncharged) capacitor. The resulting inrush current can be calculated by equation below:

Inrush Current, 
$$I = C \frac{dV}{dt}$$
,

where

C - is the total load capacitance;

 $\frac{dV}{dt}$  - the GreenFET load switch's V<sub>OUT</sub> slew rate during voltage ramp up.

This inrush current leads to the voltage drop at  $V_{\text{IN}}$  during a load switch power up. Also, this current builds a magnetic field in the parasitic inductance caused by wires from the power supply. When the voltage drop occurs, the magnetic field changes in strength and collapses. This leads to voltage spikes that appear at  $V_{\text{IN}}$  and, respectively, at  $V_{\text{OUT}}$ . These voltage spikes can be much larger than the initial  $V_{\text{IN}}$  voltage level and can greatly shorten the load switch's long-term reliability [2] or even damage it and any other circuit downstream of it.

One way to minimize this effect is to reduce inrush current for a given load capacitance by decreasing (or slowing down) the GreenFET load switch's V<sub>OUT</sub> slew rate. This can be achieved by using load switches with controlled slew rate [3], [3]. However, this method also leads to increasing total circuit turn-on time.

Another way is to add an input capacitor at  $V_{IN}$  to minimize the voltage drop during fast power-up events. Figure 10 shows a turn-on operation for a 10  $\mu$ F load capacitor with a 1000  $\mu$ F aluminum electrolytic capacitor at VIN. As shown, the voltage drop is much smaller, but it is still present due to ESR and ESL parasitic elements in capacitors described earlier. In the case of using a capacitor with



smaller ESR and ESL, it is possible to get the same or even better results with smaller capacitance value. For example, in Figure 11, a low ESR/ESL aluminum polymer capacitor was used to power up a 10uF load capacitor.

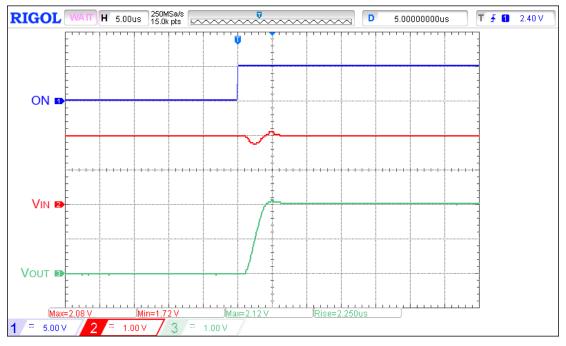


Figure 10. A Fast, Nanopower Load Switch Powering Up a 10 μF Load Capacitor with a 1000 μF Aluminum Electrolytic Capacitor at VIN (C<sub>IN</sub>).

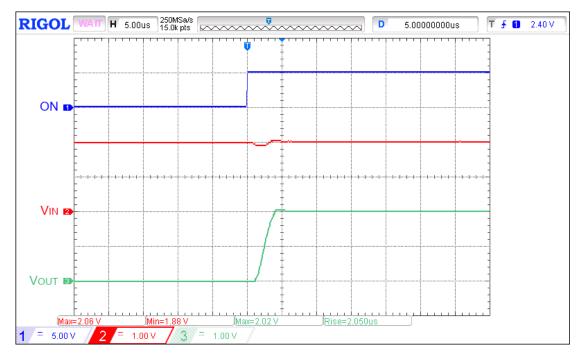


Figure 11. A Fast, Nanopower Load Switch Powering up a 10 μF Load Capacitor with a 270 μF Low ESR/ESL Aluminum Polymer Input Capacitor at VIN (C<sub>IN</sub>).



Yet another situation which requires input and output capacitors is when a load switch's active current limit (ACL) or short-circuit protection (SCL) is triggered. During these events, current through the load switch may change in large steps (eg. suddenly shut off) which if long wires (inductance) are present can cause large voltage spikes (Figure 12) that could damage the load switch and even other components powered from the same power rail. To eliminate these voltage spikes, it is necessary to a) use shorter connections from power supply to the device (Figure 13) and/or b) add or increase the corresponding capacitance (Figure 14). Since long wires tend to be more prevalent on the power supply ( $V_{IN}$ ) side, the examples below show how the voltage spikes in this situation are mitigated.

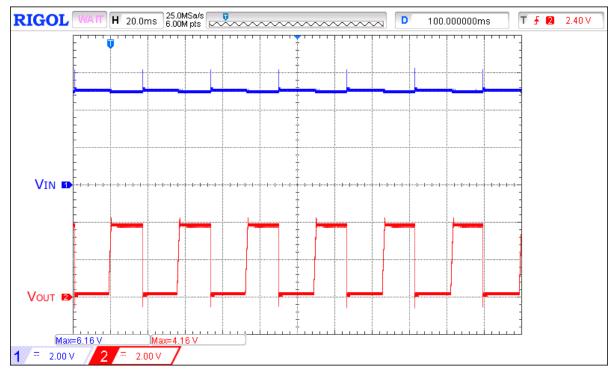


Figure 12. The Parasitic Inductance of 1.2-m Length AWG13 Wires from Power Supply Causes Voltage Spikes at V<sub>IN</sub> during Active Current Limit Operation.



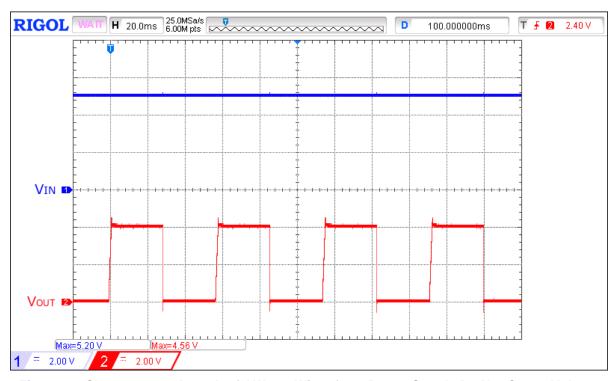


Figure 13. Shorter, 0.2-m Length of AWG13 Wires from Power Supply Do Not Cause Voltage Spikes at V<sub>IN</sub> during Active Current Limit Operation.

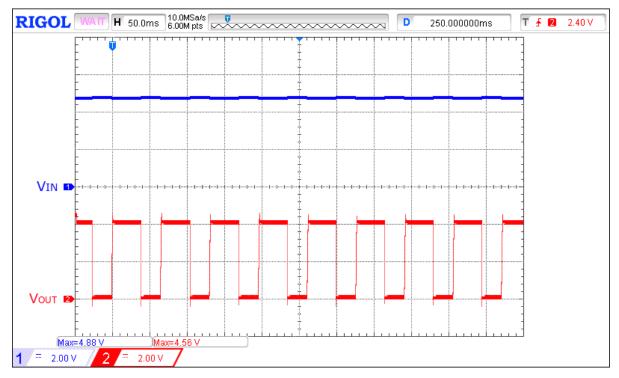


Figure 14. Adding 470 µF Low ESR/ESL Aluminum Polymer Capacitor at VIN Eliminates Voltage Spikes Caused by the Parasitic Inductance of the 1.2-m Length AWG13 Wires from Power Supply during Active Current Limit Operation.



### 6 Conclusions

Capacitors are fundamental components in most digital and analog circuits. GreenFET load switches don't inherently require input and output capacitors. However, application requirements may dictate the use of input and output capacitors. In load switch applications where input and output capacitors are needed, ceramic and/or tantalum capacitors are recommended.

#### **IMPORTANT NOTICE AND DISCLAIMER**

RENESAS ELECTRONICS CORPORATION AND ITS SUBSIDIARIES ("RENESAS") PROVIDES TECHNICAL SPECIFICATIONS AND RELIABILITY DATA (INCLUDING DATASHEETS), DESIGN RESOURCES (INCLUDING REFERENCE DESIGNS), APPLICATION OR OTHER DESIGN ADVICE, WEB TOOLS, SAFETY INFORMATION, AND OTHER RESOURCES "AS IS" AND WITH ALL FAULTS, AND DISCLAIMS ALL WARRANTIES, EXPRESS OR IMPLIED, INCLUDING, WITHOUT LIMITATION, ANY IMPLIED WARRANTIES OF MERCHANTABILITY, FITNESS FOR A PARTICULAR PURPOSE, OR NON-INFRINGEMENT OF THIRD-PARTY INTELLECTUAL PROPERTY RIGHTS.

These resources are intended for developers who are designing with Renesas products. You are solely responsible for (1) selecting the appropriate products for your application, (2) designing, validating, and testing your application, and (3) ensuring your application meets applicable standards, and any other safety, security, or other requirements. These resources are subject to change without notice. Renesas grants you permission to use these resources only to develop an application that uses Renesas products. Other reproduction or use of these resources is strictly prohibited. No license is granted to any other Renesas intellectual property or to any third-party intellectual property. Renesas disclaims responsibility for, and you will fully indemnify Renesas and its representatives against, any claims, damages, costs, losses, or liabilities arising from your use of these resources. Renesas' products are provided only subject to Renesas' Terms and Conditions of Sale or other applicable terms agreed to in writing. No use of any Renesas resources expands or otherwise alters any applicable warranties or warranty disclaimers for these products.

(Disclaimer Rev.1.01)

#### **Corporate Headquarters**

TOYOSU FORESIA, 3-2-24 Toyosu, Koto-ku, Tokyo 135-0061, Japan www.renesas.com

#### **Trademarks**

Renesas and the Renesas logo are trademarks of Renesas Electronics Corporation. All trademarks and registered trademarks are the property of their respective owners.

#### **Contact Information**

For further information on a product, technology, the most up-to-date version of a document, or your nearest sales office, please visit <a href="https://www.renesas.com/contact-us/">www.renesas.com/contact-us/</a>.